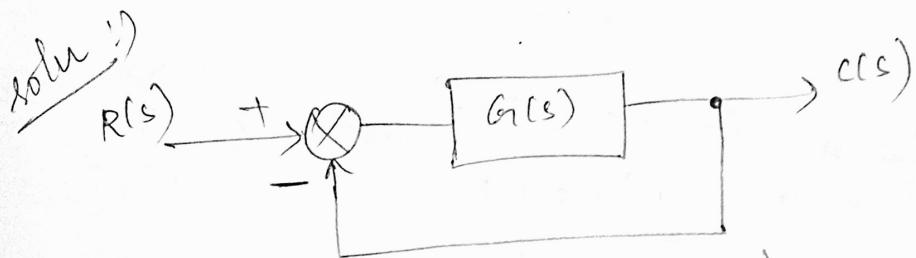


Eg) A unity feedback control system is characterized by the following open loop transfer function $G(s) = \frac{(0.4s+1)}{s(s+0.6)}$

Determine its transient response for unit step i/p & sketch the response. Evaluate the maximum overshoot and corresponding peak time.



$$\frac{C(s)}{R(s)} = \frac{G(s)}{1+G(s)} = \frac{\frac{(0.4s+1)}{s(s+0.6)}}{1 + \frac{(0.4s+1)}{s(s+0.6)}}$$

$$= \frac{0.4s+1}{s(s+0.6) + 0.4s+1}$$

$$\boxed{\frac{C(s)}{R(s)} = \frac{0.4s+1}{s^2+s+1}}$$

$$\rightarrow q(t) = u(t) \Rightarrow R(s) = 1/s$$

$$C(s) = R(s) \left(\frac{0.4s+1}{s^2+s+1} \right) = \frac{1}{s} \left(\frac{0.4s+1}{s^2+s+1} \right)$$

$$\frac{0.4s+1}{s(s^2+s+1)} = \frac{A}{s} + \frac{Bs+C}{s^2+s+1}$$

$$A = C(s) \times s \Big|_{s=0} \Rightarrow \boxed{A=1}$$

$$\rightarrow 0.4s + 1 = A(s^2 + s + 1) + (Bs + C)s$$

$$0.4s + 1 = s^2 + s + 1 + Bs^2 + Cs$$

$$1 + 0.4s = (1+B)s^2 + (1+C)s + 1$$

$$\rightarrow \text{equate} \Rightarrow 1+B=0 \Rightarrow B=-1$$

$$1+C=0.4 \Rightarrow C=-0.6$$

$$\rightarrow C(s) = \frac{1}{s} + \frac{(-1)s + (-0.6)}{s^2 + s + 1}$$

$$= \frac{1}{s} - \frac{(s+0.6)}{(s^2+s+1)}$$

$s^2 + s + 1 \Rightarrow$ to have a perfect square

$$s^2 + s + 1 + 0.25 - 0.25$$

$$(s^2 + s + 0.25) + 0.75$$

$$(s+0.5)^2 + 0.75$$

$$\rightarrow C(s) = \frac{1}{s} - \frac{(s+0.5)}{(s+0.5)^2 + 0.75} - \frac{0.1}{(s+0.5)^2 + 0.75}$$

$$= \frac{1}{s} - \frac{s+0.5}{(s+0.5)^2 + 0.75} - \frac{0.1 \times \sqrt{0.75}}{\sqrt{0.75} \left[(s+0.5)^2 + 0.75 \right]}$$

$$= \frac{1}{s} - \frac{s+0.5}{(s+0.5)^2 + 0.75} - \frac{0.1}{\sqrt{0.75}} \frac{\sqrt{0.75}}{(s+0.5)^2 + 0.75}$$

$$c(t) = L^{-1}[c(s)]$$

$$= 1 - e^{-0.5t} \cos(\sqrt{0.75}t) - 0.115 e^{-0.5t} \sin(\sqrt{0.75}t)$$

$$c(t) = 1 - e^{-0.5t} \left[\cos(\sqrt{0.75}t) + 0.115 \sin(\sqrt{0.75}t) \right]$$

→ Transient response is the response that vanishes
(is equal to zero) when $t \rightarrow \infty$.

