

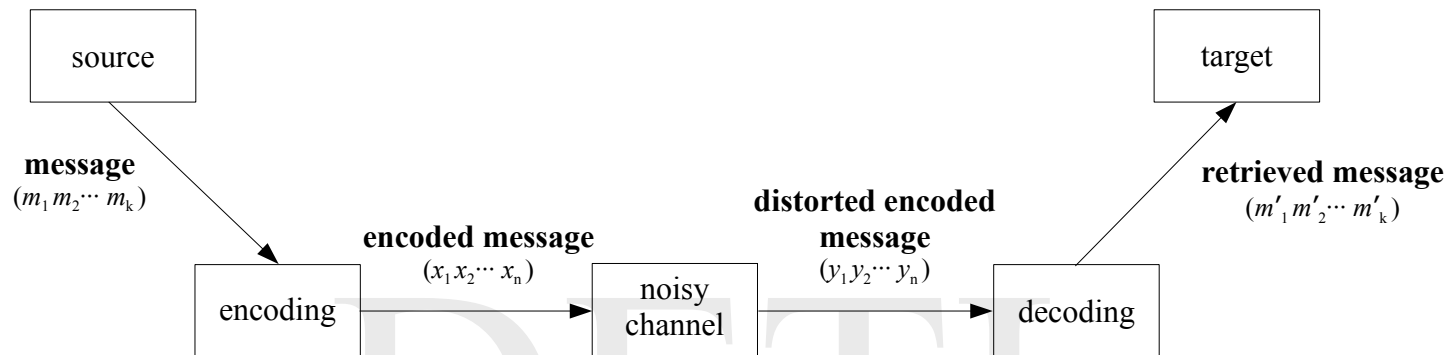


# *Arquitecturas de Alto Desempenho*

*Design Principles for Hamming Codes*

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# Theoretical Background - 1



A message  $\mathbf{m}$ , generated by a given *source*, is expressed in  $k$  symbols of the alphabet  $\Sigma$  and is further encoded into a word  $\mathbf{x}$ , using a *block code*, through expansion to length  $n$  by addition of redundant information; that is, a specific code word is built by the interspersing of symbols of the alphabet  $\Sigma$  to the symbols of the message, following some definite rule.

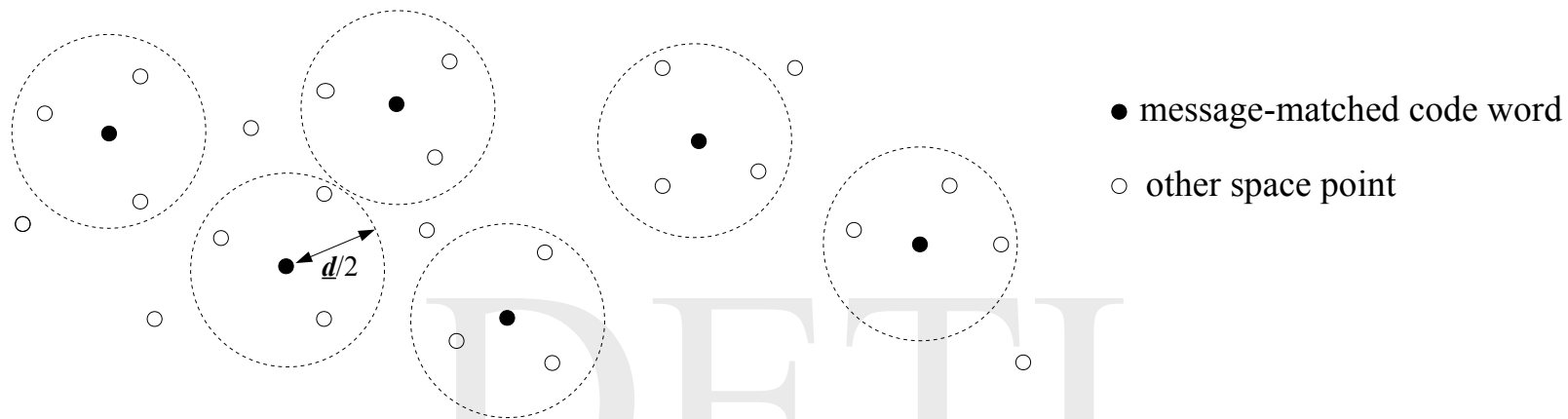
The encoded message  $\mathbf{x}$  is next transmitted over a *noisy channel*, where the symbols may be changed according to certain probabilities that are characteristic of the channel. The received message  $\mathbf{y}$  is finally decoded into the message  $\mathbf{m}'$ , which is retrieved by the *target*. Bear in mind that the *channel* concept must be taken in a broad sense: it can be a data transmitting medium in the traditional sense, such as copper, optical fiber, or air signal propagation, or a data storage device, such as a hard disk, a flash memory, or a dynamic RAM.

