

Lab 3.

Q1.  $M = \frac{20(s-1)}{(s+2)(s^2+4)}$

$\hookrightarrow$  1 LHP.  
 $\hookrightarrow$  2 imaginary poles  
 $= (-2)$  pole

$\therefore M(s)$  is marginally stable.

Q2.  $2s^4 + s^3 + 3s^2 + 5s + 10 = 0$ .

Using R-H criterion:

$s^4$	2	3	10
$s^3$	1	5	0
$s^2$	$\frac{-2}{1}$	10	0
$s$	$\frac{45}{7}$	0	0
$s^0$	10	0	0

$$-\left| \begin{array}{cc} 2 & 3 \\ 1 & 5 \end{array} \right| = -7 \quad -\left| \begin{array}{cc} 2 & 10 \\ 1 & 0 \end{array} \right| = 10$$

$$-\frac{1}{-7} \left| \begin{array}{cc} 1 & 5 \\ -7 & 10 \end{array} \right| = \frac{45}{7}$$

as there is  
a sign change  
 $\therefore$  System UNSTABLE

Q3.  $s^4 + s^3 + 2s^2 + 2s + 3 = 0$ .

$s^4$	1	2	3
$s^3$	1	2	0
$s^2$	0	3	0
$s^1$	$\frac{-3}{1}$	0	0
$s^0$	0	0	0

UNSTABLE.

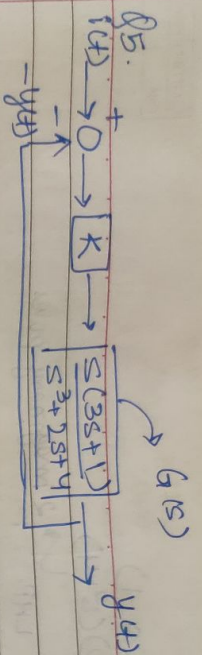
Q4.  $s^5 + 4s^4 + 8s^3 + 7s + 4 = 0$ .

$s^5$	1	8	7	0
$s^4$	4	8	4	0
$s^3$	6	6	0	0
$s^2$	4	4	0	0
$s^1$	0	0	0	0
$s^0$	4	0	0	0

No sign change

$\therefore$  System is

STABLE



K → gain of proportional controller

$$1 + K G(s) = 0$$

$$K(G(s) - (1/s) - y(t)) = y(t)$$

$$1 + \frac{s^3 + 2s + 4}{K s(3s + 1)} = 0$$

$$s^3 + 2s + 4 + K s(3s + 1) = 0$$

$$s^3 + (2 + 3K)s + 4 + Ks^2 = 0$$

$$K > 0$$

or table

$s^3$	1	$2 + K$	
$s^2$	$3K$	$4$	
$s^1$	$-\frac{4K}{3K^2 + 2K + 4}$	$0$	
$s^0$			$-\frac{1}{3K}(4 - 3K^2 - 6K)$

$$\text{as } K > 0$$

$$3K^2 + 6K - 4 > 0$$

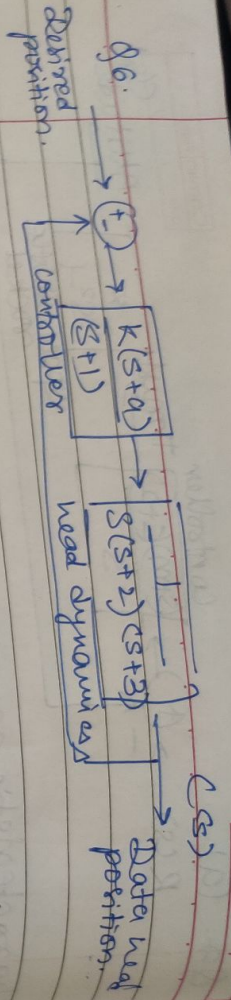
$$K > -1 + \sqrt{\frac{7}{3}} \quad \vee \quad 0.53$$

when  $K = -1 + \sqrt{\frac{7}{3}}$  both elements of 3<sup>rd</sup> row is 0.  
stable (marginally).

