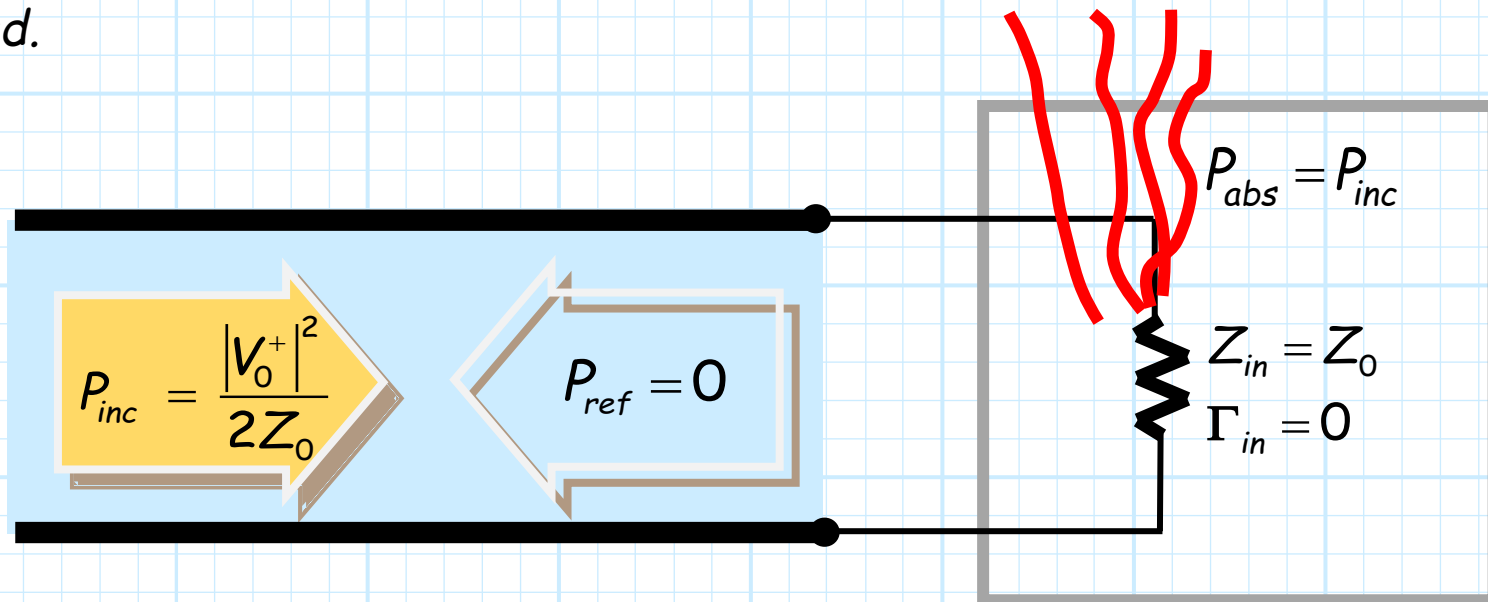


Return Loss and VSWR

Q: If the purpose of a transmission line is to transfer **energy** from some **source** (the output of some useful device) to some **load** (the **input** to some **other** useful device), then wouldn't an input impedance of Z_0 be **ideal**.

