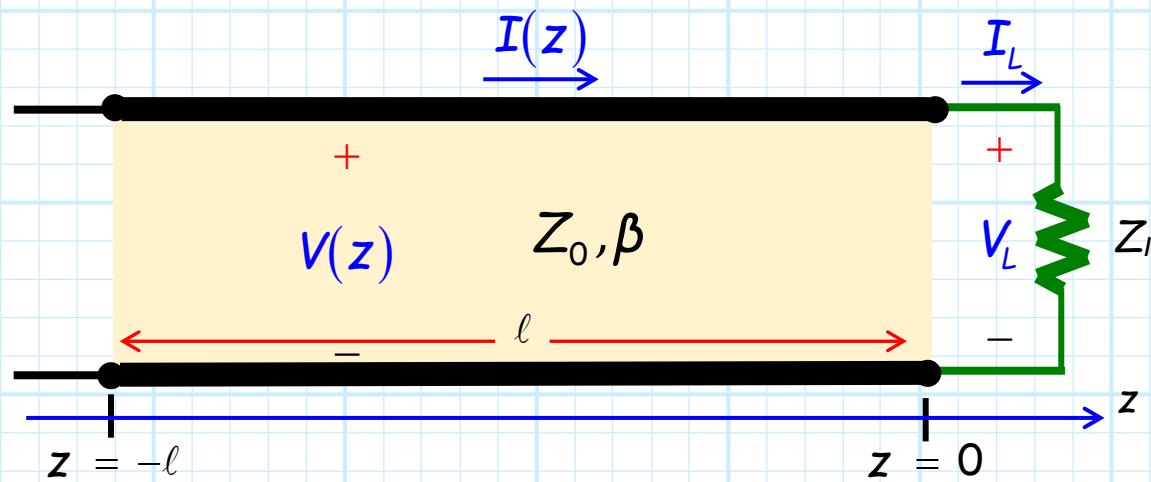


The Terminated, Lossless Transmission Line

Now let's **attach** something to our transmission line. Consider a **lossless** line, length ℓ , terminated with a **load** Z_L :



Q: What is the **current** and **voltage** at each and **every** point on the transmission line (i.e., what is $I(z)$ and $V(z)$ for all points z where $-l \leq z \leq 0$?)?

A: To find out, we must apply **boundary conditions!**

First, there's Heaviside

In other words, at the **end** of the transmission line ($z = 0$)—where the load is **attached**—we have many requirements that all must be satisfied!

Requirement 1. To begin with, the voltage and current ($I(z = 0)$ and $V(z = 0)$) must be consistent with a valid **transmission line solution** (i.e., satisfy the **telegraphers equations**):



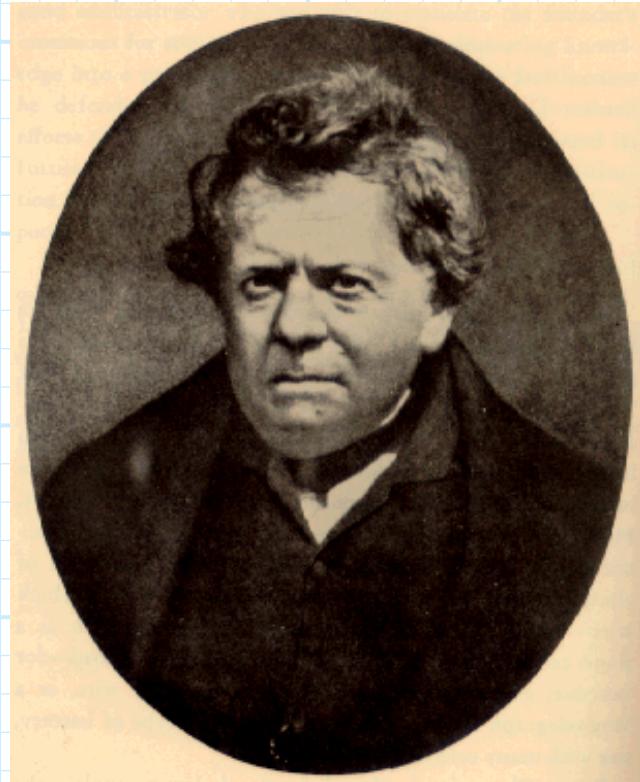
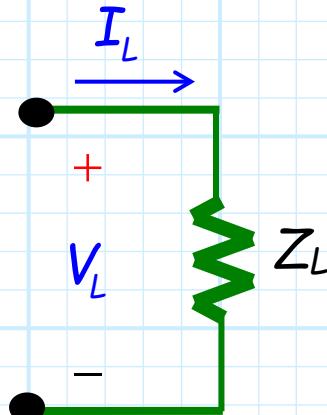
$$\begin{aligned} V(z=0) &= V^+(z=0) + V^-(z=0) \\ &= V_0^+ e^{-j\beta(0)} + V_0^- e^{+j\beta(0)} \\ &= V_0^+ + V_0^- \end{aligned}$$

