

Q1)

$$a) \quad G(s) = \frac{(1-5s)(1-2s)e^{-0.5s}}{(12s+1)(6s+1)(0.3s+1)}$$

Taylor expansion

$$e^{-0.5s} \approx (1 - 0.5s)$$

b) Taylor series approximation. FOPTD -

$$\frac{1}{12s+1} \cdot (e^{-0.5-2-5-6-0.3})^s \Rightarrow \frac{1}{12s+1} \cdot e^{-13.8s}$$

z=12

plot on matlab.

c) Skogestad FOPTD

$$\frac{1}{12s+1} \cdot \frac{1}{(12 + \frac{6}{2})s + 1} \cdot e^{(-0.5-2-5-3-0.3)s} \Rightarrow \frac{1}{(15s+1)} \cdot e^{-10.8s}$$

$$d) G(s) = \frac{K e^{-\theta s}}{s+1}$$

found  
given  $\theta = 10.3$

$$m = 0.0453$$

$$m = \frac{KM}{Z}, \text{ from graph, } K = 1$$

$$\therefore Z = 1/m = 22.1$$

$$f) \frac{(1 - Z_a s) e^{-\theta s}}{(Z_1 s + 1)(Z_2 s + 1)}$$

$$Z_a = 5$$

$$Z_1 = 12$$

$$Z_2 = 6$$

$\therefore$  by Taylor approx,

$$\frac{(1 - 5s) e^{-0.55 - 2s - 0.3s}}{(12s + 1)(6s + 1)}$$

$$\Rightarrow \frac{1 - 5s}{(12s + 1)(6s + 1)} e^{-0.55 - 2s - 0.3s}$$

by Skogestad,

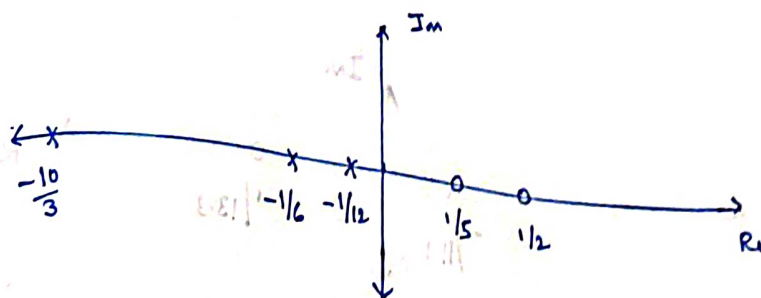
$$\frac{(1 - 5s) e^{-0.55 - 2s - 0.15s}}{(12s + 1)(6 + 0.15s + 1)}$$

$$\Rightarrow \frac{(1 - 5s) e^{-2.65s}}{(12s + 1)(6.15s + 1)}$$

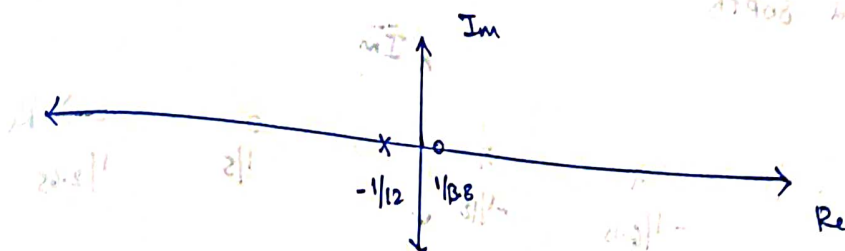
On zooming the graph,

Skogestad gives better.

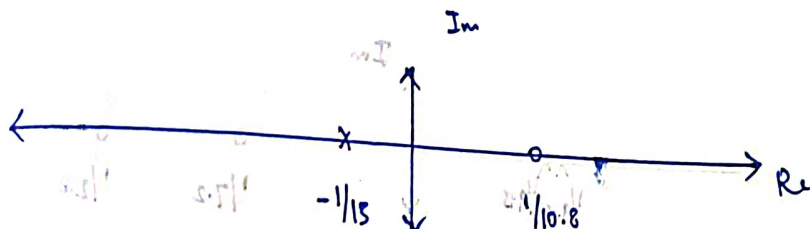
a)



b)

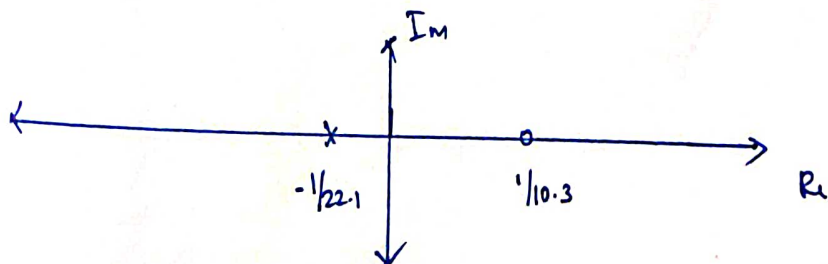


c)

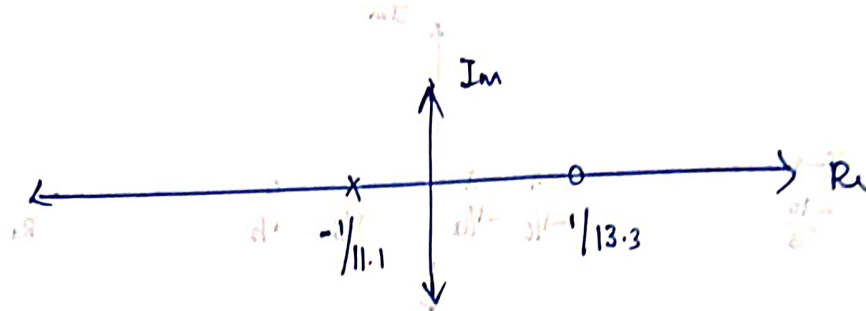


large handle on left  $\Rightarrow$  out of

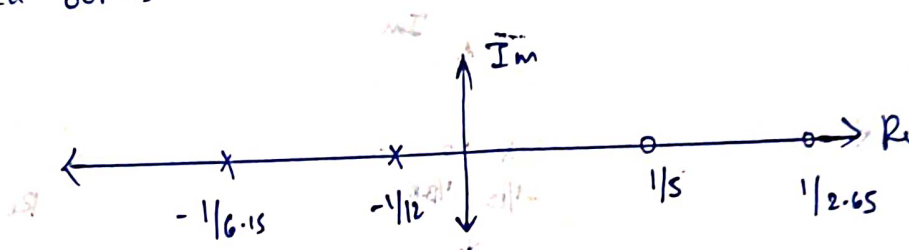
d)



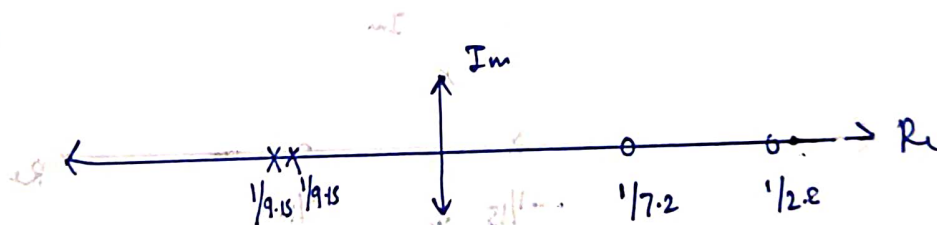
e)



f) Skogerted SOPS



g) Non linear

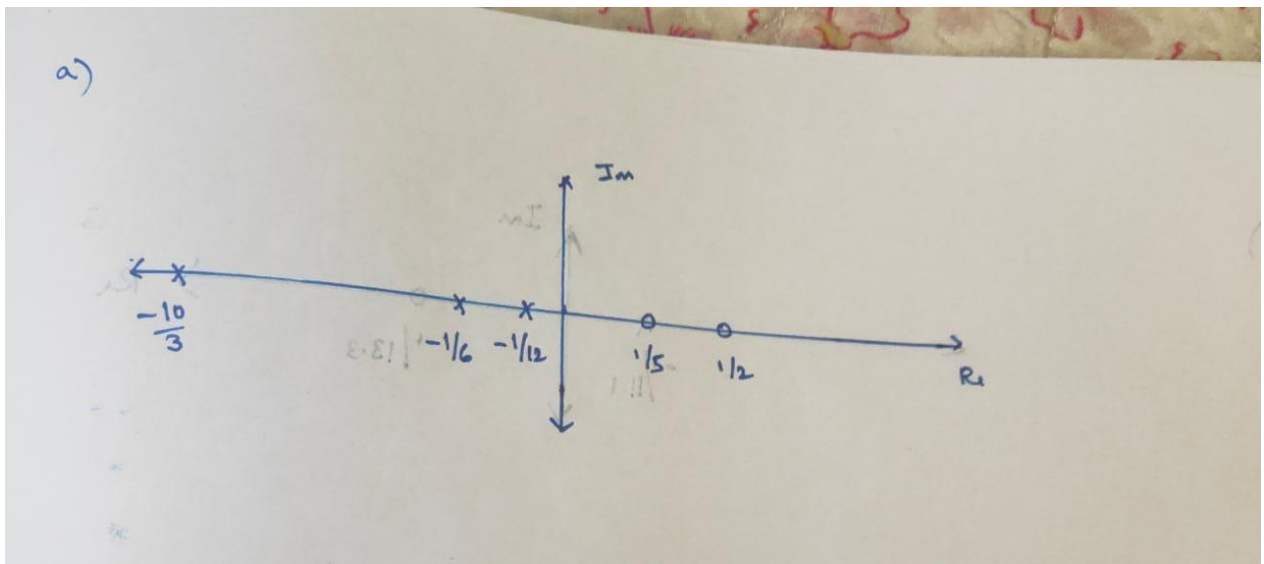
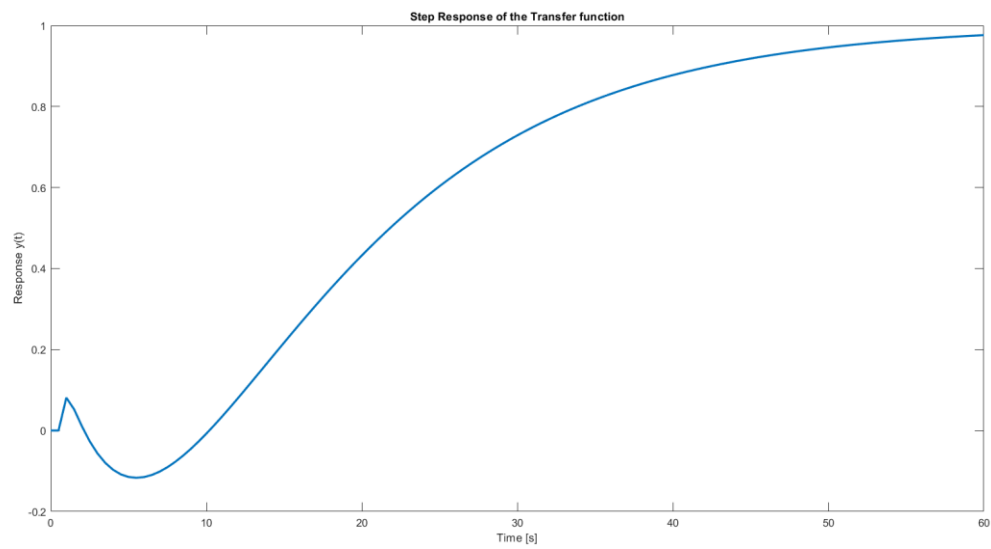


the two poles are almost equal

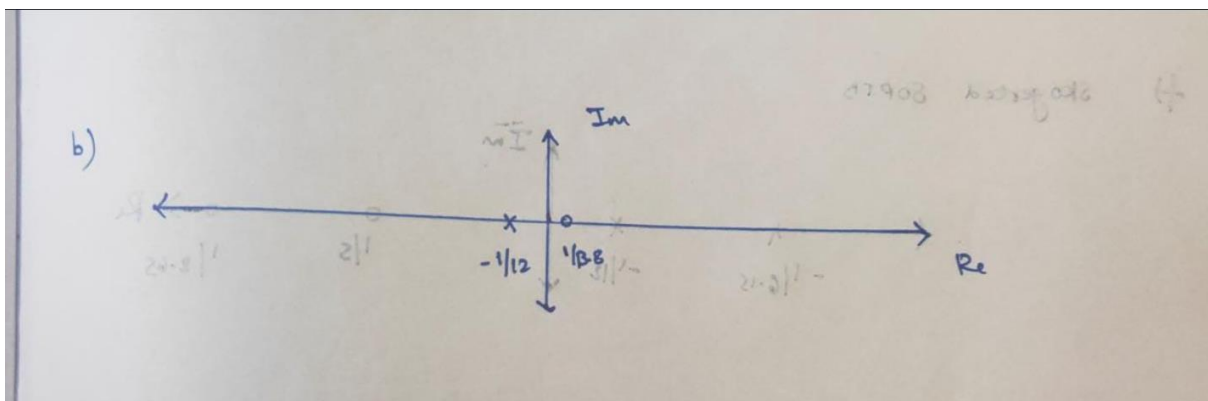
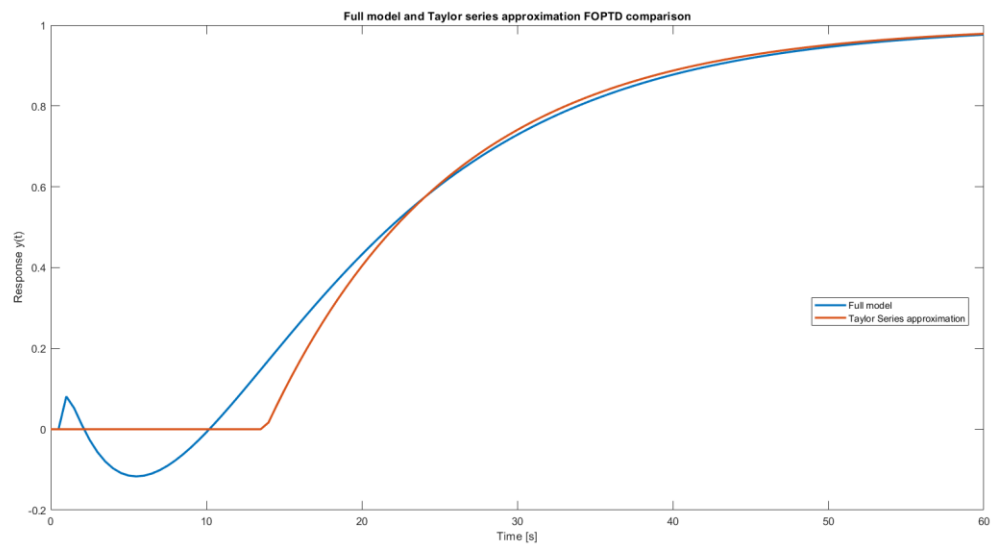


