



GYTE
Electronics Engineering

ELEC 331
Electronic Circuits 2

Fall 2014

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HW 13
Questions and Answers

Updated January 2, 2015 - 19:50

Assigned: 20141222

Due: 20141229

Answers Out: 20141230

Late Due: 20150105

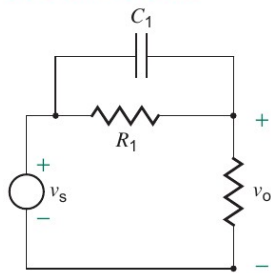
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First Order Filter**Rashid 12.2**

12.2 Determine (a) the transfer function of the network shown in Fig. P12.2 and (b) its poles and zeros.

FIGURE P12.2

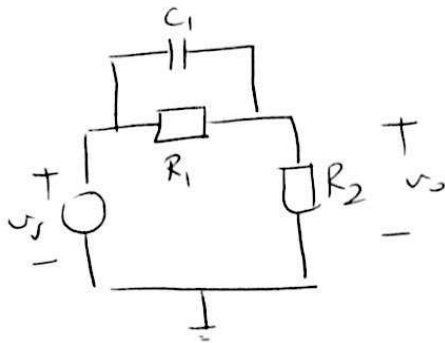


Notes: None.

Additional Tasks: Sketch the magnitude and phase Bode plots.

Necessary Knowledge and Skills: Laplace transforms, transfer functions, poles and zeros, gain, filtering operation.

Rankin
12.2



$$\begin{aligned} \frac{v_o}{v_s} &= \frac{R_2}{R_2 + R_1 \parallel \frac{1}{sC_1}} = \frac{R_2}{R_2 + \frac{R_1 \frac{1}{sC_1}}{R_1 + \frac{1}{sC_1}}} \\ &= \frac{R_2}{R_2 + \frac{R_1}{1 + sC_1 R_1}} \\ &= \frac{R_2 (1 + R_1 C_1 s)}{R_1 + R_2 + R_1 R_2 C_1 s} \\ &= \frac{R_2}{R_1 + R_2} \frac{1 + \frac{s}{1/R_1 C_1}}{1 + \frac{s}{\frac{1}{C_1 (R_1 \parallel R_2)}}} \end{aligned}$$

low frequency gain = $\frac{R_2}{R_1 + R_2}$

zero : $\frac{1}{R_1 C_1}$

pole : $\frac{1}{(R_1 \parallel R_2) C_1}$

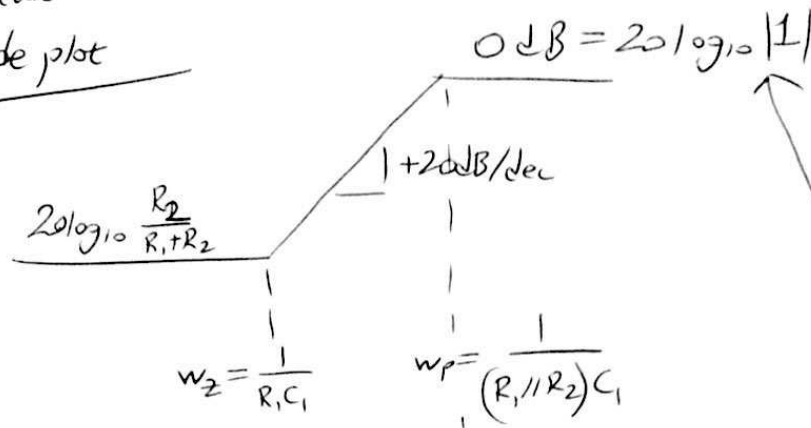
zero > pole

zero < pole \Leftarrow this is correct

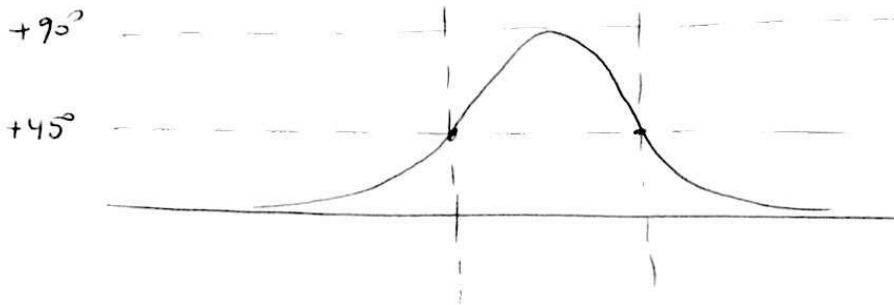
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Rashed
12-2
contin.

