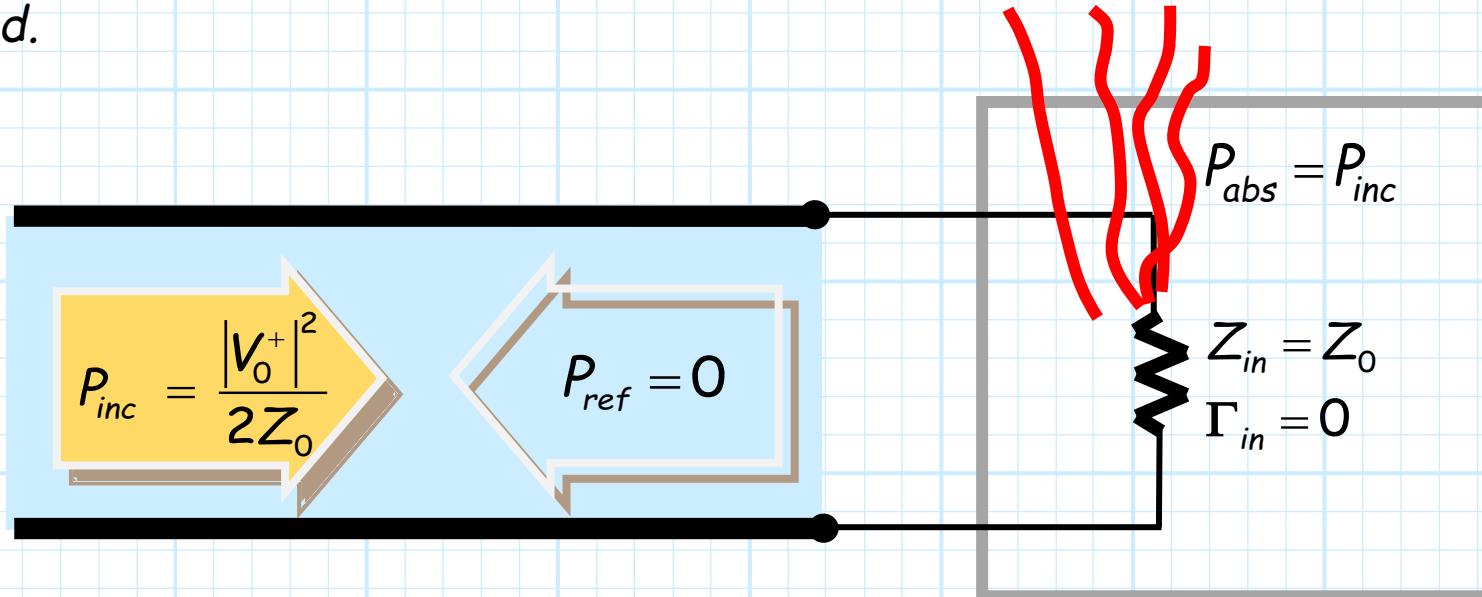


Return Loss and VSWR

Q: If the purpose of a transmission line is to transfer **energy** from some **source** (the output of some useful device) to some **load** (the input to some other useful device), then wouldn't an input impedance of Z_0 be ideal.

