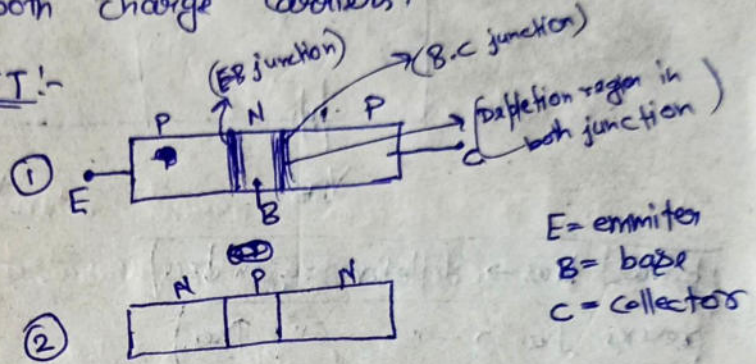


## "Unit - 2" Transistor

# BJT (Bipolar junction transistor):- Bipolar represents that the current is due to both majority and minority both charge carriers.

basic structure of BJT:-

2 types:-



Size of collector > E & B  
Size C > E > B

\* Emitter is highly doped, charge carrier is injected by emitter.  
\* Base:- width of base is very small compared to emitter and collector. It is lightly doped i.e. doping conc. of collector is low. It is used to control the flow of majority carrier injected from emitter.

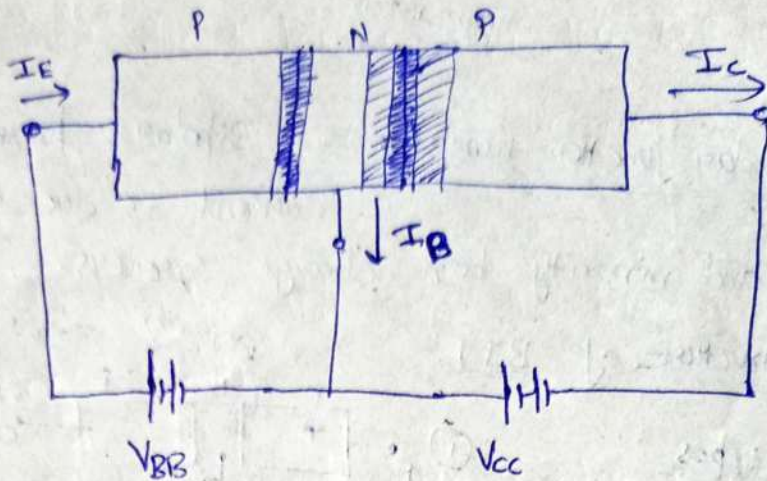
\* Collector:- Size of collector is large compared to emitter & base. because this collector is used to collect majority carriers from emitter so that may generate the heat so large area required. to dissipate heat from surface.

doping conc. is  $b/w^n$  emitter & base

It is used to collect charge carriers generated from emitter.  
It dissipate the heat.



① PNP



forward bias  $\rightarrow$  depletion region  $\downarrow$   $I_{AS}$   
 reverse bias  $\rightarrow$  " "  $\uparrow$   $I_{AR}$