Magnitude
Bode plot



phase Bode plot



$$\frac{R_2}{R_1 + R_2} \cdot \frac{R_1 \cancel{C_1}}{(R_1 \parallel R_2) \cancel{C_1}} = \frac{R_2}{R_1 + R_2} \cdot R_1 \cdot \frac{R_1 + R_2}{R_1 R_2} = 1$$

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Biquadratic Function**Rashid 12.5**

12.5 Determine (a) the pole and zero quality factors Q_p and Q_z , (b) the pole and zero resonant frequencies ω_p and ω_z , (c) the pole factor β_p , and (d) the pole angle ϕ_p . The transfer function has the general form as given by

$$H(s) = \frac{5s^2 + 15s + 100}{s^2 + 20s + 200}$$

Notes: See Section 12.5 of Rashid.

Additional Tasks: Switch to Fourier domain and use Matlab to sketch the magnitude and phase Bode plots.

Necessary Knowledge and Skills: Biquadratic function, quality factors, resonant frequencies.

Rashid
12-5

$$H(s) = 5 \frac{s^2 + 3s + 20}{s^2 + 20s + 200}$$

$$= A \frac{s^2 + 2\zeta_z \omega_{n,z} s + \omega_{n,z}^2}{s^2 + 2\zeta_p \omega_{n,p} s + \omega_{n,p}^2}$$

see other questions for the derivations of
BW and Q.

$$BW_z = 2\zeta_z \omega_{n,z} \quad Q_z = \frac{\omega_{n,z}}{BW_z}$$

$$= \frac{\sqrt{20}}{3} = \frac{2\sqrt{5}}{3}$$

$$BW_p = 2\zeta_p \omega_{n,p} \quad Q_p = \frac{\omega_{n,p}}{BW_p}$$

$$= \frac{\sqrt{200}}{20} = \frac{10\sqrt{2}}{20} = \frac{\sqrt{2}}{2}$$

Examine the denominator of $H(s)$:

$$s^2 + 20s + 200 = (s + \alpha_p + j\beta_p)(s + \alpha_p - j\beta_p)$$

$$= s^2 + 2\alpha_p s + \alpha_p^2 + \beta_p^2$$

$$\alpha_p^2 + \beta_p^2 = \omega_{n,p}^2 = 200$$

$$2\alpha_p = 20 \Rightarrow \alpha_p = 10$$

$$\beta_p = \sqrt{200 - (10)^2} = 10$$

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poles

$$p_1 = -\alpha_p - j\beta_p$$

$$p_2 = -\alpha_p + j\beta_p$$

$$\frac{\text{Rashid}}{12.5}$$

contin.

pole angle given as

$$\phi_p = \cos^{-1} \left(\frac{\alpha_p}{\sqrt{\alpha_p^2 + \beta_p^2}} \right) = \cos^{-1} \left(\frac{\alpha_p}{\omega_p} \right)$$

compute!




```

01/02/15 c12_q005_plots.m 1
!cl .

close all
clear classes
clear all

f = logspace(-16,16,10000);
w = 2*pi*f;

H = ...
    ( 5*(1j*w).^2 + 15*(1j*w) + 100 ) ...
    ./ ...
    ( 1*(1j*w).^2 + 20*(1j*w) + 200 );

% plots
myLineWidth = 4;
myFontSize = 24;

figure(101);
h1 = ...
    semilogx ...
    ( ...
    f , 20*log10(abs(H)) , ...
    'b' , ...
    'LineWidth' , myLineWidth ...
    );

grid on;

set(gca,'units','normalized')
set(gca,'Box','on','FontName','Arial',...
    'FontSize',myFontSize,'FontWeight','bold','LineWidth',4)

xlabel('Frequency (Hz)');
ylabel('Magnitude (dB)');
title('Magnitude Bode Plot')

legend(...
    h1 , ...
    'Bode Plot' , ...
    'Orientation' , 'Horizontal' , ...
    'Location' , 'South' ...
    );

figure(102);
h2 = ...
    semilogx ...
    ( ...
    f , 180 / pi * phase(H) , ...
    'r' , ...
    'LineWidth' , myLineWidth ...
    );

grid on;

axis([ 1e-20 1e+20 -20 120 ])

set(gca,'units','normalized')
set(gca,'Box','on','FontName','Arial',...
    'FontSize',myFontSize,'FontWeight','bold','LineWidth',4)

xlabel('Frequency (Hz)');
ylabel('Phase (degrees)');
title('Phase Bode Plot')

legend(...
    h2 , ...
    'Bode Plot' , ...
    'Orientation' , 'Horizontal' , ...
    'Location' , 'South' ...
    );

