### Nonlinear Control for Tracking and Obstacle Avoidance of a Wheeled Mobile Robot With Nonholonomic Constraint

Hongjiu Yang, Xiaozhao Fan, Peng Shi, Fellow, IEEE, and Changchun Hua, Member, IEEE

Abstract—This brief presents a novel control scheme for some problems on tracking and obstacle avoidance of a wheeled mobile robot with nonholonomic constraint. An extended state observer is introduced to estimate the unknown disturbances and velocity information of the wheeled mobile robot. A nonlinear controller is designed to achieve tracking target and obstacle avoidance in complex environments. Note that tracking errors converge to a residual set outside the obstacle detection region. Moreover, the obstacle avoidance is also guaranteed inside the obstacle detection region. Simulation results are given to verify the effectiveness and robustness of the proposed design scheme.

Index Terms—Extended state observer, nonholonomic constraint, obstacle avoidance, trajectory tracking, wheeled mobile robot.

#### I. INTRODUCTION

ARIOUS types of mobile robots will change our lives in the near further. Environment information is obtained by sensors in motion control for a mobile robot [1], [2]. As an important branch of mobile robots, wheeled mobile robots have better flexibility and larger work space than the traditional industrial robots [3], [4]. Therefore, they are widely used in complex environments of military and civil occasions in [5]. Some control problems on a nonholonomic wheeled mobile robot have been investigated via neural networks in [6]. Adaptive sliding mode control has been used to deal with the model uncertainty in wheeled mobile robots [7]. A radio frequency identification-based control method has also been proposed for a mobile robot [8], [9]. Moreover, wheeled mobile robots often encounter obstacles when working in complex environment [10]. Based on kinematical equations, some control problems have been studied on tracking and obstacle avoidance, please refer to [11]-[14], and so on. Note that trajectory tracking and obstacle avoidance controllers are separately designed in most

Manuscript received October 23, 2014; revised February 5, 2015 and May 17, 2015; accepted July 4, 2015. Date of publication August 6, 2015; date of current version February 17, 2016. Manuscript received in final form July 8, 2015. This work was supported in part by the National Natural Science Foundation of China under Grant 61203023, in part by the Australian Research Council under Grant DP140102180, Grant LP140100471, and Grant LE150100079, and in part by the 111 Project under Grant B12018. Recommended by Associate Editor A. G. Aghdam.

H. Yang, X. Fan, and C. Hua are with the School of Electrical Engineering, Yanshan University, Qinhuangdao 066004, China (e-mail: yanghongjiu@ysu.edu.cn; yusfxzh@163.com; cch@ysu.edu.cn).

P. Shi is with the School of Electrical and Electronic Engineering, The University of Adelaide, Adelaide, SA 5005, Australia, also with the College of Engineering and Science, Victoria University, Melbourne, VIC 8001, Australia, and also with the College of Automation, Harbin Engineering University, Harbin 150001, China (e-mail: peng.shi@adelaide.edu.au).

Color versions of one or more of the figures in this paper are available online at http://ieeexplore.ieee.org.

Digital Object Identifier 10.1109/TCST.2015.2457877

of the existing works, which easily lead to the low work efficiency and cause high frequency noise [15]. Furthermore, there is still a lot of space to improve the anti-interference ability of controllers in the current results.

The technique of active disturbance rejection control is proposed in [16]. Nowadays, it becomes a very attractive methodology in the field of automation [17]. Extended state observer is one of an important component part in the active disturbance rejection control technique [18], [19]. As a control scheme, the extended state observer has been well studied and applied successfully in [20]. The extended state observer is not dependent on the specific mathematical models of disturbances, and it also does not need to measure the effects of disturbances directly [21]. A nonsmooth feedback function, which is inherently robust against plant variations, is used to reject the disturbances in the form of orders of magnitude [22]. All these special feedback mechanisms make the active disturbance rejection control technique has a particularly satisfactory performance [23], [24]. Hence, it is an interesting idea to study the trajectory tracking and obstacle avoidance of a wheeled mobile robot via active disturbance rejection control, which motivates us to make an effort in this brief.

*Notation:* In the following, if not explicitly stated, matrices are assumed to have compatible dimensions. The shorthand diag $\{C_1, C_2, \ldots, C_n\}$  denotes a diagonal matrix.  $\Re^{n \times m}$  denotes the *n*-dimensional configuration space  $\mathbb{C}$  with generalized coordinate  $(q_1, \ldots, q_n)$  and subject to m constraints. Note that  $\|\cdot\|$  represents the two-norm of a vector. The piecewise continuous function  $\mathbf{fal}(\cdot)$  is given as follows:

$$\mathbf{fal}(\mathcal{E}, \alpha, \delta) = \begin{cases} |\mathcal{E}|^{\alpha}, & \|\mathcal{E}\| > \delta \\ \mathcal{E} & \|\mathcal{E}\| \le \delta \end{cases}$$
 (1)

where  $\alpha$  and  $\delta$  are the constants and  $\mathcal{E}$  is a variable. To relax notation, the **fal**(·) is used in the place of **fal**( $\mathcal{E}$ ,  $\alpha$ ,  $\delta$ ) in this brief.

#### II. PRELIMINARIES AND PROBLEM STATEMENT

#### A. Obstacle Avoidance Problem Description

Diagram description of obstacle avoidance problem for the mobile robot tracking the given target is shown in Fig. 1. In Fig. 1, there exist some obstacles between the target and current positions,  $[x_r \ y_r \ \theta_r]^T$  is the target position, and  $[x_i \ y_i \ \theta_i]^T$  is the real-time position and orientation of the mobile robot.