E-B = forward bias  
 B-C = reverse bias

★

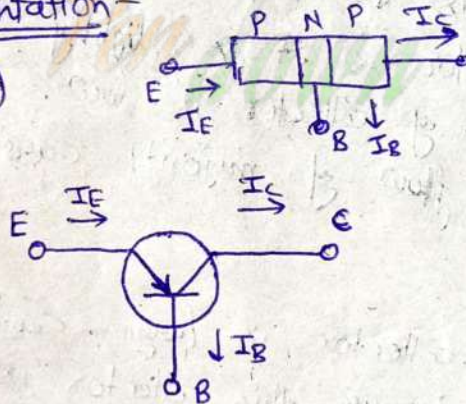
$$I_E = I_B + I_C$$

$$I_B = 2 \text{ to } 5\% \text{ of } I_E$$

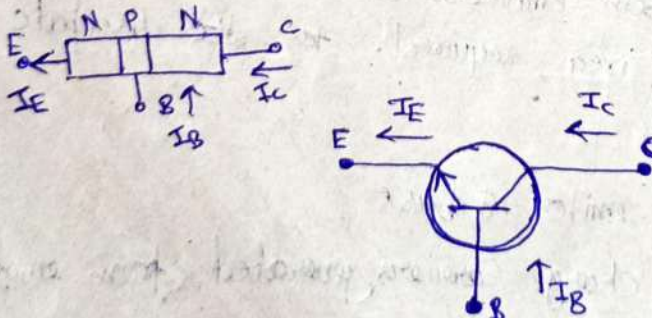
$$I_C = 95 \text{ to } 98\% \text{ of } I_E$$

★ Symbolic representation -

①

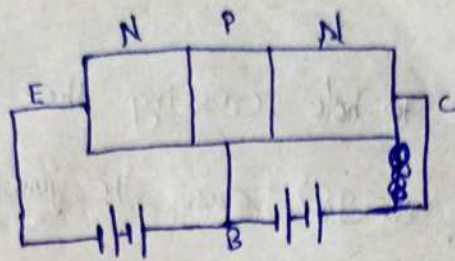


②





② NPN



(Transistor as amplifier)

# depending upon the different application these are 4 combination of voltage across diodes:



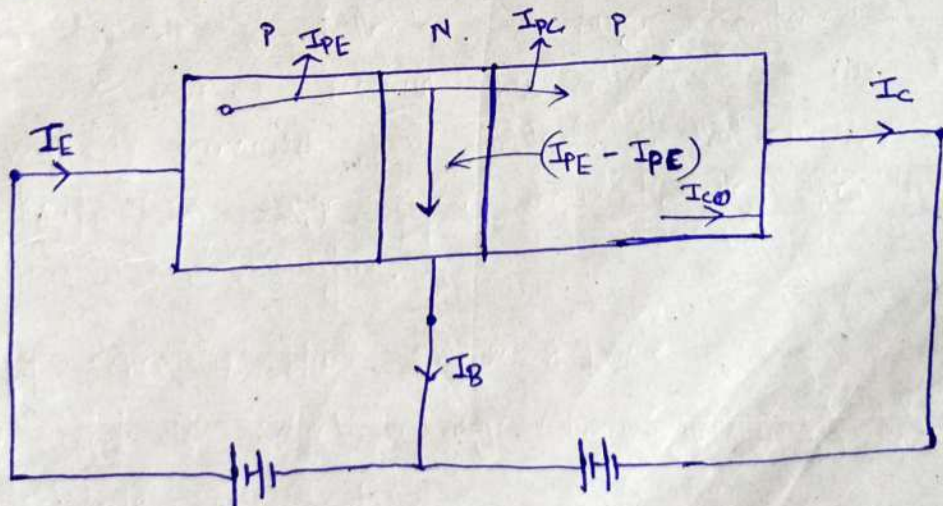
E-B Junction

B-C junction

Mode of operation

① Forward	Reverse	Active Mode	(Amplifier)
② Forward	Forward	Saturation <del>region</del> mode (Switch ON)	
③ Reverse	Reverse	Cut off mode (OFF)	
④ Reverse	Forward	Inverted mode ( )	

# Discuss on different current components -





(a)  $I_{FE}$  is current due to hole crossing the E-B junction.

(b) only 2% stops in base & 98% cross the junctions & collected by collector.

(c)  $I_{FC}$  is due to holes that cross the base and collected by collector.

(d)  $I_{CO}$  is collector current when emitter is open and is due to thermally generated minority charge carriers.

In  $I_{CO}$  there are 2 types of minority charge carriers  $\rightarrow$

(1) hole crossing base to collector region

(2) thermally generated  $e^-$  in C area passing collector to base region.

hence  $I_{CO}$  is combo of 2 current.