Q1) (Approximated time delay by taylor for the zeros)

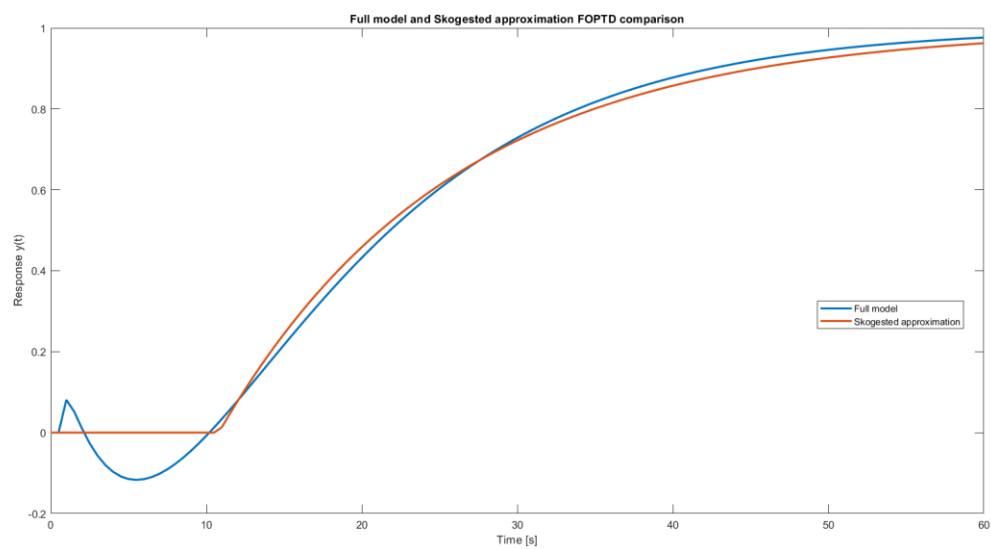
a)

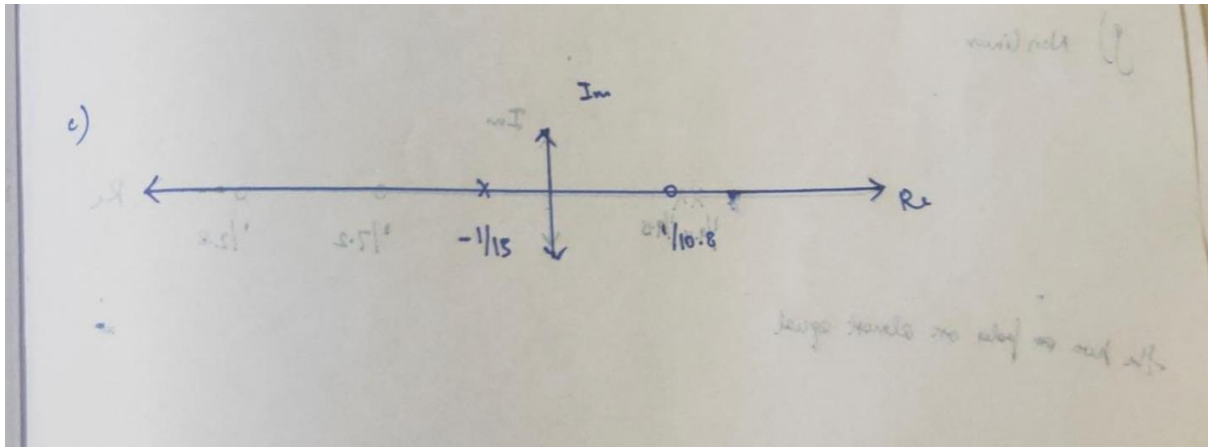


b)

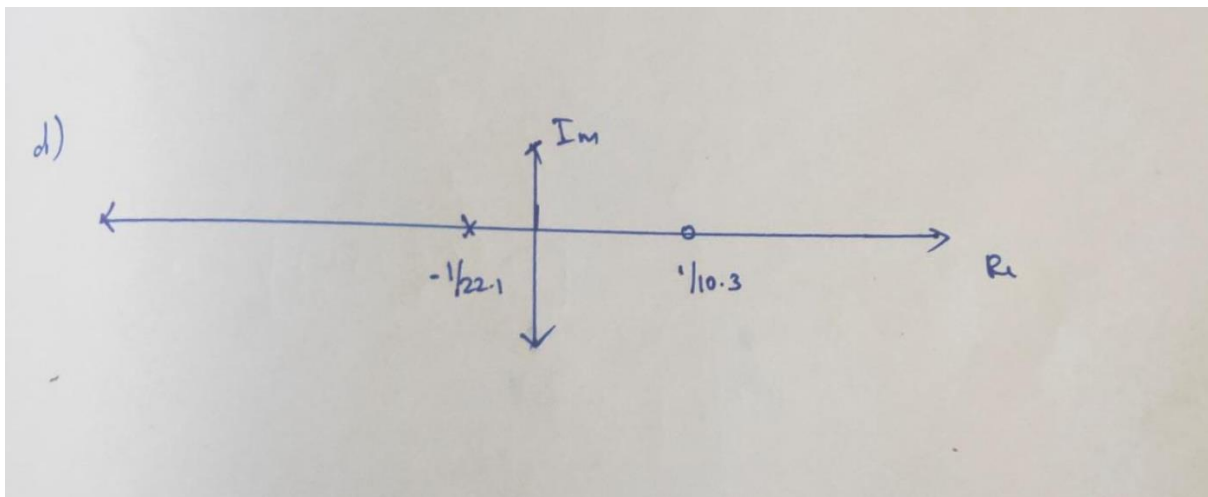
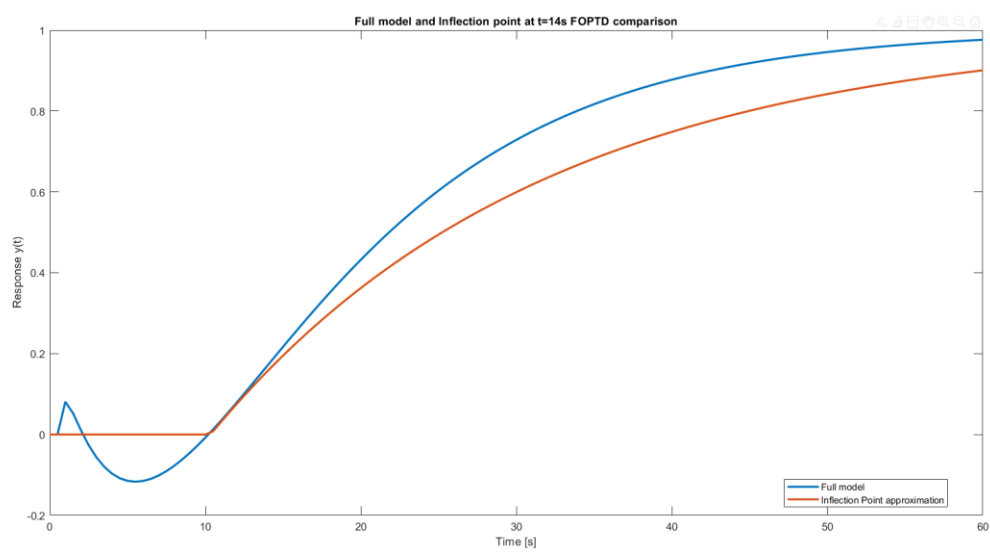


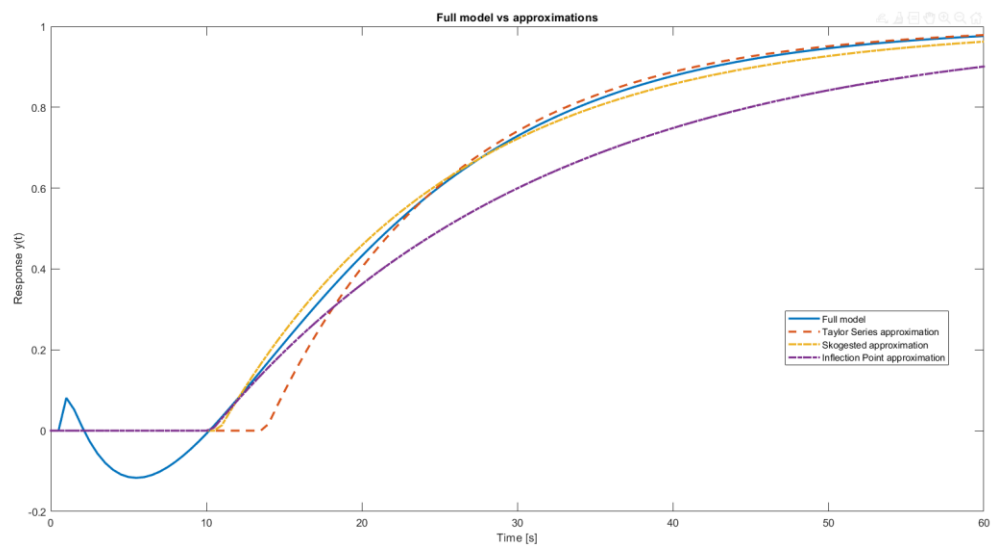
c)



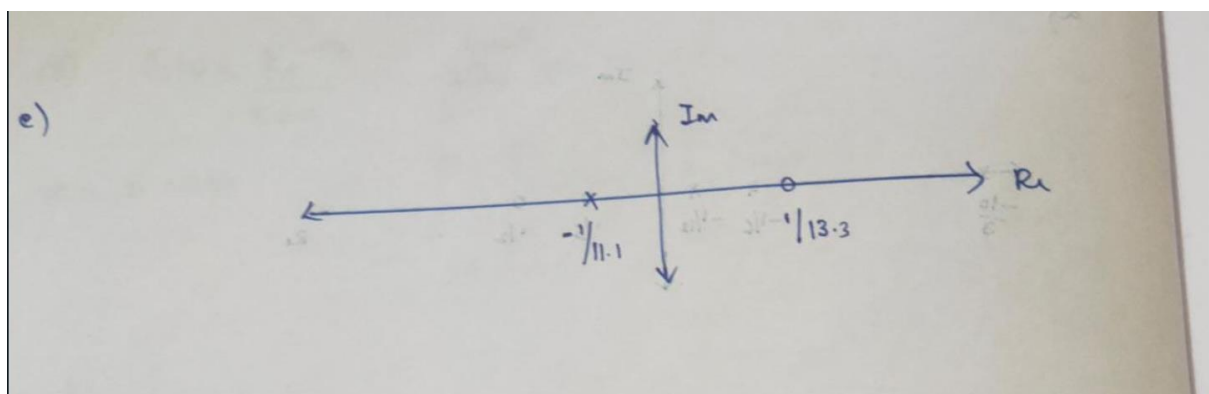
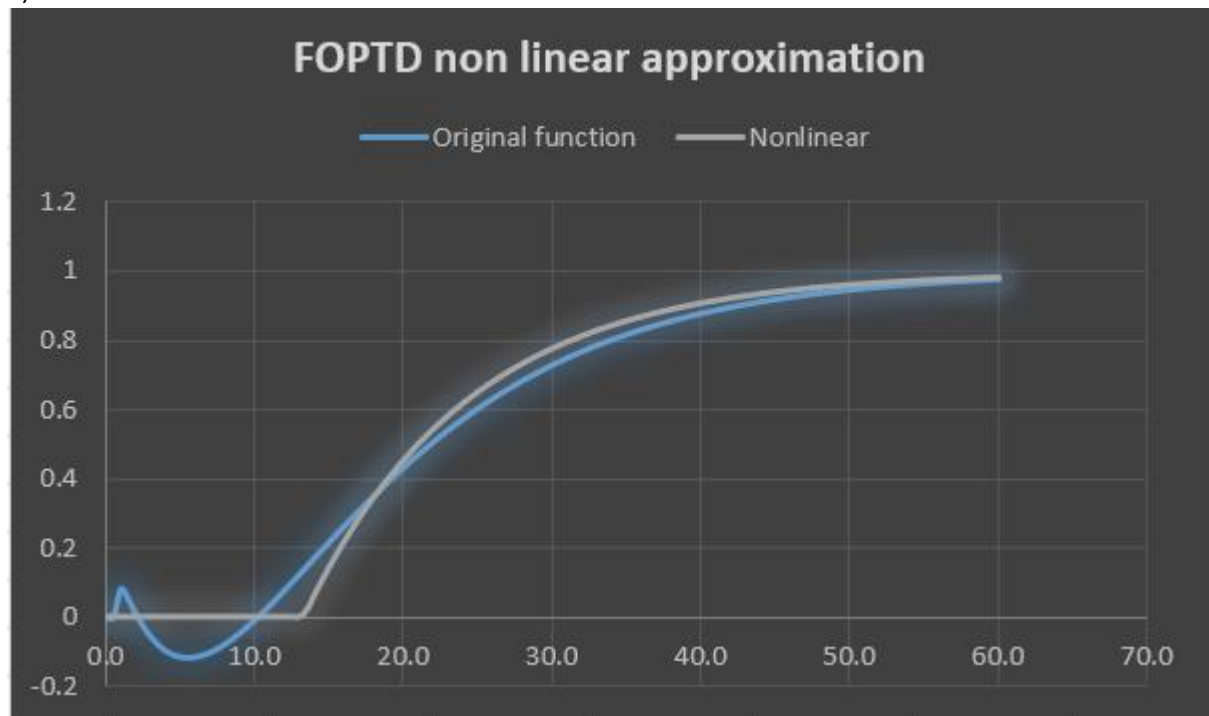


d)