$$\begin{aligned} I(z=0) &= \frac{V^+(z=0)}{Z_0} - \frac{V^-(z=0)}{Z_0} \\ &= \frac{V_0^+}{Z_0} e^{-j\beta(0)} - \frac{V_0^-}{Z_0} e^{+j\beta(0)} \\ &= \frac{V_0^+ - V_0^-}{Z_0} \end{aligned}$$

Now for Ohm

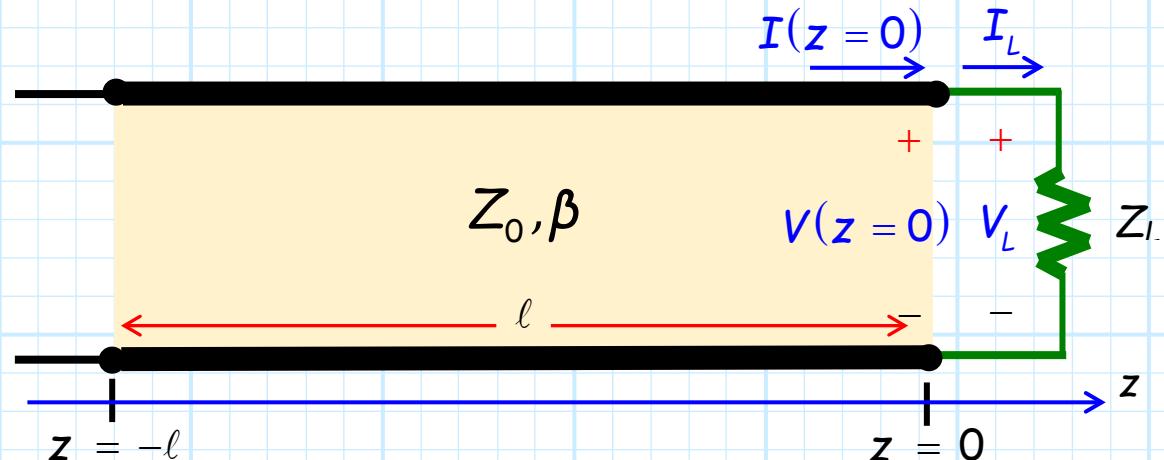
Requirement 2. Likewise, the load voltage and current must be related by the complex form of Ohm's law:

$$\frac{V_L}{I_L} = Z_L$$



Finally, we have Kirchoff

Requirement 3. Most importantly, we recognize that the values $I(z = 0)$, $V(z = 0)$ and I_L , V_L are **not** independent, but in fact are strictly related by **Kirchoff's Laws!**



From KVL and KCL we find these requirements:

$$V(z=0) = V_L \quad \text{and} \quad I(z=0) = I_L$$

→ These are our **boundary conditions!**

The result of some algebra

Combining these results, we find that the current and voltage at the end of the line must be related to the load impedance:

$$\frac{V(z=0)}{I(z=0)} = Z_L$$


Q: Hey isn't the ratio of total line voltage $V(z)$ and total line current $I(z)$ the line impedance $Z(z)$??