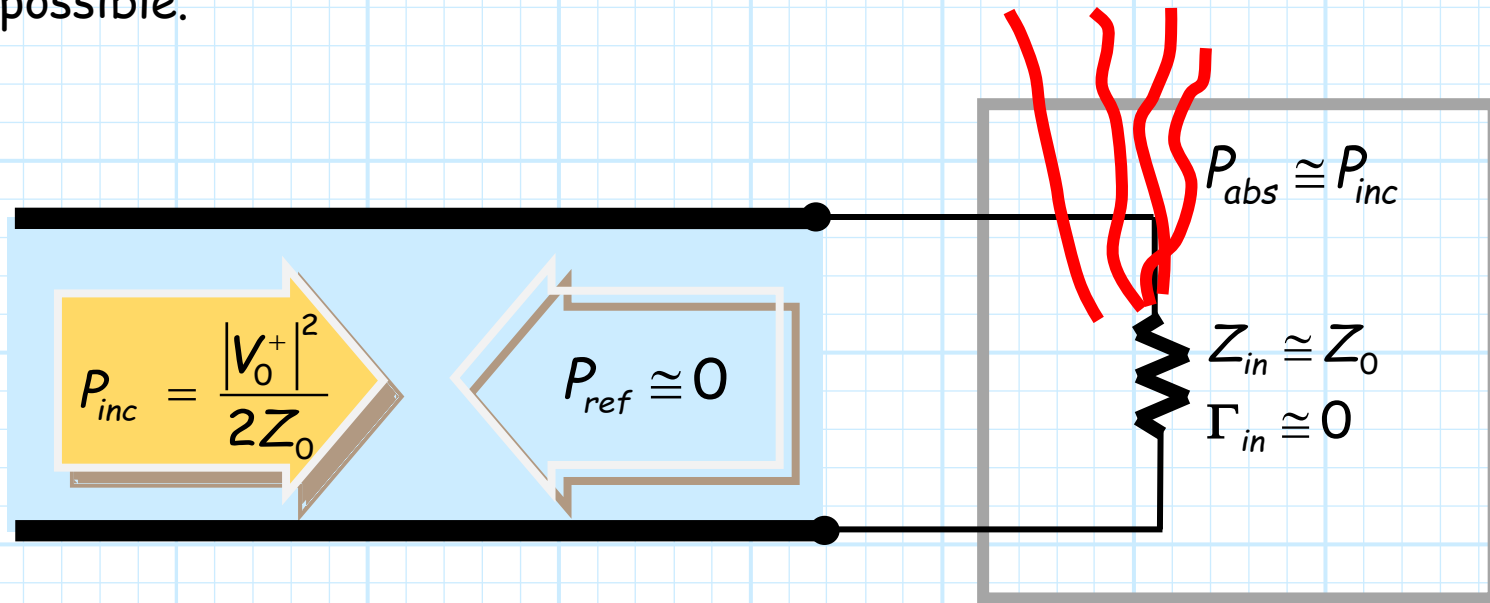
After all, **no energy is reflected—all incident energy is absorbed** by the load.



A: If connected entirely to other “**matched**” devices, it would be!

Close, but not exactly

Thus, microwave component vendors often try very hard to make the **input impedance** of their useful devices as **close** to Z_0 ($Z_0 = 50\Omega$, say) as possible.



But the value:

$$Z_{in} = 50.000000000000000000000000\dots$$

is really hard to come by!

It's not like they're lying (usually)

Q: *So do vendors tell us what Z_{in} really is?*

A: Not usually!

Remember, input **impedance**—as well as its **reflection coefficient** Γ_{in} —is a **complex** value.

Therefore microwave engineers often use **real-valued** measures to indicate how “**matched**” a given **load** (e.g., the input impedance of some useful device) really is!

Two of the most prevalent **real-valued** measures are:

1. **return loss**, and
2. **VSWR**.

Return Loss

The ratio of the incident power to the reflected power is known as **return loss**.

Typically, return loss is expressed in **decibels**:

$$R.L. = 10 \log_{10} \left[\frac{P_{inc}}{P_{ref}} \right] = -10 \log_{10} |\Gamma_{in}|^2$$

Where:

$$\Gamma_{in} = \frac{Z_{in} - Z_0}{Z_{in} + Z_0}.$$

Return Loss examples

For **example**, if the return loss is **10dB**, then:

- A.** 10% of the incident power is **reflected**, so...
- B.** the remaining **90%** is **absorbed**, thus...
- C.** we “lose” 10% of the incident power.

