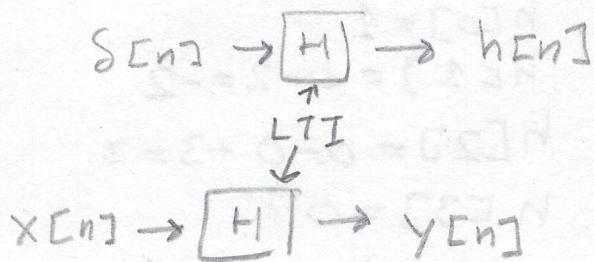


DSP Lect 3

1

Impulse response

Impulse response



$$Y[n] = X[n] * h[n] = \sum_{k=-\infty}^{\infty} x[k] h[n-k]$$

Consider this LTI System

$$Y[n] = X[n] - 2X[n-1] + 3X[n-2]$$

What is the response to

$$X[n] = \frac{1 \ 1 \ 1 \ 1}{1 \ 2} \quad ?$$

1) Direct

$$Y[-1] = 0$$

$$Y[0] = 1$$

$$Y[1] = 1 - 2(1) = -1$$

$$Y[2] = -2 + 3 = 1$$

$$Y[3] = 3$$

Instead, write

$$X[n] = \underline{\underline{1}} \ 1 \ 1 \ 1$$

(double underline means $n=0$)
what's at $n=0$? 1, 1

So, the output for $y[n] = \underline{\underline{1}} \ -1 \ 2 \ 1 \ 3$

② Using convolution sum

$$Y[n] = \sum_{k=-\infty}^{\infty} X[k] h[n-k]$$

What is the impulse response?

$$\{x[n] = [1 \underline{-2} 3]\}$$

$$h[0] = 1$$

$$h[1] = 0 - 2 = -2$$

$$h[2] = 0 - 0 + 3 = 3$$

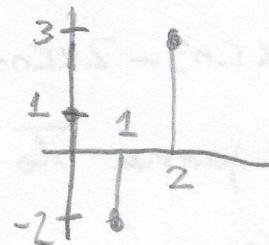
$$h[3] = 0$$

So for the impulse response; recall

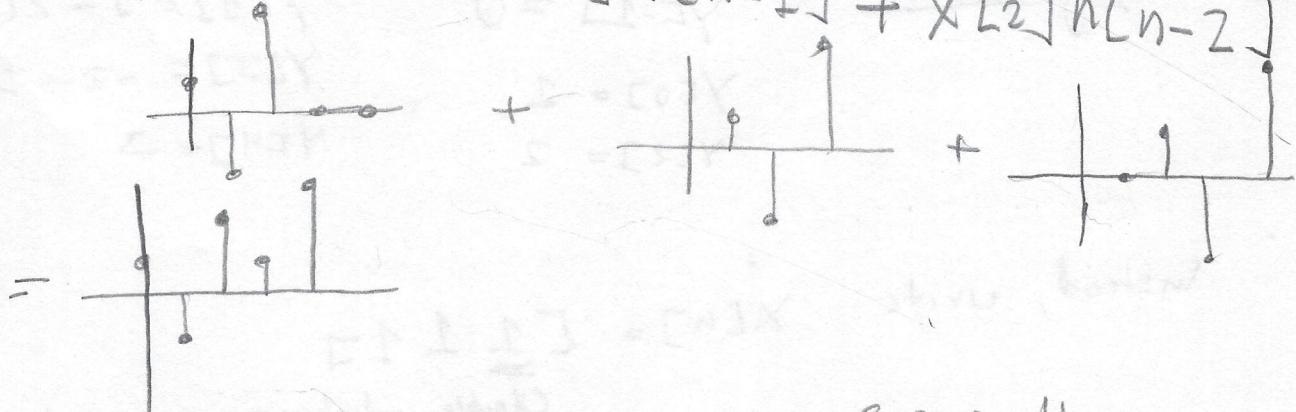
$$\xrightarrow{\text{Seq}} \boxed{1} - h[n]$$