```

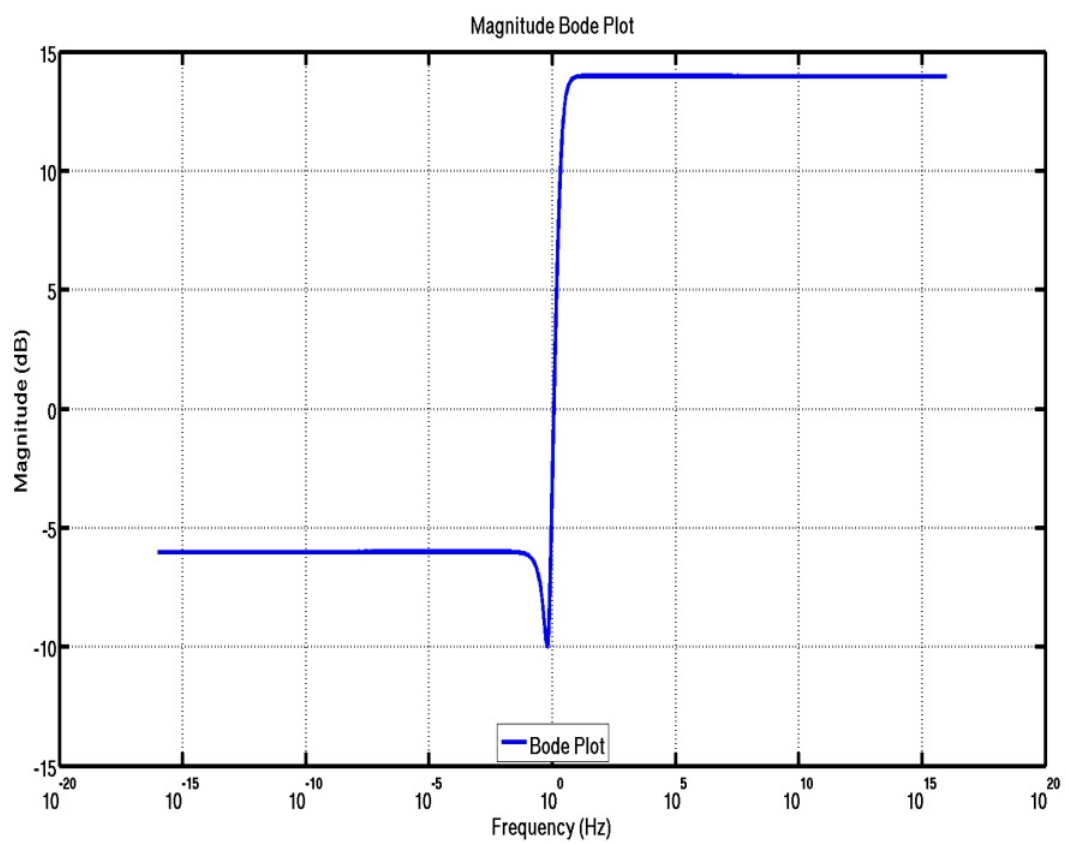
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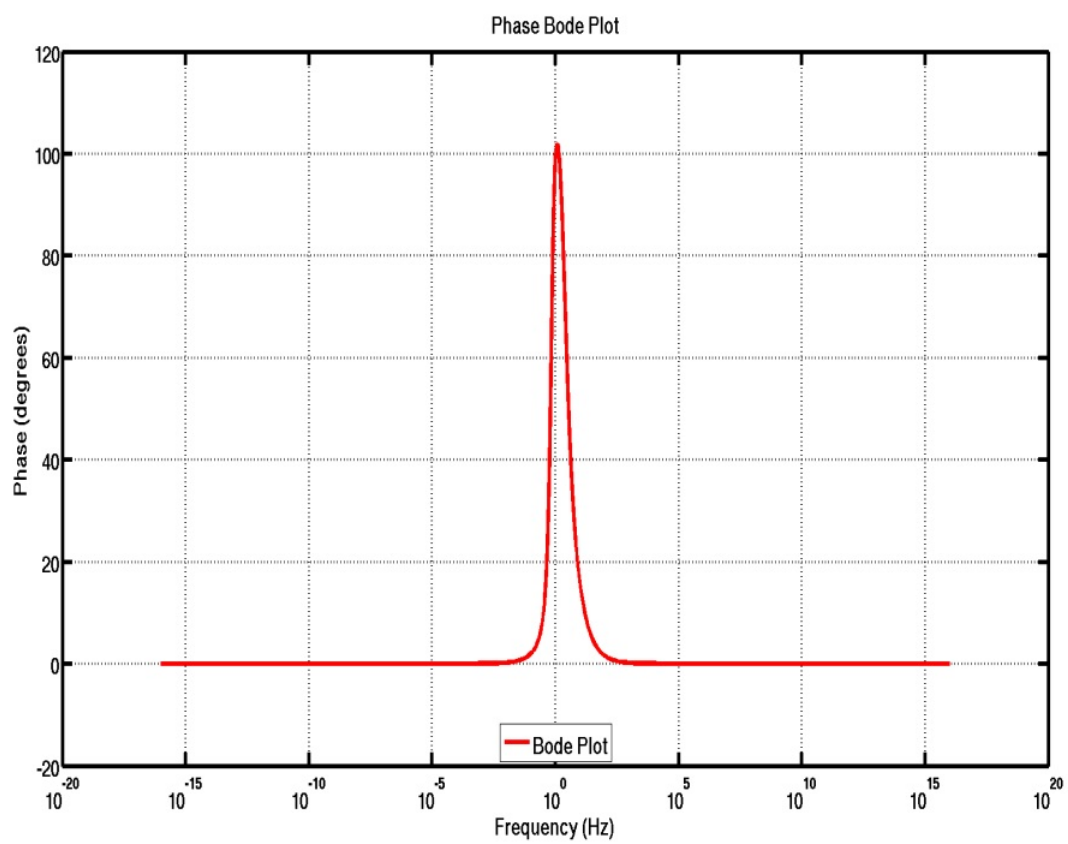
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2