Likewise, if the return loss is **30dB**, then:

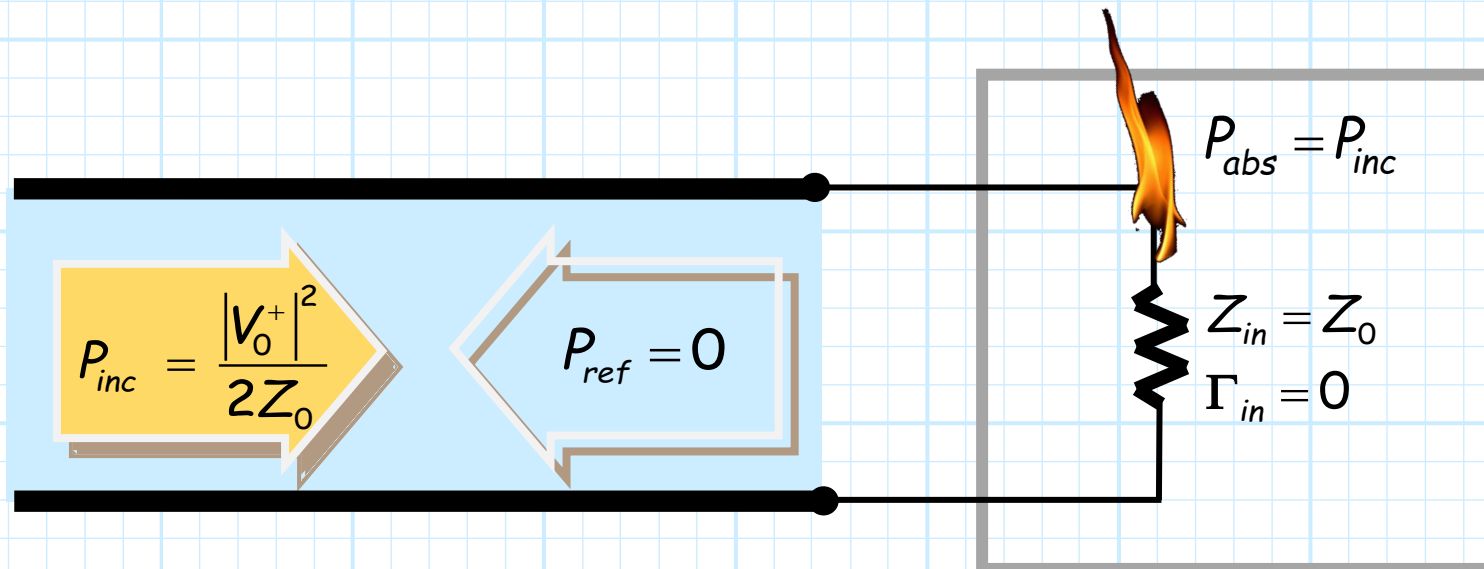
- A.** 0.1% of the incident power is **reflected**, so...
- B.** the remaining **99.9%** is **absorbed**, thus...
- C.** we “lose” 0.1% of the incident power.

Thus, a **larger** numeric value for return loss **actually** indicates **less** lost power!

More return loss is actually better

An **ideal** return loss thus would be ∞ dB—the input is a **matched load**:

$$R.L. = -10\log_{10}|\Gamma_{in}|^2 = -10\log_{10}|0|^2 = \infty$$

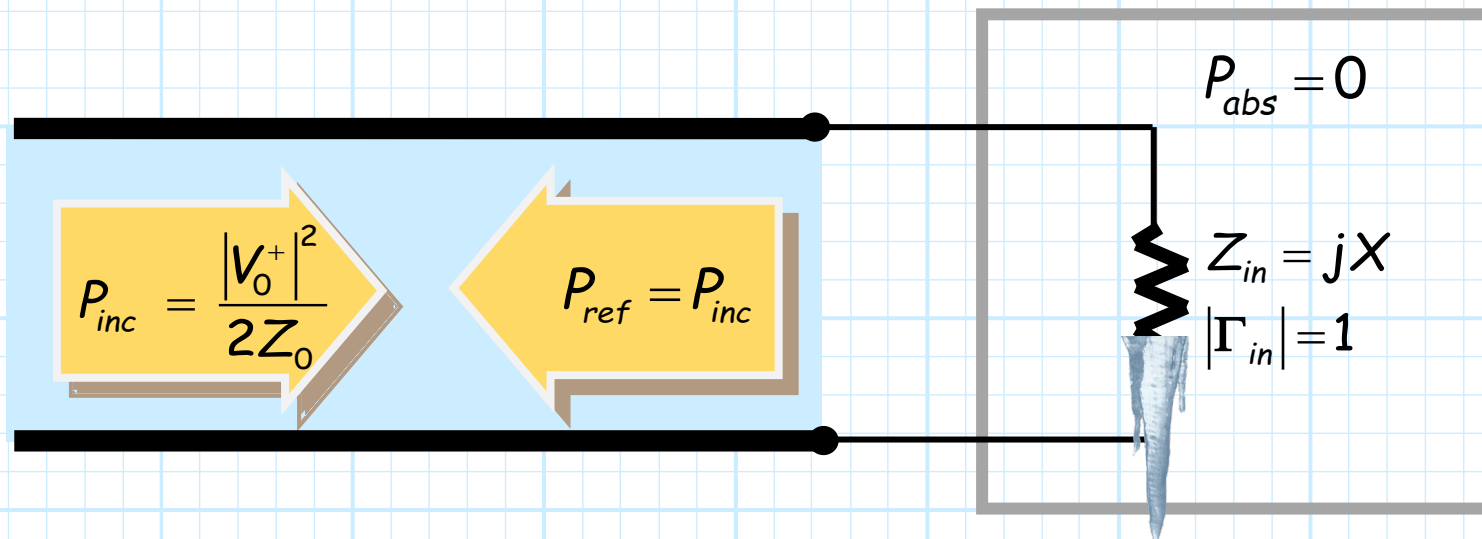


Zero return loss is the worst

The **worst case** is when **no** power is absorbed—all the incident power is **reflected**.

For this (worst) case, the return loss of is **zero**!

$$R.L. = -10 \log_{10} |\Gamma_{in}|^2 = -10 \log_{10} |1|^2 = 0$$



VSWR

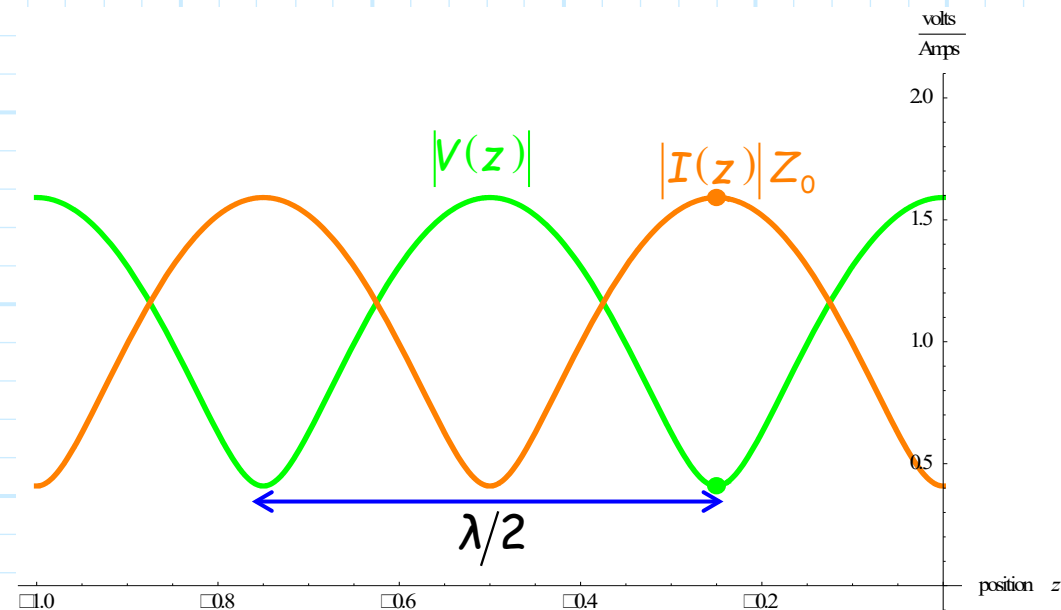
Consider again the **voltage** along a transmission line attached to a port of **some useful device** with impedance Z_{in} (where $z_L = 0$):

$$V(z) = V_0^+ [e^{-j\beta z} + \Gamma_{in} e^{+j\beta z}]$$

Now let's look at the **magnitude** only:

$$|V(z)| = |V_0^+| |e^{-j\beta z} + \Gamma_{in} e^{+j\beta z}| = |V_0^+| |1 + \Gamma_{in} e^{+j2\beta z}|$$

As this result shows—and as **you** already know—the magnitude is **not constant** with respect to position z .