$s^3$	1	$2 + K$	
$s^2$	$3K$	$4$	
$s$	$6K$	$0$	$\therefore K > -1 + \sqrt{\frac{7}{3}}$
1	$4$	$0$	

system stable for





characteristic equation:

$$1 + K(s+a) = 0$$

$$s(s+1)(s+2)(s+3)$$

$$s^4 + 6s^3 + 11s^2 + (6+K)s + Ka = 0$$

as  $K > 0 \therefore a > 0$

$s^4$	1	11	$Ka$
$s^3$	6	$6+K$	0
$s^2$	$6$	$6+K$	0
$s^1$	$6Ka - (6+K)(6+K)$	$10-K/6$	$Ka$
$s^0$	$Ka$	0	0

$$\therefore 10 - K > 0 \therefore K < 10$$

$$0 < K \leq 6.0$$

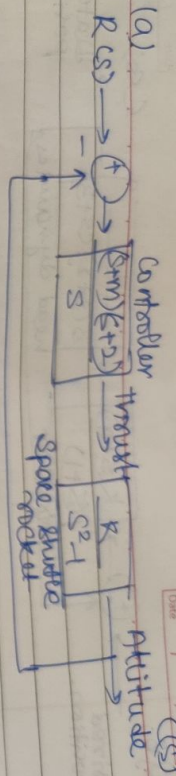
$$\text{at } K = 6.0 \quad a = 0$$

$$0 \leq a \leq \frac{10+3-K}{36}$$

system  $\rightarrow$  marginally stable

Stable region  $\therefore K \in [0, 6.0]$   
 $a \in [0, 10]$

Q7. (a)



characteristic equation:

$$1 + \frac{k(s+m)}{s(s^2-1)}(s+2) = 0$$

$$s^3 - s + ks^2 + k(m+2)s + 2km = 0$$

$$s^3 + ks^2 + (k(m+2)-1)s + 2km = 0$$

for stability:

$$k(m+2)-1 > 0 \implies m > \frac{1}{k} - 2$$

$$2km > 0 \implies m > 0$$

$$\begin{array}{c|c|c} s^3 & 1 & k(m+2)-1 \\ s^2 & k & 2km \\ s^1 & \frac{-1}{k} & \frac{k(m+2)-1}{2km} \\ s^0 & 1 & 0 \end{array}$$

$$(k-2)(m+2k-1)$$

$$\begin{cases} > 0 & \forall m \geq 0 \text{ if } k \geq 2 \\ < 0 & \forall m \leq 0 \text{ if } k < \frac{1}{2} \\ \geq 0 & \forall m \leq \frac{2k-1}{2-k} \end{cases}$$

System is marginally stable if  $m = \frac{2k-1}{2-k}$

System stable

$$k > 2, m > 0$$

$$k \in [\frac{1}{2}, 2), 0 \leq m \leq \frac{2k-1}{2-k}$$



Q 9.  
(b)  $G_1(s) = \frac{(sm)(s+2)}{s}$

$G_2(s) = \frac{k}{s^2-1}$

steady state error  $\rightarrow e = \frac{1}{\lim_{s \rightarrow 0} s G_1(s) G_2(s)}$  when system stable

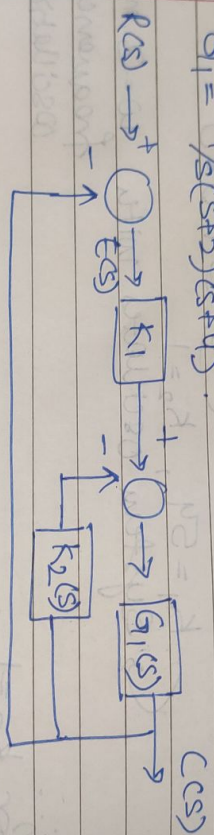
group response  $s \rightarrow 0$   
 $= (-1/2km)$

given  $\text{error} \pm 0.1$

$km$  are both +ve  
so;  
 $\frac{-1}{2km} \geq -0.1$   
 $|m| \geq 5/k$

for  
k

Q 8. (a,b)  $G_1 = 1/s(s+2)(s+4)$



$1 + k_2 s G_1(s) = 0$

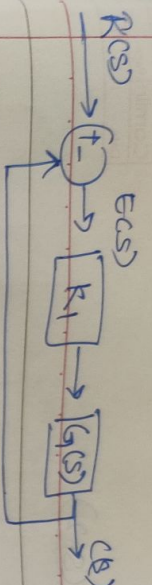
$\rightarrow$  will have 2 equal -ve roots.

