

# Two-Tone Intermodulation

**Q:** It doesn't seem to me that this dad-gum intermodulation distortion is really that much of a problem.

I mean, the first and second harmonics will likely be well outside the amplifier bandwidth, right?



**A:** True, the harmonics produced by intermodulation distortion typically are not a problem in radio system design.

There is a problem, however, that is much worse than harmonic distortion!

This problem is called **two-tone** intermodulation distortion.

# Many eigen functions at the input!

Say the input to an amplifier consists of two signals at dissimilar frequencies:

$$v_{in}(t) = a \cos \omega_1 t + a \cos \omega_2 t$$

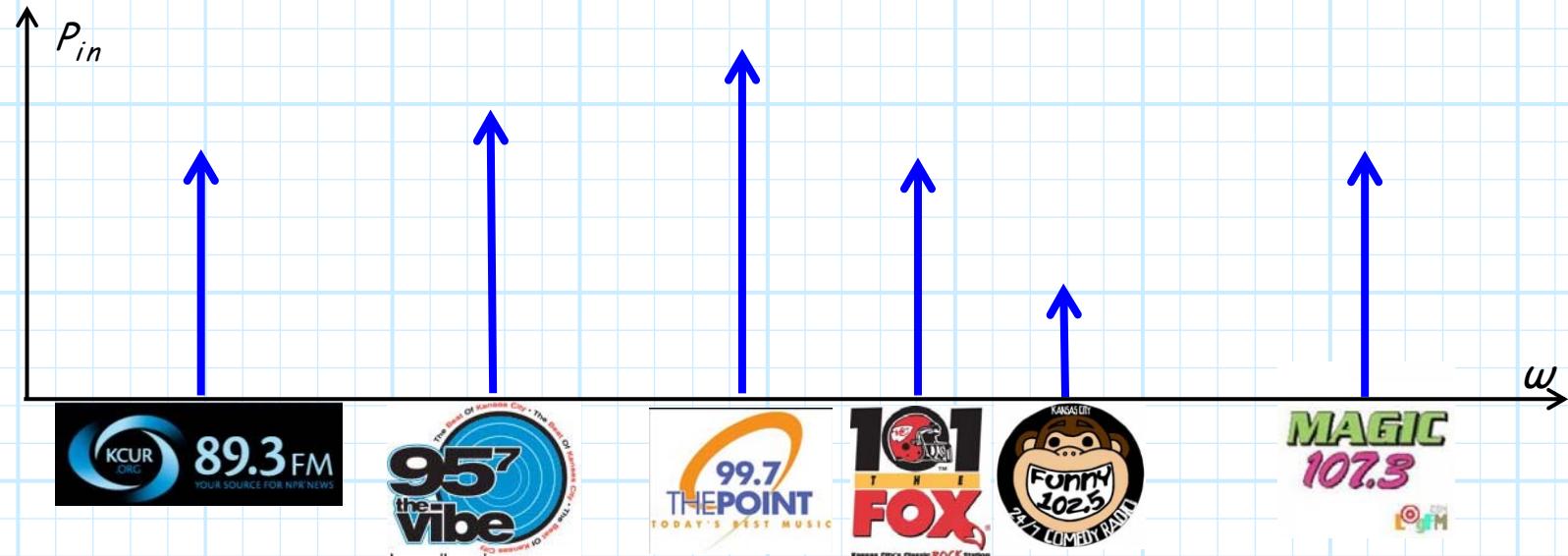
Here we will assume that both frequencies  $\omega_1$  and  $\omega_2$  are within the bandwidth of the amplifier, but are not equal to each other ( $\omega_1 \neq \omega_2$ ).

This of course is a much more realistic case, as typically there will be multiple signals at the input to an amplifier!



## For example, consider the FM band

For example, the two signals considered here could represent two **FM radio stations**, operating at frequencies within the FM band (i.e.,  $88.1 \text{ MHz} \leq f_1 \leq 108.1 \text{ MHz}$  and  $88.1 \text{ MHz} \leq f_2 \leq 108.1 \text{ MHz}$  )



## Harmonics: not the problem



*Q: My point exactly! Intermodulation distortion will produce those **dog-gone** second-order products:*

$$\frac{a^2}{2} \cos 2w_1 t \quad \text{and} \quad \frac{a^2}{2} \cos 2w_2 t$$

*and **gul-durn** third order products:*

$$\frac{a^3}{4} \cos 3w_1 t \quad \text{and} \quad \frac{a^3}{4} \cos 3w_2 t$$

*but these harmonic signals will lie well **outside** the FM band!*

**A:** True! Again, the harmonic signals are **not** the problem.

The problem occurs when the **two input signals** combine together to form **additional** second and third order products.