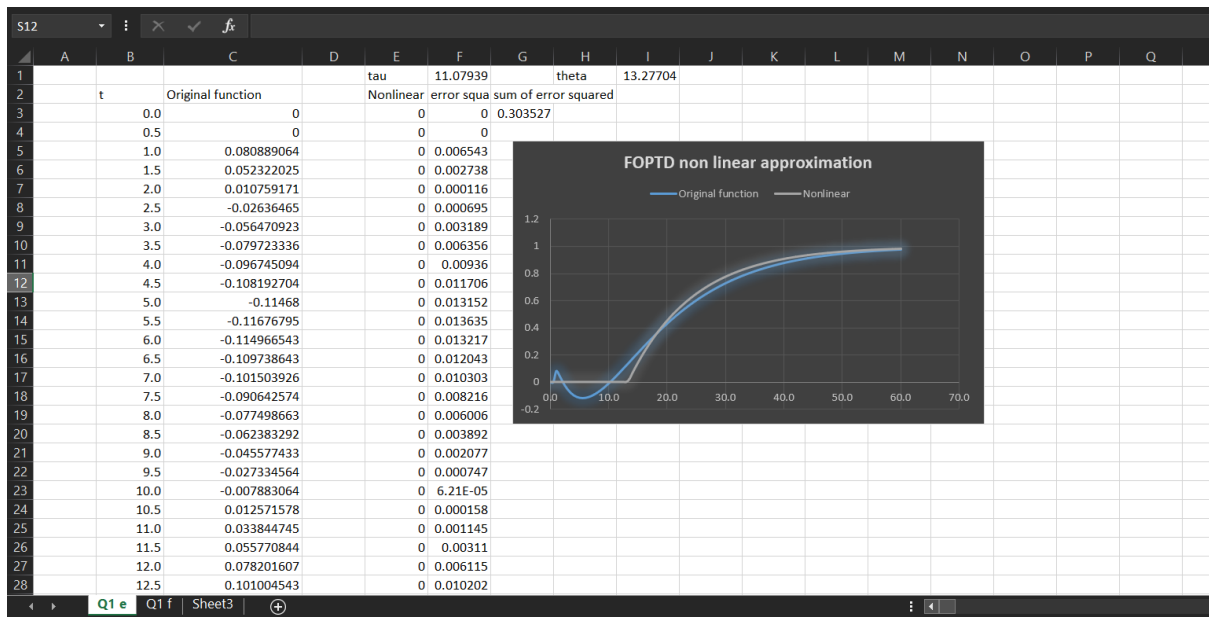




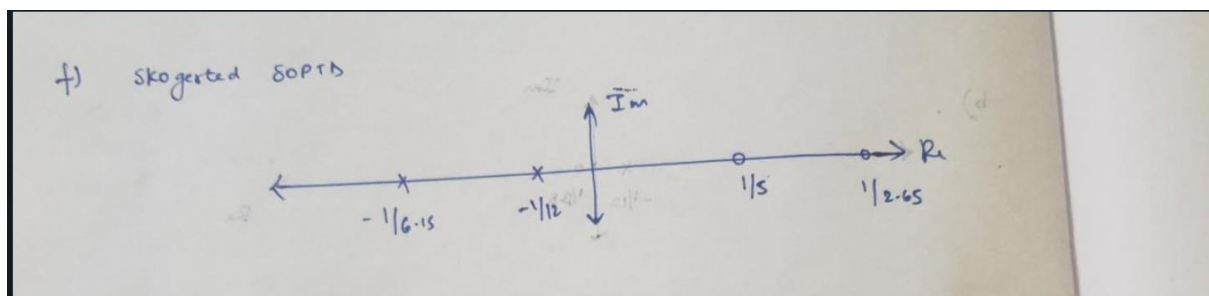
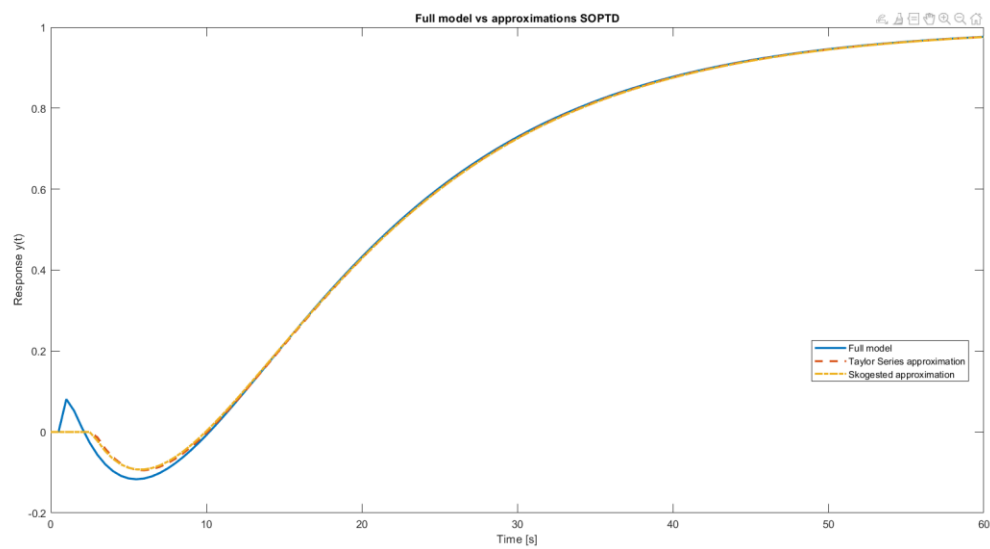
e)



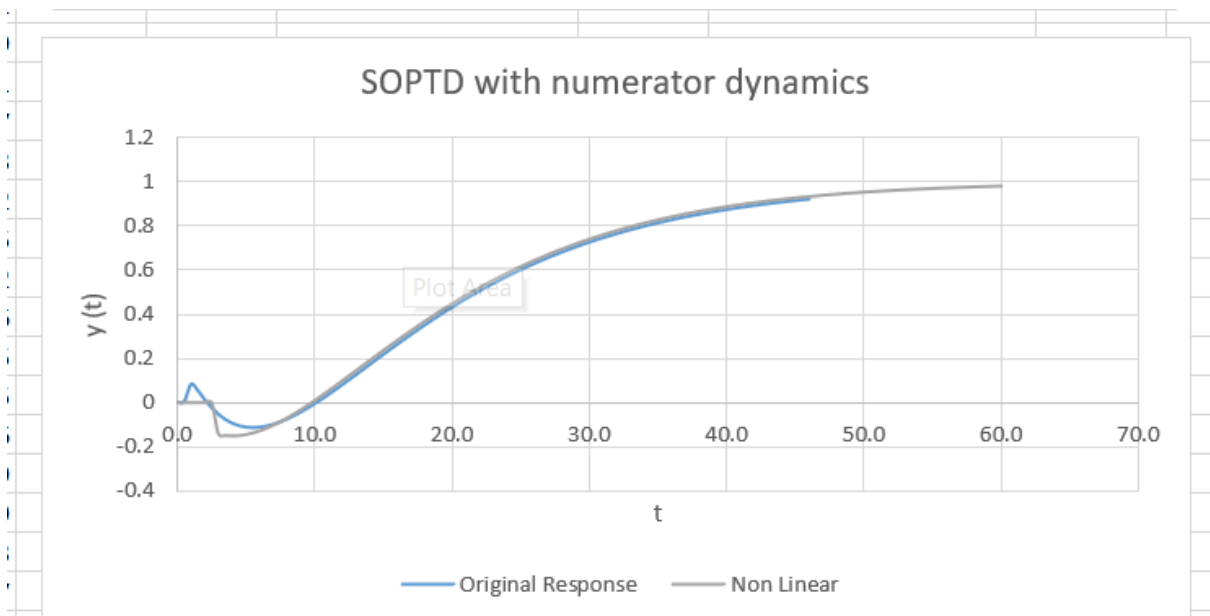
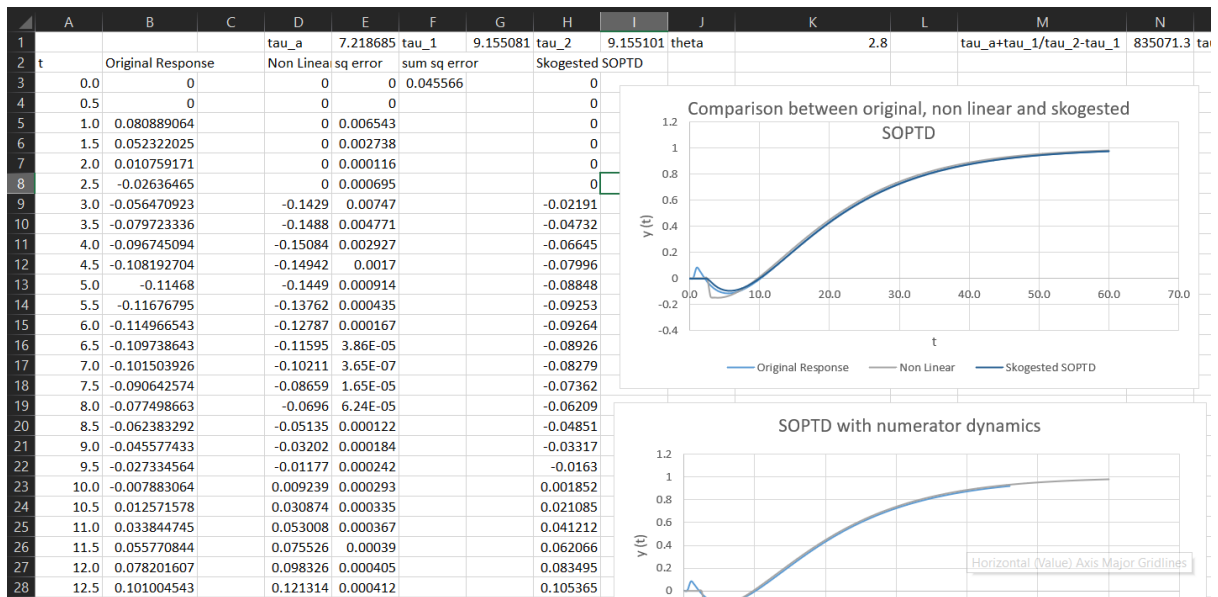




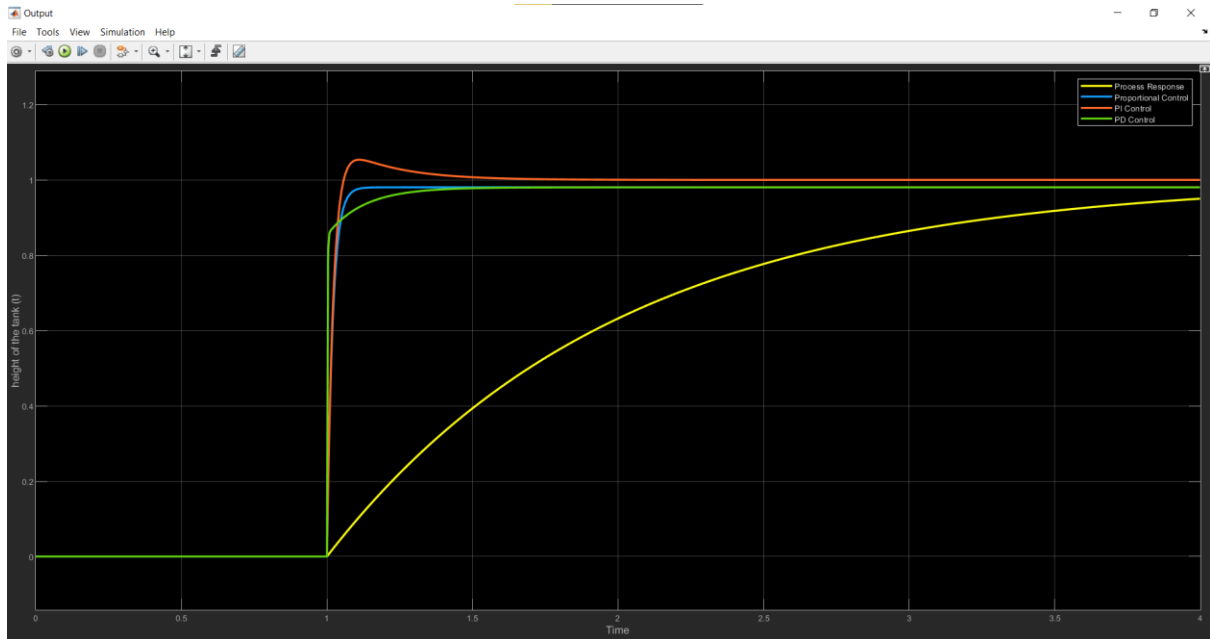
f)



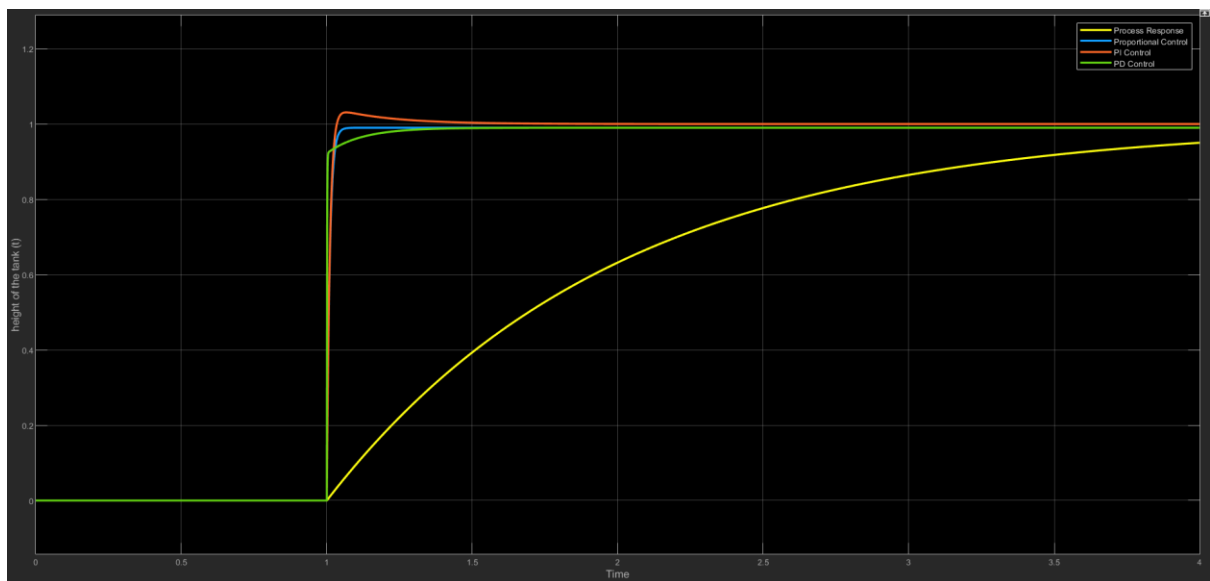
g)