$(s+2)(s+4) + k_2 = 0$

$G_1(s) = \frac{1}{s(1+(s+2)(s+4))}$

$k_2 = 1$

$s^2 + 6s + 9 = 0$   
 $\begin{cases} -3 \\ -3 \end{cases}$  2 equal roots.



$$1 + K_1 G(s) = 0$$

$$1 + K_1 = 0$$

$$s(s+2)(s+4) + 5$$

$$s^3 + 6s^2 + 9s + K_1 = 0 \quad K_1 > 0$$

$$\begin{array}{c|ccc} s^3 & 1 & 9 & \\ s^2 & 6 & K_1 & \\ s & 54 & 0 & \\ & 6 & & \end{array}$$

$0 < K_1 < 54$   
for stable system.

$$s^3 + 6s^2 + 9s + 54 = 0$$

$$(s+6)(s^2+9) = 0$$

$K_1 = 54, K_2 = 1$   
System oscillates with frequency of oscillation.

for  $K_2 = 1$   
 $\Rightarrow K_1 \in (0, 54)$

at  $K_1 = 54$ , system is marginally stable (oscillates).

if  $K_2$  not specified  
 $\omega = \sqrt{8 + K_2} \text{ rad/sec}$   
when  $K_1 = 6(8 + K_2)$ .



Q8 (c) as closed loop pole at will make system unstable, we won't be able to approximate step response as underdamped.

$$s^3 + 6s^2 + (8+k_2)s + k_1 = 0$$

$$s = -5 \rightarrow \text{pole}$$

$$k_1 = -s(6s + k_2)$$

eqn

$$k_1 = 5(k_2 + 3)$$

$$(8+5)(s^2 + s + k_2 + 3) = 0$$

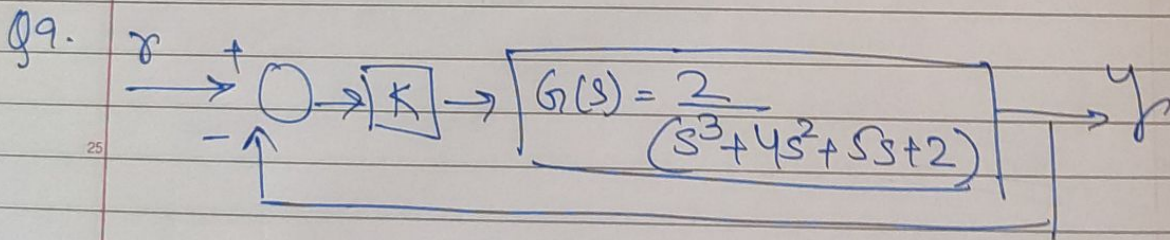
1 pole at  $s = -5$   
both complex poles have real =  $-1/2$   
real pole is 10 times away from img axis.

