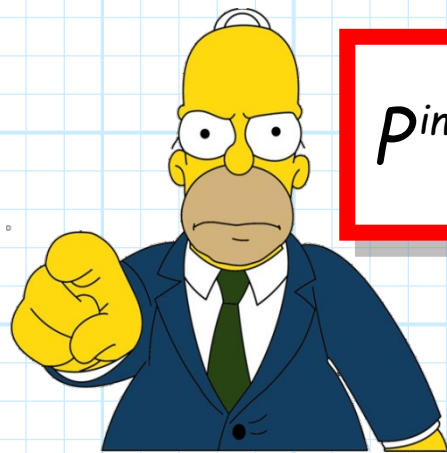


Incident and Reflected Power with "Matched" Source or Load

Q: So. You've made a big, gigantic, **dramatic** deal about the fact that the **incident** power along a transmission line is **greater** than that **delivered** by the source:



$$P^{inc} \geq P_g^{del} = P_L^{abs}$$



But. Doesn't the **inequality** \geq mean that the incident power **could be equal** to the delivered power of the source?

A: For the **right circumstance**, yes; the two **could be equal** ($P^{inc} = P_g^{del}$).

When there's no reflected power

Q: *The right circumstance?*

What circumstance is the **right** circumstance in order for $P_g^{del} = P^{inc}$?

A: Sigh.

Just **look** at the relationship:

$$P_g^{del} = P^{inc} - P^{ref}$$

Hopefully it is apparent, that **when** (and **only** when!) the **reflected** power is **zero** ($P^{ref} = 0$), the **incident** and **delivered** power will be **equal**:

$$P_g^{del} = P^{inc} \text{ when } P^{ref} = 0$$

When the load reflection coefficient is zero

Q: OK; then under *what circumstance* is the reflected power zero?

A: Sigh.

Just **look** at the relationship:

$$P^{ref} = P^{inc} |\Gamma_L|^2$$

Clearly, the reflected power will be zero **if** $|\Gamma_L|^2 = 0$ (i.e., if $\Gamma_L = 0$):

$$P_g^{del} = P^{inc} \quad \text{when} \quad \Gamma_L = 0$$

When the load is "matched"

Q: Well then. Under what @\$\$%^!&% circumstance is Γ_L zero?!?

A: Sigh.

Recall the definition of Γ_L :

$$\Gamma_L = \frac{Z_L - Z_0}{Z_L + Z_0}$$

From the **numerator**, it is apparent that Γ_L is zero if (and **only** if!) the **load** impedance Z_L is **numerically equal** to the **characteristic** impedance Z_0 of the transmission line:

$$\Gamma_L|_{Z_L=Z_0} = \frac{Z_L - Z_0}{Z_L + Z_0}|_{Z_L=Z_0} = \frac{Z_0 - Z_0}{Z_0 + Z_0} = 0$$

A special case

Q: Hey; wasn't $Z_L = Z_0$ one of the "special cases" that we studied earlier?

A: It was!

Recall that when $Z_L = Z_0$, the input impedance of our terminated transmission line is likewise numerically equal to Z_0 —regardless of line length ℓ !

We now see why this is true.

Matched load means no reflected wave

When $Z_L = Z_0$, then $\Gamma_L = 0$, and so:

$$V_0^- = \Gamma_L V_0^- = (0)V_0^- = 0$$

Meaning the **reflected** wave (if $Z_L = Z_0$) is **zero**:

$$V^-(z) = V_0^- e^{j\beta z} = 0$$

So, our **total** voltage and current along the transmission line is that of the plus (i.e., **incident**) **wave only**:

$$V(z) = V^+(z) \quad \text{and} \quad I(z) = I^+(z)$$

The **line impedance** function is thus (when $Z_L = Z_0$):

$$Z(z) = \frac{V(z)}{I(z)} = \frac{V^+(z)}{I^+(z)} = Z_0$$

"Matched" load means...

In summary, for a "matched"* load:

a) the reflected wave is **zero**:

$$V^-(z) = 0 \quad \text{if} \quad Z_L = Z_0$$

b) the power of the **reflected** wave is likewise **zero**:

$$p_{\text{ref}} = \frac{|V^-(z)|^2}{2Z_0} = 0 \quad \text{if} \quad Z_L = Z_0$$

c) meaning the incident power will be **equal** to the delivered power:

$$p_{\text{inc}} = p_g^{\text{del}} = p_L^{\text{abs}} \quad \text{if} \quad Z_L = Z_0$$

* does **not** imply a conjugate match!

Now for the "matched" source

Q: Things sure seem to *simplify* if the *load* is "matched".

What about a "matched" *source*?

Do *any* of the power relationships simplify then?

A: They sure do!

Recall for a "matched" **source**, the incident wave becomes **causal** with respect to the **source**:

$$V^+(z) = V_0^+ e^{-j\beta z} = \left(\frac{1}{2} V_g e^{-j\beta \ell} \right) e^{-j\beta z} = \frac{V_g}{2} e^{-j\beta(z+\ell)}$$

In other words, for this **special case** (and **only** for this special case!), the **indent** wave is dependent on the **source only** (the value of load Z_L is irrelevant).

All I can say is: wow!

Recall the power associated with the **incident wave** is:

$$p_{inc} = \frac{|V^+(z)|^2}{2Z_0}$$

so that for a “**matched**” source:

$$p_{inc} = \frac{|V^+(z)|^2}{2Z_0} = \frac{1}{2Z_0} \left| \frac{V_g}{2} e^{-j\beta(z+\ell)} \right|^2 = \frac{|V_g|^2}{8Z_0} \quad \leftarrow \text{WOW!!!!}$$

Q: I don't see *why* this is “wow”. Am I missing something?

A: Apparently you are. Look **closer** at the above result.

Do I have to explain everything?

Remember, this result is for the **special case** where the source impedance is a **real value**, numerically equal to Z_0 :

$$Z_g = Z_0 + j0$$

Therefore:

$$R_g = \operatorname{Re}\{Z_g\} = Z_0$$

The **source** resistance is **numerically equal** to transmission line characteristic impedance Z_0 .

Thus, the **power** of the **incident** wave can be alternatively written as:

$$p_{inc} = \frac{|V_g|^2}{8 Z_0} = \frac{|V_g|^2}{8 R_g} \quad \leftarrow \text{WOW!!!!}$$

The incident power is the available power (wow)!

Q: *This result looks vaguely familiar; haven't we seen this before?*

A: Sigh. This result is the **available power of the source**!!!!!!

Therefore:

$$P_{inc} = \frac{|V_g|^2}{8Z_0} = \frac{|V_g|^2}{8R_g} = P_g^{avl} \quad \leftarrow \text{Wow!}$$

For the special case $Z_g = Z_0 + j0$, the incident power P_{inc} is equal to the **available power** P_g^{avl} of the **source**.

"Matched" source means...

In **summary**, for a "matched"* source:

a) the power of the **incident** wave is:

$$P^{inc} = \frac{|V_g|^2}{8 Z_0} \quad \text{if} \quad Z_g = Z_0$$

b) meaning the **incident** power will be equal to the **available** power of the "matched" source:

$$P^{inc} = P_g^{avl} \quad \text{if} \quad Z_g = Z_0$$

* does **not** imply a conjugate match!