A: It sure is!

Both intuitive and satisfying

Thus, we know for $z = 0$:

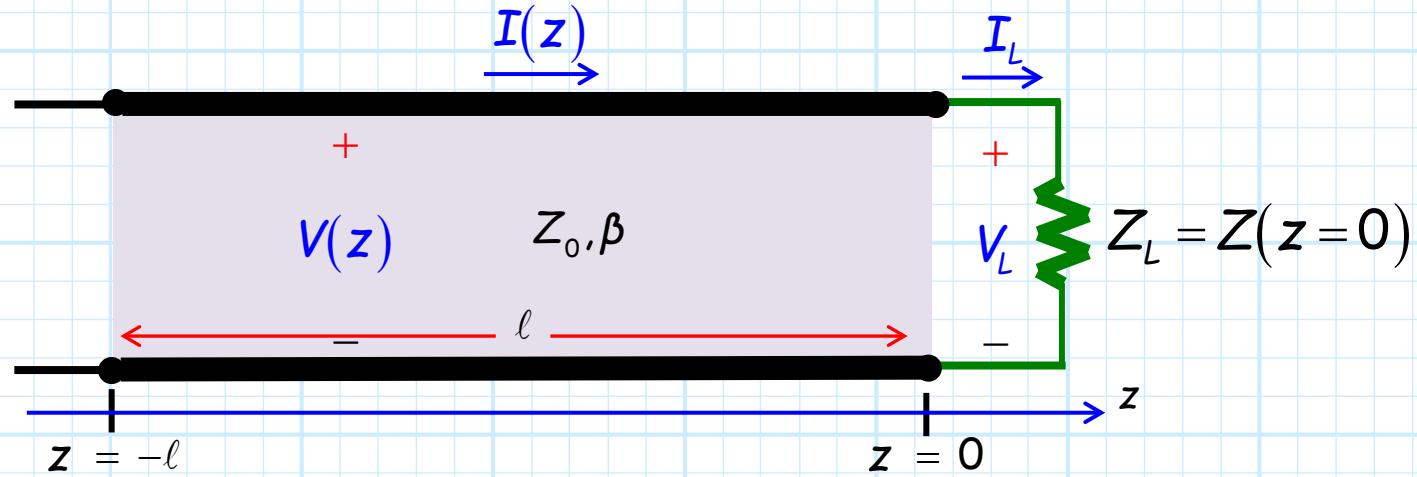
$$Z(z=0) = \frac{V(z=0)}{I(z=0)}$$

And now combining these last two expressions, we find a very **intuitive** and **satisfying** boundary condition:

$$Z(z=0) = Z_L$$

→ The **line impedance** at the end of the transmission line (i.e., at $z = 0$) must be equal to the **load impedance** attached to that end!

More than just one point!



Q: So what? Who cares about the line impedance at just the one point $z = 0$?

A: It turns out that the boundary condition determines the values of line impedance at all points z !

Here's where we prove it

To see this, we first consider the expression for line impedance:

$$Z(z) = Z_0 \left[\frac{V_0^+ e^{-j\beta z} + V_0^- e^{+j\beta z}}{V_0^+ e^{-j\beta z} - V_0^- e^{+j\beta z}} \right] = Z_0 \left[\frac{e^{-j\beta z} + \Gamma_0 e^{+j\beta z}}{e^{-j\beta z} - \Gamma_0 e^{+j\beta z}} \right]$$

Evaluating this at $z = 0$:

$$Z(z=0) = Z_0 \left[\frac{e^{-j\beta(0)} + \Gamma_0 e^{+j\beta(0)}}{e^{-j\beta(0)} - \Gamma_0 e^{+j\beta(0)}} \right] = Z_0 \left[\frac{1 + \Gamma_0}{1 - \Gamma_0} \right]$$

And so, an equivalent boundary condition statement is

$$Z(z=0) = Z_0 \left[\frac{1 + \Gamma_0}{1 - \Gamma_0} \right] = Z_L$$

A stunning result!!!!