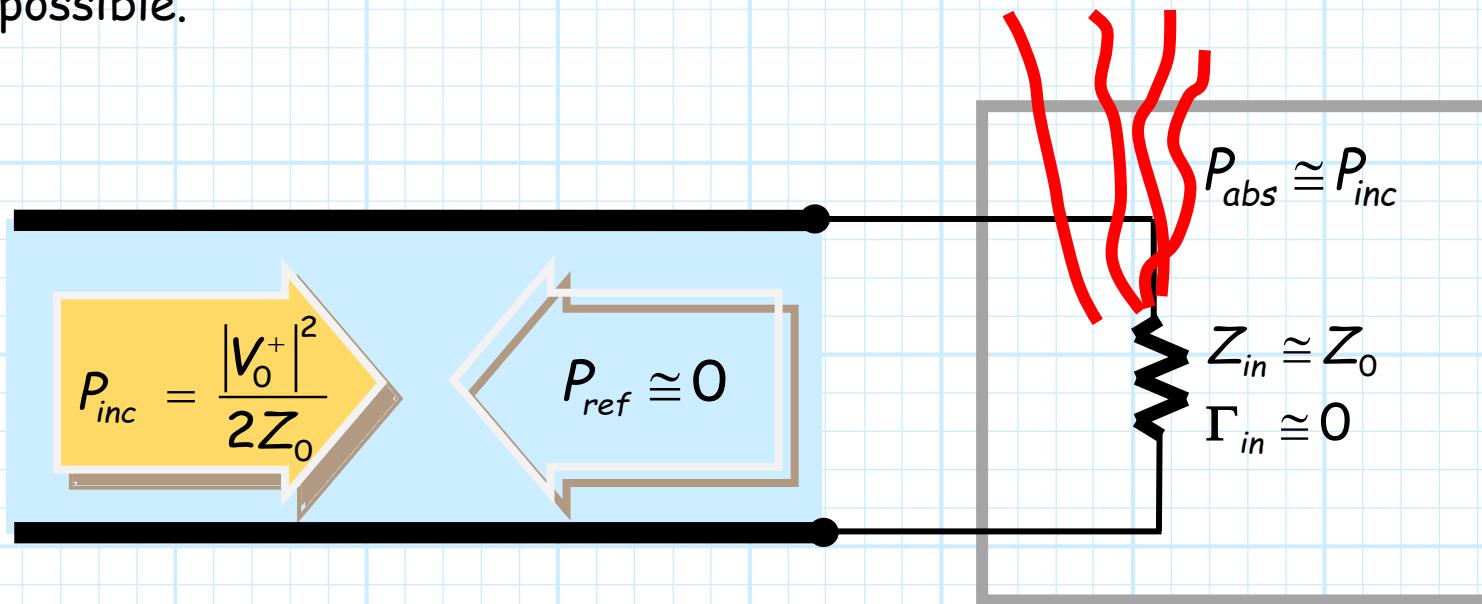
After all, no energy is reflected—all incident energy is absorbed by the load.



A: If connected entirely to other "matched" devices, it would be!

Close, but not exactly

Thus, microwave component vendors often try very hard to make the **input impedance** of their useful devices as **close** to Z_0 ($Z_0 = 50\Omega$, say) as possible.



But the value:

is really hard to come by!

It's not like they're lying (usually)

Q: So do vendors tell us what Z_{in} really is?

A: Not usually!

Remember, input **impedance**—as well as its **reflection coefficient** Γ_{in} —is a **complex** value.

Therefore microwave engineers often use **real-valued** measures to indicate how “**matched**” a given load (e.g., the input impedance of some useful device) really is!

Two of the most prevalent **real-valued** measures are:

1. **return loss**, and
2. **VSWR**.

Return Loss

The ratio of the incident power to the reflected power is known as return loss.

Typically, return loss is expressed in decibels:

$$R.L. = 10 \log_{10} \left[\frac{P_{inc}}{P_{ref}} \right] = -10 \log_{10} |\Gamma_{in}|^2$$

Where:

$$\Gamma_{in} = \frac{Z_{in} - Z_0}{Z_{in} + Z_0}.$$

Return Loss examples

For example, if the return loss is 10dB, then:

- A. 10% of the incident power is reflected, so...
- B. the remaining 90% is absorbed, thus...
- C. we "lose" 10% of the incident power.

