

Auto Regressive Processes

$$X[n] = -a_1 X[n-1] - a_2 X[n-2] + \dots - a_M X[n-M] + v[n]$$

↗
Noise

current value of X is related to previous values of X , we specify how many previous values we want to consider, M

our prediction is not 100% correct.

what should $-a_1, -a_2, \dots, -a_M$ be if given data, $v[n]$ says how off we are in the estimate.

$$E(v[n]^2) = \sigma_v^2 \quad v[n] \text{ has zero mean, } \mu=0 \text{ with some variance } \sigma_v^2$$

How do we expect

$E(X[n]X[n-k])$ to be related to each other? $L = r_{nn}$

(2)

Estimating $r(k)$ from Data

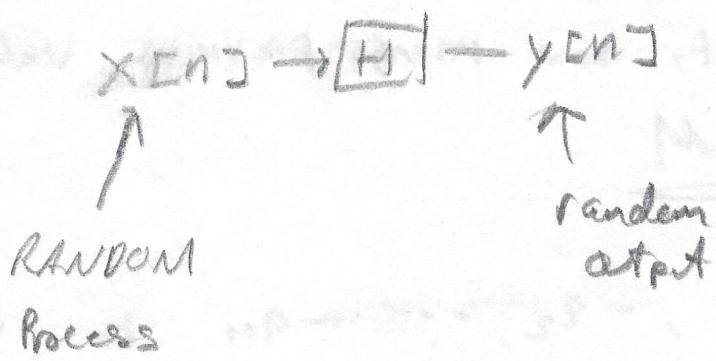
$$X[1], \dots, X[N]$$

$$r[k] = \frac{1}{N} \sum_{j=1}^N X[j]X[j-k]$$

WIENER FILTERS

OPTIMAL LINEAR DISCRETE-TIME

Filters



Goal: Design H to have $y[n]$
To a desired output $d[n]$

There will be error: $e[n] = d[n] - y[n]$

We want to minimize this statistically

example of a common signal we might want,
so say we want to predict the next value of
the signal.

(3)

given all of the x 's we have so far,
we want to predict the next value.

so the desired value is $\underline{x[n+1]}$

To do this we want to estimate
some previous value,

If I know all of the values are
correlated with each other, after seeing
enough previous data, we should be able to
do a good enough job of estimating $x[5]$
based on $x[1] - x[10]$

These are coupled together, so if $x[5]$
 $x[5]$ was noisy, we can look at the values around
and estimate a better value of $x[5]$)

when the desired value & x is one
of the previous values of x , or one of the future
values of x .

(4) Minimize J in a probabilistic way, minimize the expected value of the square of $e[n]$ over what does the filter do?

$$Y[n] = \sum_{k=0}^{\infty} h[k] x[n-k]$$

Infinite Impulse Response
causal filter

$x[n]$, $d[n]$ are zero mean

WIDE SENSE STATIONARY

(The statistics of a signal do not change as a function of time)

→ auto correlation is time invariant

- we want to minimize

$$J = E((e[n])^2)$$

$d[n]$ is likely a Past/Future Value of $x[n]$

(5)

Take the derivative and set those things equal to zero, minimize

To minimize J as a function of the $h[n]$, we take the derivative and (Gradient) set = 0

$$\frac{\partial J}{\partial h[k]} = 0 \quad \text{(Filtered coefficients)} \quad k = 0, 1, 2, \dots$$

$$J = E((e[n])^2)$$

$$\frac{\partial J}{\partial h[k]} = E \left[2e[n] \frac{\partial e[n]}{\partial h[k]} \right] \quad \text{chain rule}$$

we know

$$e[n] = d[n] - y[n] = d[n] - \sum_{n=0}^{\infty} h[n] \times [n-k]$$

$$\frac{\partial e[n]}{\partial h[k]} = -x[n-k]$$

$$\begin{aligned} \frac{\partial J}{\partial h[k]} &= E(-2e[n] \times [n-k]) \quad \text{recall we set this to zero.} \\ &= -2E(e[n] \times [n-1]) = 0 \end{aligned}$$

(6) error $e[n]$ is orthogonal to all of the inputs $x[n], x[n-1], x[n-2], \dots$
 The goal is to set up a set of equations to be satisfied.