## A mysterious pair of new signals

Recall an amplifier output is accurately described as:

$$v_{out} = A_v v_{in} + B v_{in}^2 + C v_{in}^3 + \dots$$

Consider first the second-order term if two signals are at the input to the amplifier:

$$\begin{aligned} v_2^{out} &= B v_{in}^2 \\ &= B(a \cos \omega_1 t + a \cos \omega_2 t)^2 \\ &= B(a^2 \cos^2 \omega_1 t + 2a^2 \cos \omega_1 t \cos \omega_2 t + a^2 \cos^2 \omega_2 t) \end{aligned}$$

Note the first and third terms of the above expression are precisely the same as the terms we examined on the previous handout.

They result in harmonic signals at frequencies  $2\omega_1$  and  $2\omega_2$ , respectively.

The **middle** term, however, is something new!

# More than just harmonics are created!

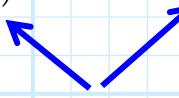
Note this middle term involves the product of  $\cos \omega_1 t$  and  $\cos \omega_2 t$ .

Again using our knowledge of **trigonometry**, we find:

$$2a^2 \cos \omega_1 t \cos \omega_2 t = a^2 \cos(\omega_2 - \omega_1)t + a^2 \cos(\omega_2 + \omega_1)t$$

Note that since  $\cos(-x) = \cos x$ , we can **equivalently** write this as:

$$2a^2 \cos \omega_1 t \cos \omega_2 t = a^2 \cos(\omega_1 - \omega_2)t + a^2 \cos(\omega_1 + \omega_2)t$$



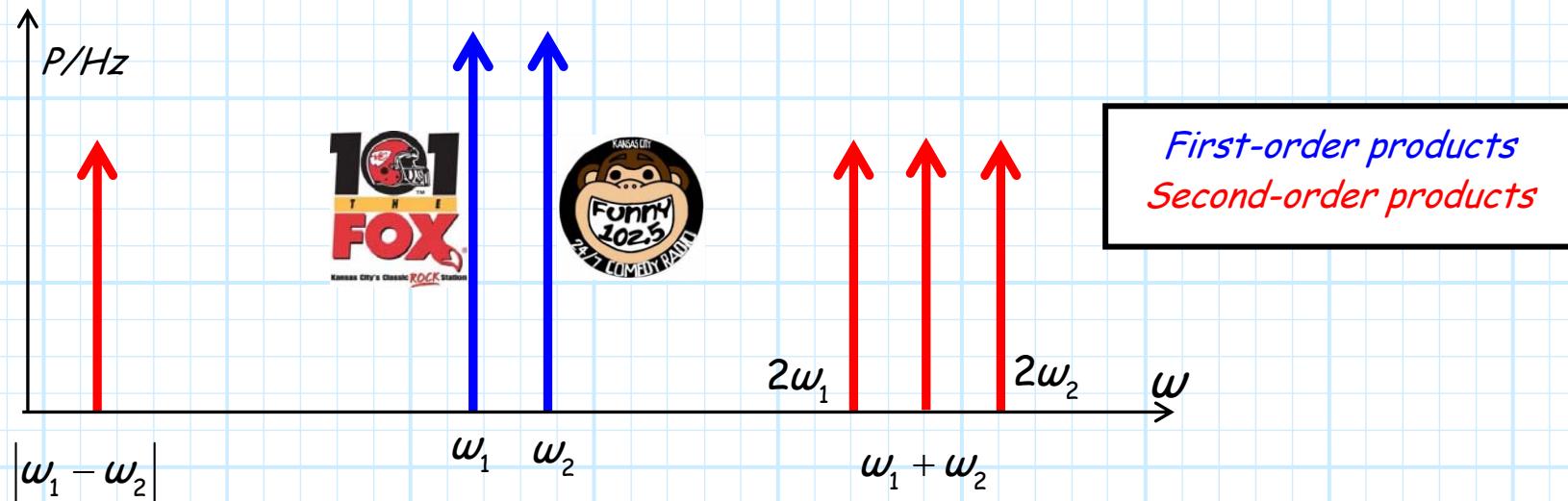
Either way, the result is obvious—we produce **two new signals!**

→ These new **second-order** signals oscillate at frequencies:

$$(\omega_1 + \omega_2) \quad \text{and} \quad |\omega_1 - \omega_2|$$

## Look at the amplifier output-Yikes!

Thus, if we looked at the **frequency spectrum** (i.e., signal power as a function of frequency) of an amplifier **output** when **two sinusoids** are at the input, we would see something like this:



Note that the new terms have a frequency that is either much **higher** than both  $\omega_1$  and  $\omega_2$  (i.e.,  $(\omega_1 + \omega_2)$ ), or much **lower** than both  $\omega_1$  and  $\omega_2$  (i.e.,  $|\omega_1 - \omega_2|$ ).

