

for $N=4$,

$$\begin{aligned} s_1 &= e^{j\left(\frac{\pi}{2} + \frac{\pi}{8}\right)} = e^{j\left(\frac{5\pi}{8}\right)} = -0.3827 + j0.9239 \\ s_2 &= e^{j\frac{7\pi}{8}} = -0.9239 + j0.3827 \\ s_3 &= e^{j\frac{9\pi}{8}} = -0.9239 - j0.3827 \\ s_4 &= e^{j\frac{11\pi}{8}} = -0.3827 - j0.9239 \end{aligned}$$

Now, the denominator polynomial of $H(s)$ is

$$\left\{ (s+0.3827)^2 + (0.9239)^2 \right\} \left\{ (s+0.9239)^2 + (0.3827)^2 \right\} = 0$$

$$(s^2 + 1.84776s + 1)(s^2 + 0.76536s + 1) = 0$$

\therefore for 4^{th} order Butterworth filter with $\omega_c = 1 \text{ rad/sec}$,

$$H(s) = \frac{1}{(s^2 + 0.76536s + 1)(s^2 + 1.84776s + 1)}$$

All the poles obtained from equations ④, ⑤ &

⑥ are normalized poles (ie. for $\omega_c = 1 \text{ rad/sec}$)

The unnormalized poles are given by

$$s_k' = \omega_c \cdot s_k$$

\therefore The transfer function of un-normalized Butterworth filter is obtained by substituting

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$s \rightarrow s/r_c$ in the transfer function
of normalized Butterworth filter.

List of Butterworth Polynomials:

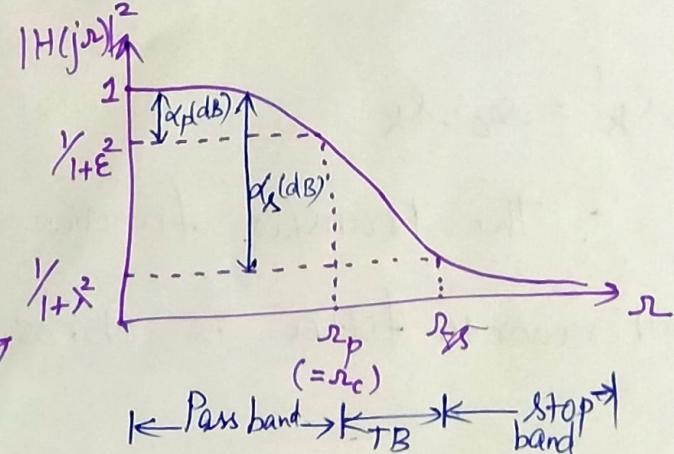
N	denominator of $H(s)$
1	$s+1$
2	$s^2 + \sqrt{2}s + 1$
3	$(s+1)(s^2+s+1)$
4	$(s^2+0.76537s+1)(s^2+1.8477s+1)$
5	$(s+1)(s^2+0.61803s+1)(s^2+1.61803s+1)$

Order of Butterworth filter (N):

Let us consider approximate magnitude response of Butterworth filter which has maximum passband attenuation (α_p) at passband frequency (r_p) and minimum stop band attenuation (α_s) at stopband frequency (r_s) as shown in fig.

shown in

fig.



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17.

$$|H(j\omega)|^2 = \frac{1}{1 + \varepsilon^2 \left(\frac{\omega}{\omega_p}\right)^{2N}} \longrightarrow ①a$$

(or)

$$|H(j\omega)| = \frac{1}{\sqrt{1 + \varepsilon^2 \left(\frac{\omega}{\omega_p}\right)^{2N}}} \longrightarrow ①b$$

Take logarithm on both sides of eqn ①a ,

$$20 \log |H(j\omega)| = \underbrace{10 \log 1}_{(=0)} - 10 \log \left[1 + \varepsilon^2 \left(\frac{\omega}{\omega_p}\right)^{2N} \right]$$

$$20 \log |H(j\omega)| = -10 \log \left[1 + \varepsilon^2 \left(\frac{\omega}{\omega_p}\right)^{2N} \right] \longrightarrow ②$$