#### B. Nonholonomic Wheeled Mobile Robot Model

The model of a wheeled mobile robot is shown in Fig. 2. In this brief, we consider a two-wheeled mobile robot that



Fig. 1. Mobile robot avoidance obstacle problem description.

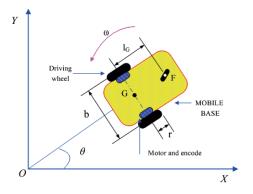


Fig. 2. Model of two-wheeled nonholonomic mobile robot.

is described by the following nonlinear generalized dynamics system:

$$C(\mathbf{q})\ddot{\mathbf{q}} + B_m(\mathbf{q}, \dot{\mathbf{q}})\dot{\mathbf{q}} + F(\dot{\mathbf{q}}) + \tau_d = E(\mathbf{q})\tau - A^T(\mathbf{q})\lambda$$
 (2)

in which  $C(\mathbf{q}) \in \mathbb{R}^{n \times n}$  is a symmetric positive definite inertia matrix,  $B_m(\mathbf{q}, \dot{\mathbf{q}}) \in \mathbb{R}^{n \times n}$  is the centripetal and Coriolis matrix,  $F(\dot{\mathbf{q}}) \in \mathbb{R}^{n \times 1}$  is the surface friction,  $\tau_d$  is the bounded unknown disturbances,  $E(\mathbf{q}) \in \mathbb{R}^{n \times r}$  is the input transformation matrix,  $\tau = [\tau_r \tau_l] \in \mathbb{R}^{r \times 1}$  is the input vector,  $A(\mathbf{q}) \in \mathbb{R}^{m \times n}$  is the matrix associated with constraints, and  $\lambda \in \mathbb{R}^{m \times 1}$  is the constraint force vector. There exist some parameter relations of Fig. 2 to system (2) (see [6]).

#### C. Structural Properties of a Mobile Robot

In this section, an appropriate dynamic model of the mobile robot than system (2) is obtained to achieve control purpose. The kinematic equation is given as follows:

$$\dot{\mathbf{q}} = \mathbf{S}(\mathbf{q})\mathbf{v} \tag{3}$$

where  $\mathbf{v} = [v \ w]^T$ ,  $\mathbf{v}$  is bounded.  $\mathbf{S}(\mathbf{q})$  is a Jacobian matrix. Taking the time derivative of (3), it is obtained that

$$\ddot{\mathbf{q}} = \mathbf{S}_q \dot{\mathbf{v}} + (\mathbf{S}_q \odot \dot{\mathbf{q}}) \mathbf{v} \tag{4}$$

where  $\mathbf{S}_q$  is the partial derivative of each component of Jacobian matrix  $\mathbf{S}$  with respect to  $\mathbf{q}$ ,  $\odot$  is the multiplication of each component (which is a row vector) of the matrix  $\mathbf{S}_q$  with  $\dot{\mathbf{q}}$ . Now, substituting (4) in system (2), we rewrite system (2) as follows:

$$\overline{C}(\mathbf{q})\dot{\mathbf{v}} + \overline{B}_m(\mathbf{q}, \dot{\mathbf{q}})\mathbf{v} + \overline{F} + \overline{\tau}_d = \overline{\tau}$$
 (5)

in which  $\overline{C}(\mathbf{q}) \in \Re^{2\times 2}$  is the symmetric positive definite inertia matrix,  $\overline{B}_m(\mathbf{q}, \dot{\mathbf{q}}) \in \Re^{2\times 2}$  is the centripetal and Coriolis matrix,  $\overline{F} \in \Re^{2\times 1}$  is the surface friction that is assumed to be bounded,  $\overline{\tau}_d$  denotes the bounded unknown disturbances,

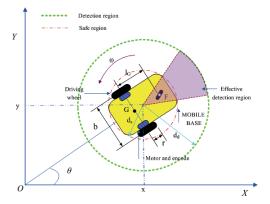


Fig. 3. Wheeled mobile robot with avoidance and detection region.

and  $\overline{\tau} \in \Re^{2 \times 1}$  is the input vector. Then, some fundamental properties regarding system (5) are summarized as follows.

Property 1: [6]  $\overline{\underline{C}}(\mathbf{q})$  and  $\|\overline{\boldsymbol{B}}_{m}(\mathbf{q},\dot{\mathbf{q}})\|$  are bounded.

Property 2: [6]  $\overline{C}(\mathbf{q}) - 2\overline{B}_m(\mathbf{q}, \dot{\mathbf{q}})$  is skew symmetric.

## III. DYNAMIC TRACKING AND AVOIDANCE OBSTACLE CONTROL FOR MOBILE ROBOT

#### A. Potential Function for Obstacle Avoidance

From [10], the following potential function is introduced to solve the problems of obstacle avoidance for a wheeled mobile robot in this brief:

$$V_{\text{ob}} = \left(\min\left\{0, \ \left(L_{\text{rod}}^2 - d_d^2\right) \left(L_{\text{rod}}^2 - d_s^2\right)^{-1}\right\}\right)^2 \tag{6}$$

where  $d_d > d_s > 0$ . Note that,  $d_s$  and  $d_d$  are the two radii of the avoidance and detection regions, respectively. The avoidance obstacle and detection region of a wheeled mobile robot are shown in Fig. 3.