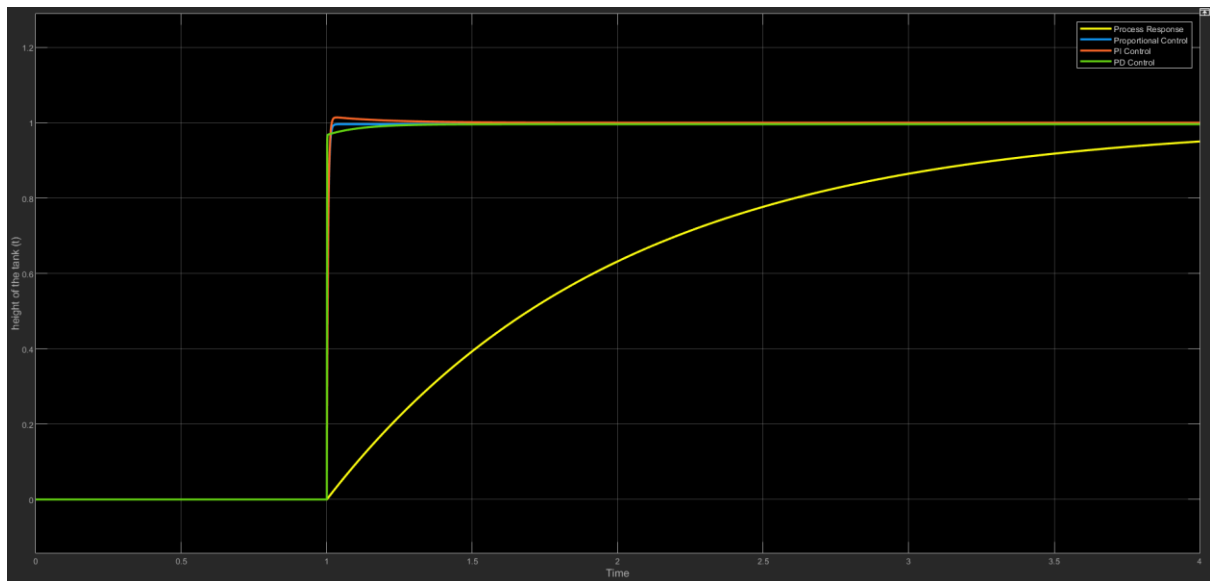
Q2)