## *Theoretical Background - 2*

One can define the *information rate*  $R$ , which measures the slowdown of the effective data transmission, as  $k/n$ . On the other hand, given the channel characteristics, one defines the *capacity*  $C$  of the channel as something which, as Claude Shannon has shown, has the property that, for  $R < C$ , it is possible to find an *encoding/decoding* scheme such that the probability that  $\mathbf{m} \neq \mathbf{m}'$  can be made arbitrarily small. If, however,  $R > C$ , no such scheme exists.

Claude Shannon, however, did not show how to build such *encoding/decoding* schemes. This has been the pioneering work of Richard Hamming. Presently, however, the case for the best codes in terms of the maximal number of errors that one can correct for a given information rate and code length is not clear. Existence theorems are known, but the exact bound is still an open problem.

## Theoretical Background - 3



To understand how one can achieve an *error-correcting* code by increasing the number of symbols in the code word relative to the message word, one may think of each code word as a point in a  $n$ -dimensional space where a *metric* is defined, that is, a real-valued function  $d(\mathbf{x}, \mathbf{y})$  which, given two arbitrary space points,  $\mathbf{x}$  and  $\mathbf{y}$ , computes the distance between them.

Now, if one can find an *encoding* scheme which matches each message word to code words that are as widely-spaced as possible within the  $n$ -dimensional space, then all space points in the neighborhood of radius  $\underline{d}/2$  of each message-matched code word, where  $\underline{d}$  represents the minimum value of the distance between two arbitrary message-matched code words, can be *decoded* as matching the corresponding message word through the use of a minimum distance criterion.

## *Hamming codes - 1*

Hamming codes are able to detect and correct one symbol of the received message. The metric,  $d(\mathbf{x}, \mathbf{y})$  is here defined as the number of positions where the symbols of the code words  $\mathbf{x}$  and  $\mathbf{y}$  differ, the so-called *Hamming distance*.

In the binary case, the alphabet  $\Sigma$  is the set  $\{0, 1\}$  and the space of code words is  $F_2^n$ , where  $F_2$  is the Galois field of two elements. By taking the number of redundant symbols, *parity bits*, to be  $r$ , one gets

$$\begin{aligned} n &= 2^r - 1 && \text{for the code length} \\ k &= n - r && \text{for the message length} \end{aligned}$$

giving rise to a  $[n, k]$  block code, with a minimum distance between code words of 3.

## *Hamming codes - 2*

The encoding process is described by

$$\mathbf{x} = \mathbf{m} \cdot \mathbf{G}$$

where  $\mathbf{m}$  is the message,  $\mathbf{x}$  its code word and  $\mathbf{G}$  the  $k \times n$  code generating matrix, defined as

$$\mathbf{G} = \left\| \mathbf{I}_k \mid -\mathbf{A}^T \right\|.$$

The *decoding* process, on the other hand, is described by

$$\mathbf{H} \cdot \mathbf{y}^T = \mathbf{0} \Rightarrow \text{message bits of } \mathbf{y} \text{ are correct (there is no error)}$$

$$\mathbf{H} \cdot \mathbf{y}^T \neq \mathbf{0} \Rightarrow \text{error on bit whose position is the column number whose contents is } \mathbf{H} \cdot \mathbf{y}^T$$

where  $\mathbf{y}$  is the received encoded message and  $\mathbf{H}$  the  $r \times n$  parity check matrix, defined as


$$\mathbf{H} = \left\| \mathbf{A} \mid \mathbf{I}_r \right\|$$

and has as columns all the pairwise linear independent vectors of length  $r$ .


## *Computation of a parity-check matrix for [15,11]*

$$\mathbf{H} = \left\| \mathbf{A} \mid \mathbf{I}_4 \right\| =$$

$$= \left\| \begin{array}{cccccccccc} 1 & 1 & 1 & 0 & 0 & 0 & 1 & 1 & 1 & 0 & 1 \\ 1 & 0 & 0 & 1 & 1 & 0 & 1 & 1 & 0 & 1 & 1 \\ 0 & 1 & 0 & 1 & 0 & 1 & 1 & 0 & 1 & 1 & 1 \\ 0 & 0 & 1 & 0 & 1 & 1 & 0 & 1 & 1 & 1 & 1 \end{array} \right\| \begin{array}{cccc} 1 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 \\ 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \end{array} \right\|$$