$\alpha_p \rightarrow \max_{\text{passband}} \text{attenuation}$ at $\omega = \omega_p$

To find α_p , sub $\omega = \omega_p$ in equation ②

$$20 \log |H(j\omega_p)| = -10 \log (1 + \varepsilon^2) = -\alpha_p \text{ (dB)}$$

$$\boxed{\alpha_p = 10 \log (1 + \varepsilon^2)} \longrightarrow ③$$

$$0.1 \alpha_p = \log (1 + \varepsilon^2)$$

Take antilog on both sides

$$10^{0.1 \alpha_p} = 1 + \varepsilon^2$$

$$\boxed{\varepsilon = \sqrt{10^{0.1 \alpha_p} - 1}} \longrightarrow ④$$

$\alpha_s \rightarrow$ minimum stop band attenuation at $\omega = \omega_s$,
to find α_s , sub $\omega = \omega_s$ in eqn ②,

$$20 \log |H(j\omega_s)| = \underbrace{10 \log 1}_{=0} - 10 \log \left[1 + \varepsilon^2 \left(\frac{\omega_s}{\omega_p} \right)^{2N} \right] = -\alpha_s$$

$$\alpha_s = 10 \log \left[1 + \varepsilon^2 \left(\frac{\omega_s}{\omega_p} \right)^{2N} \right] \rightarrow ⑤$$

$$0.1 \alpha_s = \log \left[1 + \varepsilon^2 \left(\frac{\omega_s}{\omega_p} \right)^{2N} \right]$$

Take antilog,

$$10^{0.1 \alpha_s} = 1 + \varepsilon^2 \left(\frac{\omega_s}{\omega_p} \right)^{2N}$$

$$\varepsilon^2 \left(\frac{\omega_s}{\omega_p} \right)^{2N} = 10^{\alpha_s(0.1)} - 1 \rightarrow ⑥$$

Sub ε^2 in eqn ⑥,

$$10^{0.1 \alpha_p} - 1 \left(\frac{\omega_s}{\omega_p} \right)^{2N} = 10^{0.1 \alpha_s} - 1$$

$$\left(\frac{\omega_s}{\omega_p} \right)^{2N} = \sqrt{\frac{10^{0.1 \alpha_s} - 1}{10^{0.1 \alpha_p} - 1}} \rightarrow ⑥a$$

Take log on both sides,

$$\log \left(\frac{\omega_s}{\omega_p} \right)^{2N} = \log \left\{ \sqrt{\frac{10^{0.1 \alpha_s} - 1}{10^{0.1 \alpha_p} - 1}} \right\}$$

$$N = \frac{\log \sqrt{\frac{10^{0.1 \alpha_s} - 1}{10^{0.1 \alpha_p} - 1}}}{\log (\omega_s / \omega_p)} \rightarrow ⑦$$

$N \rightarrow$ order of the filter (integer)

equation ⑦ result will not be an integer value.

so round off N to next higher integer.

$$\text{i.e. } N \geq \frac{\log \sqrt{\frac{10^{0.1\alpha_s} - 1}{10^{0.1\alpha_p} - 1}}}{\log (\omega_s/\omega_p)}$$

$$\boxed{N \geq \frac{\log (\lambda/\varepsilon)}{\log (\omega_s/\omega_p)}} \rightarrow ⑧$$

$$\text{where } \lambda = \sqrt{10^{0.1\alpha_s} - 1}$$

$$\varepsilon = \sqrt{10^{0.1\alpha_p} - 1}$$

let $\frac{\omega_p}{\omega_s} = k$ (transition ratio), & $A = \lambda/\varepsilon$

$$\boxed{N \geq \frac{\log A}{\log (\gamma_k)}} \rightarrow ⑨$$

Cut-off frequency (ω_c):

magnitude squared response of ideal ^{BWooth} filter,

$$|H(j\omega)|^2 = \frac{1}{1 + \left(\frac{\omega}{\omega_c}\right)^{2N}} \rightarrow ⑩$$