Dominant pole approximation

$$k_2 = 1 \quad k_1 = 20$$

$$G(s) = \frac{20}{s^3 + 6s^2 + 9s + 20}$$

$$G_{app}(s) = \frac{4}{s^2 + s + 4}$$



$$Y(s) = \frac{K G(s)}{1 + K G(s)} X(s)$$

$$E(s) = \frac{R(s)}{1 + K G(s)}$$

$$R(s) = 1/s$$

$$e = \frac{1}{1 + K \lim_{s \rightarrow 0} G(s)}$$



$$e = \frac{1}{1+K}$$

$$G(s) = (-2, -1, -1) \text{ in LHP.}$$

$$\frac{1}{1+K} \leq 0.15$$

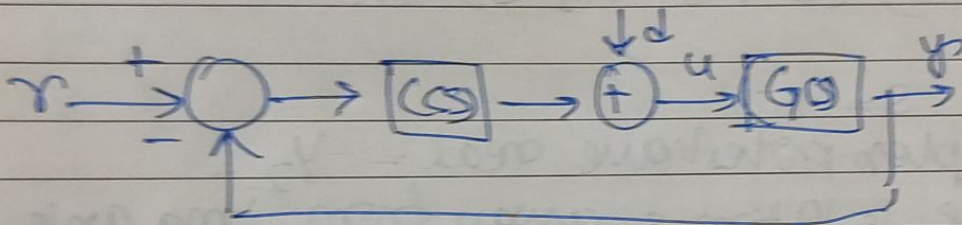
$$K \geq 17/3$$

Characteristic eqn<sup>n</sup>

$$1+K G(s)=0$$

$$s^3 + 4s^2 + 5s + 2 + 2K = 0$$

Q10

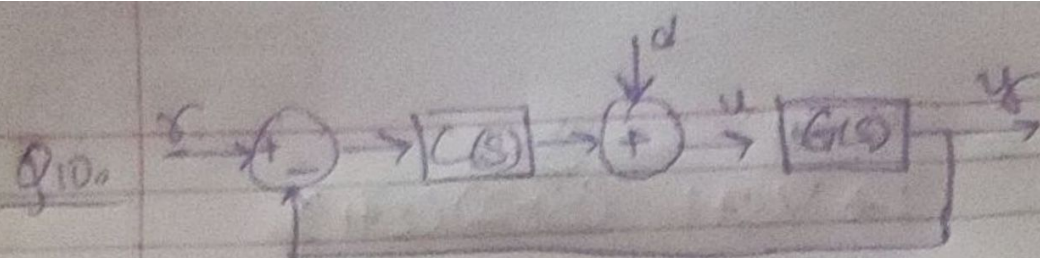


$$G(s) = 1/s+1$$

$$C(s) = \frac{s+1}{s+1}$$

Case 1





$$R(s) = D(s) \cdot \frac{1}{s}$$

$$G(s)(C(s)(R(s) - Y(s)) + D(s)) = Y(s)$$

$$Y(s) = U(s)G(s)$$

$r \rightarrow y$

$$G(s)C(s)R(s) = Y(s)(1 + G(s)C(s))$$

$$\boxed{\frac{Y(s)}{R(s)} = \frac{GC}{1 + GC}}$$

$d \rightarrow y$

$$GD = Y(1 + GC)$$

$$\boxed{\frac{Y}{D} = \frac{G}{1 + GC}}$$

$r \rightarrow u$

$$\frac{UG}{R} = \frac{GE}{1 + GE}$$

$$\boxed{\frac{U}{R} = \frac{C}{1 + GC}}$$

$d \rightarrow u$

$$\frac{GU}{D} = \frac{G}{1 + GC}$$

$$\boxed{\frac{U}{D} = \frac{1}{1 + GC}}$$

Case (1)

$$C = \frac{s+1}{s+1} \quad G = \frac{1}{s+1}$$

$$r \rightarrow y = \frac{Y}{R} = \frac{1/(s+1)}{1 + 1/(s+1)} = \frac{1}{s+2}$$

$$Y = \frac{1}{s+2} \cdot \frac{1}{s}$$

$$Y = \frac{1}{2} \left( \frac{1}{s} - \frac{1}{s+2} \right)$$

$$\boxed{y(t) = \frac{1}{2} (u(t) - e^{-2t} u(t))}$$



$d \rightarrow y$

$$\frac{Y}{D} = \frac{Y(s+1)}{1+Y(s+1)} = \frac{s+1}{(s+2)(s+1)}$$

$$Y = \frac{s+1}{(s+2)(s+1)(s)} \rightarrow Y = \frac{1}{2s} + \frac{2}{3(s-1)} - \frac{1}{6(s+2)}$$

$$y(t) = \left( \frac{1}{2} + \frac{2e^t}{3} - \frac{1}{6}e^{-2t} \right) u(t)$$

$d \rightarrow u$

$$\frac{U}{D} = \frac{1}{1+Y(s+1)} = \frac{s+1}{s+2} \quad U = \frac{s+1}{s(s+2)}$$

$$U = \frac{1}{2} \left( \frac{1}{s+2} + \frac{1}{s} \right) \quad u(t) = \left( \frac{1}{2} + \frac{1}{2}e^{-2t} \right) u(t)$$

$\sigma \rightarrow u$

$$\frac{U}{R} = \frac{s+1}{1+Y(s+1)} = \frac{s+1}{s+2}$$

$$U = \frac{s+1}{(s+2)s} = \frac{3}{2} \frac{1}{s+2} - \frac{1}{2s} \quad u(t) = \left( \frac{3e^{-2t}}{2} - \frac{1}{2} \right) u(t)$$



Case ②  $G(s) = \frac{s+1}{s+1}$   $C(s) = \frac{1}{s+1}$

$d \rightarrow u$

$$\frac{U}{D} = \frac{1}{1+GC} \rightarrow u(t) = \left(\frac{1}{2} + \frac{1}{2}e^{-2t}\right)u(t)$$

remains the same.

$r \rightarrow y$

$$\frac{Y}{R} = \frac{GC}{1+GC} \rightarrow y(t) = \left(\frac{1}{2} - \frac{1}{2}e^{-2t}\right)u(t)$$

remains the same

$r \rightarrow u$

$$\frac{U}{R} = \frac{1/s+1}{1+1/s+1}$$

Same as case ①  
 $d \rightarrow y$ .

$$u(t) = \left(-\frac{1}{2} + \frac{2}{3}e^{-t} - \frac{e^{-2t}}{6}\right)u(t)$$

$d \rightarrow y$

$$\frac{Y}{D} = \frac{s+1/s+1}{1+s+1/s+1}$$

Same as case ①  
 $r \rightarrow u$ .

$$y(t) = \left(\frac{3}{2}e^{-2t} - \frac{1}{2}\right)u(t)$$