$\mathbf{A}$



$\mathbf{I}_4$

## *Computation of the associated code generating matrix*

$$\mathbf{G} = \left\| \mathbf{I}_{11} \mid -\mathbf{A}^T \right\| =$$

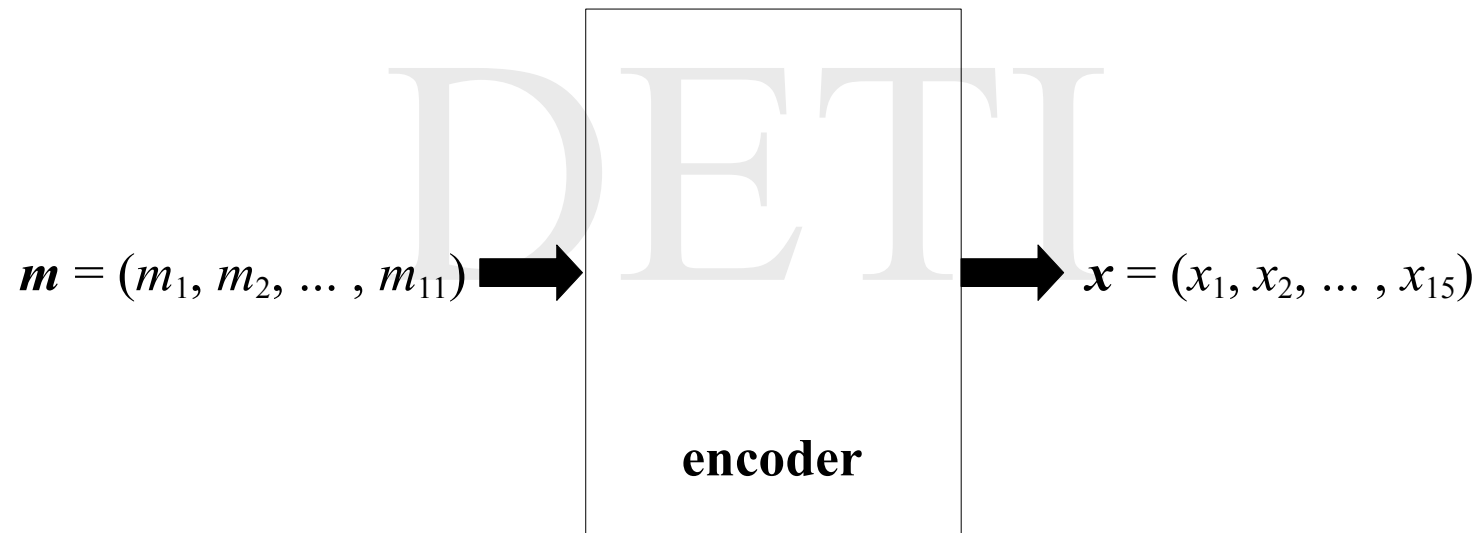
$$= \left\| \begin{array}{cccccccccccc|cccc} 1 & 0 & 0 & 0 & 0 & 0 & 0 & 0 & 0 & 0 & 0 & 1 & 1 & 0 & 0 \\ 0 & 1 & 0 & 0 & 0 & 0 & 0 & 0 & 0 & 0 & 0 & 1 & 0 & 1 & 0 \\ 0 & 0 & 1 & 0 & 0 & 0 & 0 & 0 & 0 & 0 & 0 & 1 & 0 & 0 & 1 \\ 0 & 0 & 0 & 1 & 0 & 0 & 0 & 0 & 0 & 0 & 0 & 0 & 1 & 1 & 0 \\ 0 & 0 & 0 & 0 & 1 & 0 & 0 & 0 & 0 & 0 & 0 & 0 & 1 & 0 & 1 \\ 0 & 0 & 0 & 0 & 0 & 1 & 0 & 0 & 0 & 0 & 0 & 0 & 0 & 1 & 1 \\ 0 & 0 & 0 & 0 & 0 & 0 & 1 & 0 & 0 & 0 & 0 & 1 & 1 & 1 & 0 \\ 0 & 0 & 0 & 0 & 0 & 0 & 0 & 1 & 0 & 0 & 0 & 1 & 1 & 0 & 1 \\ 0 & 0 & 0 & 0 & 0 & 0 & 0 & 0 & 1 & 0 & 0 & 1 & 0 & 1 & 1 \\ 0 & 0 & 0 & 0 & 0 & 0 & 0 & 0 & 0 & 1 & 0 & 0 & 1 & 1 & 1 \\ 0 & 0 & 0 & 0 & 0 & 0 & 0 & 0 & 0 & 0 & 1 & 1 & 1 & 1 & 1 \end{array} \right\|$$