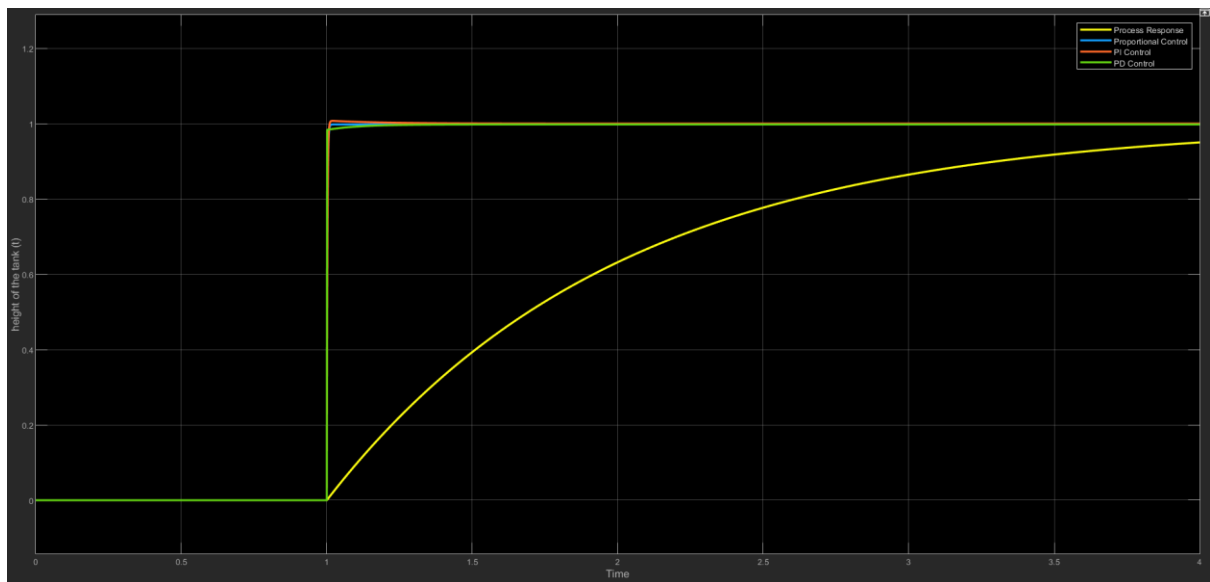
K=50



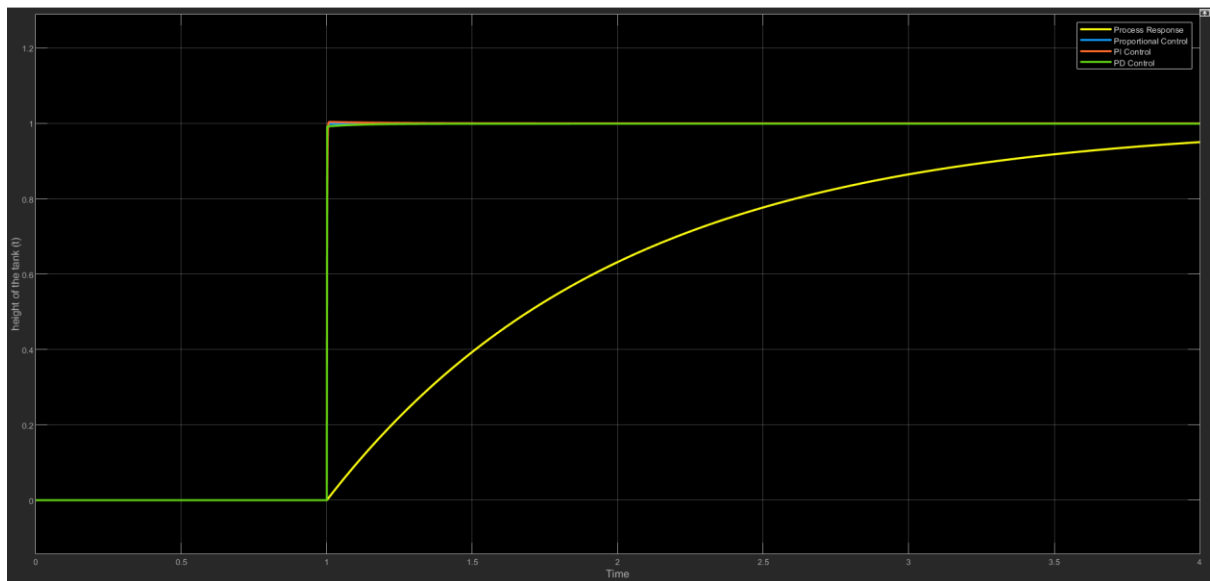
K=100



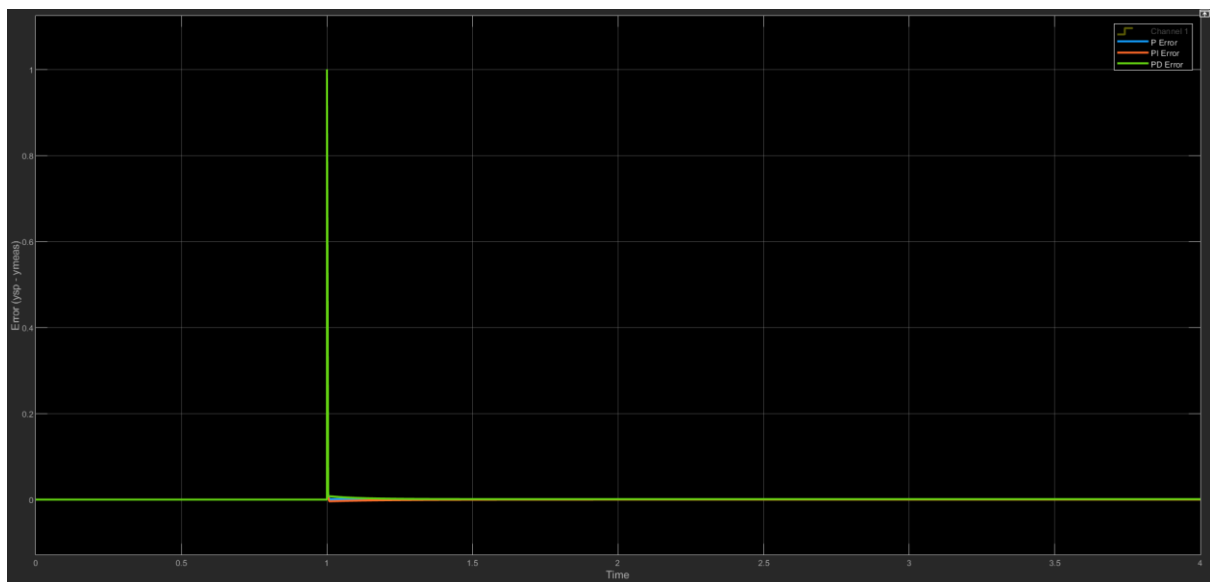
K=250



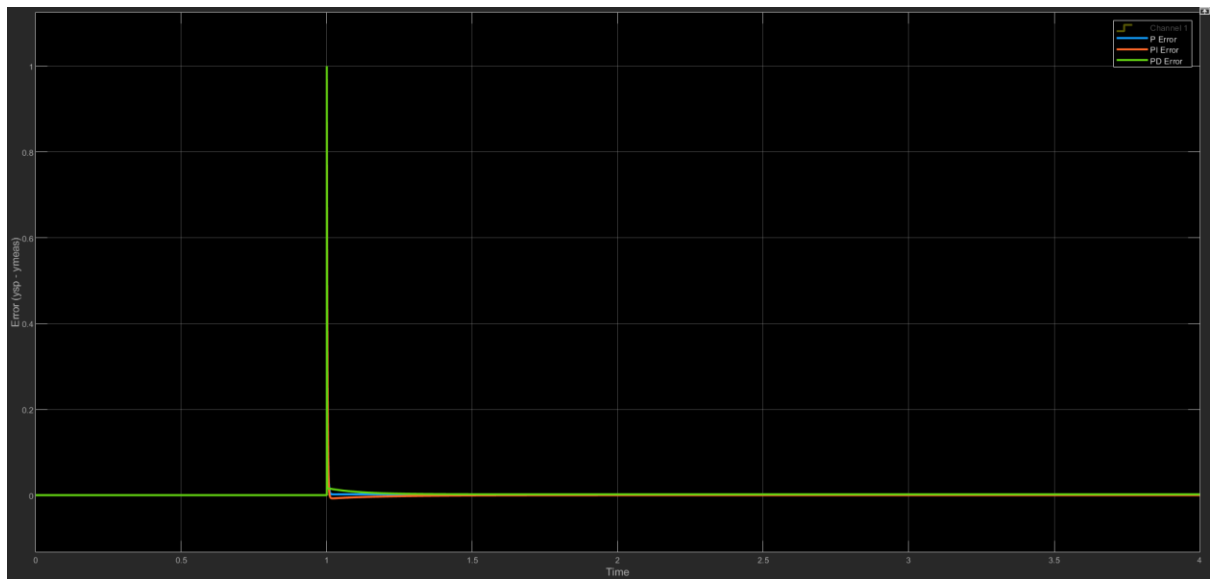
K=500



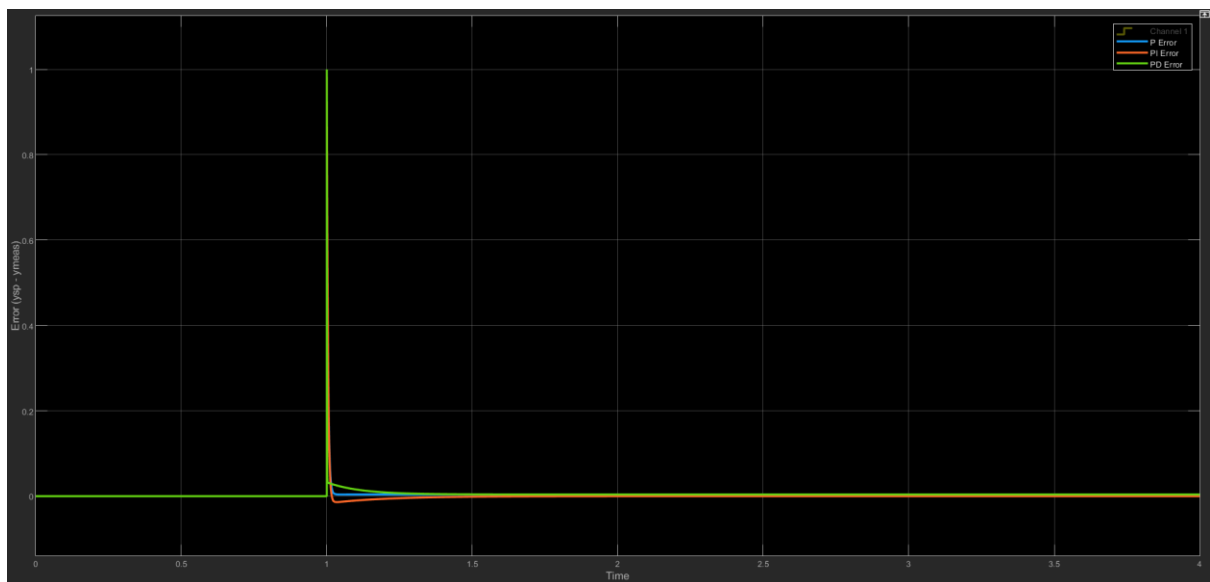
K=1000



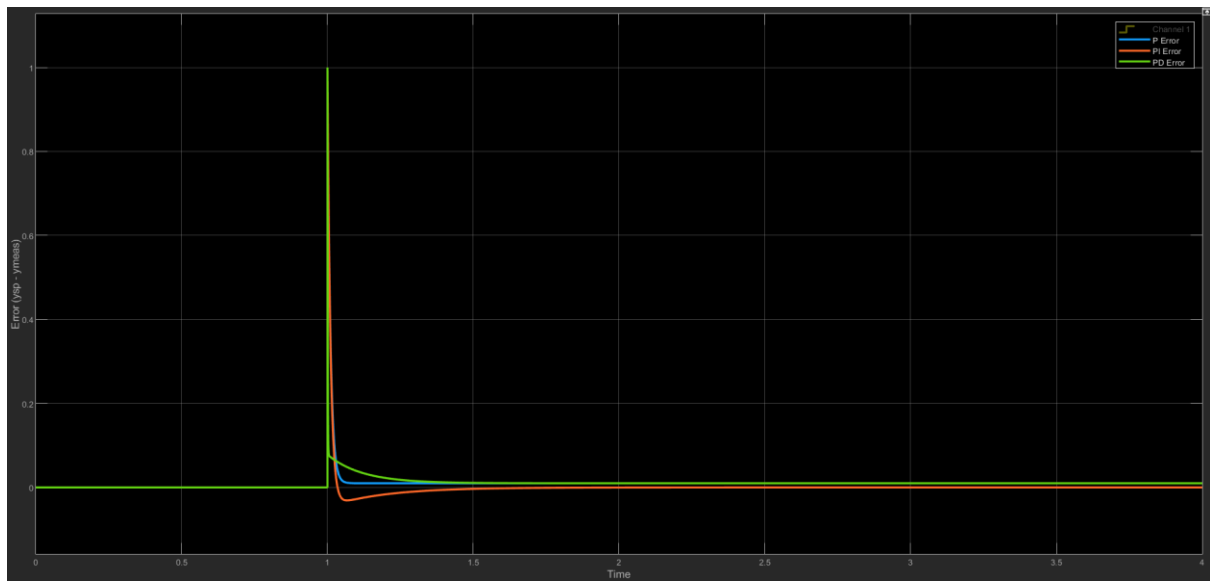
K=1000



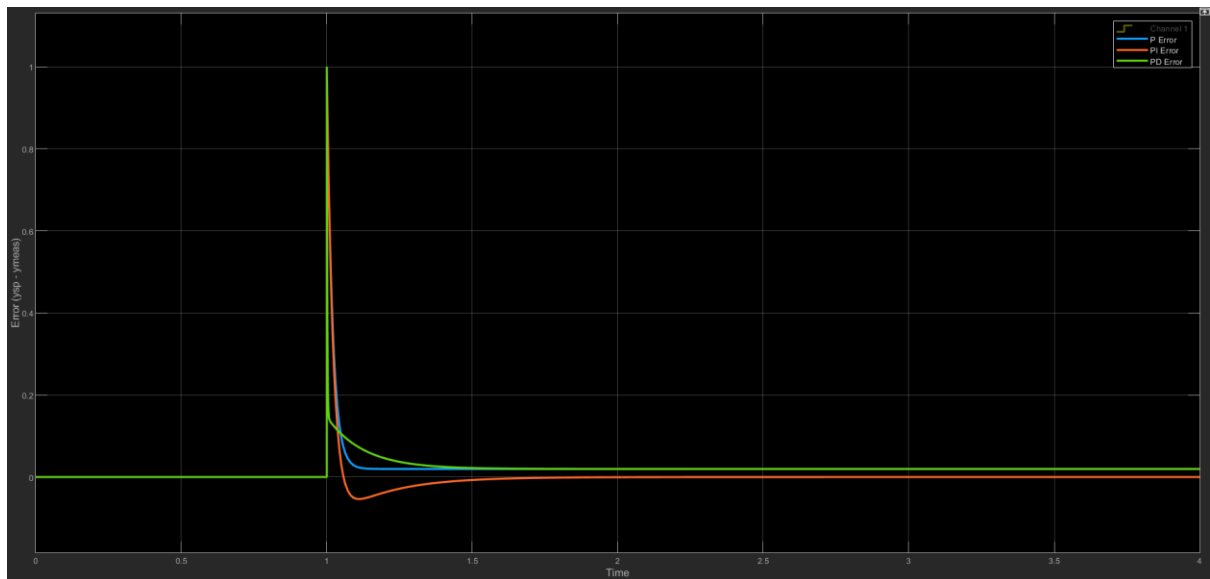
K=500



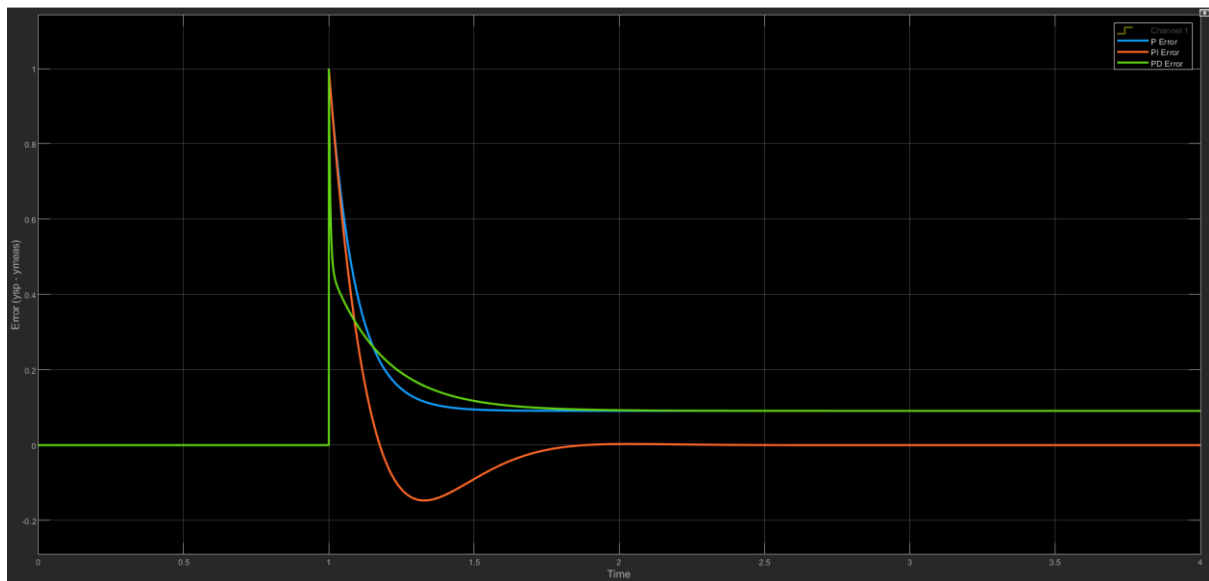
K=250



K=100

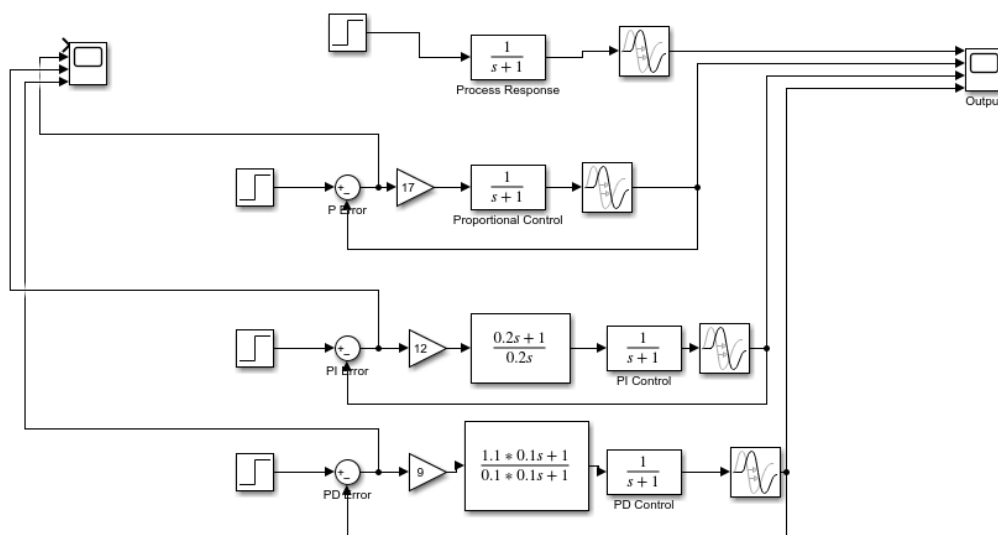


K=50



K=10

b) P, PI, PD controller together from top to bottom respectively

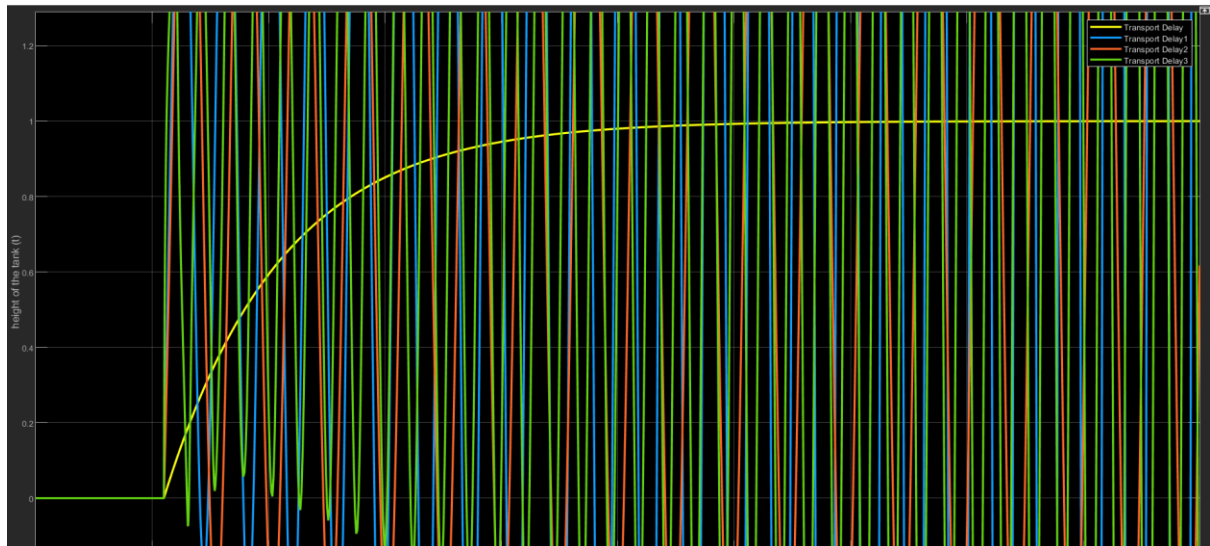


For P controller the value of  $K_c \geq 16$  gives unstable output

For PI controller, the value of  $K_c \geq 12$  gives unstable output

For PD controller, the value of  $K_c \geq 9$  gives unstable output



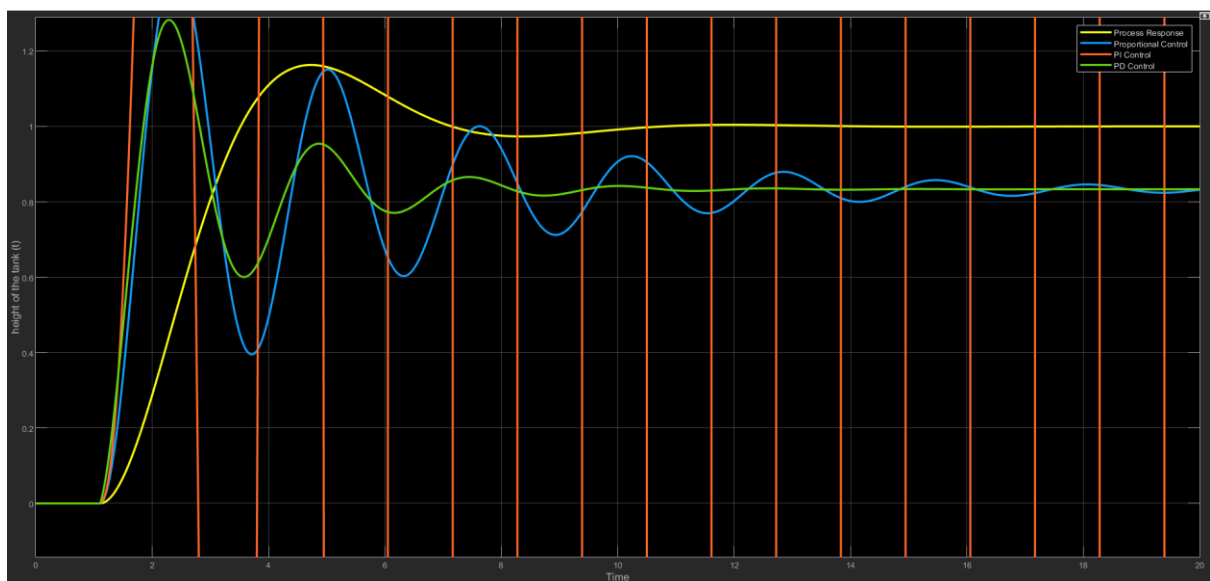


All are unstable outputs

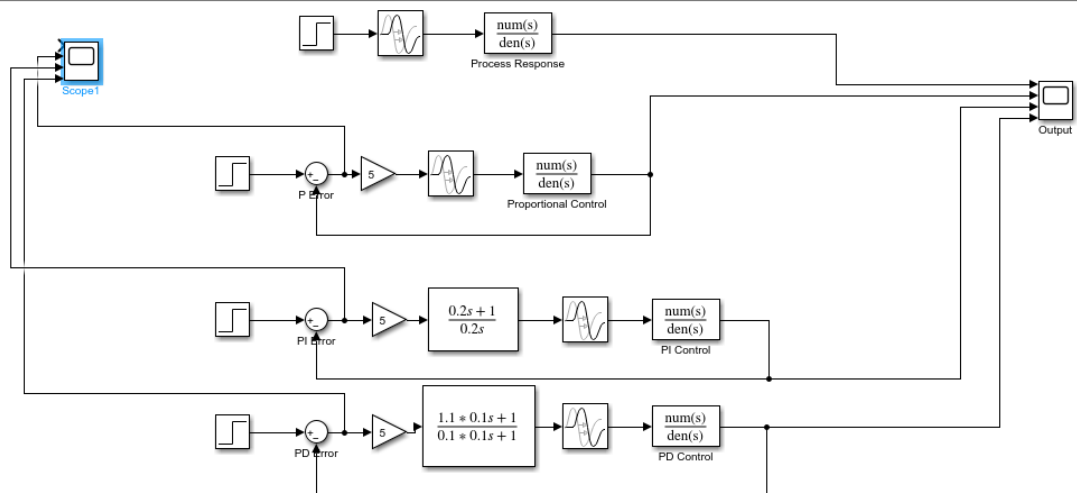
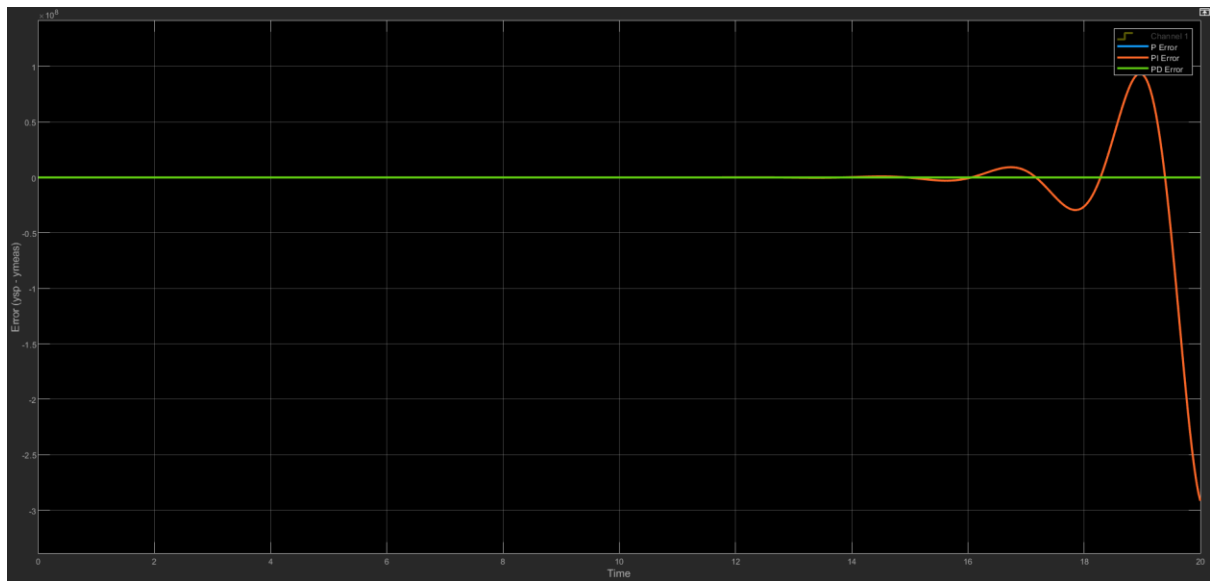
c)