①

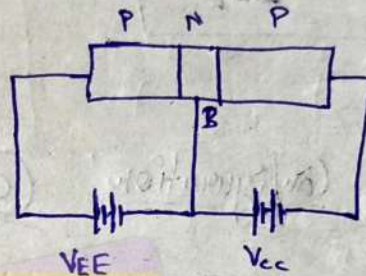
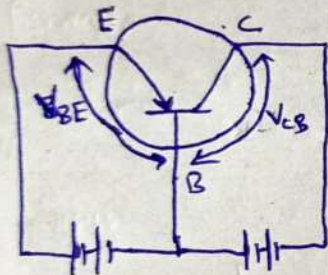


# # Different Configuration of Transistor!

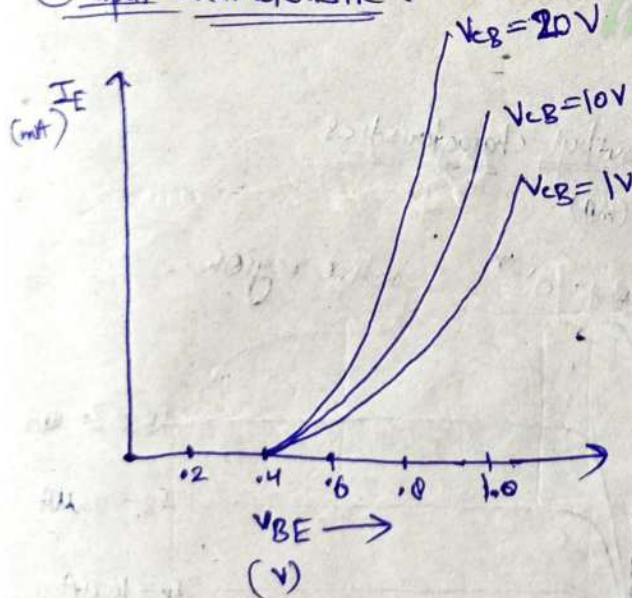
11/10/22

- ① CB (Common Base Configuration)
- ② CE (" emitter confi ")
- ③ CC (" collector " )

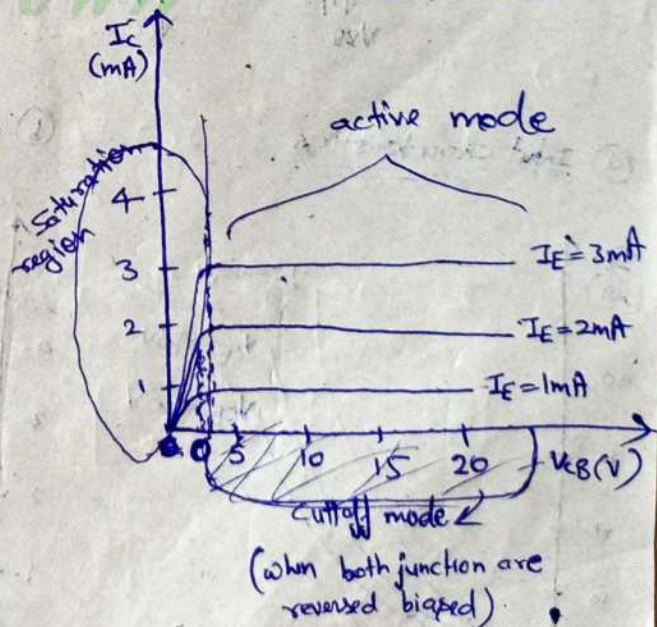
## ① CB!:-



### ① Input characteristic :-



### ② output characteristics



(when both junction are reversed biased)