Definition

It can be shown that the **largest** and **smallest** values of this magnitude are:

$$|V(z)|_{\max} = |V_0^+|(1 + |\Gamma_{in}|) \quad |V(z)|_{\min} = |V_0^+|(1 - |\Gamma_{in}|)$$

The ratio of $|V(z)|_{\max}$ to $|V(z)|_{\min}$ is known as the **Voltage Standing Wave Ratio (VSWR)**:

$$VSWR \doteq \frac{|V(z)|_{\max}}{|V(z)|_{\min}} = \frac{1 + |\Gamma_{in}|}{1 - |\Gamma_{in}|} \quad \therefore \quad 1 \leq VSWR \leq \infty$$

As with **return loss**, VSWR is a real value that is on dependent on the **magnitude** of Γ_{in} (i.e., $|\Gamma_{in}|$) **only** !

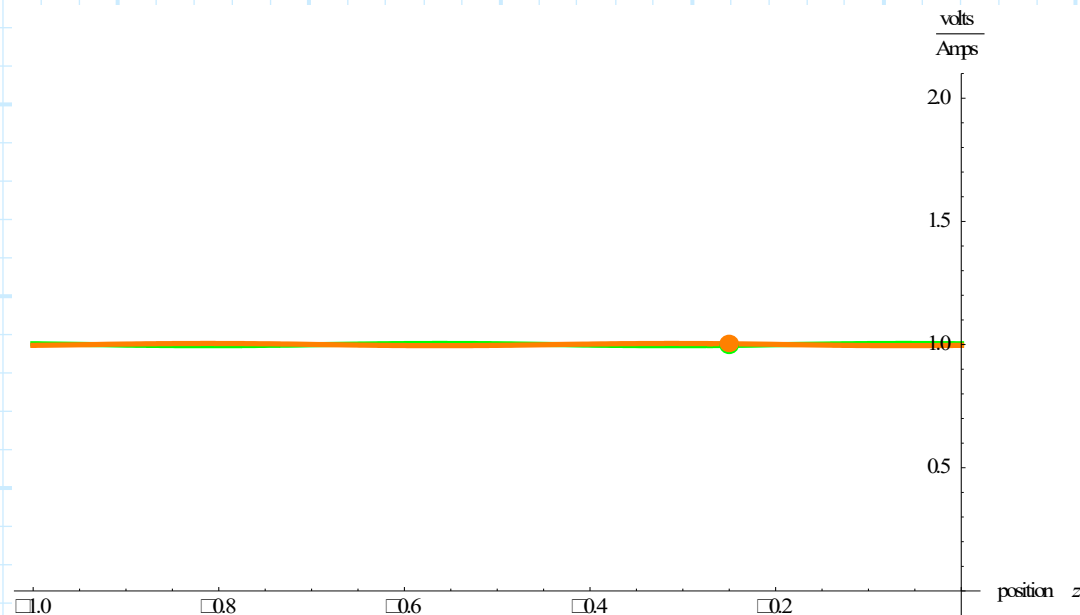
The best VSWR is 1.0

Note for the **ideal** input impedance, where $|\Gamma_{in}| = 0$ (i.e., $Z_{in} = Z_0$), then **VSWR = 1.0**.

We find for this case:

$$|V(z)|_{max} = |V(z)|_{min} = |V_0^+|$$

In other words, the voltage magnitude is a **constant** with respect to position z .



The worst VSWR is infinite

Conversely, for the **worst** input impedance where $|\Gamma_{in}| = 1$, then $VSWR = \infty$.

We find for **this** case:

$$|V(z)|_{min} = 0 \quad \text{and} \quad |V(z)|_{max} = 2 |V_0^+|$$

Recall for this case the **voltage magnitude** $|V(z)|$ varies **greatly** with respect to position z !