In the above eqn, the 2nd term, as $t \rightarrow \infty$, becomes

Zero,
 $\therefore c(t) \Big|_{\text{transient}} = e^{-0.5t} \left[\cos(\sqrt{0.75}t) + 0.115 \sin(\sqrt{0.75}t) \right]$

~~Note:~~
 $c(t) \Rightarrow$ Total Response

$$\text{total response} = \text{Transient + Steady state Response}$$

$$c(t) = c_t(t) + c_{ss}(t)$$

$$\boxed{c(t) \Big|_{t \rightarrow \infty} = c_{ss}(t)}$$

since $\boxed{c_t(t) \Big|_{t \rightarrow \infty} = 0}$

→ Maximum overshoot (M_p)

$$\therefore M_p = e^{-\frac{\zeta \pi}{\sqrt{1-\zeta^2}}} \times 100 \%$$

$$s^2 + s + 1 \Rightarrow s^2 + 2\zeta \omega_n s + \omega_n^2$$

$$\omega_n^2 = 1 \Rightarrow \omega_n = 1 \text{ rad/sec}$$

$$2\zeta \omega_n = 1$$

$$\zeta = \frac{1}{2} = 0.5 //$$

$$\therefore M_p = e^{-\frac{0.5 \pi}{\sqrt{1-0.5^2}}} \times 100 \%$$

$$= e^{-1.813} \times 100 \%$$

$$\boxed{M_p = 16.32 \%}$$

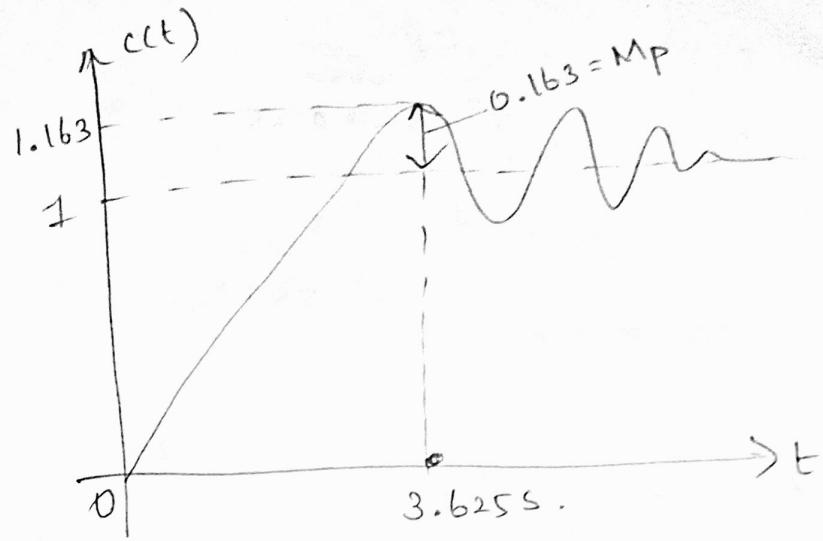
$$M_p = 0.163.$$

$$\rightarrow \text{peak time } t_p = \frac{\pi}{\omega_d} = \frac{\pi}{\omega_n \sqrt{1-\zeta^2}}$$