Case 1 – Underdamped ( $1/s^2 + s + 1$ )

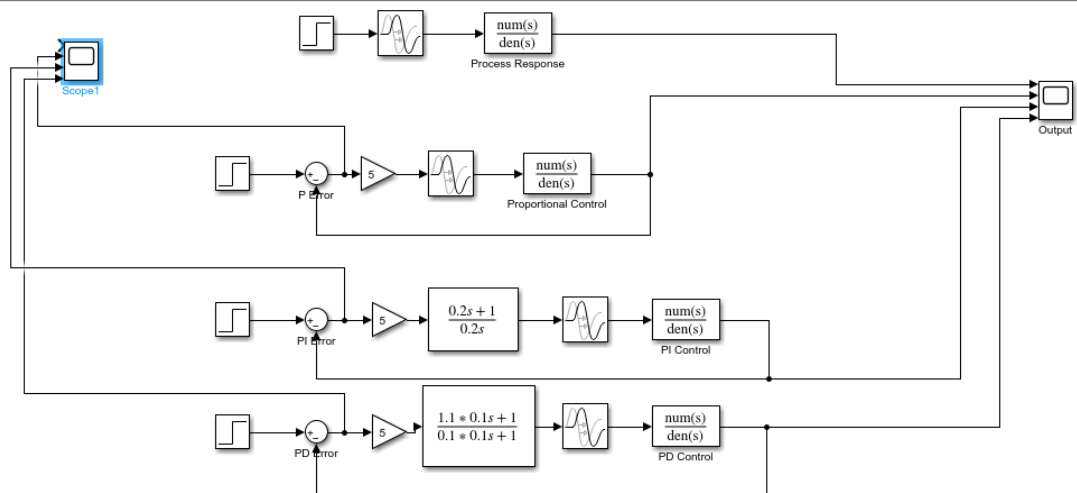
Output



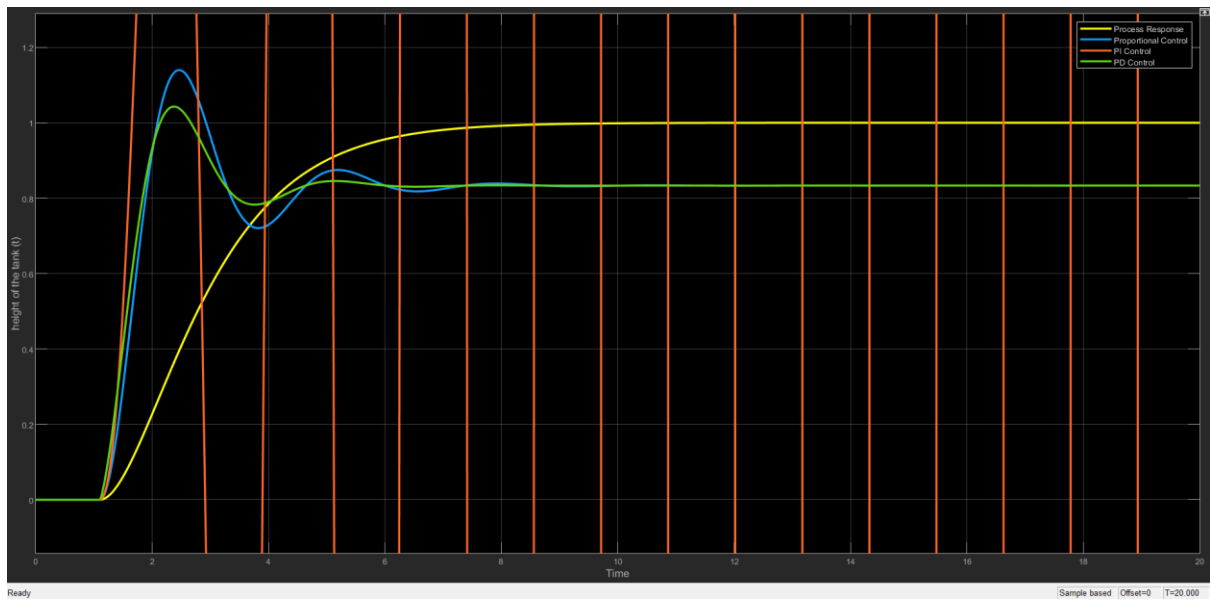
Error



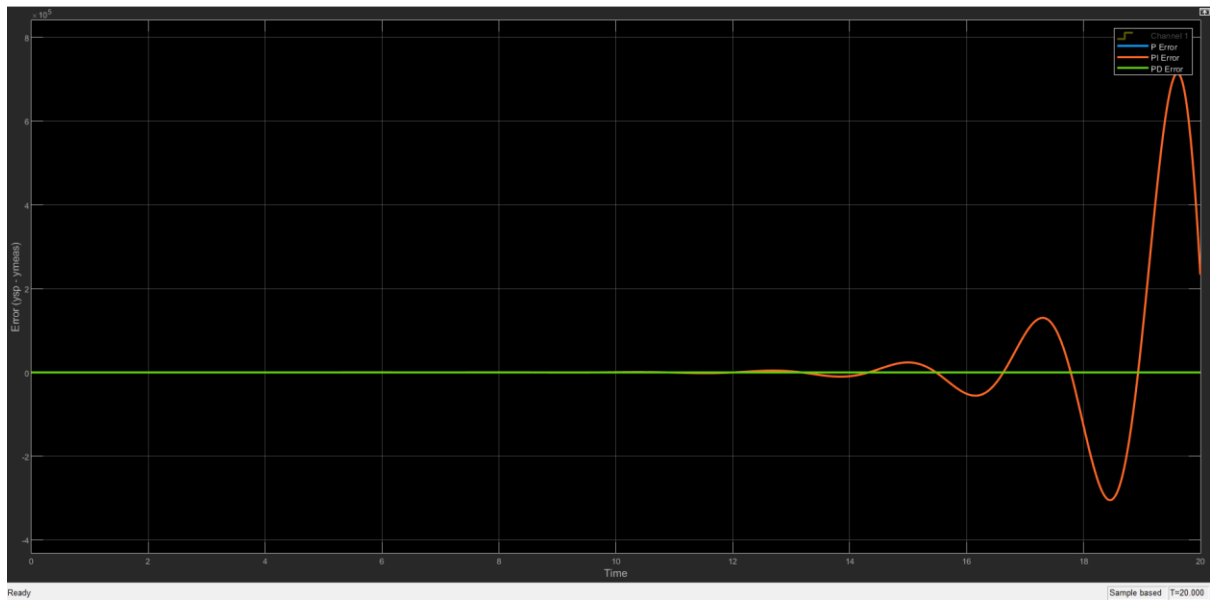
Case 2 : Critically damped ( $1/(s+1)^2$ )



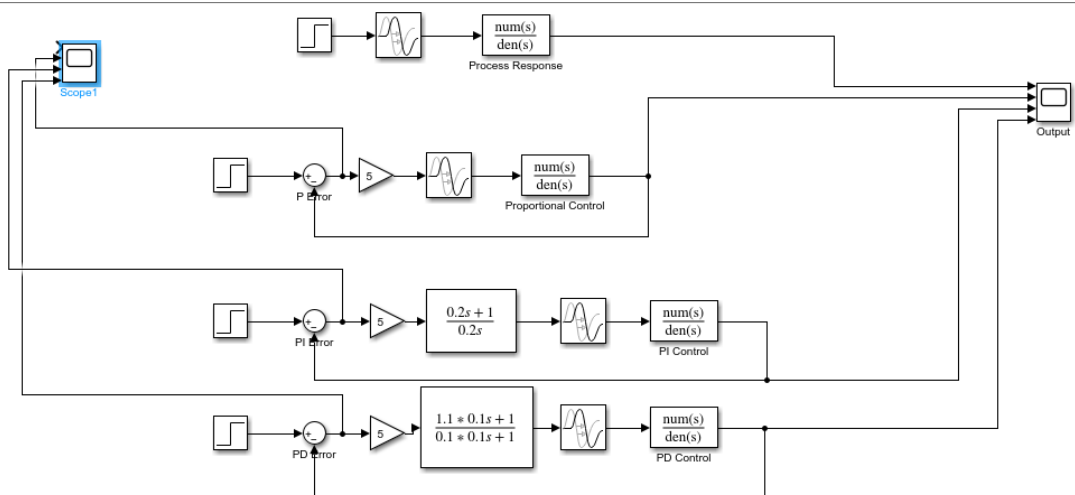
Output



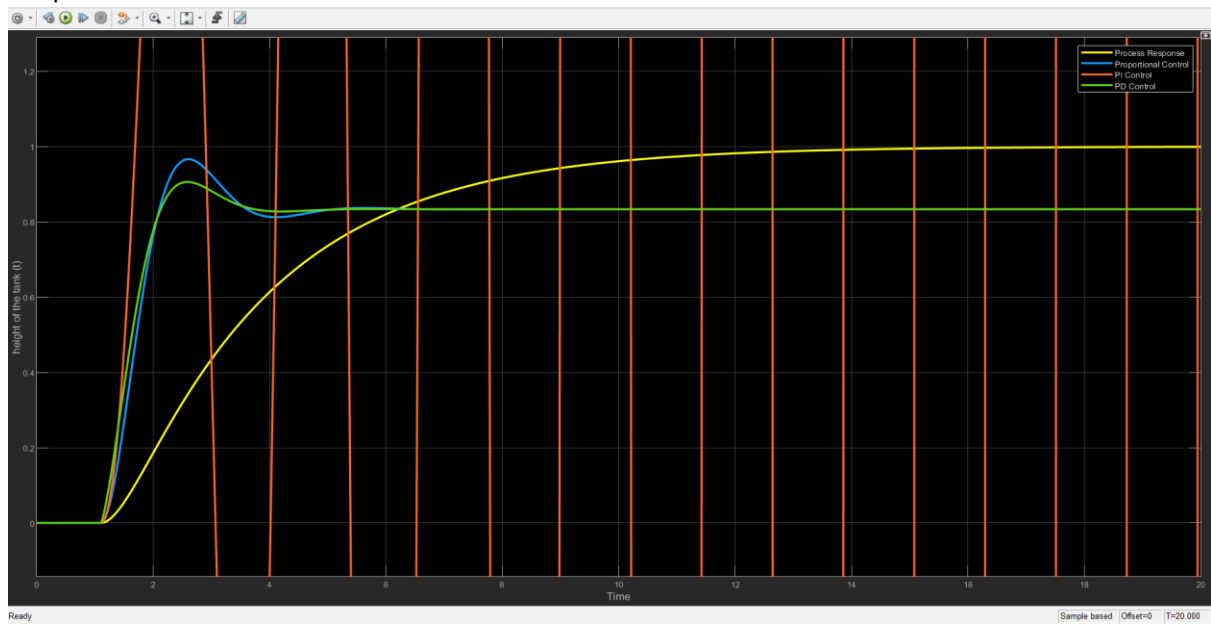
## Error



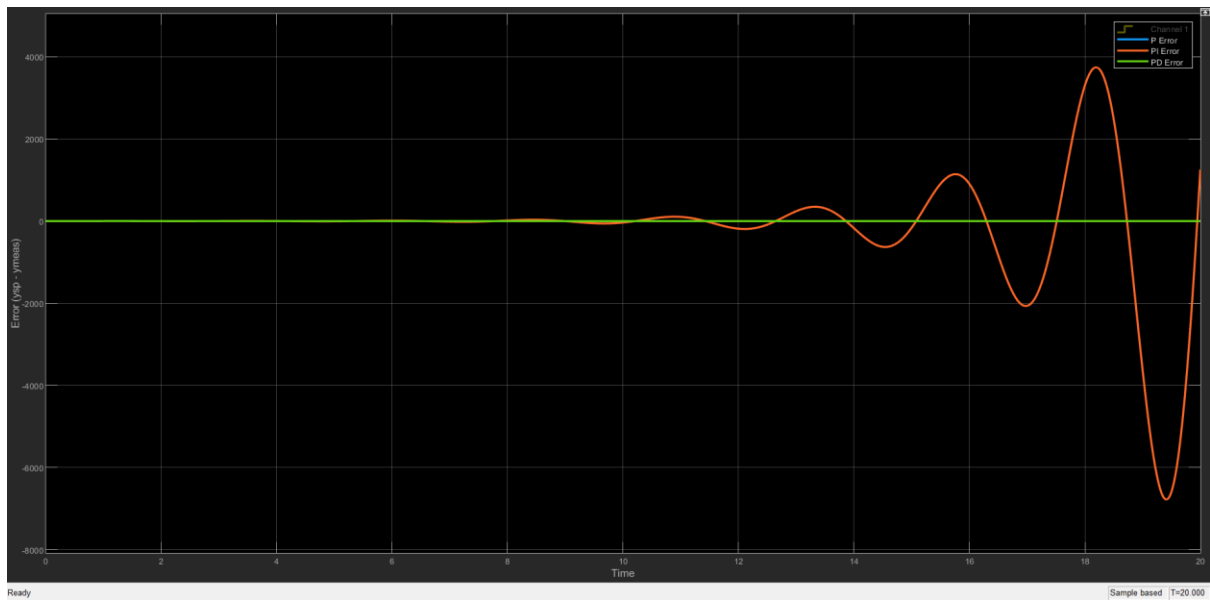
Case 3 : Over Damped ( $1/(s+1)(s+2)$ )