$\underbrace{\hspace{15em}}_{\mathbf{I}_{11}}$

$\underbrace{\hspace{5em}}_{-\mathbf{A}^T}$



## *Parallel implementation of the encoder for [15,11] - 1*



## *Parallel implementation of the encoder for [15,11] - 2*

### **Straightforward implementation**

$\mathbf{x} = \mathbf{m} \times \mathbf{G}$  , where matrix  $\mathbf{G}$  is fixed

$$x_i = m_i \text{ , for } i = 1, 2, \dots, 11$$

$$x_{12} = m_1 \oplus m_2 \oplus m_3 \oplus m_7 \oplus m_8 \oplus m_9 \oplus m_{11}$$

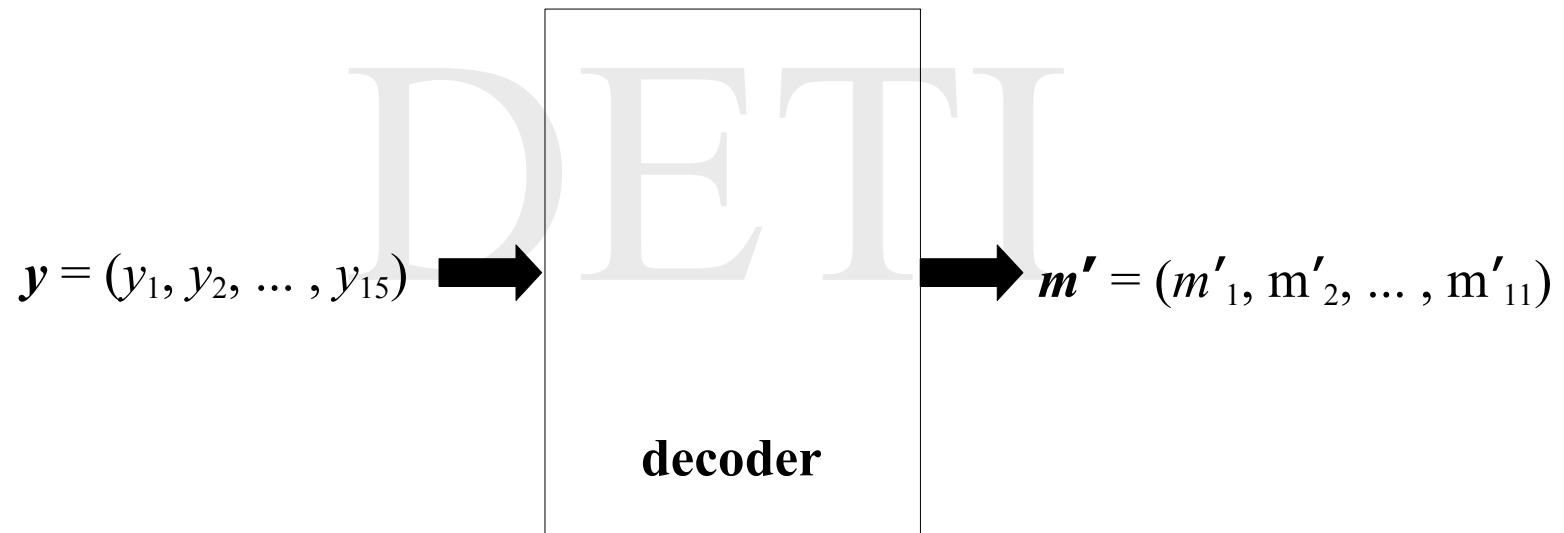
$$x_{13} = m_1 \oplus m_4 \oplus m_5 \oplus m_7 \oplus m_8 \oplus m_{10} \oplus m_{11}$$

$$x_{14} = m_2 \oplus m_4 \oplus m_6 \oplus m_7 \oplus m_9 \oplus m_{10} \oplus m_{11}$$

$$x_{15} = m_3 \oplus m_5 \oplus m_6 \oplus m_8 \oplus m_9 \oplus m_{10} \oplus m_{11}$$

- 24 x-ors are needed
- 6 x-or propagation time delays in the worst case.

