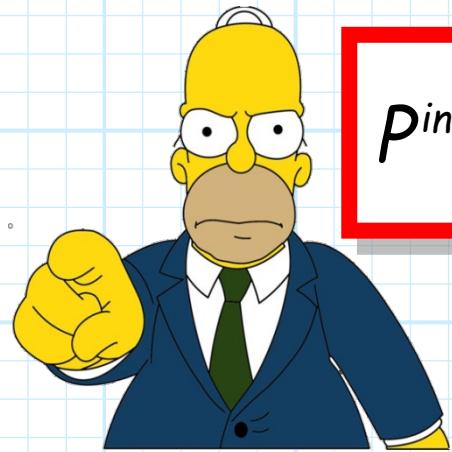


Incident and Reflected Power with “Matched” Source or Load

Q: So. You've made a big, gigantic, dramatic deal about the fact that the *incident power along a transmission line is greater than that delivered by the source*:



$$P_{\text{inc}} \geq P_g^{\text{del}} = P_L^{\text{abs}}$$



But. Doesn't the *inequality \geq* mean that the *incident power could be equal to the delivered power of the source?*

A: For the right circumstance, yes; the two could be equal ($P_{\text{inc}} = P_g^{\text{del}}$).

When there's no reflected power

Q: The right circumstance?

What circumstance is the right circumstance in order for $P_g^{del} = P^{inc}$?

A: Sigh.

Just look at the relationship:

$$P_g^{del} = P^{inc} - P^{ref}$$

Hopefully it is apparent, that when (and only when!) the reflected power is zero ($P^{ref} = 0$), the incident and delivered power will be equal:

$$P_g^{del} = P^{inc} \text{ when } P^{ref} = 0$$

When the load reflection coefficient is zero

Q: OK; then under what circumstance is the reflected power zero?

A: Sigh.

Just look at the relationship:

$$P^{\text{ref}} = P^{\text{inc}} |\Gamma_L|^2$$

Clearly, the reflected power will be zero if $|\Gamma_L|^2 = 0$ (i.e., if $\Gamma_L = 0$):

$$P_g^{\text{del}} = P^{\text{inc}} \quad \text{when} \quad \Gamma_L = 0$$

When the load is “matched”

Q: Well then. Under what @#\$%^!&% circumstance is Γ_L zero?!?

A: Sigh.

Recall the definition of Γ_L :

$$\Gamma_L = \frac{Z_L - Z_0}{Z_L + Z_0}$$

From the numerator, it is apparent that Γ_L is zero if (and **only if!**) the load impedance Z_L is **numerically equal to the characteristic impedance Z_0** of the transmission line:

$$\Gamma_L|_{Z_L=Z_0} = \left. \frac{Z_L - Z_0}{Z_L + Z_0} \right|_{Z_L=Z_0} = \frac{Z_0 - Z_0}{Z_0 + Z_0} = 0$$

A special case

Q: Hey; wasn't $Z_L = Z_0$ one of the "special cases" that we studied earlier?

A: It was!

Recall that when $Z_L = Z_0$, the **input impedance** of our terminated transmission line is likewise numerically equal to Z_0 —**regardless of line length ℓ !**

We now see why this is true.

Matched load means no reflected wave

When $Z_L = Z_0$, then $\Gamma_L = 0$, and so:

$$V_0^- = \Gamma_L V_0^- = (0)V_0^- = 0$$

Meaning the **reflected wave** (if $Z_L = Z_0$) is **zero**:

$$V^-(z) = V_0^- e^{j\beta z} = 0$$

So, our **total voltage** and current along the transmission line is that of the plus (i.e., **incident**) wave only:

$$V(z) = V^+(z) \quad \text{and} \quad I(z) = I^+(z)$$

The **line impedance function** is thus (when $Z_L = Z_0$):

$$Z(z) = \frac{V(z)}{I(z)} = \frac{V^+(z)}{I^+(z)} = Z_0$$

"Matched" load means...

In summary, for a "matched"^{*} load:

a) the reflected wave is **zero**:

$$V^-(z) = 0 \quad \text{if} \quad Z_L = Z_0$$

b) the power of the **reflected** wave is likewise **zero**:

$$P_{\text{ref}} = \frac{|V^-(z)|^2}{2Z_0} = 0 \quad \text{if} \quad Z_L = Z_0$$

c) meaning the incident power will be **equal** to the delivered power:

$$P_{\text{inc}} = P_g^{\text{del}} = P_L^{\text{abs}} \quad \text{if} \quad Z_L = Z_0$$

* does **not** imply a conjugate match!

Now for the “matched” source

Q: Things sure seem to *simplify* if the load is “matched”.

What about a “matched” source?

Do any of the power relationships simplify then?

A: They sure do!

Recall for a “matched” source, the incident wave becomes **causal** with respect to the source:

$$V^+(z) = V_0^+ e^{-j\beta z} = \left(\frac{1}{2} V_g e^{-j\beta \ell}\right) e^{-j\beta z} = \frac{V_g}{2} e^{-j\beta(z+\ell)}$$

In other words, for this **special case** (and **only** for this special case!), the indent wave is dependent on the **source only** (the value of load Z_L is irrelevant).

All I can say is: wow!

Recall the power associated with the **incident wave** is:

$$P^{inc} = \frac{|V^+(z)|^2}{2Z_0}$$

so that for a "matched" source:

$$P^{inc} = \frac{|V^+(z)|^2}{2Z_0} = \frac{1}{2Z_0} \left| \frac{V_g}{2} e^{-j\beta(z+\ell)} \right|^2 = \frac{|V_g|^2}{8Z_0} \quad \leftarrow \text{WOW!!!!}$$

Q: I don't see why this is "wow". Am I missing something?

A: Apparently you are. Look closer at the above result.

Do I have to explain everything?

Remember, this result is for the **special case** where the source impedance is a **real value**, numerically equal to Z_0 :

$$Z_g = Z_0 + j0$$

Therefore:

$$R_g = \operatorname{Re}\{Z_g\} = Z_0$$

The **source resistance** is numerically equal to transmission line characteristic impedance Z_0 .

Thus, the **power of the incident wave** can be alternatively written as:

$$P^{inc} = \frac{|V_g|^2}{8Z_0} = \frac{|V_g|^2}{8R_g}$$

← WOW!!!!

The incident power is the available power (wow)!

Q: This result looks vaguely familiar; haven't we seen this before?

A: Sigh. This result is the available power of the source!!!!!!

Therefore:

$$P^{inc} = \frac{|V_g|^2}{8Z_0} = \frac{|V_g|^2}{8R_g} = P_g^{avl} \quad \leftarrow \text{Wow!}$$

For the special case $Z_g = Z_0 + j0$, the incident power P^{inc} is equal to the available power P_g^{avl} of the source.

"Matched" source means...

In summary, for a "matched"^{*} source:

a) the power of the incident wave is:

$$P^{inc} = \frac{|V_g|^2}{8 Z_0} \quad \text{if} \quad Z_g = Z_0$$

b) meaning the incident power will be equal to the available power of the "matched" source:

$$P^{inc} = P_g^{avl} \quad \text{if} \quad Z_g = Z_0$$

* does not imply a conjugate match!