);





LC Resonator

Sedra 12.73

12.73 A coil having an inductance of $10\ \mu\text{H}$ is intended for applications around 1-MHz frequency. Its Q is specified to be 200. Find the equivalent parallel resistance R_p . What is the value of the capacitor required to produce resonance at 1 MHz? What additional parallel resistance is required to produce a 3-dB bandwidth of 10 kHz?

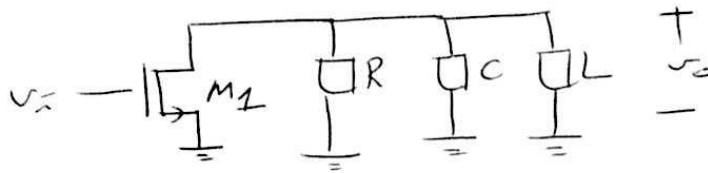
Notes: None.

Additional Tasks: None.

Necessary Knowledge and Skills: LC resonators, bandwidth, quality factor Q , resonance frequency.

See pg 1141-1142 of Sedra and Smith.
Consider the following tuned amplifier.

Sedra
12-73



$$\frac{v_o}{v_i} \approx -g_m \left[R \parallel \frac{1}{sC} \parallel sL \right]$$

$$= -g_m \frac{1}{\frac{1}{R} + sC + \frac{1}{sL}}$$

$$= -g_m \frac{sL}{s \frac{L}{R} + s^2 LC + 1}$$

$$= -g_m \frac{s}{\frac{s}{R} + s^2 C + \frac{1}{L}}$$

$$= -\frac{g_m}{C} \frac{s}{s^2 + \frac{s}{RC} + \frac{1}{LC}}$$

$$(\text{resonance freq})^2 = \omega_0^2 = \frac{1}{LC}$$

$$2\delta\omega_0 = \frac{1}{RC}$$

$$2\delta \frac{1}{\sqrt{LC}} = \frac{1}{RC} \Rightarrow \delta = \frac{1}{2R} \sqrt{\frac{L}{C}}$$