Magnitude squared response of practical filter,

$$|H(j\omega)|^2 = \frac{1}{1 + \epsilon^2 \left(\frac{\omega}{\omega_p}\right)^{2N}} \rightarrow ②$$

Compare ① & ②,

$$1 + \left(\frac{\omega}{\omega_c}\right)^{2N} = 1 + \epsilon^2 \left(\frac{\omega}{\omega_p}\right)^{2N}$$

$$\epsilon^2 = \left(\frac{\omega_p}{\omega_c}\right)^{2N} = 10^{0.1\alpha_p - 1}$$

$$\frac{\omega_p}{\omega_c} = [10^{0.1\alpha_p - 1}]^{\frac{1}{2N}}$$

$$\boxed{\omega_c = \frac{\omega_p}{[10^{0.1\alpha_p - 1}]^{\frac{1}{2N}}}} \rightarrow ③$$

$$\omega_c = \frac{\omega_p}{(\epsilon^2)^{\frac{1}{2N}}} = \boxed{\frac{\omega_p}{\epsilon^{\frac{1}{N}}} = \omega_c} \rightarrow ④$$

Refer eqn ⑥a in page no 18.

$$\left(\frac{\omega_s}{\omega_p}\right)^{2N} = \frac{10^{0.1\alpha_s} - 1}{10^{0.1\alpha_p} - 1}$$

$$\omega_s = (\omega_p) \left[\frac{10^{0.1\alpha_s} - 1}{10^{0.1\alpha_p} - 1} \right]^{\frac{1}{2N}}$$

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$$21.$$

$$R_s = R_c \left[\frac{10^{0.1\alpha_p} - 1}{10^{0.1\alpha_s} - 1} \right]^{\frac{1}{2N}} \cdot \frac{\left[10^{0.1\alpha_s} - 1 \right]^{\frac{1}{2N}}}{\left[10^{0.1\alpha_p} - 1 \right]^{\frac{1}{2N}}}$$

$$\boxed{R_c = \frac{R_s}{\left[10^{0.1\alpha_s} - 1 \right]^{\frac{1}{2N}}}} \rightarrow ⑤$$

wkt $X^2 = 10^{0.1\alpha_s} - 1$

$$R_c = \frac{R_s}{(X^2)^{\frac{1}{2N}}} = \boxed{\frac{R_s}{X^{\frac{1}{N}}} = R_c} \rightarrow ⑥$$

$$\boxed{R_c = \frac{R_p}{\left(10^{0.1\alpha_p} - 1 \right)^{\frac{1}{2N}}} = \frac{R_s}{\left[10^{0.1\alpha_s} - 1 \right]^{\frac{1}{2N}}}} \rightarrow ⑦$$

At $\alpha_p = 3\text{dB}$, $\boxed{R_c = R_p}$

Ques Given the specification $\alpha_p = 1\text{dB}$, $\alpha_s = 30\text{dB}$,

$R_p = 200 \text{ rad/sec}$, $R_s = 600 \text{ rad/sec}$, find 'N' (order of the filter)

$$N \geq \frac{\log A}{\log(\frac{1}{k})}$$

$$A = \frac{\lambda}{\varepsilon} = \sqrt{\frac{10^{0.1\alpha_s} - 1}{10^{0.1\alpha_p} - 1}} = \left[\frac{10^3 - 1}{10^1 - 1} \right]^{0.5}$$

$$A = 62.115$$

$$k = \frac{\omega_p}{\omega_s} = \frac{200}{600} = \frac{1}{3}$$

$$N \geq \frac{\log(62.115)}{\log(3)}$$

$N \geq 3.758$ (round off to next higher integer)

$$\boxed{N=4}$$

Pbm Determine the order and poles of Low Pass

Butterworth filter that has a $-3dB$ Bandwidth or attenuation at 500 Hz and attenuation of 40dB at 1000 Hz.