A reference trajectory is given as  $(x_r, y_r, \theta_r)$ , which is bounded. Define the tracking errors  $e_x = x_i - x_r$ ,  $e_y = y_i - y_r$ , and  $e_\theta = \theta_i - \theta_r$ . Moreover,  $L_{\text{rod}}$  is the distance between the robot and the obstacle. Hence, we have the following exact mathematical expression:

$$L_{\text{rod}} = \sqrt{(x_i - x_o)^2 + (y_i - y_o)^2}$$
 (7)

where  $(x_o, y_o)$  denotes the obstacle position, and  $(x_i, y_i)$  denotes the real-time location position of a mobile robot.

Remark 1: The given trajectory is smooth, and satisfies  $|e_{\theta}| \neq (\pi/2)$ . The reference trajectory does not initiate sharp turns of 90° with respect to the current orientation of a mobile robot. We allow a perturbed desired orientation  $\bar{\theta}_r$  be instead of the desired orientation  $\theta_r$  to solve singularity problem. In the case, there has  $\bar{\theta}_r = \theta_r + \bar{\epsilon}_{\theta}$ , in which  $\bar{\epsilon}_{\theta} \neq 0$  is a small perturbation value.

#### B. Extended State Observer Design

From (5), we obtain

$$\dot{\mathbf{v}} = -\overline{\mathbf{C}}^{-1}(\mathbf{q})[\overline{\mathbf{B}}_m(\mathbf{q}, \dot{\mathbf{q}})\mathbf{v} + \overline{\mathbf{F}} + \overline{\mathbf{\tau}}_d] + \overline{\mathbf{C}}^{-1}(\mathbf{q})\overline{\mathbf{\tau}}.$$
 (8)

Allowing  $\mathbf{x}_2 = -\overline{C}^{-1}(\mathbf{q})[\overline{B}_m(\mathbf{q},\dot{\mathbf{q}})\mathbf{x}_1 + \overline{F} + \overline{\tau}_d]$ ,  $\mathbf{x}_1 = \mathbf{v}$ , and  $\mathbf{u} = \overline{C}^{-1}(\mathbf{q})\overline{\tau}$ , one has that

$$\begin{cases} \dot{\mathbf{x}}_1 = \mathbf{x}_2 + \mathbf{u} \\ \dot{\mathbf{x}}_2 = \mathbf{h} \end{cases} \tag{9}$$

in which **h** is defined as the first-order derivative of extended state  $\mathbf{x}_2$ . An extended state observer for system (9) is structured by

$$\begin{cases} \dot{\hat{\mathbf{x}}}_1 = \hat{\mathbf{x}}_2 - \beta_{01}\mathbf{r}_1 + \mathbf{u} \\ \dot{\hat{\mathbf{x}}}_2 = -\beta_{02}\mathbf{fal} \end{cases}$$
(10)

in which  $\hat{\mathbf{x}}_1$  and  $\hat{\mathbf{x}}_2$  are the two estimated values of  $\mathbf{x}_1$  and  $\mathbf{x}_2$ , respectively, and  $\mathbf{r}_1 = \hat{\mathbf{x}}_1 - \mathbf{x}_1$  is extended state error. Parameters  $\beta_{01}$  and  $\beta_{02}$  are the regulable gain constants. According to (9) and (10), the error system is written as follows:

$$\begin{cases} \dot{\mathbf{r}}_1 = \mathbf{r}_2 - \beta_{01} \mathbf{r}_1 \\ \dot{\mathbf{r}}_2 = -\beta_{02} \mathbf{f} \mathbf{a} \mathbf{l} - \mathbf{h} \end{cases}$$
(11)

in which  $\mathbf{r}_2 = \hat{\mathbf{x}}_2 - \mathbf{x}_2$  is the extended state error.

Assumption 1: The first-order derivative of  $\mathbf{x}_2$  is existed and bounded. That is,  $\dot{\mathbf{x}}_2 = \mathbf{h}$  is bounded.

Remark 2: The variable **h** denotes the change rate of external force, it is bounded in the wheeled mobile robot with nonholonomic constraint. Therefore, **h** is bounded in Assumption 1 as [21], [23], and practical.

#### C. Nonlinear Controller Design

The control objective is to design an appropriate controller that assures a mobile robot accurately tracking the given trajectory with obstacles in complex environment. Let  $\hat{\theta}_r$  be  $\hat{\theta}_r = [E_x(t)\hat{E}_y - E_y(t)\hat{E}_x](E_x^2 + E_y^2)^{-1}$ , where  $\hat{E}_x \approx [E_x(t+T) - E_x(t)]T^{-1}$  and  $\hat{E}_y \approx [E_y(t+T) - E_y(t)]T^{-1}$ , for some small T > 0. Note that, both  $E_x$  and  $E_y$  are smooth. Let an auxiliary velocity control input  $\mathbf{v}_r$  be given by

$$\mathbf{v}_r = \left[ -k_1 \sqrt{E_x^2 + E_y^2} \cos e_\theta, -k_2 e_\theta + \hat{\theta}_r \right]^T$$
 (12)

in which  $k_1 > 0$  and  $k_2 > 0$  are the two adjustable parameters. Then, the auxiliary velocity error is obtained as follows:

$$\mathbf{e}_v = \mathbf{v}_r - \mathbf{v}.\tag{13}$$

Using (5) and differentiating (13), the mobile robot dynamics is rewritten as

$$\overline{C}(\mathbf{q})\dot{\mathbf{e}}_v = -\overline{B}_m(\mathbf{q},\dot{\mathbf{q}})\mathbf{e}_v - \overline{\tau} + \mathbf{x}_2$$
 (14)

in which  $\mathbf{x}_2 = \overline{C}(\mathbf{q})\dot{\mathbf{v}}_r + \overline{B}_m(\mathbf{q},\dot{\mathbf{q}})\mathbf{v}_r - \overline{F}(\mathbf{q}) + \overline{\tau}_d$ . Hence, the actual controller is designed as

$$\bar{\tau} = \hat{\mathbf{x}}_2 + k_3 \mathbf{fal}(\mathbf{e}_v, \alpha, \delta) \tag{15}$$

where  $k_3$  is a sufficiently large positive constant.

