Massive Notes 1.3

Key Points: Ways to Improve the Spectral Efficiency; Increase the Transmit Power

1. Ways to Improve the Spectral Efficiency

1. There are different ways to improve the per-cell SE in cellular networks. Consider $\bf a$ two-cell network where the average channel gain between a BS and every UE in a cell is identical, as shown below.

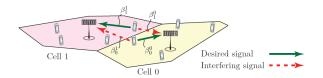


Figure 1: A two-cell network.

- 2.1. β_0^0 denotes the average channel gain from a UE in cell 0 to its serving BS; 2.2. β_1^0 denotes the average channel gain from a UE in cell 1 to BS in cell 0 (i.e., β_1^0 is the interfering signals);
- 2.3. β_1^1 denotes the average channel gain from a UE in cell 1 to its serving BS;
- 2.4. β_0^1 denotes the average channel gain from a UE in cell 0 to BS in cell 1 (i.e., β_0^1 is the interfering signals).
- 3. The average channel gains are **positive dimensionless quantities** that are often very small since the signal energy decays quickly with the propagation distance; values in the range from -70 dB to -120 dB are common within the serving cell, while the interfering signals are even smaller.
- 4. Suppose $\beta_0^0 = \beta_1^1$ and $\beta_1^0 = \beta_0^1$, we can define the **ratio** $\bar{\beta}$ between the inter-cell and intra-cell channel gains as

$$\bar{\beta} = \frac{\beta_1^0}{\beta_0^0} = \frac{\beta_0^1}{\beta_0^0} = \frac{\beta_1^0}{\beta_1^1} = \frac{\beta_0^1}{\beta_1^1} \tag{1}$$

5. Typically we have $0 \le \bar{\beta} \le 1$, where $\bar{\beta} \approx 0$ corresponds to a **negligibly weak inter-cell** interference and $\bar{\beta} \approx 1$ means that the inter-cell interference is as strong as the desired signals (which may happen for UEs at the cell edge).

2. Increase the Transmit Power(1)

6. The SE naturally depends on the strength of the received desired signal, represented by the average SNR (i.e., $\frac{p\mathbb{E}\{|h|^2\}}{\sigma^2}$), the average SNR of a UE in cell 0 therefore is

$$SNR_0 = \frac{p}{\sigma^2} \beta_0^0 \tag{2}$$

where p denotes the UE's transmit power and σ^2 is the noise power. SNR₀ plays a key role in this topic.

7. Suppose there is one active UE per cell and that each BS and UE is equipped with a single patch antenna with a size that is smaller than the wavelength. Channel is flat-fading (i.e., the coherence bandwidth of the channel is larger than the signal bandwidth), so the symbol-sampled complex-baseband signal $y_0 \in \mathbb{C}$ received at the BS in cell 0 is

$$y_0 = \underbrace{h_0^0 s_0}_{\text{Desired signal Interfering signal}} + \underbrace{h_1^0 s_1}_{\text{Noise}} + \underbrace{n_0}_{\text{Noise}}$$
(3)

where $n_0 \sim \mathcal{N}_{\mathbb{C}}(0, \sigma^2)$, and the scalars $s_0, s_1 \sim \mathcal{N}_{\mathbb{C}}(0, p)$ represents the information signals (e.g., QAM signals) transmitted by the desired and interesting UEs, respectively. We consider one model of LoS propagation and one model of NLoS propagation. In single-antenna LoS propagation, we have

$$h_i^0 = \sqrt{\beta_i^0} \quad \text{for} \quad i = 0, 1 \tag{4}$$

- 8. Since **phase rotation** does not affect the SE, it can be neglected. The channel gain β_i^0 can be interpreted as the macroscopic **large-scale fading** in LoS propagation, caused by distance-dependent pathloss. This parameter is **constant** if the transmitter and receiver are **fixed**, while it **changes** if the transmitter and/or receiver **move**.
- 9. Microscopic movements (at the order of the wavelength) can be modeled as phase-rotations in h_i^0
- 10. Large movements (at the order of **meters**) lead to **substantial** changes in β_i^0 . We often consider a fixed value of h_i^0 .