$$\text{Given: } \alpha_p = 3dB, \quad \alpha_s = 40dB$$

$$f_p = 500 \text{ Hz} \quad f_s = 1000 \text{ Hz}$$

$$\omega_p = 2\pi f_p = 1000\pi \text{ rad/sec}, \quad \omega_s = 2\pi f_s = 2000\pi \text{ rad/sec}$$

$$N \geq \frac{\log A}{\log(\gamma_k)}$$

$$N \geq 6.6$$

$$\boxed{N=7}$$

The poles of Butterworth filter are given by,

$$s_k' = \omega_c \cdot e^{j\phi_k}$$

$$\text{where } \phi_k = \frac{\pi}{2} + \frac{(2k-1)\pi}{2N}, \quad k=1,2,\dots,7$$

23.

$$s_k' = 1000\pi \cdot e^{j\phi_k} = 1000\pi \cdot s_k$$

$$k=1, e^{j(\frac{\pi}{2} + \frac{\pi}{14})} = e^{j(\frac{9\pi}{14})} = -0.22 + j0.974 = s_1$$

$$k=2, e^{j\frac{10\pi}{14}} = -0.623 + j0.781 = s_2$$

$$k=3, e^{j\frac{12\pi}{14}} = -0.9009 + j0.433 = s_3$$

$$k=4, e^{j\pi} = -1 = s_4$$

$$k=5, e^{j\frac{14\pi}{14}} = -0.901 - j0.433 = s_5$$

$$k=6, e^{j\frac{16\pi}{14}} = -0.623 - j0.781 = s_6$$

$$k=7, e^{j\frac{18\pi}{14}} = -0.22 - j0.974 = s_7$$

Design of an analog Low Pass Butterworth filter:

Procedure:

Step 1: From the given specifications, find the order of the filter (N). \leftarrow integer value

Step 2: Find the transfer function $H(s)$ for $\omega_c = 1 \text{ rad/sec}$ for the value of N .

Step 3: Calculate the value of cut off frequency (ω_c)

Step 4: Find the transfer function $H_a(s)$ for the above value of ω_c by substituting $s \rightarrow \frac{s}{\omega_c}$ in $H(s)$.

Pbm Design an analog Butterworth filter that has a -2dB passband attenuation at a frequency of 20 rad/sec and atleast -10dB stopband attenuation at 30 rad/sec.

$$\text{Given: } \alpha_p = 2\text{dB} , \quad \alpha_s = 10\text{dB}$$

$$\omega_p = 20\text{rad/sec}, \quad \omega_s = 30\text{rad/sec}.$$

Solution:

Step 1: $N = ?$

$$N \geq \frac{\log \sqrt{\frac{10^{0.1\alpha_s}}{10^{0.1\alpha_p}} - 1}}{\log (\omega_s/\omega_p)}$$

$$N \geq 3.37$$

$$\boxed{N = 4}$$

Step 2: $H(s) = ?$ for $\omega_c = 1\text{rad/sec}$

from table,

Using denominator of $H(s)$ for $N=4$,

$$H(s) = \frac{1}{(s^2 + 0.76537s + 1)(s^2 + 1.8477s + 1)}$$

Step 3: $\omega_c = ?$

$$\omega_c = \frac{\omega_p}{(10^{0.1\alpha_p} - 1)^{1/2N}} = \frac{20}{(10^{0.2} - 1)^{1/8}} = 21.3868$$

Step 4:

$$H_a(s) = H(s)/s = s/s_{rc}$$

$$H_a(s) = \frac{1}{\left\{ \left(\frac{s}{21.3868} \right)^2 + 0.76537 \left(\frac{s}{21.3868} \right) + 1 \right\} \left\{ \left(\frac{s}{21.3868} \right)^2 + 1.8477 \left(\frac{s}{21.3868} \right) + 1 \right\}}$$

$$0.20921 \times 10^6$$

$$H_a(s) = \frac{0.20921 \times 10^6}{(s^2 + 16.3686s + 457.394)(s^2 + 39.5176s + 457.394)}$$