So write

$$\frac{1}{\sum h[n]} \mathcal{E} = -2 \mathcal{E}(e[n] x[n-k])$$

$$= -2 \mathcal{E}\left(d[n] - \sum_{i=0}^{\infty} h[i] x[n-i]\right) x[n-k]$$

$$= -2 \mathcal{E}(d[n] x[n-k]) - \sum_{i=0}^{\infty} h[i] \mathcal{E}(x[n-i] x[n-k]) = 0$$

$$= -2 \mathcal{E}(d[n] x[n-k]) + 2 \sum_{i=0}^{\infty} h[i] \frac{\mathcal{E}(x[n-i] x[n-k])}{R(i-k)}$$

$$\xrightarrow{20}$$

We are minimizing the error

$\xrightarrow{\text{auto correlation}}$

So Now

$$\sum_{i=0}^{\infty} h[i] R[i-k] = P[-k]$$

$$= \mathcal{E}(x[n-k] d[n])$$

We define $P[-k]$ as the expected

value

$$P[-k] = E(x[n-k]d[n])$$

(7)

Note the WIENER - HOPF equations

$$\sum_{i=0}^{\infty} h[i] r[i-k] = P[-k]$$

If d is a different value of memory

same sign

then this becomes one of the auto-correlation numbers

Some other correlations if related to x (Not X)
through

If we have an IIR filter then it's hard to think about $h[i]$ being infinite (too much estimation)

If we allow ourselves to deal only with an FIR filter (causal Wc)

If we only consider a length - M FIR filter

The Wiener - Hopf equations are simpler

$$\sum_{i=0}^{M-1} h[i] r[i-k] = P[-k]$$

Looks like Yule - Walker equations

⑧

matrix form

 $R = 0, 1$

$$\begin{bmatrix} r(0) & r(1) & r(2) & \cdots & r(M-1) \\ r(1) & r(0) & r(1) & \cdots & r(M-2) \\ r(2) & r(2) & r(0) & \cdots & r(M-3) \\ \vdots & \vdots & \vdots & \ddots & \vdots \\ r(M-1) & r(M-1) & r(M-2) & \cdots & r(0) \end{bmatrix} \begin{bmatrix} h(0) \\ h(1) \\ h(2) \\ \vdots \\ h(M-1) \end{bmatrix} = \begin{bmatrix} p(0) \\ p(-1) \\ p(-2) \\ \vdots \\ p(-(M-1)) \end{bmatrix}$$

Autocorrelation

$$R_h = P_R$$

optimal filter Taps

$$h = R^{-1} P = R \setminus P$$

$$\text{RMS error } J = E((e[n])^2)$$

$$= E \left((d[n] - \sum_{k=0}^{n-1} h[k] x[n-k]) (d[n] - \sum_{k=0}^{n-1} h[k] x[n-k]) \right)$$

$$= E((d[n])^2) - \sum_{k=0}^{n-1} h[k] \underbrace{E(x[n-k] d[n])}_P$$

$$+ \sum_{k=0}^{n-1} \sum_{i=0}^{n-1} h[k] h[i] \underbrace{E(x[n-k] x[n-i])}_{R \text{- autocorrelation sys}}$$

$$= \underline{\sigma_d^2} - \underline{2h^T p} + \underline{h^T R h}$$

scalar ↓ scalar scalar
 scalar as well
 → dot-product

error for
any chosen
h

best h,

At the optimal h , $\hat{h} = R^{-1}p$

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$$\hat{J} = \underline{\sigma_d^2} - \underline{2\hat{h}^T p} + \underline{\hat{h}^T R \hat{h}}$$

$$= \sigma_d^2 - 2(R^{-1}p)^T p + (R^{-1}p)^T R (R^{-1}p)$$

$$= \sigma_d^2 - 2p^T R^{-1}p + p^T R^{-1}p$$

$$= \sigma_d^2 - p^T R^{-1}p \quad \begin{matrix} \text{This is the best} \\ \text{error we can get} \end{matrix}$$

We can show \nwarrow variance of desired signal \downarrow minimum error

$$J = \sigma_d^2 - 2h^T p + h^T R h$$

Error for my h = $\frac{(\sigma_d^2 - p^T R^{-1}p) + (p^T R^{-1}p - 2h^T p + h^T R h)}{\hat{J}}$

\downarrow $\hat{J} = \hat{J} + (h - \hat{h})^T R (h - \hat{h})$ - some positive #
 (R is positive definite matrix)

→ Error for optimal \hat{h})

10

optimal filter is unique

as $(h - \hat{h})$ must be zero

for \hat{f}

suppose the desired signal, $x[n]$, is one of the original signal input values.