↑
impulse ↑
 response

$$\rightarrow [1 \underline{-2} 3]$$



$$y[n] = x[0]h[n] + x[1]h[n-1] + x[2]h[n-2]$$



$$\begin{aligned}
 & [1 \underline{-2} 3] \\
 & + [0 \underline{1} -2 3] \\
 & + [0 \underline{0} 1 -2 3] \\
 \hline
 & [1 \underline{-1} 2 1 3]
 \end{aligned}$$

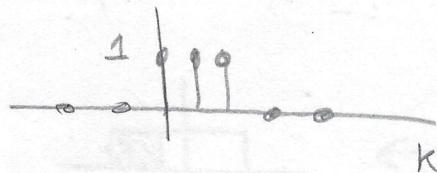
Easier than
writing the stuck

3) FLIP and SLIDE one signal 12:49

(3)

$$\sum_{k=-\infty}^{\infty} x[k] h[n-k]$$

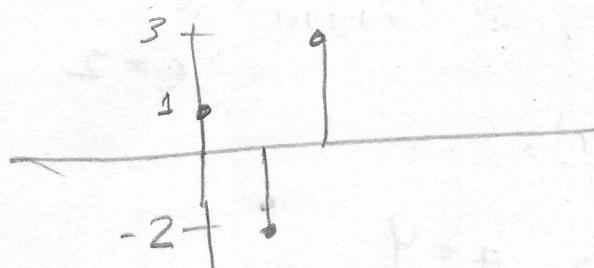
$x[k]$



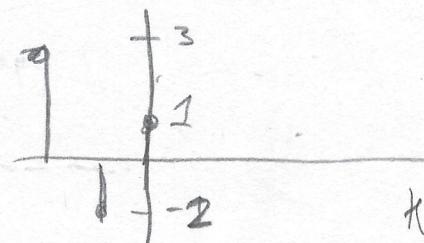
Well what's $h[n-k]$?

k is like my dependant variable, n is const.

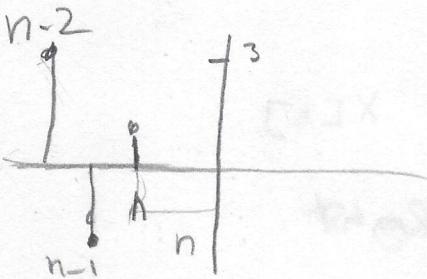
$h[k]$



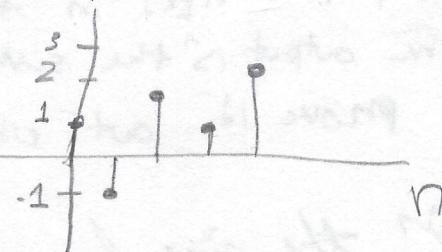
and $\Sigma h[-k]$



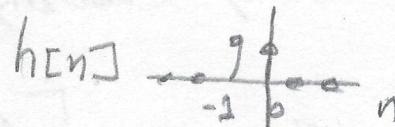
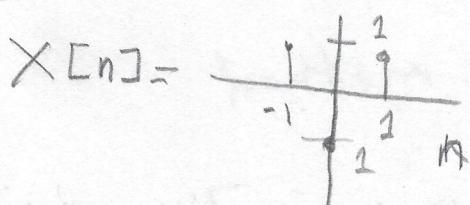
$\Sigma h[n-k]$