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$$\frac{v_o}{v_i} = - \frac{g_m}{C} G(s) \frac{1}{2f\omega_0}$$

Sedra
12-73
Contin.

where $G(s) = \frac{2f\omega_0 s}{s^2 + 2f\omega_0 s + \omega_0^2}$

Now will compute the bandwidth of the tuned amplifier by calculating high- and low-frequency cut-offs.

$$G(j\omega) = \frac{j\omega 2f\omega_0}{(\omega_0^2 - \omega^2) + j2f\omega_0 \omega}$$

Note that the midband gain is

$$\begin{aligned} \frac{v_o}{v_i} &= - \frac{g_m}{C} \frac{1}{2f\omega_0} \\ &= - \frac{g_m}{C} R \cancel{\phi} \\ &= -g_m R \end{aligned}$$

$$\left| G(j\omega_{-3dB}) \right|^2 = \frac{(2f\omega_0)^2 \omega_{-3dB}^2}{(\omega_0^2 - \omega_{-3dB}^2)^2 + (2f\omega_0 \omega_{-3dB})^2} = \frac{1}{2}$$

If the above eqn is solved for ω_{-3dB} (as done for the series RLC network)

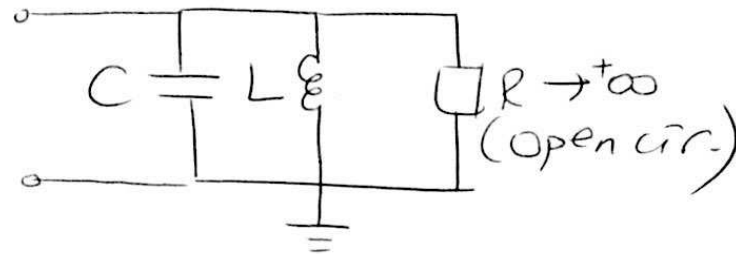
$$\begin{aligned} \text{low-freq cutoff } \omega_L &= \omega_0 \left[-\delta + \sqrt{1 + \delta^2} \right] \\ \text{high-freq cutoff } \omega_H &= \omega_0 \left[\delta + \sqrt{1 + \delta^2} \right] \end{aligned}$$

$$\begin{aligned} \text{Bandwidth } BW &= \omega_H - \omega_L = 2f\omega_0 \\ &= \cancel{\frac{1}{RC}} \end{aligned}$$

Note that $R \rightarrow +\infty \Rightarrow BW \rightarrow 0$

Sedra
12-73
contin.

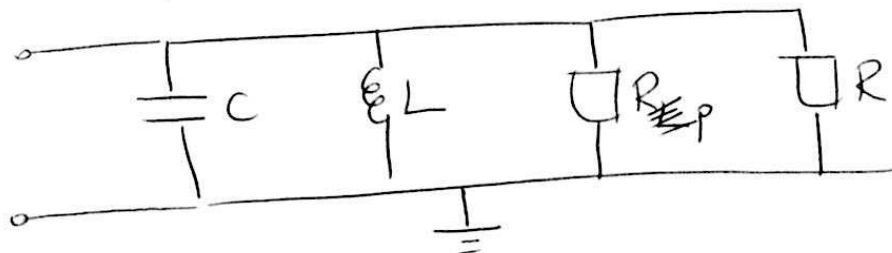
The ideal LC tank is



$$\text{quality factor } Q = \frac{\omega_0}{BW} = \frac{1/\sqrt{LC}}{1/RC} = R \sqrt{\frac{C}{L}}$$

Note that $R \rightarrow +\infty \Rightarrow Q \rightarrow +\infty$ (ideal)

In the question we have



$L = 1 \mu\text{H}$ (given)

$$\omega_0 = \frac{1}{\sqrt{LC}} = 2\pi \cdot 1\text{MHz (required)}$$

compute C.

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(with R_{open} cir.) $Q = R_p \sqrt{\frac{C}{L}} = 200$ (given)

C computed
 L known

compute R_p .

Sabra
12-73
contin.

(with R not open cir.)

$$BW = 2\pi \cdot 10 \text{ KHz}$$

$$= \frac{1}{(R_p // R) C}$$

C computed.

R_p computed.

R compute \Leftarrow