Special case

Linear prediction

We want to predict $x[n+1]$

$x[n], x[n-1], \dots, x[n-(M-1)]$

One-Step-Forward Linear Predictor

$$\hat{x}[n+1] = \sum_{k=0}^{M-1} h[k] x[n-k]$$

Predictions

If $x[n]$ what is the optimal set of filter taps that give me the best prediction of x

The above case is for the situation where the desired signal is one of the original filter values. (11)

This corresponds to $d[n] = x[n+1]$

Take all of the stuff that happened just before the signal in order to predict the next one

$$d[n] = x[n] \quad \text{Filter}$$

$$d[n] = x[n + \text{positive } \#] \quad \text{prediction}$$

$$d[n] = x[n - \text{positive } \#] \quad \text{smoothing}$$

For ONE-STEP Ahead Prediction we can find the answer with a Wiener filter!

$$d[n] = x[n+1] \quad \text{our prediction}$$

$$e[n] = x[n+1] - \hat{x}[n+1]$$

$$\text{We want to minimize } J = E((e[n])^2)$$

② Wiener Hoff equations regime

$$R \hat{h} = P_R \quad k = 0, 1, 2, \dots, (M-1)$$

$$\mathcal{E}(x[n-k] \cdot x[n+1])$$

So set up matrix equations

$$\begin{bmatrix} r(0) & r(1) \\ r(1) & r(0) \\ \vdots & \vdots \\ r(M-1) & r(0) \end{bmatrix} \begin{bmatrix} h[0] \\ h[1] \\ \vdots \\ h[M-1] \end{bmatrix} = \begin{bmatrix} p(1) \\ r(M) \end{bmatrix}$$

$$P = \begin{bmatrix} & & \\ & & \\ & & \end{bmatrix}$$

Predict the taps our p_i 's are
other values
of auto correlations

Wiener Hoff for one-step
ahead prediction

Error for the optimal \hat{h}

(13)

$$\hat{J} = \sigma_d^2 - \underline{P}^T \underline{R}^{-1} \underline{P}$$

$$r = P = [r(0) \dots r(M)]$$

$$E((d[n])^2)$$

$$= E((x[n+1])^2) = r(0)$$

$$\therefore \hat{J} = r(0) - r^T \hat{h} \quad \hat{h} = \underline{R}^{-1} \underline{P}$$

(5)

Best error and best filter (\hat{h})

$$\begin{bmatrix} r(0) & r^T \\ r & R \end{bmatrix} \begin{bmatrix} 1 \\ -\hat{h} \end{bmatrix} = \begin{bmatrix} \hat{J} \\ 0 \end{bmatrix}$$

$$r(0) - r^T \hat{h} = \hat{J}$$

$$r - R \hat{h} = 0 \rightarrow r = \tilde{R} \hat{h}$$

Likewise for auto correlation

matrix. $(m+1) \times (m+1)$

- (14) Say we have an M tap 1-step forward predictor, OK I can give one more tap.
 You can look at one more element in the past to help predict things.
 Can we get a next-best filter by updating the filter we already have
 (we don't want to calculate the filter we already have)
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Levinson-Durbin Algorithm

How can we take the optimal Length- M predictor and easily determine the OPTIMA Length $M+1$ Predictor?

IF $a_m = \begin{bmatrix} 1 \\ -h \end{bmatrix}$ ^{length M OPTIMAL}
 filter

$$\sum_{l=0}^n c_{ml} r(l-i) = \begin{cases} 1 & i=0 \\ 0 & i=1 \dots M \end{cases}$$

Different way of writing the matrix equations

If we chose $d[n] = x[n-M]$ and predicted this from the samples

$x[n], x[n-1], \dots, x[n-(M-1)]$, this would be a similar problem

(Smoothing, or backwards prediction)

Wiener - Neft in this case gives a similar looking system for the optimal filter

$$g[0], g[1], \dots, g[M-1]$$

Forward Prediction

$$\begin{bmatrix} r[0] & r[1] & \dots & r[M-1] \\ r[M-1] & r[0] & \dots & r[M-2] \end{bmatrix} \begin{bmatrix} h[0] \\ h[1] \\ \vdots \\ h[M-1] \end{bmatrix} = \begin{bmatrix} r[0] \\ r[1] \\ \vdots \\ r[M] \end{bmatrix}$$

FIR
Upside down

Backward Prediction

$$\begin{bmatrix} R \end{bmatrix} \begin{bmatrix} g[0] \\ g[1] \\ \vdots \\ g[M-1] \end{bmatrix} = \begin{bmatrix} r[M] \\ r[M-1] \\ \vdots \\ r[0] \end{bmatrix}$$

dowm
Upside down

(16) The optimal g_s are the optimal
hs flipped backwards

$$\hat{g}^r[k] = \hat{h}[(n+1)-k]$$

$$\hat{g}[k] = \hat{h}[M-k]$$

$\hat{\int}$ (Best expected error)

B the same