With some algebraic elbow grease we rearrange this last expression to find:

$$\Gamma_0 = \frac{Z_L - Z_0}{Z_L + Z_0}$$

Now inserting this into the general expression for line impedance:

$$\begin{aligned} Z(z) &= Z_0 \left[\frac{e^{-j\beta z} + \Gamma_0 e^{+j\beta z}}{e^{-j\beta z} - \Gamma_0 e^{+j\beta z}} \right] \\ &= Z_0 \left[\frac{e^{-j\beta z} + \left(\frac{Z_L - Z_0}{Z_L + Z_0} \right) e^{+j\beta z}}{e^{-j\beta z} - \left(\frac{Z_L - Z_0}{Z_L + Z_0} \right) e^{+j\beta z}} \right] \\ &= Z_0 \left[\frac{(Z_L + Z_0) e^{-j\beta z} + (Z_L - Z_0) e^{+j\beta z}}{(Z_L + Z_0) e^{-j\beta z} - (Z_L - Z_0) e^{+j\beta z}} \right] \end{aligned}$$

→ How. About. That.

Q: ????



Make sure you see why this is stunning

A: Look carefully at this result:

$$Z(z) = Z_0 \left[\frac{(Z_L + Z_0) e^{-j\beta z} + (Z_L - Z_0) e^{+j\beta z}}{(Z_L + Z_0) e^{-j\beta z} - (Z_L - Z_0) e^{+j\beta z}} \right]$$

Notice the unknown **wave amplitudes** V_0^+ and V_0^- are not in this result!

Q: ????

A: The load impedance Z_L not only determines the value of line impedance $Z(z)$ at the **end** of the transmission line (i.e., at $z = 0$).

It likewise sets the value of line impedance at all locations z !

What about the other viewpoint?

Q: So, this is the boundary condition result for the voltage $V(z)$, current $I(z)$, impedance $Z(z)$ viewpoint.

What about the plus-wave $V^+(z)$, minus-wave $V^-(z)$, reflection coefficient $\Gamma(z)$?

A: First, recall the reflection coefficient function is:

$$\Gamma(z) = \Gamma_0 e^{+j2\beta z}$$

where the complex constant Γ_0 is:

$$\Gamma_0 \doteq \Gamma(z=0) = \frac{V^-_0}{V^+_0}$$

The resulting reflection coefficient function

Now, recall that we just determined for a transmission line terminated in a load Z_L that:

$$\Gamma_0 = \frac{Z_L - Z_0}{Z_L + Z_0}$$

Meaning the reflection coefficient function for a lossless terminated transmission line is:

$$\Gamma(z) = \Gamma_0 e^{+j2\beta z} = \left[\frac{Z_L - Z_0}{Z_L + Z_0} \right] e^{+j2\beta z}$$

Now we know the value of Γ everywhere

As with line impedance, we see that the load not only determines the value of the reflection coefficient function at the end of the line:

$$\Gamma_0 = \Gamma(z=0) = \left[\frac{Z_L - Z_0}{Z_L + Z_0} \right]$$

but also determines the reflection coefficient at all points z along the line:

$$\Gamma(z) = \left[\frac{Z_L - Z_0}{Z_L + Z_0} \right] e^{+j2\beta z}$$

This shows up a lot

Q: This complex value:

$$\frac{Z_L - Z_0}{Z_L + Z_0}$$

sure seems to show up a lot. Does it have any significance?