Ques For the given specifications design an analog Butterworth filter.

$$0.9 \leq |H(j\omega)| \leq 1, \quad 0 \leq \omega \leq \omega_p$$

$$|H(j\omega)| \leq 0.2, \quad \omega_s \leq \omega \leq \pi$$

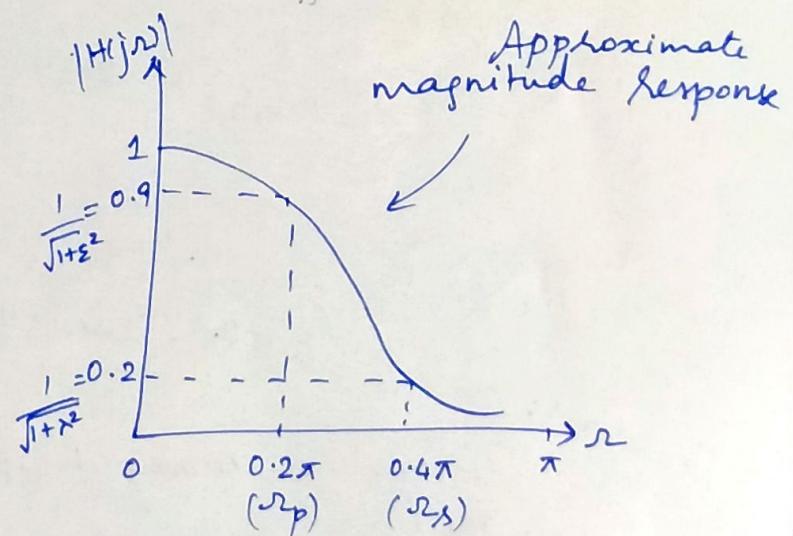
Given:

$$\omega_p = 0.2\pi$$

$$\omega_s = 0.4\pi$$

$$\frac{1}{\sqrt{1+\varepsilon^2}} = 0.9$$

$$\frac{1}{\sqrt{1+\lambda^2}} = 0.2$$



$$\varepsilon = 0.484, \quad \lambda = 4.898$$

Step 1: $N = ?$

$$N \geq \frac{\log(\lambda/\varepsilon)}{\log(r_s/r_p)}$$

$$N \geq 3.34 \quad \boxed{N=4}$$

Step 2: $H(s) = \frac{1}{(s^2 + 0.76537s + 1)(s^2 + 1.8477s + 1)}$
 $H(\lambda) = ?$
 $\text{at } r_c = \text{rad/sec}$

Step 3: $r_c = ?$

$$\begin{aligned} r_c &= \frac{r_p}{(10^{0.1\alpha_p} - 1)^{1/2N}} = \frac{r_p}{\varepsilon^{1/N}} \\ &= \frac{0.2\pi}{(0.484)^{1/4}} = 0.24\pi \end{aligned}$$

Step 4: $H_a(s) = ?$

$$H_a(s) = H(s)/s = s/0.24\pi$$

$$H_a(s) = \frac{1}{\left\{ \left(\frac{s}{0.24\pi} \right)^2 + 0.76537 \left(\frac{s}{0.24\pi} \right) + 1 \right\} \left\{ \left(\frac{s}{0.24\pi} \right)^2 + 1.8477 \left(\frac{s}{0.24\pi} \right) + 1 \right\}}$$

$$H_a(s) = \frac{0.323}{(s^2 + 0.577s + 0.0576\pi^2)(s^2 + 1.393s + 0.0576\pi^2)}$$

Homework: Design an analog Butterworth LPF for the following specifications.

$$\alpha_p = 0.5 \text{ dB}, \alpha_s = 22 \text{ dB}, f_p = 10 \text{ kHz}, f_s = 25 \text{ kHz}$$