→ Either way, these **new** signals will typically be **outside** the amplifier bandwidth!

## Don't forget third-order products

**Q:** *Outside the bandwidth! I thought you said these "two-tone" intermodulation products were some "big problem".*

*These sons of a gun appear to be no more a problem than the harmonic signals!*

**A:** This observation is indeed correct for **second**-order, two-tone intermodulation products.



But, we have **yet** to examine the **third**-order terms! I.E.,

$$V_3^{out} = C V_{in}^3 = C (\alpha \cos \omega_1 t + \alpha \cos \omega_2 t)^3$$

→ If we multiply this all out, and again apply our trig knowledge, we find that a bunch of new **third-order** signals are created!!

## Even more mysterious new signals

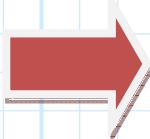
Among these signals, of course, are the second **harmonics**  $\cos 3\omega_1 t$  and  $\cos 3\omega_2 t$ .

Additionally, however, we get these **new signals**:

$$\cos(2\omega_2 - \omega_1)t \text{ and } \cos(2\omega_1 - \omega_2)t$$

Note since  $\cos(-x) = \cos x$ , we can **equivalently** write these terms as:

$$\cos(\omega_1 - 2\omega_2)t \text{ and } \cos(\omega_2 - 2\omega_1)t$$

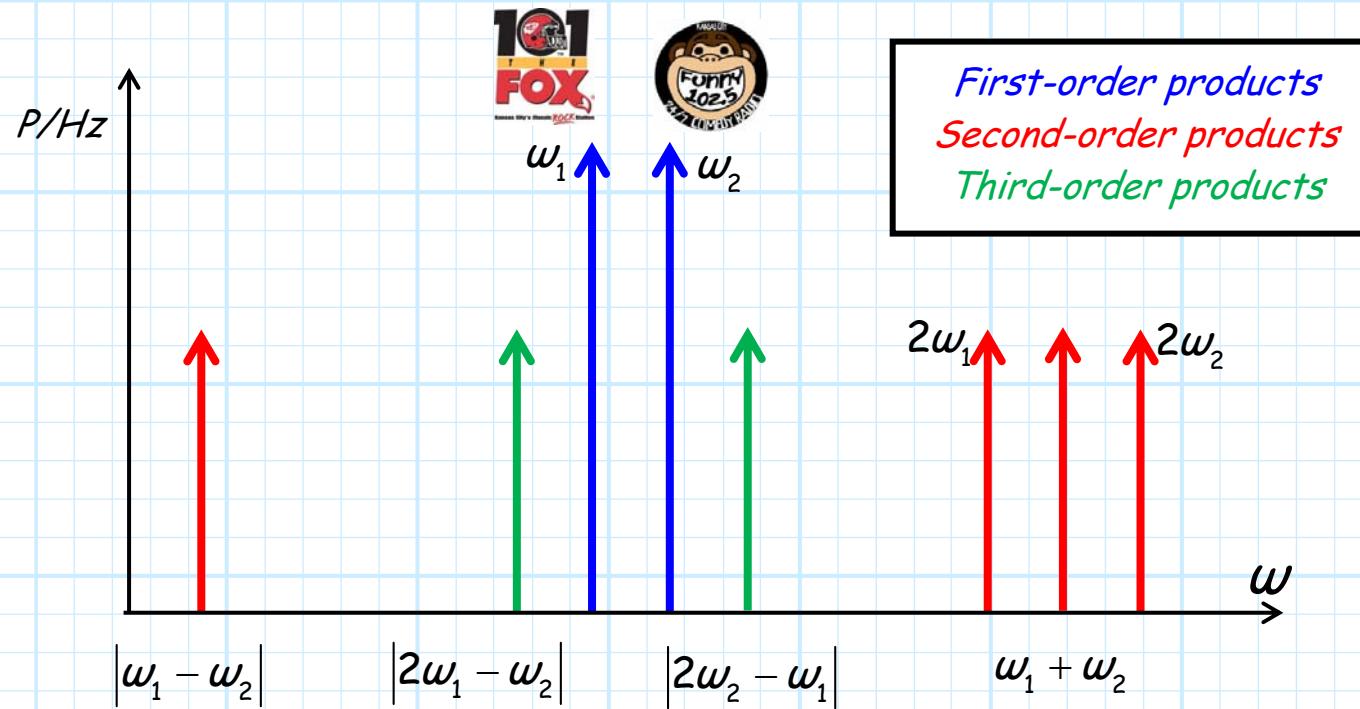


Either way, it is apparent that the **third-order** products include signals at frequencies:

$$|\omega_1 - 2\omega_2| \text{ and } |\omega_2 - 2\omega_1|.$$

# Look at the amplifier output—double Yikes!

Now let's look at the output spectrum with these new third-order products included:



Now you should see the problem!