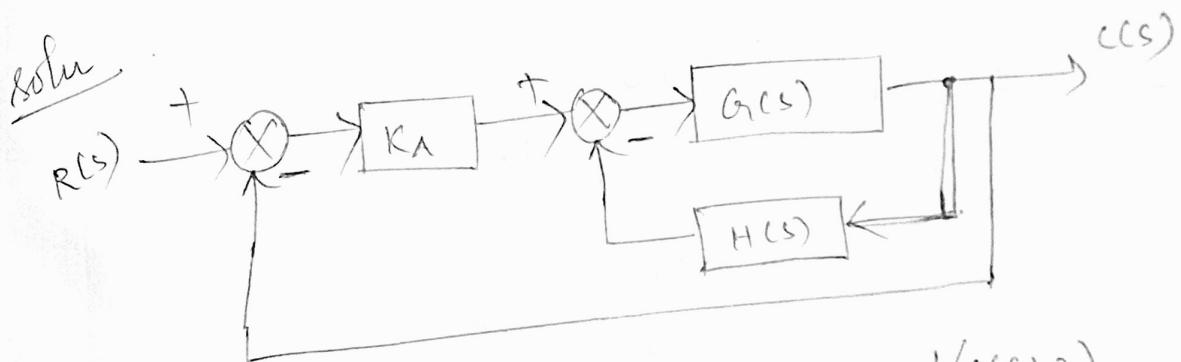
$$= \frac{\pi}{(1) \sqrt{1-0.25}}$$

$$= \pi / 0.866$$

$$\boxed{t_p = 3.625 \text{ sec}}$$



2) A unity feedback control s/m. has an amplifier with gain $K_A = 10$ f gain ratio $G(s) = 1/s(s+2)$ in the feed forward path. A derivative feedback, $H(s) = SK_0$ is introduced as a minor loop around $G(s)$. Determine the derivative feedback constant, (K_0) so that the s/m damping factor is 0.6.



$$\text{Inner fb loop} \Rightarrow \frac{G(s)}{1 + G(s)H(s)} = \frac{\frac{1}{s(s+2)}}{1 + \frac{1}{s(s+2)} \cdot SK_0}$$

$$= \frac{1}{s(s+2) + SK_0}$$

$$= \frac{1}{s^2 + 2s + SK_0}$$

$$\text{Multiplied with } K_A = \frac{K_A}{s^2 + 2s + SK_0}$$

$$\rightarrow \text{unity fb} \Rightarrow \frac{C(s)}{R(s)} = \frac{\frac{K_A}{s^2 + 2s + K_0 s}}{1 + \frac{K_A}{s^2 + 2s + K_0 s}}$$

$$\boxed{\frac{C(s)}{R(s)} = \frac{K_A}{s^2 + (2+K_0)s + K_A}}$$

$$\rightarrow K_A = 10$$

$$\frac{C(s)}{R(s)} = \frac{10}{s^2 + (2+K_0)s + 10}$$

$$\omega_n^2 = 10 \Rightarrow \omega_n = \sqrt{10} = \underline{3.162 \text{ rad/sec}}$$

$$\rightarrow 2c_0 \omega_n = (2+K_0)$$

$$2 \times 0.6 \times 3.162 = 2+K_0$$

$$K_0 = 3.7944 - 2$$

$$\boxed{K_0 = 1.7944}$$

Steady State error :-

- The steady state error is the value of error signal $[e(t)]$, which t tends to infinity.
- The errors arise from the nature of I/p's, type of S/m & from non-linearity of S/m components.
- The performance of the S/m is studied from the steady state error with step, ramp & parabolic I/p's.



$$\rightarrow E(s) = R(s) - c(s)H(s).$$

$$\text{if } c(s) = E(s)G(s)$$

$$\rightarrow \therefore E(s) = R(s) - E(s)G(s)H(s)$$

$$E(s) + E(s)G(s)H(s) = R(s)$$

$$\boxed{E(s) = \frac{R(s)}{1 + G(s)H(s)}}.$$

→ $e(t) \rightarrow$ error signal in time domain.

$$\therefore e(t) = L^{-1}[E(s)]$$

$$= L^{-1}\left[\frac{R(s)}{1 + G(s)H(s)}\right]$$

$\rightarrow e_{ss} \Rightarrow$ Steady state error is defined as the value of $e(t)$ as t tends to infinity.

$$e_{ss} = \lim_{t \rightarrow \infty} e(t)$$

\rightarrow By final value theorem of Laplace transform,

$$e_{ss} = \lim_{t \rightarrow \infty} e(t) \stackrel{LT}{\Leftrightarrow} s \lim_{s \rightarrow 0} E(s)$$

$$\rightarrow \text{X} \quad e_{ss} = \lim_{t \rightarrow \infty} e(t) = \lim_{s \rightarrow 0} s E(s) = \lim_{s \rightarrow 0} \frac{s R(s)}{1 + G(s)H(s)}$$

Static Error constants:-

\rightarrow The steady state error value depends on the type number and the i/p signal.