# IV. EFFECTIVENESS ANALYSIS OF SYSTEM CONTROLLER A. Convergence of Second-Order Extended State Observer

Theorem 1: Consider the error system given by (11) under Assumption 1. Choose the nonsmooth function **fal** as in (1), which closes to the derivative of the system's nonlinearities. There exist appropriate positive matrices  $\beta_{01}$  and  $\beta_{02}$ , which are defined in (10). By adjusting parameter  $\beta_{02}$  such that  $\mathbf{r}_1 \simeq (\mathbf{h}\beta_{02}^{-1})^2$  holds, which shows that estimate error  $\mathbf{r}_1$  is bounded. Based on (11), it is easy to know that  $\mathbf{r}_2$  is

also bounded. Then, the observation accuracy of extended state observer is guaranteed.

*Proof:* Consider the following Lyapunov function:

$$V(\mathbf{r}_{1}, \mathbf{r}_{2}) = M|\mathbf{r}_{1}|^{\frac{3}{2}} - O\mathbf{r}_{1}\mathbf{r}_{2} + N\mathbf{r}_{2}^{2}$$
 (16)

in which M, O, and N are constants that are satisfied with

$$M > 0$$
,  $O > 0$ ,  $N > 0$ ,  $O^2 - 4MN < 0$ . (17)

Hence, it is obtained that (16) is positive. The partial derivatives of (16) with respect to  $\mathbf{r}_1$  and  $\mathbf{r}_2$  are given as follows:  $(\partial V/\partial \mathbf{r}_1) = (3/2)M|\mathbf{r}_1|^{(1/2)}\mathrm{sign}(\mathbf{r}_1) - O\mathbf{r}_2$  and  $(\partial V/\partial \mathbf{r}_2) = -O\mathbf{r}_1 + 2N\mathbf{r}_2$ . One has that  $\dot{V}(\mathbf{r}_1, \mathbf{r}_2) = ((3/2)M - 2N\beta_{02} + O\beta_{01}|\mathbf{r}_1|^{(1/2)})|\mathbf{r}_1|^{-(1/4)}|\mathbf{r}_1|^{(3/4)}\mathrm{sign}(\mathbf{r}_1)\mathbf{r}_2 - O\mathbf{r}_2^2 - ((3/2)M\beta_{01} - O\beta_{02})|\mathbf{r}_1|^{2(3/4)} + Oh\mathbf{r}_1 - 2Nh\mathbf{r}_2$ . The above equation is regarded as the form of a quadratic function with variables  $|\mathbf{r}_1|^{(3/4)}\mathrm{sign}(\mathbf{r}_1)$  and  $\mathbf{r}_2$ . It is rewritten as the following form:

$$\dot{V} = -p|\mathbf{r}_1|^{2\left(\frac{3}{4}\right)} + q|\mathbf{r}_1|^{\frac{3}{4}}\operatorname{sign}(\mathbf{r}_1)\mathbf{r}_2 - o\mathbf{r}_2^2 + \Theta \qquad (18)$$

in which p is a constant, q is a function of the variable  $\mathbf{r}_1$ , with  $q=((3/2)M-2N\beta_{02}+O\beta_{01}|\mathbf{r}_1|^{(1/2)})|\mathbf{r}_1|^{-(1/4)}$ ,  $p=((3/2)M\beta_{01}-O\beta_{02})$  and o=O, and  $\Theta$  denotes the variable  $O\mathbf{h}\mathbf{r}_1-2N\mathbf{h}\mathbf{r}_2$ . The quadratic part of (18) is negative, if and only if p>0, q>0, o>0,  $q^2-4po<0$ . Let constants  $a=O\beta_{01}$ ,  $c=(3/2)M-2N\beta_{02}$ , and  $b=(O((3/2)M\beta_{01}-O\beta_{02}))^{(1/2)}$ , and variable  $\phi=|\mathbf{r}_1|^{(1/4)}$ . There exists  $c+a\phi^2<2b\phi$  that holds for any  $\phi$ , if and only if inequality  $b^2-ac>0$  is satisfied. Inequality  $a^2-4po<0$  holds, if and only if  $\phi$  in the quadratic equation  $a\phi^2-2b\phi+c=0$  is taken as the value in the following set  $\{\phi|\phi_1<\phi<\phi_2\}$ , where  $\phi_1=a^{-1}(b-(b^2-ac)^{(1/2)})$  and  $\phi_2=a^{-1}(b+(b^2-ac)^{(1/2)})$  are the two roots of  $a\phi^2-2b\phi+c=0$ . Form inequality  $b^2-ac>0$ , we have