Complete area of active mode is active region

In saturation region both junction forward biased

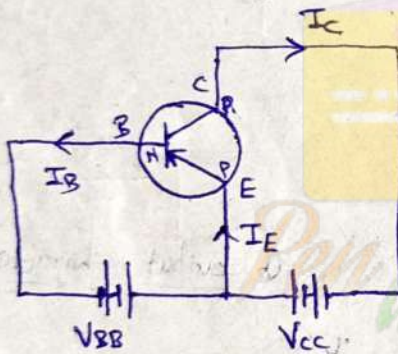


#  $I_E = I_C + I_B$

①  $\alpha = \frac{I_C}{I_E}$  (ratio of output / input current)

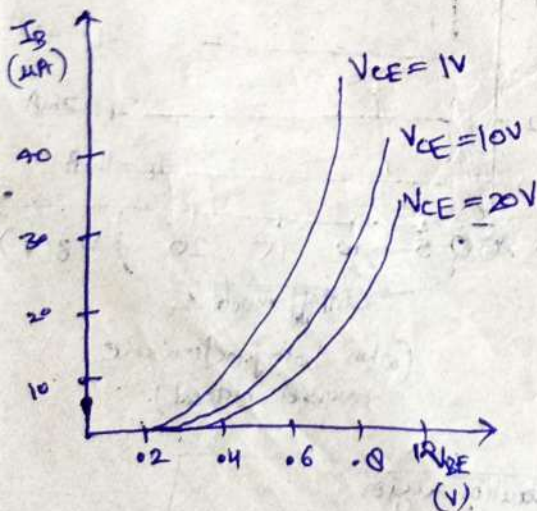
②  $\alpha_{ac} = \frac{\Delta I_C}{\Delta I_E}$  (ratio of change of o/i current)