A: Note that it is the value of the reflection coefficient at the end of the transmission line (i.e., where the load is attached).

$$\Gamma(z=0) = \left[\frac{Z_L - Z_0}{Z_L + Z_0} \right]$$

Thus, this complex value is known as the load reflection coefficient Γ_L :

$$\Gamma_L \doteq \frac{Z_L - Z_0}{Z_L + Z_0}$$

Both intuitive and satisfying

Thus, (if the load is located at $z = 0$) we can recast our reflection coefficient math as:

$$\Gamma_0 = \frac{V_0^-}{V_0^+} = \Gamma_L$$

and:

$$\Gamma(z) = \Gamma_L e^{+j2\beta z}$$

Perhaps most importantly, the boundary condition for reflection coefficient is again very intuitive and satisfying:

$$\Gamma(z=0) = \Gamma_L$$

Now for Kirchoff

Now, one important result of this boundary condition is that the value of wave amplitude V_0^- and wave amplitude V_0^+ are coupled.

→ If ya' know one, ya' know the other.

$$\Gamma_L = \Gamma_0 = \frac{V_0^-}{V_0^+} \quad \Rightarrow \quad V_0^- = V_0^+ \Gamma_L$$

And thus, the propagating minus-wave can be expressed in terms of the plus-wave amplitude:

$$V^-(z) = V_0^- e^{+j\beta z} = V_0^+ \Gamma_L e^{+j\beta z}$$

The Bottom Line

And so finally, the **voltage** and **current** along the terminated transmission line can be expressed in terms of **load reflection coefficient** Γ_L :

$$V(z) = V_0^+ \left[e^{-j\beta z} + \Gamma_L e^{+j\beta z} \right]$$

$$I(z) = \frac{V_0^+}{Z_0} \left[e^{-j\beta z} - \Gamma_L e^{+j\beta z} \right]$$

$$Z(z) = Z_0 \left(\frac{e^{-j\beta z} + \Gamma_L e^{+j\beta z}}{e^{-j\beta z} - \Gamma_L e^{+j\beta z}} \right)$$

Note the above expressions are accurate **ONLY** if the load Z_L is located at position $z = 0$.

Two waves and a gamma

We can alternatively express the solutions as:

$$V^+(z) = V_0^+ e^{-j\beta z}$$

$$V^-(z) = V_0^+ \Gamma_L e^{+j\beta z}$$

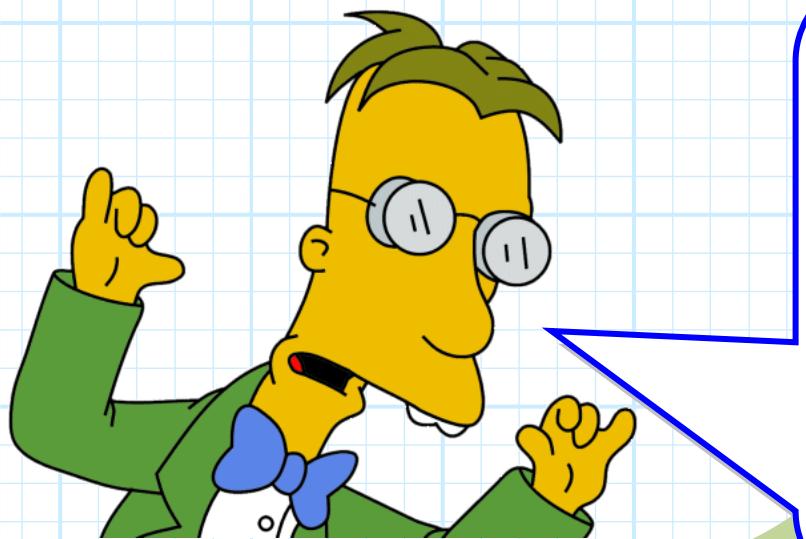
$$\Gamma(z) = \frac{V^-(z)}{V^+(z)} = \Gamma_L e^{+j2\beta z}$$

Where:

$$\Gamma_L \doteq \frac{Z_L - Z_0}{Z_L + Z_0}$$

What about V_0^+ ??

Q: But, how do we determine V_0^+ ??



A: We require a **second** boundary condition to determine V_0^+ . The only boundary left is at the **other end** of the transmission line.

Typically, a **source** of some sort is located there. This makes physical sense, as something must generate the **incident wave** !