\rightarrow Type-0 \Rightarrow constant steady state error (e_{ss}) with step i/p

Type-1 \Rightarrow constant e_{ss} for ramp i/p or velocity signal

Type-2 \Rightarrow constant e_{ss} for parabolic signal or acceleration signal

\rightarrow errors:-

(i) positional error constant (K_p) $= \lim_{s \rightarrow 0} G(s)H(s)$

(ii) velocity error constant (K_v) $= \lim_{s \rightarrow 0} s G(s)H(s)$

$$(iii) \text{ Acceleration error constant } (K_a) = \lim_{s \rightarrow 0} s^2 G(s) H(s).$$

→ The errors K_p, K_v & K_a are called as static error constants.

(1) I/p :- unit Step signal

$$e_{ss} = \lim_{s \rightarrow 0} s E(s)$$

$$= \lim_{s \rightarrow 0} s \cdot \frac{R(s)}{1 + G(s)H(s)}$$

$$= \lim_{s \rightarrow 0} \frac{s_0 Y_s}{1 + G(s)H(s)}$$

$$= \frac{1}{1 + \lim_{s \rightarrow 0} G(s)H(s)}$$

$$e_{ss} = \frac{1}{1 + K_p}$$

case (i) : Type '0' s/m

$$K_p = \lim_{s \rightarrow 0} G(s)H(s) = \lim_{s \rightarrow 0} \frac{K (s+z_1)(s+z_2)\dots}{(s+p_1)(s+p_2)\dots}$$

$$K_p = \frac{K (z_1)(z_2)\dots}{(p_1)(p_2)\dots}$$

= constant.

$$\therefore \boxed{e_{ss} = \frac{1}{1 + K_p}} \Rightarrow \text{constant}$$

case(ii) Type - I S/m

$$K_p = \lim_{s \rightarrow 0} G(s) H(s)$$

$$= \lim_{s \rightarrow 0} K \frac{(s+z_1)(s+z_2) \dots}{s(s+p_1)(s+p_2) \dots}$$

$$K_p = \infty.$$

$$e_{ss} = \frac{1}{1+K_p} = \frac{1}{1+\infty} = 0$$

$$\boxed{\therefore e_{ss} = 0.}$$

→ ∴ The value of 'K_p' is infinity for type 'I'
for above S/m's with unit step as i/p.

2) Ramp i/p

$$e_{ss} = \lim_{s \rightarrow 0} \frac{s R(s)}{1 + G(s) H(s)}$$

$$r(t) = t \Rightarrow R(s) = \frac{1}{s^2}.$$

$$e_{ss} = \lim_{s \rightarrow 0} \frac{s \cdot \frac{1}{s^2}}{1 + G(s) H(s)}$$

$$= \lim_{s \rightarrow 0} \frac{1}{s [1 + G(s) H(s)]}$$

$$e_{ss} = \lim_{s \rightarrow 0} \frac{1}{s + G(s)H(s)}.$$

$$= \lim_{s \rightarrow 0} \frac{1}{s G(s)H(s)}$$

$$e_{ss} = \lim_{s \rightarrow 0} \frac{1}{K_V}$$

case (i) - Type 0 s/m.

$$K_V = \lim_{s \rightarrow 0} s G(s) H(s)$$

$$= \lim_{s \rightarrow 0} \frac{s K (s+z_1)(s+z_2)}{(s+p_1)(s+p_2)}$$

$$K_V = 0$$

$$\therefore e_{ss} = \lim_{s \rightarrow 0} \frac{1}{K_V} = \infty$$

case (ii) type-1 s/m.

$$K_V = \lim_{s \rightarrow 0} s G(s) H(s)$$

$$= \lim_{s \rightarrow 0} \frac{s K (s+z_1)(s+z_2)}{s (s+p_1)(s+p_2)}$$

K_V = constant value

$$e_{ss} = \frac{1}{K_V} \Rightarrow \text{constant}$$

case (iii) - Type - 2 s/m

$$K_V = \lim_{s \rightarrow 0} s G(s) H(s)$$

$$= \lim_{s \rightarrow 0} \frac{s \cdot K(s+z_1)(s+z_2)}{s^2(s+p_1)(s+p_2)}$$

$$= \infty.$$

$$\therefore e_{ss} = \frac{1}{K_V} = \frac{1}{\infty} = 0.$$

Above 'Type 2' - the e_{ss} will be zero
for ramp I/P.