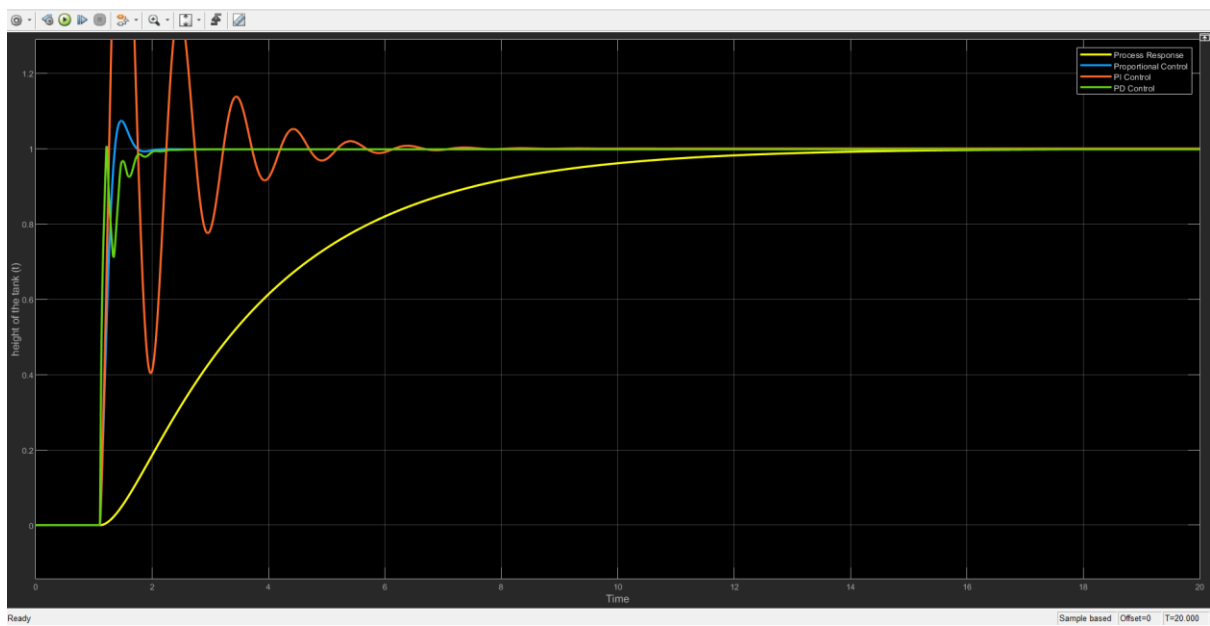
## Output



## Error



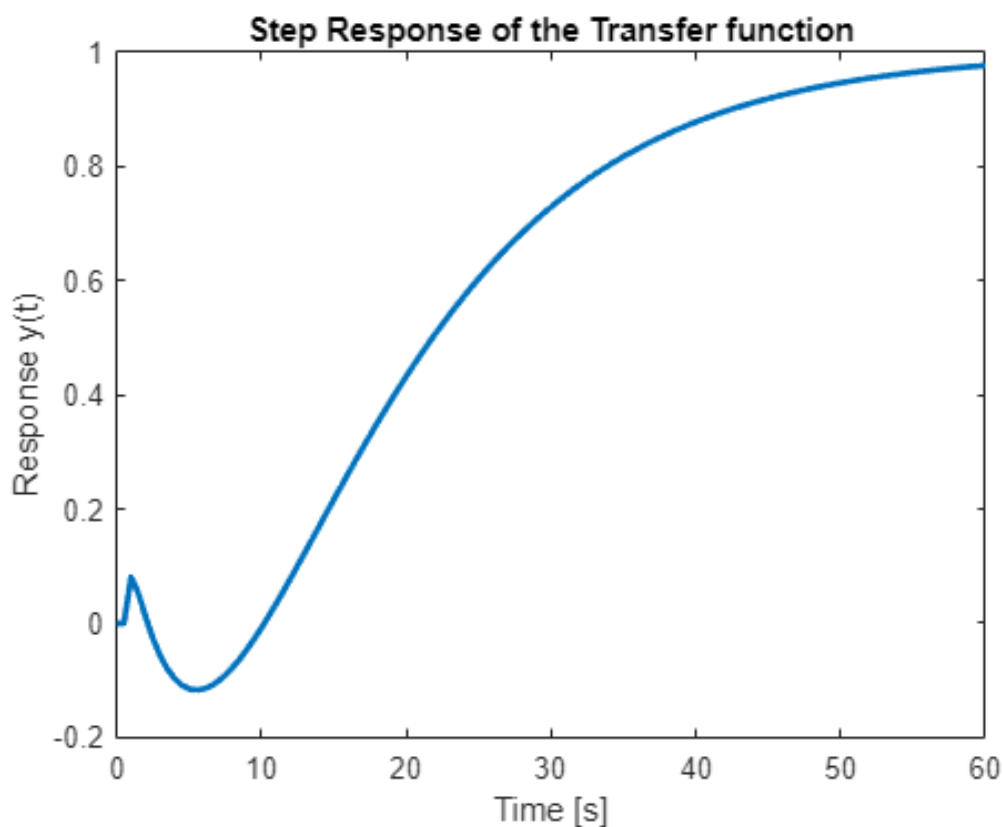
f) Yes, For underdamped systems, the  $K_c = 500$  is also stable for the transfer function of  $1/(s^2+100s+1)$  but for  $1/s^2+10s+1$ , the response became unstable



```

% defining variables
syms s
G = ((1-5*s)*(1-2*s)*exp(-0.5*s))/((12*s+1)*(6*s+1)*(0.3*s+1)); % the transfer function
U = 1/s; % step input
y=ilaplace(G*U); % response of step input
dy=diff(y);
t=0:0.5:60;
t=t'; % for easier copying
y=subs(y);
doubley= double(y); % made for copying
plot(t,y,'linewidth',2)
title("Step Response of the Transfer function")
ylabel("Response y(t)")
xlabel("Time [s]")

```

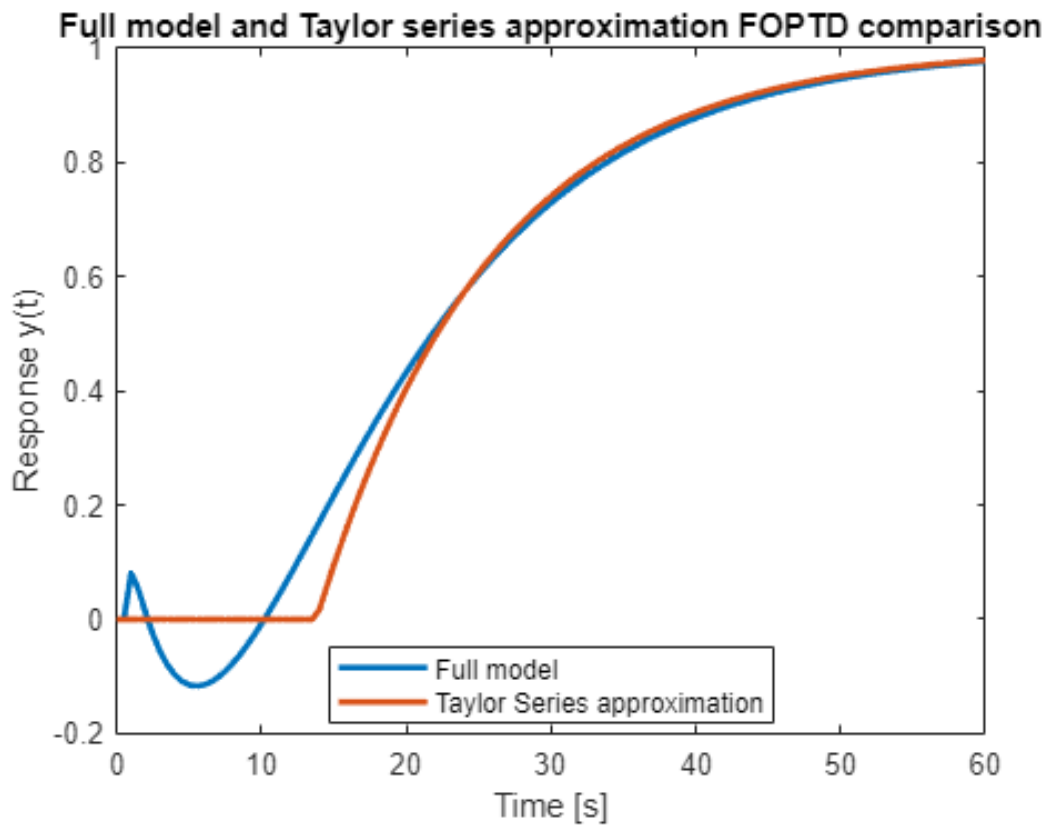


```

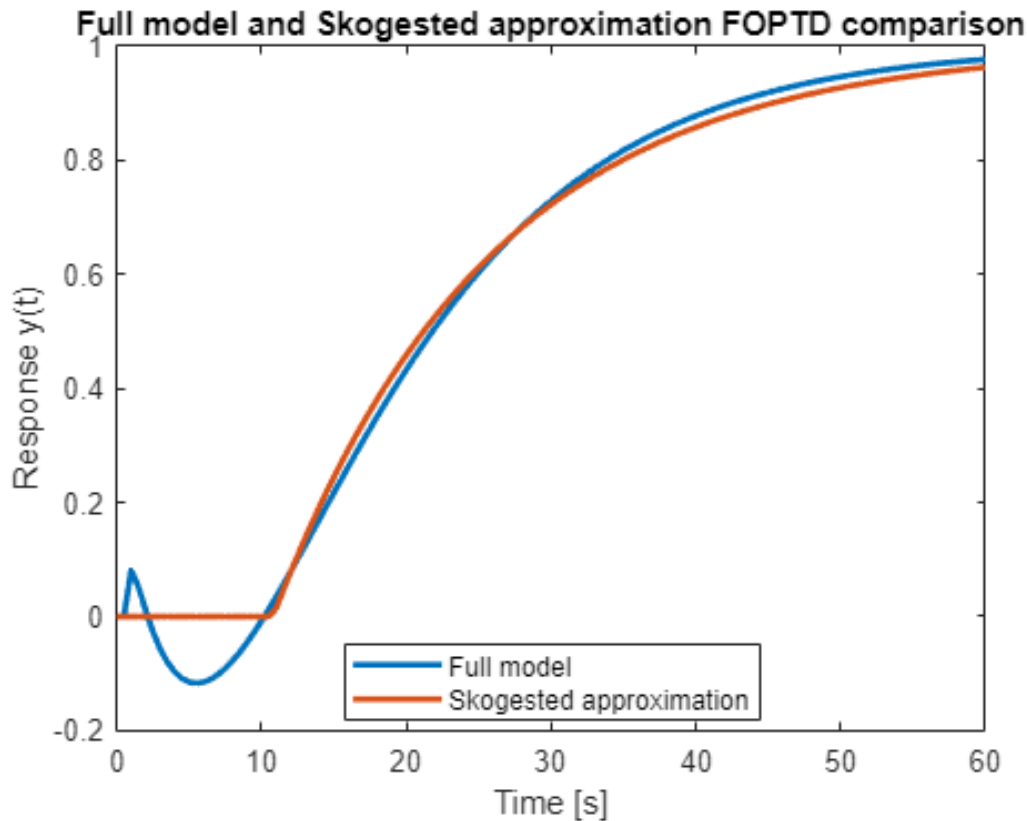
%b
G_b = exp(-13.8*s)/(12*s+1); % obtained by solving on paper
y_b=ilaplace(G_b*U);
y_b=subs(y_b);
plot(t,y,'linewidth',2)
hold on
plot(t,y_b,'linewidth',2)
hold off
ylabel("Response y(t)")
xlabel("Time [s]")
title("Full model and Taylor series approximation FOPTD comparison")

```

```
legend("Full model","Taylor Series approximation", Location="best")
```



```
%c
G_c = exp(-10.8*s)/(15*s+1); % obtained by solving on paper
y_c=ilaplace(G_c*U);
y_c=subs(y_c);
plot(t,y,'linewidth',2)
hold on
plot(t,y_c,'linewidth',2)
hold off
ylabel("Response y(t)")
xlabel("Time [s]")
title("Full model and Skogested approximation FOPTD comparison")
legend("Full model","Skogested approximation", Location="best")
```



```
%d
t_d=14;
m = subs(dy,t_d);
```

m =

$$\frac{595 e^{-\frac{9}{8}}}{2106} - \frac{220 e^{-\frac{9}{4}}}{513} + \frac{12190 e^{-45}}{20007}$$

```
p=double(y(find(t==t_d)))
```

p = 0.1705

```
theta = (t_d*m-p)/m
```

theta =

$$-\frac{\frac{3080 e^{-\frac{9}{4}}}{513} - \frac{4165 e^{-\frac{9}{8}}}{1053} - \frac{170660 e^{-45}}{20007} + \frac{6143957457824883}{36028797018963968}}{\frac{595 e^{-\frac{9}{8}}}{2106} - \frac{220 e^{-\frac{9}{4}}}{513} + \frac{12190 e^{-45}}{20007}}$$