11. In NLoS propagation environments, the channel responses are **random variables** that change over **time** and frequency, and it there is **sufficient scattering** between the UEs and the BS, then

$$h_i^0 \sim \mathcal{N}_{\mathbb{C}}(0, \beta_i^0) \quad \text{for} \quad i = 0, 1$$
 (5)

- (5) is called **Rayleigh Fading** because the magnitude $|h_i^0|$ is a **Rayleigh distributed** random variable.
- 12. The transmitted signal reaches the receiver through **many different paths** and the **superimposed** received signals can either **reinforce** or **cancel** each other. When the number of paths is **large**, the **central limit theorem** motivates the use of a **Gaussian distribution**.
- 13. **Small-scale fading**: a microscopic effect caused by **small variations** in the propagation environment (e.g., movement of the transmitter, receiver, or other objects). The phenomenon that 12 talk about is small-scale fading.
- 14. Large-scale fading: the variance β_i^0 is interpreted as the macroscopic large-fading, which includes distance-dependent pathloss, shadowing, antenna gains, and penetration losses in NLoS propagation.

3. Increase the Transmit Power(2): LoS case

15. Suppose the BS in cell 0 knows the channel responses. An **achievable** UL SE for the desired UE in the LoS case is

$$SE_0^{LoS} = \log_2 \left(1 + \frac{1}{\bar{\beta} + \frac{1}{SNR_0}} \right) = \log_2 \left(1 + \frac{1}{\bar{\beta} + \frac{\sigma^2}{p\beta_0^0}} \right)$$
 (6)

where $\bar{\beta} = \frac{\beta_1^0}{\beta_0^0}$ and $SNR_0 = \frac{p}{\sigma^2}\beta_0^0$, so we have the SINR expression of SE

$$SE_0^{LoS} = \log_2 \left(1 + \frac{p\beta_0^0}{p\beta_0^0} + \frac{\sigma^2}{N_{\text{oise power}}} \right)$$
Interference power Noise power (7)

16. If we increase p, the SE can be improved. However, the SE will NOT increase indefinitely with p. In the LoS, we have

$$SE_0^{LoS} \to \log_2\left(1 + \frac{1}{\overline{\beta}}\right) = \log_2\left(1 + \frac{\beta_0^0}{\beta_1^0}\right) \quad \text{as} \quad p \to \infty$$
 (8)

where the limit is completely determined by the strength of the interference. This is due to the fact that the desired UE and the interfering UE both increase their transmit powers.

Program 1: LoS

Listing 1: LoS.m

```
%Empty workspace and close figures
 2 close all;
 3 | clear;
4
 5 Spefine the SNR range for analytical curves
  |SNRdB| = -10 : 0.1 : 50;
   SNR = 10 .^ (SNRdB/10);
9
   %Define the different beta_bar values (strength of inter—cell interference)
10 | betabar = [1e-1, 1e-3]';
11
12 %Preallocate matrices for storing the simulation results
13 | SE_LoS = zeros(length(betabar), length(SNR));
14
15 % Go through different strengths of the interference
   for b = 1 : length(betabar)
16
17
        %Compute SE under line—of—sight (LoS) propagation as in (1.17)
18
        SE_LoS(b, :) = log2(1 + 1 ./ (betabar(b) + 1 ./ SNR));
19
20
21
   end
22
23
   % Plot the simulation results
24 | figure;
25
   hold on; box on;
26
27
   plot(SNRdB, SE_LoS(1, :), 'r—', 'LineWidth', 1.5);
   plot(SNRdB, SE_LoS(2, :), 'b-.', 'LineWidth', 1.5);
28
29
30 | xlabel('SNR [dB]');
31 | ylabel('Average SE [bit/s/Hz]');
32
33 | legend('beta = 0.1, LoS', 'beta = 0.001, LoS', 'Location', 'SouthEast');
34 |ylim([0, 10]);
```

Output 1:

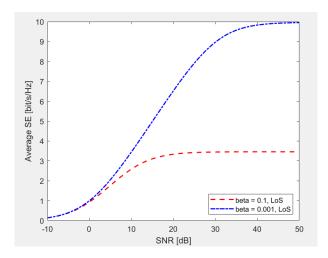


Figure 2: Average UL SE as a function of the SNR for different cases of inter-cell interference strength in LoS model. (END)