## *Parallel implementation of the decoder for [15,11] - 1*



## *Parallel implementation of the decoder for [15,11] - 2*

### **Straightforward implementation of the error detecting part**

$$\mathbf{p} = \mathbf{H} \times \mathbf{y}^T, \text{ where matrix } \mathbf{H} \text{ is fixed}$$

$$p_1 = y_1 \oplus y_2 \oplus y_3 \oplus y_7 \oplus y_8 \oplus y_9 \oplus y_{11} \oplus y_{12}$$

$$p_2 = y_1 \oplus y_4 \oplus y_5 \oplus y_7 \oplus y_8 \oplus y_{10} \oplus y_{11} \oplus y_{13}$$

$$p_3 = y_2 \oplus y_4 \oplus y_6 \oplus y_7 \oplus y_9 \oplus y_{10} \oplus y_{11} \oplus y_{14}$$

$$p_4 = y_3 \oplus y_5 \oplus y_6 \oplus y_8 \oplus y_9 \oplus y_{10} \oplus y_{11} \oplus y_{15}$$

- 28 x-ors are needed
- 7 x-or propagation time delays in the worst case.

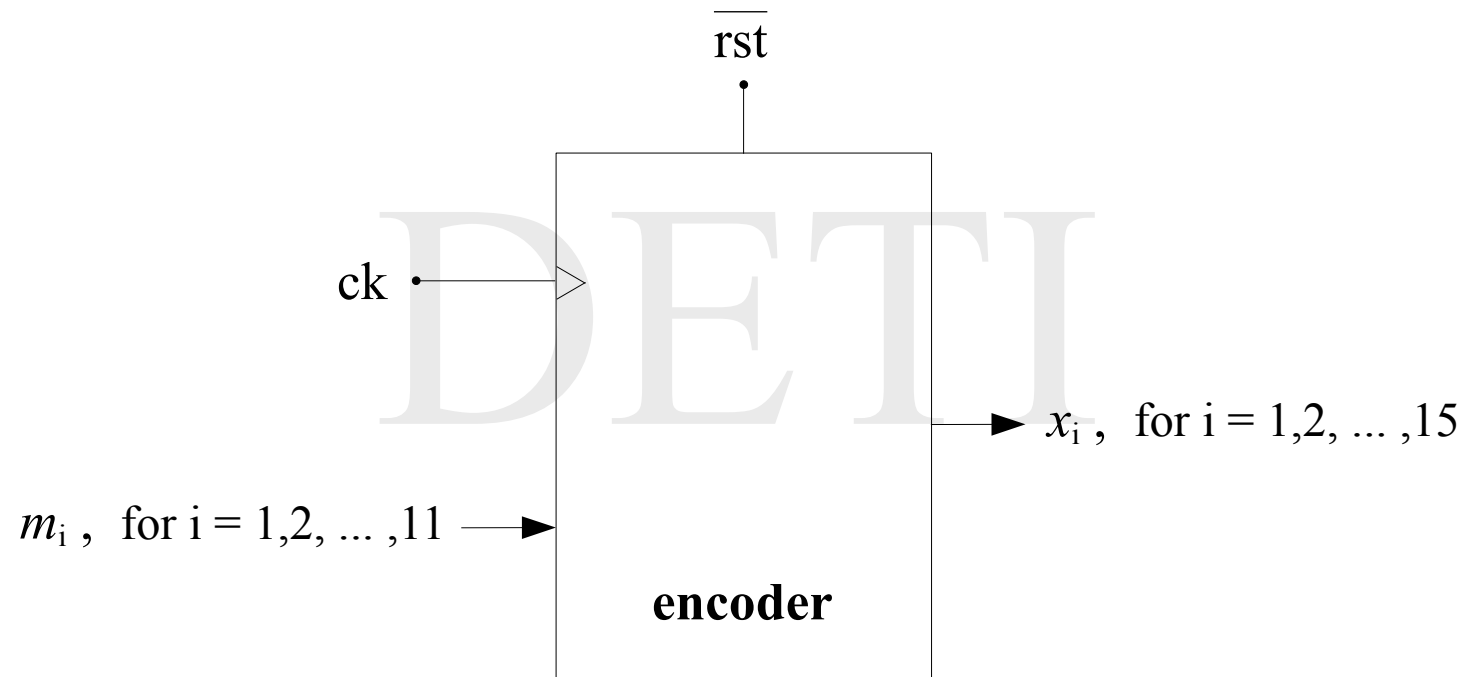
## *Parallel implementation of the decoder for [15,11] - 3*

### **Implementation of the error correcting part**

$$\mathbf{p} = \mathbf{H} \times \mathbf{y}^T \neq 0 \Rightarrow \exists_k \, ml_k = \overline{y_k}$$

$$ml_i = sel_i(\mathbf{p}) \oplus y_i, \text{ for } i = 1, 2, \dots, 11$$

## *Bit-serial implementation of the encoder for [15,11]*



## *Bit-serial implementation of the decoder for [15,11]*