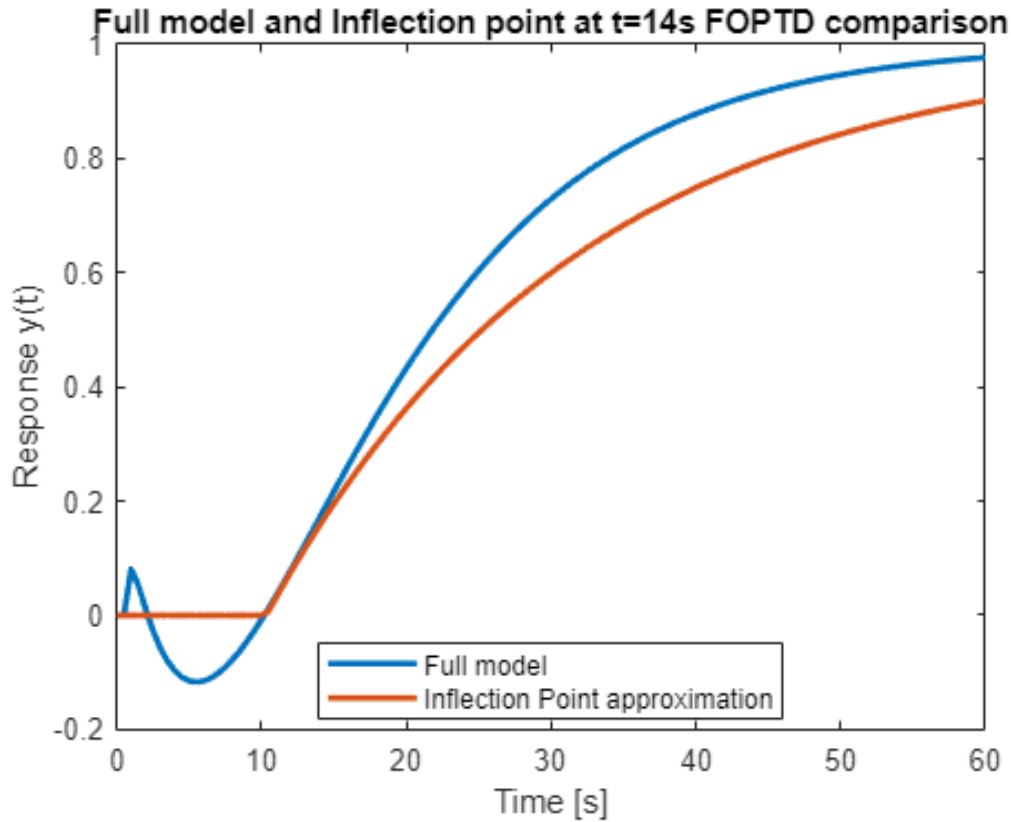
```
tau_d=1/m;
G_d = exp(-theta*s)/(tau_d*s+1);
y_d=ilaplace(G_d*U);
y_d=subs(y_d);
plot(t,y,'linewidth',2)
```



```

hold on
plot(t,y_d,'linewidth',2)
hold off
ylabel("Response y(t)")
xlabel("Time [s]")
title("Full model and Inflection point at t=14s FOPTD comparison")
legend("Full model","Inflection Point approximation", Location="best")

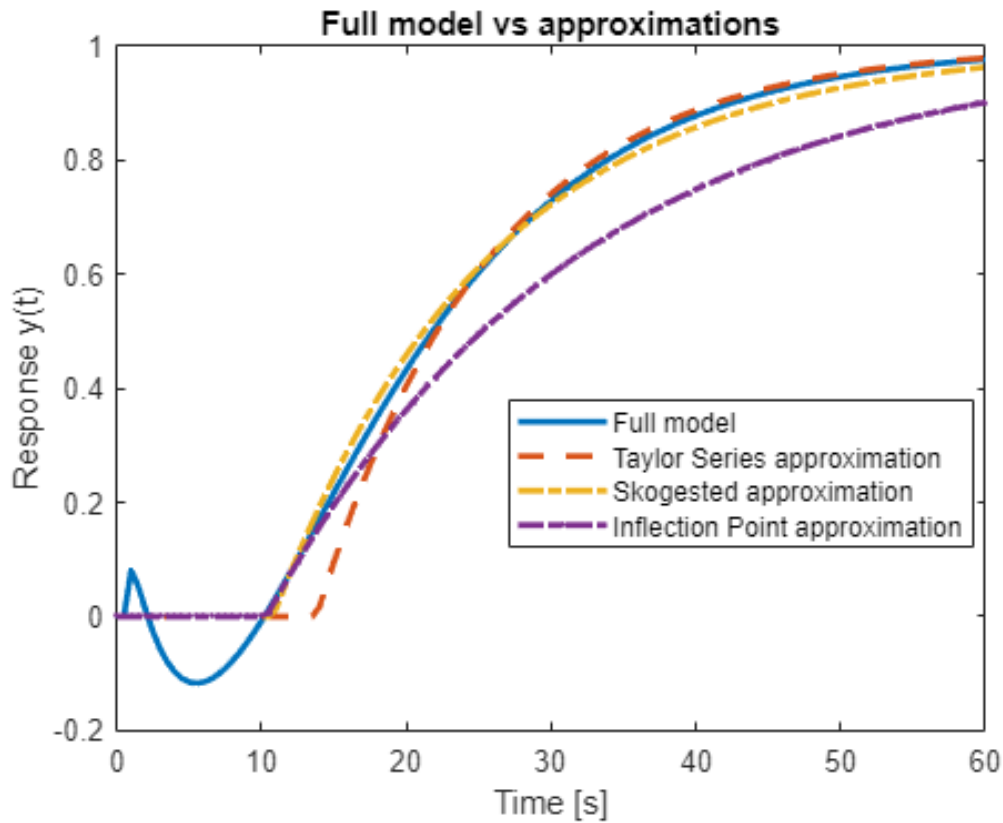
```



```

plot(t,y,'linewidth',2)
hold on
plot(t,y_b,'--','linewidth',2)
plot(t,y_c,'-.','linewidth',2)
plot(t,y_d,'-..','linewidth',2)
hold off
title("Full model vs approximations")
ylabel("Response y(t)")
xlabel("Time [s]")
legend("Full model","Taylor Series approximation","Skogested approximation","Inflection Point a

```



%e in excel

% f

```
G_2t = ((1-5*s)*exp(-2.8*s))/((12*s+1)*(6*s+1));
```

```
G_2s = ((1-5*s)*exp(-2.65*s))/((12*s+1)*(6.15*s+1));
```

```
y_2t=ilaplace(G_2t*U);
```

```
y_2t=subs(y_2t);
```

```
y_2s=ilaplace(G_2s*U);
```

```
y_2s=subs(y_2s);
```

```
y2s=double(y_2s);
```

```
plot(t,y,'linewidth',2)
```

```
hold on
```

```
plot(t,y_2t,'--','linewidth',2)
```

```
plot(t,y_2s,'-.','linewidth',2)
```

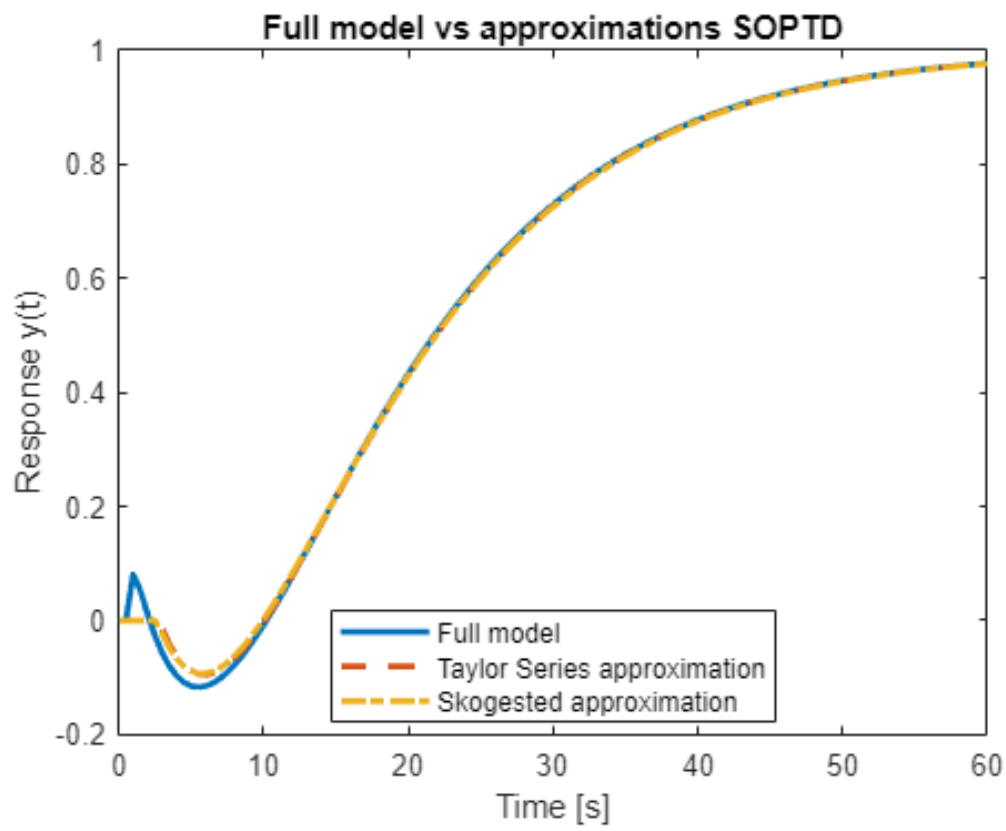
```
hold off
```

```
title("Full model vs approximations SOPTD")
```

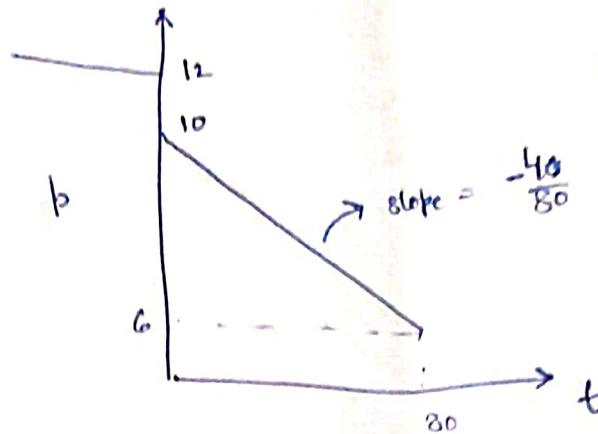
```
ylabel("Response y(t)")
```

```
xlabel("Time [s]")
```

```
legend("Full model","Taylor Series approximation","Skogested approximation", Location="best")
```



# 801



Given step change = 3mA

$$E(s) = 3/s$$

for PI,

$$\frac{P(s)}{E(s)} = K_c \left( 1 + \frac{1}{T_I s} \right)$$

$$E(s) = M/s$$

$$P(s) = K_c \frac{M}{s} + \frac{K_c M}{T_I s^2}$$

$$p'(t) = K_c M + \frac{K_c M}{T_I} t$$

$$\text{At } t=0, p'(t) = -2$$

$$M = 3$$

$$K_c = -2/3$$

$$\text{slope} = -1/20 = \frac{K_c M}{T_I}$$

$$T_I = 40 \text{ min}$$

Q # 8.2

11

(a)

$$\frac{P'(s)}{E(s)} = \frac{k_1}{T_I s + 1} + k_2$$

$$= \frac{(k_1 + k_2) + s k_2 T_I}{T_I s + 1}$$

$$(k_1 + k_2) \left[ \frac{1 + \frac{s k_2 T_I}{k_1 + k_2}}{T_I s + 1} \right]$$

b) for PD,  $\frac{P'(s)}{E(s)} = K_c \left[ \frac{1 + T_D s}{\alpha T_D s + 1} \right]$   
by comparing

$$K_c = k_1 + k_2$$

$$\alpha T_D = T_I$$

$$T_D = \frac{k_2}{k_1 + k_2} T_I$$

$$T_D = \frac{k_2 T_I}{k_1 + k_2}$$

$$\alpha = 1$$

$$\Rightarrow \alpha k_2 = k_1 + k_2$$

$$k_2 = \frac{K_c}{\alpha}$$

$$\& k_1 = K_c \left[ 1 - \frac{1}{\alpha} \right]$$

$$T_D = \frac{T_I}{\alpha}$$

8.83

9

a) Eq. 8.14  $\frac{P'(s)}{E(s)} = K_c^+ \left[ 1 + \frac{1}{T_I^+} + Z_D^+ s \right]$  parallel form

Eq. 8.15

$$\frac{P'(s)}{E(s)} = K_c \left( \frac{Z_I s + 1}{Z_I s} \right) \left( \frac{Z_D s + 1}{1 + s Z_D} \right)$$

Expanding #8.15

$$K_c \left[ 1 + \frac{Z_D}{Z_I} + \frac{1}{T_I s} + Z_D s \right]$$

Comparing  $s^0$  coeff

$$K_c^+ = K_c \left[ 1 + \frac{Z_D}{Z_I} \right]$$

Comparing  $s^{-1}$  coeff

$$\frac{K_c^+}{T_I^+} = \frac{K_c}{T_I}$$

$$T_I^+ = T_I \left[ 1 + \frac{Z_D}{Z_I} \right]$$

$$K_c^+ Z_D^+ = K_c Z_D$$

$$Z_D^+ = \frac{Z_D}{1 + \frac{Z_D}{Z_I}}$$

(c) b) Since  $z_D > 0$  &  $z_I > 0$

$$\left(1 + \frac{z_D}{z_I}\right) \geq 1$$

$$K_c^+ \geq K_c, \quad z_I^+ \geq z_I, \quad z_D^+ \leq z_D$$

c)  $K_c = 4$        $z_I = 10 \text{ min}$

$$z_D = 2 \text{ min}$$

$$K_c^+ = 1 + \frac{2}{10} = 1.2$$

$$K_c^+ = 4.8$$

$$z_I^+ = 10 \times 1.2 = 12 \text{ min}$$

$$z_D^+ = \frac{2}{1.2} = 1.67 \text{ min}$$

✓  
(c)



Q87 ~~amp~~ output = 12 mA

$$y_{sp} = T_0$$

$$y_m = T_0$$

$$K_c = 2$$

$$T_I = 1.5 \text{ min}$$

$$T_D = 0.5 \text{ min}$$

$$e(t) = 0.5t$$

$$E(t) = \frac{0.5}{s^2}$$

PIB controller,  $p(t)$  is given by

$$p(t) = \underbrace{\bar{p}(t)}_{12 \text{ mA}} + K_c \left[ e(t) + \frac{1}{T_I} \int_0^t e(t') dt + T_D \frac{de(t)}{dt} \right]$$

$$p(t) = 12 \text{ mA} + K_c \left[ 0.5t + \frac{1}{1.5} \cdot \frac{0.5t^2}{2} + 0.5 \times 0.5 \right]$$

$$p(t) = 12 \text{ mA} + t + \frac{t^2}{2} + 0.5 \quad \text{for } t > 0$$

$$p(t) = 12 \text{ mA} \quad \text{for } t < 0$$

(b) for PI,  $T_D = 0$

$$\therefore p(t) = 12 + t + \frac{t^2}{3}$$

(c) plot done separately.

Q1

Q2



Q 8.6  $Y_m(s) = \frac{2}{s}$

$$y_{sp} = 0$$

$$y_m(t) = 2$$

$$e(t) = y_{sp} - y_m = -2$$

$$= \ominus$$

for PI controller,

$$p'(t) = K_c \left[ e(t) + \frac{1}{\tau_i} \int_0^t e(t) dt \right]$$

$$= K_c \left[ -2 + \frac{1}{\tau_i} \int_0^t [-2] dt \right]$$

$$= -2K_c - \frac{2t K_c}{\tau_i}$$

at  $t=0$ ,

$$p'(t) = 6 = -2K_c$$

$$K_c = -3$$

$$\text{slope} = 1.2 = -\frac{2K_c}{\tau_i}$$

$$\tau_i = \frac{-2 \times -3}{1.2} = 5 \text{ min}$$