Likewise, if the return loss is 30dB, then:

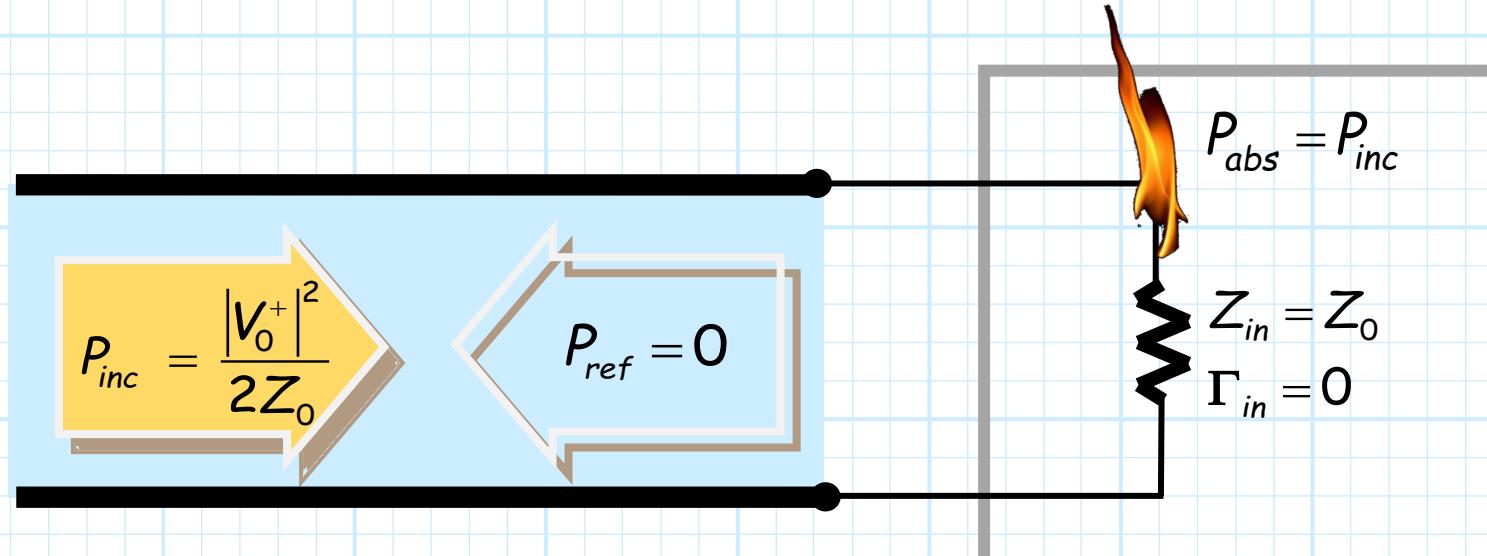
- A. 0.1% of the incident power is reflected, so...
- B. the remaining 99.9% is absorbed, thus...
- C. we "lose" 0.1% of the incident power.

Thus, a larger numeric value for return loss actually indicates less lost power!

More return loss is actually better

An ideal return loss thus would be ∞ dB—the input is a matched load:

$$R.L. = -10 \log_{10} |\Gamma_{in}|^2 = -10 \log_{10} |0|^2 = \infty$$

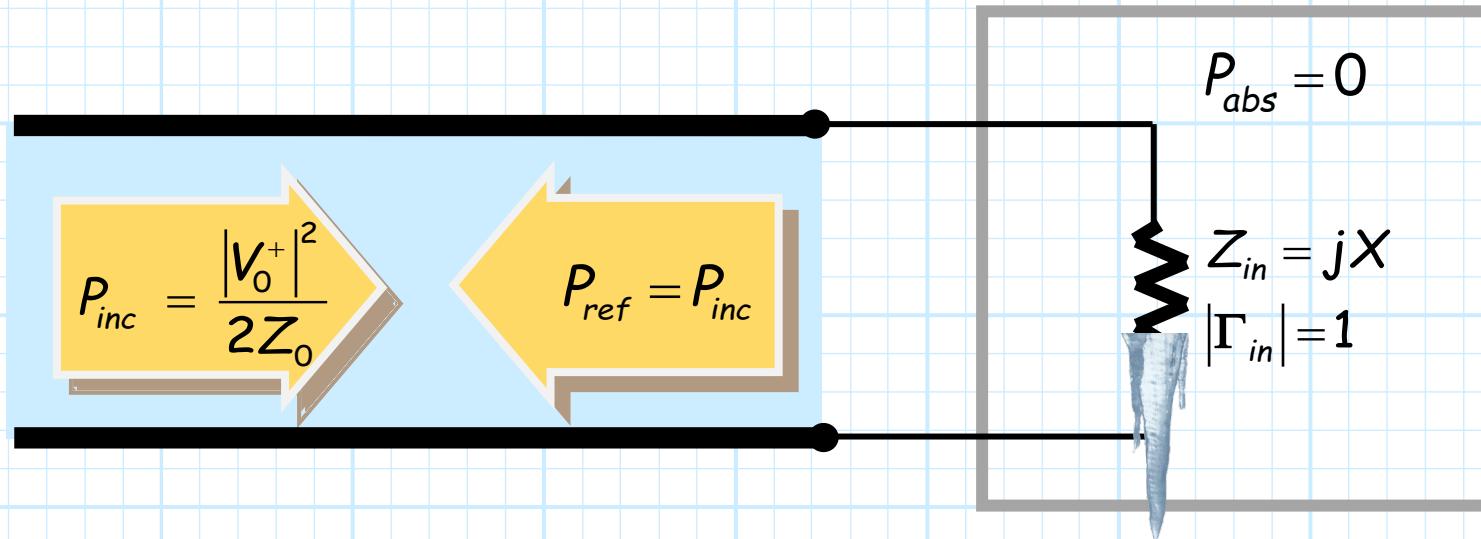


Zero return loss is the worst

The **worst case** is when no power is absorbed—all the incident power is reflected.

For this (worst) case, the return loss of is **zero!**

$$R.L. = -10 \log_{10} |\Gamma_{in}|^2 = -10 \log_{10} |1|^2 = 0$$



VSWR

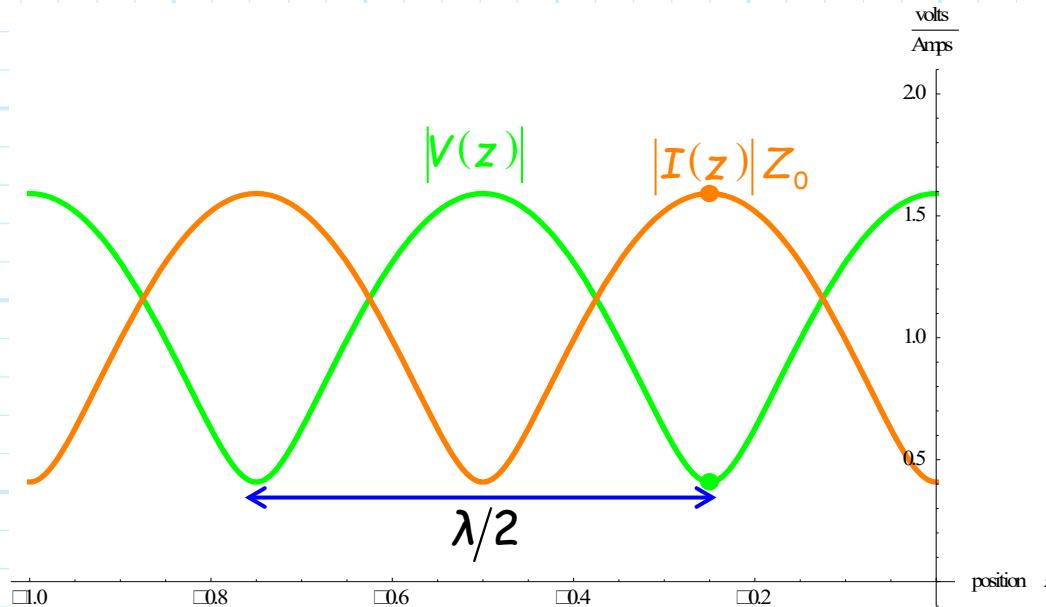
Consider again the **voltage** along a transmission line attached to a port of **some useful device** with impedance Z_{in} (where $z_L = 0$):

$$V(z) = V_0^+ \left[e^{-j\beta z} + \Gamma_{in} e^{+j\beta z} \right]$$

Now let's look at the **magnitude** only:

$$|V(z)| = |V_0^+| |e^{-j\beta z} + \Gamma_{in} e^{+j\beta z}| = |V_0^+| |1 + \Gamma_{in} e^{+j2\beta z}|$$

As this result shows—and as you already know—the magnitude is **not constant** with respect to position z .