## ② Common emitter Configuration: (CE)

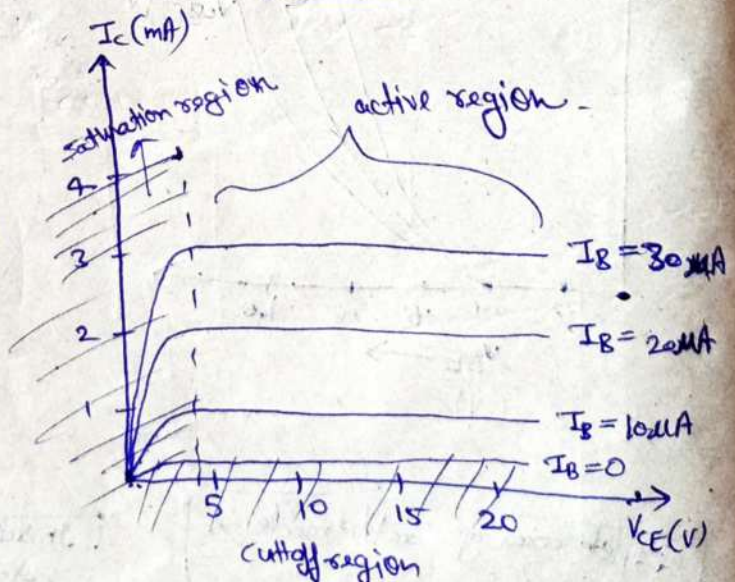


$I_E = I_C + I_B$

### ① Input characteristics



### ② output characteristics



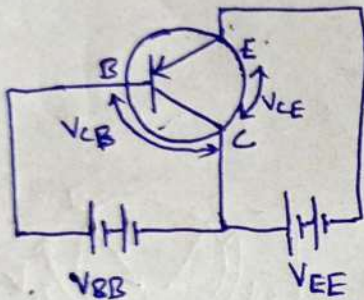


# Current gain ( $\beta$ )

$$\textcircled{1} \beta_{DC} = \frac{I_C}{I_B}$$

$$\textcircled{2} \beta_{AC} = \frac{\Delta I_C}{\Delta I_B}$$

③ CC:-



# Current gain ( $\gamma$ )

$$\textcircled{1} \gamma_{DC} = \frac{I_E}{I_B}$$

$$\textcircled{2} \gamma_{AC} = \frac{\Delta I_E}{\Delta I_B}$$

$$I_E = \alpha I_C$$

$$I_C = \beta I_B$$

$$I_E = \gamma I_B$$

$$\alpha = \frac{I_C}{I_E}$$

$$\beta = \frac{I_C}{I_B}$$

$$\gamma = \frac{I_E}{I_B}$$

# rel<sup>n</sup> b/w  $\alpha, \beta$  &  $r$

$$\alpha = \frac{I_c}{I_E} = \frac{I_c}{I_c + I_B}$$

$$(I_E = I_c + I_B)$$

$$\frac{1}{\alpha} = 1 + \frac{I_B}{I_c}$$

$$\boxed{\frac{1}{\alpha} = 1 + \frac{1}{\beta} = \frac{1+\beta}{\beta}}$$

$$\alpha = \frac{\beta}{1+\beta}$$

Similarly,

$$\boxed{\beta = \frac{\alpha}{1-\alpha}}$$

Now

$$r = \frac{I_E}{I_B} = \frac{I_c + I_B}{I_B} = 1 + \frac{I_c}{I_B}$$

$$1 + \frac{1}{\beta}$$

$$\boxed{r = 1 + \beta}$$

hence

$$\boxed{r = \frac{1}{1-\alpha}}$$



#  $I_C = \alpha I_E + I_{CBO}$  → for common base

#  $I_C = \beta I_B + I_{CEO}$  → for common emitter

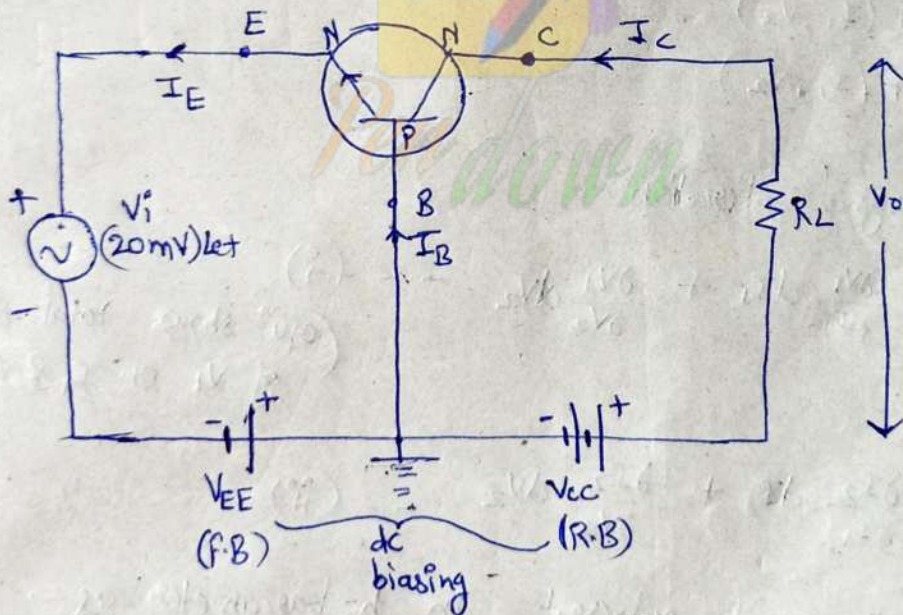
$I_{CBO} = I_{CO}$  is the collector to base current when emitter is open

$\alpha$  is the factor will collected by collector

ie  $\alpha$  is amount of majority carriers cross barrier

# Transistor as amplifier :- (in Active region)

$I_1$  = Forward bias  
 $I_2$  = Reverse bias



★ Voltage Amplification factor :-

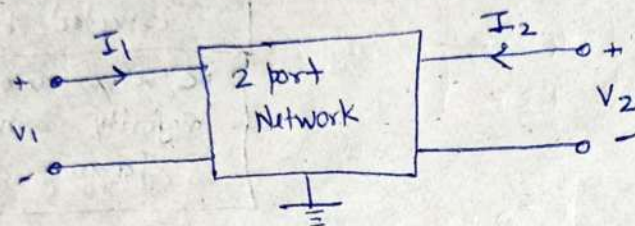
$$A_v = \frac{V_o}{V_i}$$

★ Current Amplification factor =  $\alpha \in (0.95 \text{ to } 0.98)$   
( $dI =$ )



## # Hybrid Parameters (H-Parameters)

h-parameters are applicable to any 2-port Network.