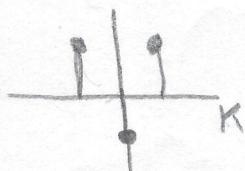
So the plotted Σ



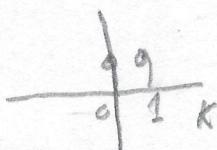
Show it and join it through $x[k]$



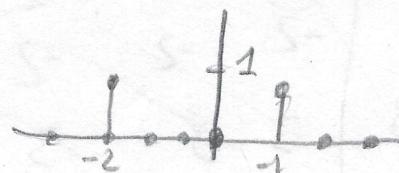
$x[k]$



$h[-k]$



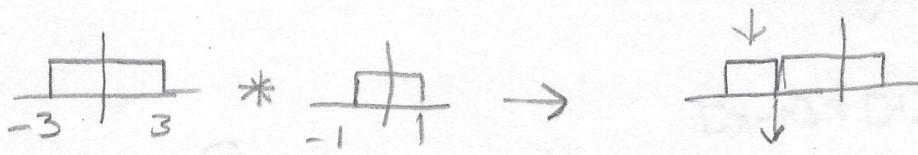
$x[n]*h[n]$



$n = -1$ at $n[n]$ 1st one!

$\therefore n+1 = -1$

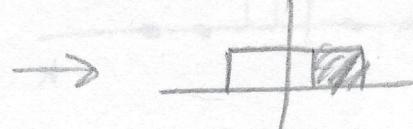
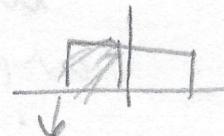
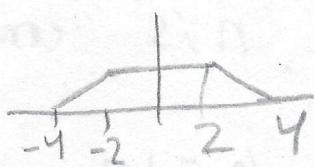
(4) recall the picture method



$$t+1 \quad t-1$$

$$t+1 = -3$$

$$t = -4$$



$$t+1 = -3 \quad t = -2$$

$$t+1 =$$

$$t+1 = 3$$

$$t = 2$$



for $\begin{array}{|c|} \hline 1 & 1 & 1 \\ \hline \end{array}$ and $\begin{array}{|c|} \hline 1 & 1 \\ \hline \end{array}$ $t-1 = 3 \quad t = 4$

Put $h[k]$ in the middle of $x[k]$
→ The output is the sum

move it out either Left and Right

Or, the signal meets at, -2

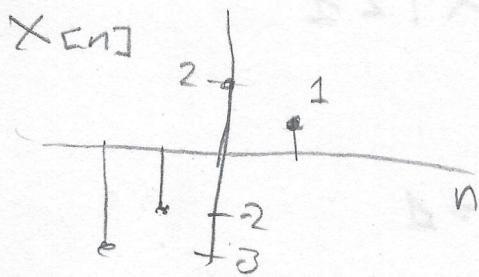
4) convolution away (Wozny) method

h	X	
1	1 2 1	
1	1 1 1	
-2	-2 -2 -2	
3	3 3 3	

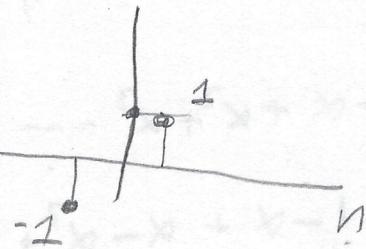
zero element of output

$$\text{Ex 1} \quad \begin{array}{c|cccc} x & 1 & -1 & 1 \\ \hline 1 & 1 & -2 & 1 \\ 1 & 1 & -1 & 1 \end{array} \rightarrow [1 \ 0 \ \underline{\underline{0}} \ 1] \quad \text{correct}$$

Ex 2



$h[n]$



$$\begin{array}{c|cccc} x & -3 & -2 & 2 & 1 \\ \hline -1 & +3 & 2 & -2 & -1 \\ 1 & -3 & -2 & 2 & 1 \\ 1 & -3 & -2 & 2 & 1 \end{array} = [+3 \ -1 \ -7 \ \underline{\underline{1}} \ 3 \ 1]$$