$$O\beta_{02}(2N\beta_{01} - O) > 0. (19)$$

In order to obtain a large interval between the two roots  $\phi_1$  and  $\phi_2$  such that the smaller root approaches to zero, inequality (19) has to be large enough. Therefore, we need to take a large O, and choose N to make inequality  $2N\beta_{01} - O$  be large. According to the above analysis, we have the conclusion that coefficients M, O, and N of (16) are satisfied with  $3M\beta_{01} > 2O\beta_{02}$ ,  $3M > 4N\beta_{02}$ , and  $2N\beta_{01} > O$ , where  $\beta_{01}$  and  $\beta_{02}$  are the two given parameters. Giving a large O > 0, and closing  $N > (O/2\beta_{01})$ ,  $M > (4/3)N\beta_{02}$ , which also satisfy inequality (17), one has that  $M > (2O\beta_{02}/3\beta_{01})$ holds. Hence, (16) is positive and its derivative along the system trajectories is negative in a wide range of variable  $\mathbf{r}_1$ and arbitrary  $\mathbf{r}_2$ . Therefore, the error system (11) is stable at its equilibrium point. Based on (18), we suppose  $\dot{V} = \dot{V}_1 - \dot{V}_2$ , where  $\dot{V}_1 = -p|\mathbf{r}_1|^{2(3/4)} + q|\mathbf{r}_1|^{(3/4)} \operatorname{sign}(\mathbf{r}_1)\mathbf{r}_2 - o\mathbf{r}_2^2$ , and  $\dot{V}_2 = -O\mathbf{h}\mathbf{r}_1 + 2N\mathbf{h}\mathbf{r}_2$ . After selecting the coefficients M, O, and N by the above-described manner, function (18) is positive in the region of the intersected upper portion of parabolic  $V_1$ and flat plane  $V_2$ . Both the magnitude of  $\mathbf{r}_1$  and the root of equation  $p|\mathbf{r}_1|^{2(3/4)} = O\mathbf{h}\mathbf{r}_1$  are in the same order, i.e.,  $\mathbf{r}_1 \simeq ((O\mathbf{h}p^{-1})^2$ . Furthermore, it is known that p and  $O\beta_{02}$  are in the same order from  $p=((3/2)M\beta_{01}-O\beta_{02})$ . Hence, there exists  $\mathbf{r}_1\simeq (O\mathbf{h}p^{-1})^2\simeq (\mathbf{h}\beta_{02}^{-1})^2$ . The error  $\mathbf{r}_1$  is bounded. Thus, there exists a constant  $R_1$  with  $R_1\leq 1$  such that  $\mathbf{r}_1\leq R_1\leq 1$  by adjusting parameter  $\beta_{02}$ . There exists  $\dot{\mathbf{r}}_1=0$  when the error system (11) is stable. Note that,  $\mathbf{r}_1$  and  $\mathbf{r}_2$  are satisfied with  $\mathbf{r}_2=\beta_{01}\mathbf{r}_1$ , so the estimation error  $\mathbf{r}_2$  is also bounded. There also exists a known  $R_{2\,\mathrm{max}}$  such that  $\|\mathbf{r}_2\|\leq R_{2\,\mathrm{max}}$ .

#### B. Stability Analysis of Closed-Loop System

Theorem 2: Consider the system of a nonholonomic mobile robot (5), the extended state observer (10), the error system (11), the nonlinear controller  $\bar{\tau}$  in (15), and the nonsmooth function **fal** as in (1) and Theorem 1. By adjusting parameters, such that the following two inequalities hold:

$$\|\mathbf{e}\| > \|\mathbf{d}\| \mathcal{K}_{\min}^{-1}(\mathcal{M}), \quad \|e_{\theta}\| > \epsilon_{\theta} K_{2\min}^{-1} \|\mathbf{e}_{v}\| > \max \left\{ \delta^{1-\alpha} R_{2\max} K_{3\min}^{-1}, \quad R_{2\max} K_{3\min}^{-1} \right\}.$$

Then, both **e** and **e**<sub> $\theta$ </sub> converge in a residual set. That is, we have: 1) the signals in the closed-loop system are bounded; 2) the auxiliary velocity error is arbitrarily small value; and 3) the steady-state tracking errors  $e_x$ ,  $e_y$ , and  $e_\theta$  are uniformly ultimately bounded.

Proof: Consider the following Lyapunov function as:

$$V_3 = \frac{1}{2} \left( e_x^2 + e_y^2 + e_\theta^2 \right) + \left( \min \left\{ 0, \frac{L_{\text{rod}}^2 - d_d^2}{L_{\text{rod}}^2 - d_s^2} \right\} \right)^2 + V_4$$

where  $V_4 = (1/2)\mathbf{e}_v^T \overline{\mathbf{C}}(\mathbf{q})\mathbf{e}_v$ . The derivative of the Lyapunov function  $V_3$  is given by

$$\dot{V}_3 = e_x \dot{e}_x + e_y \dot{e}_y + e_\theta \dot{e}_\theta + \frac{\partial V_{\text{ob}}}{\partial x_i} \dot{x}_i + \frac{\partial V_{\text{ob}}}{\partial y_i} \dot{y}_i + \dot{V}_4. \quad (20)$$

Applying (14) and (15), taking the manipulate of derivative for  $V_4$ , one has that

$$\dot{V}_4 = \frac{1}{2} \mathbf{e}_v^T (\overline{\mathbf{C}}(\mathbf{q}) - 2\overline{\mathbf{B}}_m(\mathbf{q}, \dot{\mathbf{q}})) \mathbf{e}_v - \mathbf{e}_v^T k_3 \text{fal} + \mathbf{e}_v^T \mathbf{r}_2.$$