$V_1$  and  $I_2$  are dependable variables  
 $V_2$  and  $I_1$  are independent variables

so

$$V_1 = f(I_1, V_2)$$

$$I_2 = f(I_1, V_2)$$

for small signal operation

$$dV_1 = \frac{\partial V_1}{\partial I_1} dI_1 + \frac{\partial V_1}{\partial V_2} dV_2 \quad \text{--- (i)}$$

(eqn shows total dependence of  $V_1$  on  $I_1$  &  $V_2$ )

$$dI_2 = \frac{\partial I_2}{\partial I_1} dI_1 + \frac{\partial I_2}{\partial V_2} dV_2 \quad \text{--- (ii)}$$

these 2 eqn can be represent in h-parameter as—

$$V_1 = h_{11} I_1 + h_{12} V_2$$

$$I_2 = h_{21} I_1 + h_{22} V_2$$



## Calculation of h-parameters -

step (i) short ckt output i.e.  $V_2 = 0$  for a.c.

then eqn (i)

$$h_{11} = \left. \frac{V_1}{i_1} \right|_{V_2=0} = \text{Input impedance } (h_i)$$

in eqn (ii)

~~scribble~~

$$h_{21} = \left. \frac{i_2}{i_1} \right|_{V_2=0} = \text{forward current gain } (h_f)$$

step (ii)

$$i_1 = 0$$

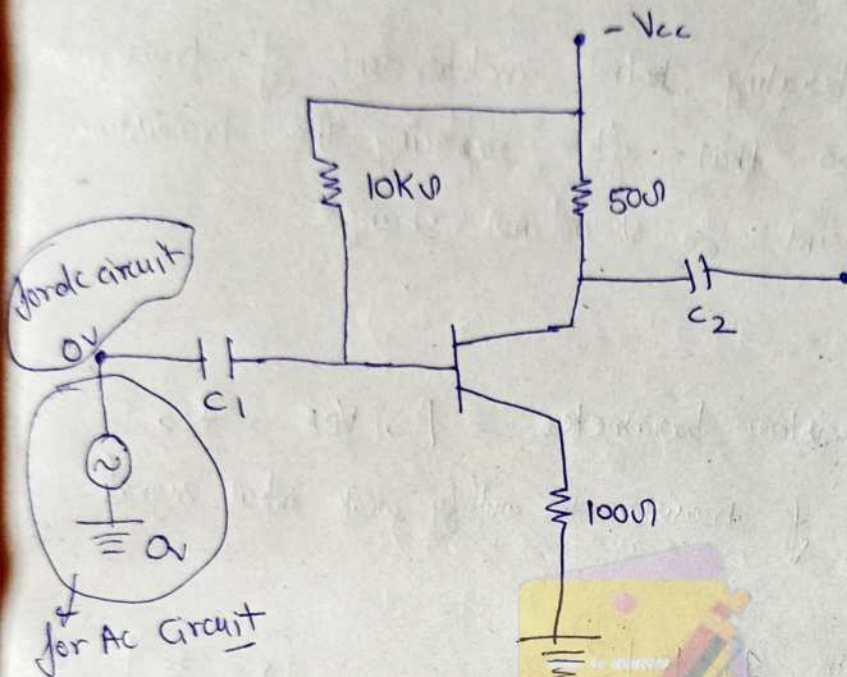
(input is open ckt)

$$h_{12} = \left. \frac{V_1}{V_2} \right|_{i_1=0} = \text{Reverse voltage gain } (h_r)$$

$$h_{22} = \left. \frac{i_2}{V_2} \right|_{i_1=0} = \text{output Admittance } (h_o)$$