These third-order products are very close in frequency to  $w_1$  and  $w_2$ .

They will likely lie within the bandwidth of the amplifier!

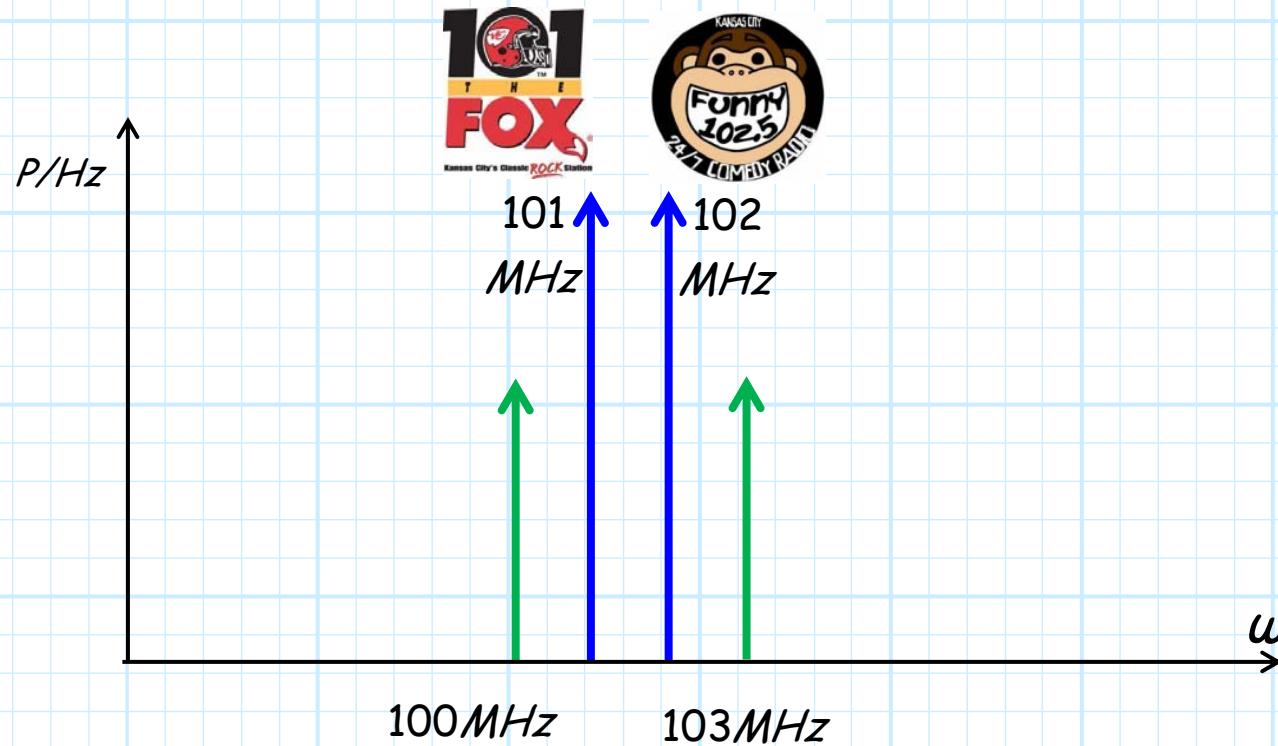
# And there might be a station at 100 or 103

For example, if  $f_1 = 101 \text{ MHz}$  and  $f_2 = 102 \text{ MHz}$ , then:

$$2f_1 - f_2 = 100 \text{ MHz} \text{ and,}$$

$$2f_2 - f_1 = 103 \text{ MHz.}$$

→ All frequencies are well within the FM radio bandwidth!



## Make sure IP3 is high!

Thus, these **third-order**, two-tone intermodulation products are the most significant distortion terms.

→ This is why we are most concerned with the **third-order** intercept point of an amplifier!



*I select amplifiers with  
the highest possible  
3<sup>rd</sup> order intercept point!*