By Property 2, the first term of the right-hand side in  $\dot{V}_4$  is zero. Thus, there exists

$$\dot{V}_4 = -\mathbf{e}_p^T k_3 \mathbf{fal} + \mathbf{e}_p^T \mathbf{r}_2. \tag{21}$$

Substituting (1) into (21), one has that

$$\dot{V}_{4a} < -\|\mathbf{e}_v\|^2 K_{3\min} \delta^{\alpha-1} + \|\mathbf{e}_v\| R_{2\max}, \quad \|\mathbf{e}_v\| \le \delta 
\dot{V}_{4b} < -\|\mathbf{e}_v\|^2 K_{3\min} + \|\mathbf{e}_v\| R_{2\max}, \quad \|\mathbf{e}_v\| > \delta$$

in which  $R_{2 \text{ max}}$  is the maximum value of  $\mathbf{r}_2$  and  $K_{3 \text{ min}}$  is the minimum value of  $k_3$ . We have that  $\dot{V}_{4a}$  is guaranteed negative as long as the following condition holds:

$$\|\mathbf{e}_v\| > \delta^{1-\alpha} R_{2\max} K_{3\min}^{-1}.$$
 (22)

Furthermore, it is obtained that  $\dot{V}_{4b}$  is guaranteed negative as long as the following condition holds:

$$\|\mathbf{e}_v\| > R_{2\max} K_{3\min}^{-1}.$$
 (23)

Based on (22) and (23), we take

$$\|\mathbf{e}_v\| > \max \left\{ \delta^{1-\alpha} R_{2\max} K_{3\min}^{-1}, R_{2\max} K_{3\min}^{-1} \right\}$$
 (24)

in this section.

TABLE I
COMPARISON BETWEEN PD CONTROLLER AND
NONLINEAR CONTROLLER

	$e_x(m)$	$e_y(m)$	$e_{\theta}(rad)$	Jitter
PD	0.447	0.446	3.1400	Yes
This paper	0.400	0400	3.1400	No

Considering the error dynamics  $\dot{e}_x$ ,  $\dot{e}_y$ , and  $\dot{e}_\theta$ , using the expressions  $\cos(e_\theta) = E_x(E_x^2 + E_y^2)^{-1}$ ,  $\sin(e_\theta) = E_y(E_x^2 + E_y^2)^{-1}$  and substituting (12) into the first five terms of (20), we have that

$$\dot{V}_5 \le k_1 \cos^2(e_\theta) \left( E_x^2 + E_y^2 \right) - \|e_\theta\| (k_2 \|e_\theta\| - \epsilon_\theta) - \zeta$$
 (25)

in which  $\zeta = e_x \dot{x}_r - e_y \dot{y}_r$ . When the robot is outside the detection region, i.e.,  $L_{\rm rod} > d_d$ , we have  $(\partial V_{\rm ob}/\partial x_i) = (\partial V_{\rm ob}/\partial y_i) = 0$  and the inequality (25) becomes

$$\dot{V}_{5} = - \begin{bmatrix} e_{x} \\ e_{y} \\ e_{\theta} \end{bmatrix}^{T} \mathcal{M} \begin{bmatrix} e_{x} \\ e_{y} \\ e_{\theta} \end{bmatrix} + \begin{bmatrix} e_{x} \\ e_{y} \\ e_{\theta} \end{bmatrix} \begin{bmatrix} \dot{x}_{r} \\ \dot{y}_{r} \\ -\epsilon_{\theta} \end{bmatrix}$$

for the reason of that  $\cos^2(e_\theta)$  is not equal to zero. Hence,  $\dot{V}_5 < 0$  holds, if the following inequality holds:

$$\|\mathbf{e}\| > \|\mathbf{d}\| \mathcal{K}_{\min}^{-1}(\mathcal{M}) \tag{26}$$

where  $\mathbf{e} = [e_x \ e_y \ e_\theta]^T$ ,  $\mathbf{d} = [\dot{x}_r \ \dot{y}_r \ \epsilon_\theta]$ , and

$$\mathcal{M} = \text{diag}\{k_1 \cos^2(e_\theta), k_1 \cos^2(e_\theta), k_1\}.$$

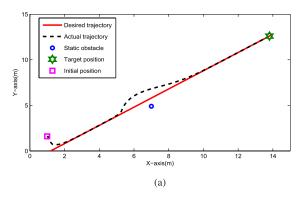
From the analysis above, we know that the tracking errors are bounded when the conditions (24) and (26) are satisfied. Thus, the stability of the error dynamics (14) is guaranteed outside the detection region. When the robot is inside the detection region, i.e.,  $d_s \leq L_{\text{rod}} < d_d$ , there exists  $\dot{x}_r = \dot{y}_r = 0$  such that inequality (25) becomes  $\dot{V}_5 \leq k_1 \cos^2(e_\theta)(E_x^2 + E_y^2) - \|e_\theta\|(k_2\|e_\theta\| - \epsilon_\theta)$ . We have that  $\dot{V}_5$  is negative if and only if

$$||e_{\theta}|| > \epsilon_{\theta} K_{2 \min}^{-1} \tag{27}$$

where  $K_{2 \, \text{min}}$  is the minimum value of  $k_2$ . As shown in [10], if  $\dot{V}$  is negative definite, then V is nonincreasing inside the detection region. Since  $\lim V_{\text{ob}} = \infty$  as  $\|\mathbf{z} - \mathbf{z}_o\| \to r^+$ , where  $\mathbf{z} = [x_i \ y_i]^T$  and  $\mathbf{z}_o = [x_o \ y_o]^T$ . Collision avoidance is guaranteed inside the detection region when the conditions (24) and (27) are satisfied.