## # Transistor biasing -



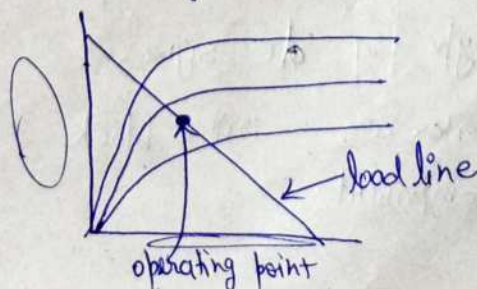
① why biasing required?

Ans

To active an transistor. so that we can apply input signal the operating point doesnot move either in Saturation or Cutoff mode.

i.e we make Q-point independent of device parameters variation so that it doesnot shift.

② (i) stabilize operating point in center of active region





(ii) stabilise the collector current against the temperature variation. ( $10^\circ\text{C}$  in temp doubles the current)

(iii) make the operating point independent of transistor parameters so that after replacing the transistor operational point ~~do~~ should not change

② what are transistor parameter  $\beta$ ,  $V_{BE}$ ,  $I_{CO}$ ?

Ans)  $\beta$  = capability of transistor to amplify weak input signal

③ why  $\beta \uparrow$  with  $\uparrow$  in temp?

Ans) as  $T \uparrow$   $\beta \uparrow$ ,  $V_{BE} \downarrow$ ,  $I_{CO} \uparrow$

$$\beta = \frac{I_C}{I_B}$$

so we try to make it temp independent

as on  $\uparrow$  temp, the transport factor reduces in turn causing decrease in base current therefore increase in  $\beta$

④ what is the state of  $C_1$  &  $C_2$  in given ques circuit?

Ans) we get AC input ( $C_1$ ) then amplified AC signal at output ( $C_2$ ) and capacitor ( $C_1$ ) block any input type of DC signal

Similarly capacitor  $C_2$  will block any output DC signal component



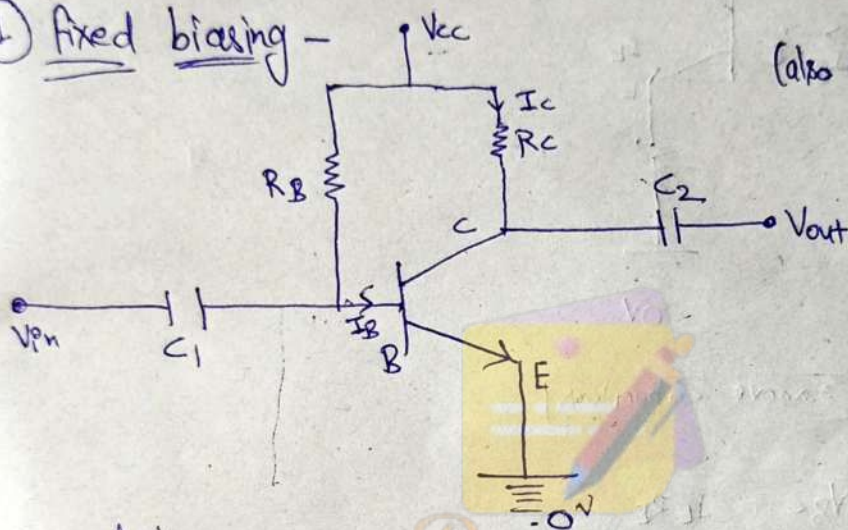
for AC component both  $C_1$  &  $C_2$  will behave as  
conducting wire



## # Types of biasing in Transistors

- ① fixed
- ② self bias
- ③ voltage divider biasing

### ① fixed biasing -



(also known as base-bias ckt)

$$V_{cc} = I_B R_B$$

$$V_{CE} = V_{cc} - I_C R_C$$

dc analysis -

$$V_{cc} = I_B R_B + V_{BE}$$

~~$$I_B = \frac{V_{cc} - V_{BE}}{R_B}$$~~

$$I_B = \frac{V_{cc} - V_{BE}}{R_B}$$

$$V_{BE} \approx 0$$

$$I_B = \frac{V_{cc}}{R_B}$$

This is called fixed biasing  
because the value of  $I_B$  is fixed.