But we may deal with infinitely long inputs

Ex $x[n] = \sum_{k=0}^n u[k] \quad x \in (0,1)$
 $h[n] = u[n]$

$$h[n] = \underbrace{1 \ 1 \ 1 \ 1 \dots}_{\text{doesn't matter which signal we flip}}$$

$$x[n] = \underbrace{\alpha^n}_{\text{choose } h[n]}$$

flip $h[n]$ then rem it into $x[n]$

so we get $y[0]=1$, $y[1]=1+\alpha$, $y[2]=1+\alpha+\alpha^2$

$$⑥ \sum_{n=0}^{\infty} x^n = \frac{1}{1-x}, \quad n \geq 0$$

$$\text{or write} \quad = \left(\sum_{k=0}^n \alpha^k \right) u[n]$$

$$\rightarrow \text{recall } \sum_{k=0}^{\infty} \alpha^k = \frac{1}{1-\alpha} \quad |\alpha| < 1$$

$$(1 + x + x^2 + x^3 + \dots) \in (1-x)^{-1}$$

$$= 1 - \alpha + \alpha - \alpha^2 + \alpha^2 - \alpha^3 + \alpha^3 - \alpha^4,$$

$$= 1 + \alpha + \alpha^2 + \alpha^3 + \dots$$

$$-\alpha - \alpha^2 - \alpha^3 - \alpha^4 = 1$$

$$\begin{aligned} \sum_{k=0}^n x^k &= \sum_{k=0}^{\infty} x^k - \sum_{k=n+1}^{\infty} x^k = \sum_{k=0}^{\infty} x^k - q^{n+1} \left(\sum_{k=0}^{\infty} x^k \right) \\ &= \frac{1}{1-x} - \frac{q^{n+1}}{1-x} = \frac{1-q^{n+1}}{1-x} \end{aligned}$$

so Smally

$$Y[n] = \left(\frac{1 - \alpha^{n+1}}{1 - \alpha} \right) u[n]$$

what should this look like?

So in certain situations you have to use
the convolution sum

$$Y[n] = \sum_{k=-\infty}^{\infty} x[k] h[n-k]$$

$$= \sum_{k=-\infty}^{\infty} \underbrace{x[k]}_u u[n-k] h[n-k]$$

$$= \sum_{k=0}^{\infty} x[k] u[n-k]$$

$$= \sum_{k=0}^n x[k] \quad \text{if } n < 0 \quad \text{then force it}$$

\rightarrow add $d[n]$

$$\rightarrow \left(\sum_{k=0}^n x[k] \right) u[n]$$

Properties of LTI Systems (How convolution
behaves)

1) LTI system determined by Impulse response
Not true for non LTI systems

Like $y[n] = n x[n]$ not linear

\rightarrow use Stability test (for impulse response)

$\int |y| \rightarrow 0$ but not for other

if we know the response to signals

Some special input, Other signals in terms of that special input

\rightarrow that also characterizes the System too

⑧ 2) Commutative Property

Special Note!
 we could put the
 step function into a
 system → find the
 step response → as long as
 we know how to take the
 arbitrary input and decompose it
 in terms of step functions, then
 we could characterise things
 in the step response

→ in a similar way with the
 impulse response

Another note

what does the impulse
 response mean?

→ want to know the impulse
 response of a concert Hall?

fire a shot in it,

→ laid shot to hear the
 reverberations, that's the
 impulse response

→ this may ^{be} difficult ^a (to produce)
 response may just be to turn it on like a step response.
 Step response may be easier to produce

→ then find impulse response from that

→ or do pulses

2) Commutative

$$x[n] * h[n] = h[n] * x[n]$$

$$\sum_{k=-\infty}^{\infty} x[k] h[n-k] \quad \begin{aligned} m &= n - k \\ k &= n - m \end{aligned}$$

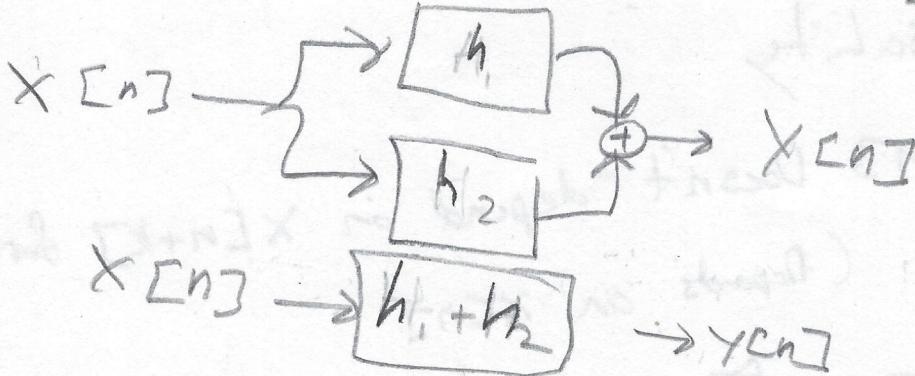
$$\rightarrow \sum_{m=-\infty}^{\infty} x[n-m] h[m]$$