Definition

It can be shown that the **largest** and **smallest** values of this magnitude are:

$$|V(z)|_{max} = |V_0^+|(1 + |\Gamma_{in}|) \quad |V(z)|_{min} = |V_0^+|(1 - |\Gamma_{in}|)$$

The ratio of $|V(z)|_{max}$ to $|V(z)|_{min}$ is known as the **Voltage Standing Wave Ratio (VSWR)**:

$$\text{VSWR} \doteq \frac{|V(z)|_{max}}{|V(z)|_{min}} = \frac{1 + |\Gamma_{in}|}{1 - |\Gamma_{in}|} \quad \therefore \quad 1 \leq \text{VSWR} \leq \infty$$

As with return loss, VSWR is a real value that is dependent on the magnitude of Γ_{in} (i.e., $|\Gamma_{in}|$) only !

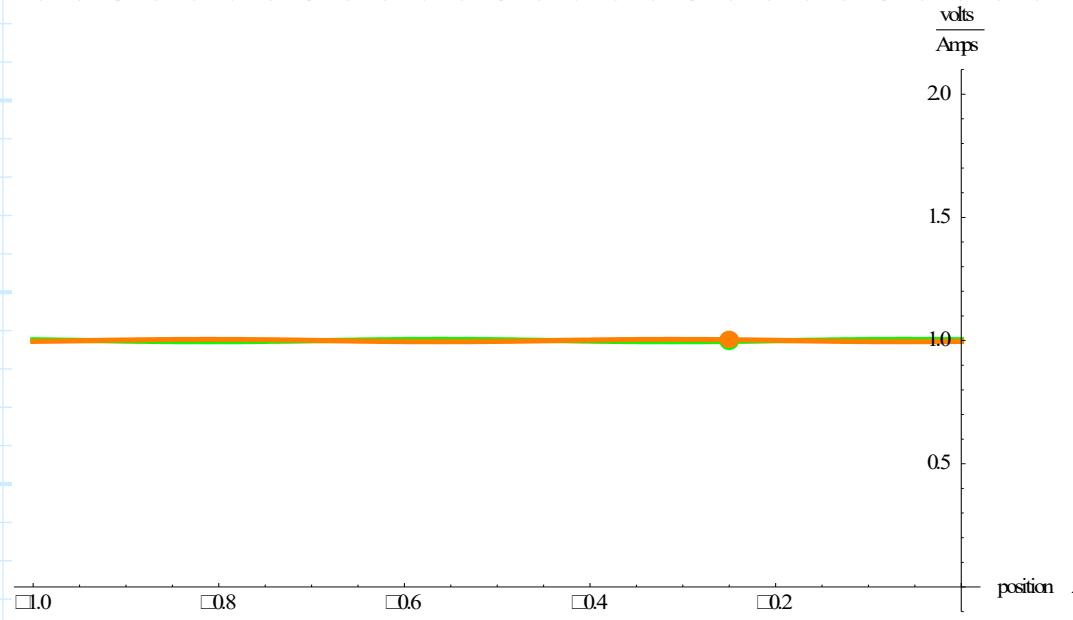
The best VSWR is 1.0

Note for the ideal input impedance, where $|\Gamma_{in}|=0$ (i.e., $Z_{in}=Z_0$), then
VSWR = 1.0 .

We find for this case:

$$|V(z)|_{max} = |V(z)|_{min} = |V_0^+|$$

In other words, the voltage magnitude is a **constant** with respect to position z .



The worst VSWR is infinite

Conversely, for the **worst** input impedance where $|\Gamma_{in}|=1$, then $VSWR=\infty$.

We find for this case:

$$|V(z)|_{min} = 0 \quad \text{and} \quad |V(z)|_{max} = 2 |V_0^+|$$

Recall for this case the voltage magnitude $|V(z)|$ varies greatly with respect to position z !