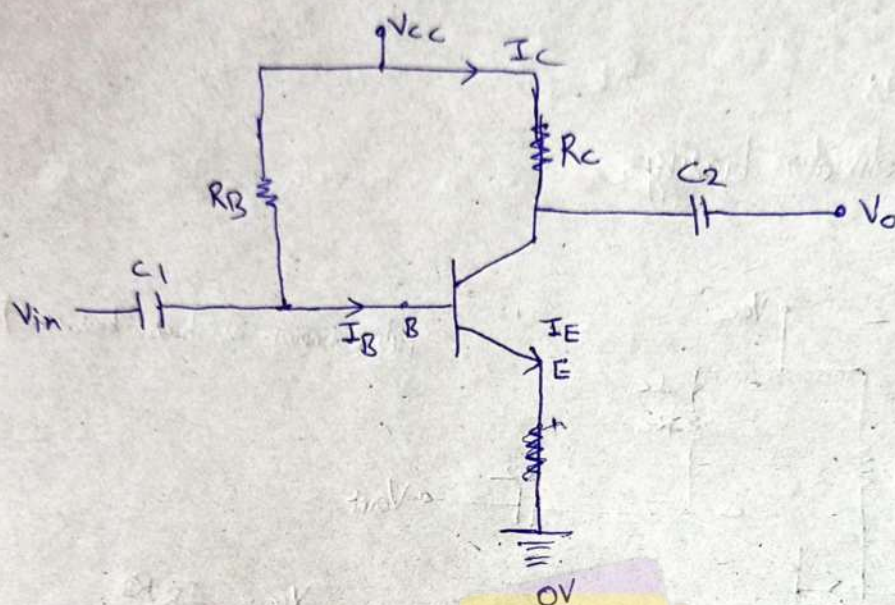
$$V_{cc} = I_C R_C + V_{CE} \rightarrow \text{Terminal voltage}$$

$$V_{CE} = V_{cc} - I_C R_C$$

$$I_C = \beta I_B$$



## ② Emitter self biasing -



dc analysis → remove capacitors

$$V_{CC} = I_B R_B + V_{BE} + I_E R_E$$

$$\boxed{I_E = (1 + \beta) I_B}$$

$$V_{CC} = I_B R_B + V_{BE} + (1 + \beta) I_B R_E$$

$$\boxed{I_B = \frac{V_{CC} - V_{BE}}{R_B + (\beta + 1) R_E}} \rightarrow \text{To calculate value of } I_B$$

then

$$\boxed{I_C = \beta I_B + (1 + \beta) I_{CBO}}$$

if  $I_{CBO}$  not given

$$\boxed{I_{CBO} = 0}$$

$$V_{CC} = I_C R_C + V_{CE} + I_E R_E$$

$$V_{CE} = V_{CC} - R_C I_C - I_E R_E$$



$$\text{If } I_C \approx I_E$$

$$V_{CE} = V_{CC} - I_C R_C + R_E$$

then the role of feedback is to provide output from input.

$$I_E = (1 + \beta) I_B$$

$$I_B = \frac{V_{CC} - I_{BE}}{R_B + (1 + \beta) R_C}$$

$$V_{CC} = I_B R_B + I_E R_E$$

$$V_{CE} = V_{CC} - I_C R_C - I_{BE}$$

$$R = \frac{V_{CC} - V_{BE}}{R_B + R_E (1 + \beta)}$$