⑨

3) Distributive property

$$x[n] * (h_1[n] + h_2[n])$$

$$= x[n] * h_1[n] + x[n] * h_2[n]$$



→ Like putting in parallel

4) Associativity

$$\begin{aligned} & x[n] * (h_1[n] * h_2[n]) \\ & (x[n] * h_1[n]) * h_2[n] \end{aligned}$$

Like $x[n] \rightarrow [h_1] \rightarrow [h_2] \rightarrow y[n]$

equivalently

order
can be

$$x[n] \rightarrow [h_1 * h_2] \rightarrow y[n]$$

interchanged

$$x[n] \rightarrow [h_2] \rightarrow [h_1] \rightarrow y[n]$$

→ can put systems in a plot and many
order

→ Not true for non-linear
systems

(10)

$$x \rightarrow [2] \rightarrow [x^2] \rightarrow (2x)^2 = 4x^2$$

$$x \rightarrow [x^2] \rightarrow [2] \rightarrow 2x^2$$

see?

5) Causality

$y[n]$ Doesn't depend on $x[n+k]$ for $k > 0$
 (Depends on $x[n]$)

$$y[n] = \sum_{k=-\infty}^{\Delta} h[k]x[n-k]$$

if $k > -|n|$

→ so for a causal system, $h[k]$ must be zero
 for future values of x

For causal system $h[k] = 0$ for $k < 0$
 impulse response

can't

→ Look generally like
 something takes the
 step function.

→ impulse response has to be zero before the
 impulse occurs should be casing
 before zero

b) STEP Response.

$$S[n] = u[n] - u[n-1]$$

1111

-1111

$$h[n] = S[n] - S[n-1]$$

↗ Step response

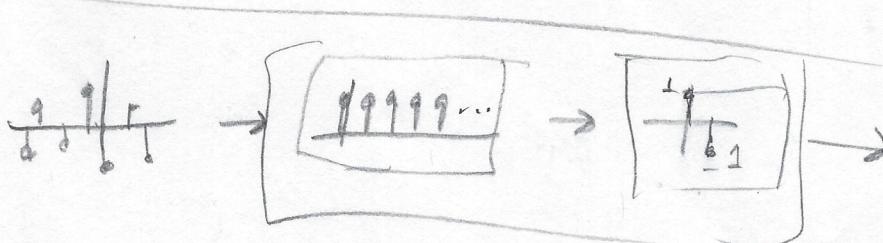
How to get impulse response $H(u[n])$

58:58

Since $u[n] = \sum_{k=-\infty}^n \delta[k]$

$$S[n] = \sum_{k=-\infty}^n h[k] *$$

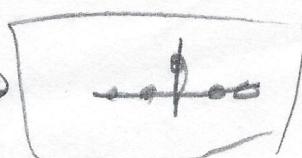
Step response also characterizes the LTI system.



(use associative property)

Don't do a series of convolution,

→ convolute both systems →



The Identity system

(12) many electrical and mechanical systems are described by Diff. eq's

Discrete versions are Difference equations

$$\sum_{k=0}^N a_k Y[n-k] = \sum_{k=0}^M b_k X[n-k]$$

$$Y[n] = Y_h[n] + Y_p[n]$$

homogeneous \nearrow
of particular solution

$$Y[n] = \sum b_k X[n-k] \quad (\text{finite impulse response system})$$