#### V. SIMULATION RESULTS

To illustrate the effectiveness of the designed control scheme, some simulation results are implemented based on the system (5), the extended state observer (10), and nonlinear controller (14). The physical parameters of wheeled robot in Fig. 2 are set as m=10.0 kg, b=0.22 m,  $l_G=0.17$  m, r=0.05 m, and  $J_d=0.75$  kg × m<sup>2</sup>. The reference trajectory is a straight line with initial coordinates (1.2, 0.0) and orientation 45°, respectively. The desired velocity and angular are, respectively,  $v_r=0.2$  m/s and  $w_r=0.0$  rad/s and the initial velocity and angular are, respectively,  $v_o=0.35$  m/s and  $w_o=0.2$  rad/s.



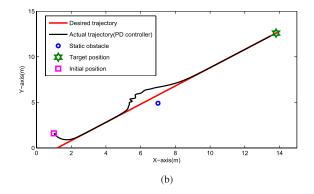
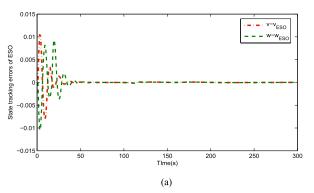


Fig. 4. Tracking and obstacle avoidance results of the mobile robot. (a) Nonlinear controller. (b) PD controller.

TABLE II
COMPARISONS OF THE OBTAINED RESULTS WITH OTHERS

References	Contributions
[11]	Adaptive control no considering external disturbances at the kinematic level
[12]	Model-reference control without external disturbances
[13]	An integrated method and path following no considering external disturbances
[14]	Binary logic controller and FLC without external disturbances
This paper	Nonlinear controller with considering external disturbances at the dynamic level



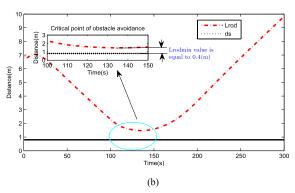
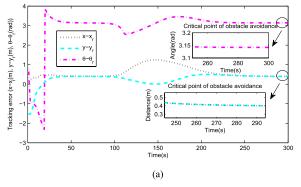


Fig. 5. State tracking errors of extended state observer and the distance between the robot and the obstacle, and the avoidance region. (a) State tracking errors. (b) Schematic diagram.



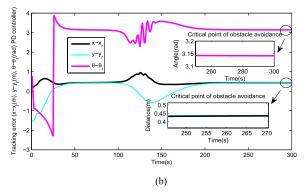
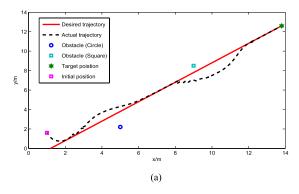


Fig. 6. Trajectory tracking errors of the mobile robot. (a) Tracking errors based on nonlinear controller. (b) Tracking errors based on PD controller.

The initial coordinates and orientation of the vehicle are (1.0, 1.5) and 60°, respectively. The parameters of observer are chosen as  $\beta_{01} = \text{diag}\{50, 50\}$  and  $\beta_{02} = \text{diag}\{135, 135\}$ . The simulation results are shown in Figs. 4–6.

The position errors based on the proposed nonlinear controller and proportion differentiation controller are shown in Table I.

Remark 3: In Fig. 7, the position, the shape, and the size of the obstacle can be arbitrarily set through changing m, n, and r according to the actual needs. In this brief, an extended state observer is introduced to estimate the unknown disturbances that are compensated in controller (15). The extended state observer has strong anti-interference ability [24], so the proposed control scheme in this brief has strong robustness.



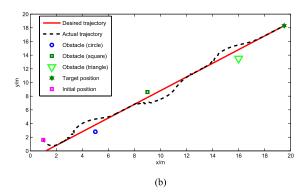


Fig. 7. Tracking and obstacle avoidance results based on nonlinear controller in complex environment. (a) Two obstacles with different shapes. (b) Three obstacles with different shapes and sizes.

Some comparison results are given in Table II. Therefore, there exists some interesting content in this brief.

#### VI. CONCLUSION

In this brief, we have proposed a nonlinear controller for tracking and obstacle avoidance of a wheeled mobile robot with nonholonomic constraint. The proposed nonlinear control scheme for a wheeled mobile robot at the dynamics level has been designed without the assumption on the prefect velocity tracking. Simulation results for the proposed scheme have been indeed feasible and effective.

#### ACKNOWLEDGMENT

The authors would like to thank the handling Associate Editor and reviewers for all their helpful and constructive comments and suggestions.

#### REFERENCES

- [1] T. C. Lee and Z.-P. Jiang, "Uniform asymptotic stability of nonlinear switched systems with an application to mobile robots," *IEEE Trans. Autom. Control*, vol. 53, no. 5, pp. 1235–1252, Jun. 2008.
- [2] P. Sun and S. Wang, "Redundant input guaranteed cost switched tracking control for omnidirectional rehabilitative training walker," *Int. J. Innov*ative Comput., Inf., Control, vol. 10, no. 3, pp. 883–895, Jun. 2014.
- [3] K. D. Do, "Formation tracking control of unicycle-type mobile robots with limited sensing ranges," *IEEE Trans. Control Syst. Technol.*, vol. 16, no. 3, pp. 527–538, May 2008.
- [4] P. Mukilan and A. Wahi, "An efficient human object detection and tracking with the aid of morphological operation and optimization algorithm," *Int. J. Innovative Comput.*, *Inf.*, *Control*, vol. 11, no. 4, pp. 1139–1153, 2015.
- [5] D. Sun, S. Hu, X. Shao, and C. Liu, "Global stability of a saturated nonlinear PID controller for robot manipulators," *IEEE Trans. Control Syst. Technol.*, vol. 17, no. 4, pp. 892–899, Jul. 2009.
- [6] R. Fierro and F. L. Lewis, "Control of a nonholonomic mobile robot using neural networks," *IEEE Trans. Neural Netw.*, vol. 9, no. 4, pp. 589–600, Jul. 1998.
- [7] B. S. Park, S. J. Yoo, J. B. Park, and Y. H. Choi, "Adaptive neural sliding mode control of nonholonomic wheeled mobile robots with model uncertainty," *IEEE Trans. Control Syst. Technol.*, vol. 17, no. 1, pp. 207–214, Jan. 2009.
- [8] M. S. Miah and W. Gueaieb, "RFID-based mobile robot trajectory tracking and point stabilization through on-line neighboring optimal control," J. Intell. Robot. Syst., vol. 78, nos. 3–4, pp. 377–399, Jun. 2015.

- [9] M. S. Miah and W. Gueaieb, "Mobile robot trajectory tracking using noisy RSS measurements: An RFID approach," ISA Trans., J. Autom., vol. 53, no. 2, pp. 433–443, Mar. 2014.
- [10] S. Wen, W. Zheng, J. Zhu, X. Li, and S. Chen, "Elman fuzzy adaptive control for obstacle avoidance of mobile robots using hybrid force/position incorporation," *IEEE Trans. Syst., Man, Cybern. C, Appl. Rev.*, vol. 42, no. 4, pp. 603–608, Jul. 2012.
- [11] M. Cui, D. Sun, W. Liu, M. Zhao, and X. Liao, "Adaptive tracking and obstacle avoidance control for mobile robots with unknown sliding," *Int. J. Adv. Robot. Syst.*, vol. 9, no. 1, pp. 1–14, 2012.
- [12] N. Uchiyama, T. Hashimoto, S. Sano, and S. Takagi, "Model-reference control approach to obstacle avoidance for a human-operated mobile robot," *IEEE Trans. Ind. Electron.*, vol. 56, no. 10, pp. 3892–3896, Oct. 2009.
- [13] A. Sgorbissa and R. Zaccaria, "Integrated obstacle avoidance and path following through a feedback control law," *J. Intell. Robot. Syst.*, vol. 72, nos. 3–4, pp. 409–428, Dec. 2013.
- [14] I. Ullah, F. Ullah, Q. Ullah, and S. Shin, "Integrated tracking and accident avoidance system for mobile robots," *Int. J. Control, Autom.*, Syst., vol. 11, no. 6, pp. 1253–1265, Dec. 2013.
- [15] M. Quoy, S. Moga, and P. Gaussier, "Dynamical neural networks for planning and low-level robot control," *IEEE Trans. Syst., Man, Cybern. A, Syst. Humans*, vol. 33, no. 4, pp. 523–532, Jul. 2003.
- [16] J. Han, Active Disturbance Rejection Control Technique: Technique for Estimating and Compensating Uncertainties. Beijing, China: National Defense Industry, 2008.
- [17] J. Han, "From PID to active disturbance rejection control," *IEEE Trans. Ind. Electron.*, vol. 56, no. 3, pp. 900–906, Mar. 2009.
- [18] H.-L. Xing, J.-H. Jeon, K. C. Park, and I.-K. Oh, "Active disturbance rejection control for precise position tracking of ionic polymer–metal composite actuators," *IEEE/ASME Trans. Mechatronics*, vol. 18, no. 1, pp. 86–95, Feb. 2013.
- [19] W. Wang and Z. Gao, "A comparison study of advanced state observer design techniques," in *Proc. Amer. Control Conf.*, Jun. 2003, pp. 4754–4759.
- [20] Z. Zhu, D. Xu, J. Liu, and Y. Xia, "Missile guidance law based on extended state observer," *IEEE Trans. Ind. Electron.*, vol. 60, no. 12, pp. 5882–5891, Dec. 2013.
- [21] H. Liu and S. Li, "Speed control for PMSM servo system using predictive functional control and extended state observer," *IEEE Trans. Ind. Electron.*, vol. 59, no. 2, pp. 1171–1183, Feb. 2012.
- [22] Y. Xia, P. Shi, G.-P. Liu, J. Han, and D. Rees, "Active disturbance rejection control for uncertain multivariable systems with timedelay," *IET Control Theory Appl.*, vol. 1, no. 1, pp. 75–81, Jan. 2007.
- [23] Q. Zheng, L. Dong, D. H. Lee, and Z. Gao, "Active disturbance rejection control for MEMS gyroscopes," *IEEE Trans. Control Syst. Technol.*, vol. 17, no. 6, pp. 1432–1438, Nov. 2009.
- [24] X. Yang and Y. Huang, "Capabilities of extended state observer for estimating uncertainties," in *Proc. Amer. Control Conf.*, Jun. 2009, pp. 3700–3705.