

Homework #9 of 「類比積體電路導論」

作業繳交截止日期: **Dec. 17, 2024 12:00** (上傳E3數位平台繳交)

本次作業共兩大題, 9.1~9.2

請將作業轉成一個 PDF 檔案(file size 小於 10MB), 檔名請使用

「AIC_HW9_自己的學號」(例如: AIC_HW9_109700018), 於作業繳

交截止日期/時間前, 上傳到指定的E3 數位平台繳交。

Unless otherwise stated, in the following problems, use the device data shown in Table 1 and assume that $V_{DD}=3V$. The Dielectric constant of gate oxide is 3.9 and $\epsilon_0=8.854\times 10^{-12}F/m$.

Table 1. Level 1 SPICE models for NMOS and PMOS devices.

NMOS Model			
LEVEL = 1	VTO = 0.7	GAMMA = 0.45	PHI = 0.9
NSUB = 9e+14	LD = 0.08e-6	UO = 350	LAMBDA = 0.1
TOX = 9e-9	PB = 0.9	CJ = 0.56e-3	CJSW = 0.35e-11
MJ = 0.45	MJSW = 0.2	CGDO = 0.4e-9	JS = 1.0e-8
PMOS Model			
LEVEL = 1	VTO = -0.8	GAMMA = 0.4	PHI = 0.8
NSUB = 5e+14	LD = 0.09e-6	UO = 100	LAMBDA = 0.2
TOX = 9e-9	PB = 0.9	CJ = 0.94e-3	CJSW = 0.32e-11
MJ = 0.5	MJSW = 0.3	CGDO = 0.3e-9	JS = 0.5e-8

VTO: threshold voltage with zero V_{SB} (unit: V)

GAMMA: body-effect coefficient (unit: $V^{1/2}$)

PHI: $2\Phi_F$ (unit: V)

TOX: gate-oxide thickness (unit: m)

NSUB: substrate doping (unit: cm^{-3})

LD: source/drain side diffusion (unit: m)

UO: channel mobility (unit: $cm^2/V/s$)

LAMBDA: channel-length modulation coefficient (unit: V^{-1})

CJ: source/drain bottom-plate junction capacitance per unit area (unit: F/m²)

CJSW: source/drain sidewall junction capacitance per unit length (unit: F/m)

PB: source/drain junction built-in potential (unit: V)

MJ: exponent in CJ equation (unitless)

MJSW: exponent in CJSW equation (unitless)

CGDO: gate-drain overlap capacitance per unit width (unit: F/m)

CGSO: gate-source overlap capacitance per unit width (unit: F/m)

JS: source/drain leakage current per unit area (unit: A/m²)

9.1 The circuit of Fig.1 is designed with $2 * \left(\frac{W}{L}\right)_3 = 4 * K * \left(\frac{W}{L}\right)_4 = K * \left(\frac{W}{L}\right)_5$,
 $R_1 = 1.5k\Omega$, and $R_2 = 2k\Omega$. Assume that $\lambda = \gamma = 0$. Q_1 , Q_2 , and Q_3 are identical.
Determine K such the V_{out} has a zero TC. Use the parameters listed below.

$$\frac{\partial V_{BE}}{\partial T} = -1.5mV / K \text{ and } \frac{\partial V_T}{\partial T} \approx +0.087mV / K$$

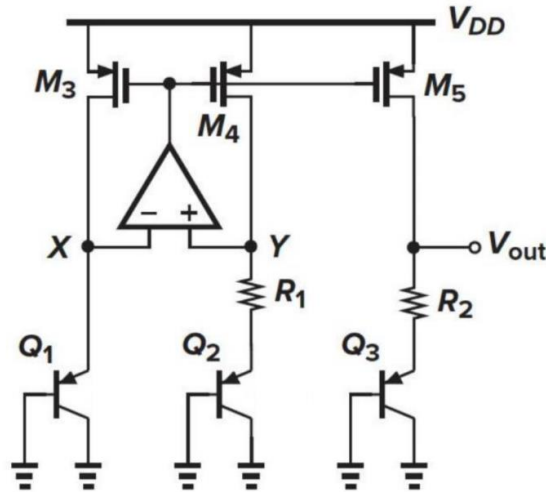


Fig.1

9.2 Consider the bandgap reference shown below. The opamp is ideal. For Q_1 and Q_2 ,
 $\beta_F = \infty$, $V_A = \infty$, $I_{S2} = 6 \text{ pA} = 6I_{S1}$. At $T = T_0$, let $V_T = kT_0/q = 26 \text{ mV}$,
 $dV_T/dT = +0.087 \text{ mV}/^\circ\text{C}$, and $d|V_{BE}|/dT = -1.74 \text{ mV}/^\circ\text{C}$.
(a) Let $V_{os} = 0 \text{ mV}$, $R_1 = R_2 = R$. Find R to make $dV_o/dT = 0$ at $T = T_0$.
(b) Let $V_{os} = 20 \text{ mV}$. Calculate V_o in terms of V_{BE2} . Use the R found in (a).

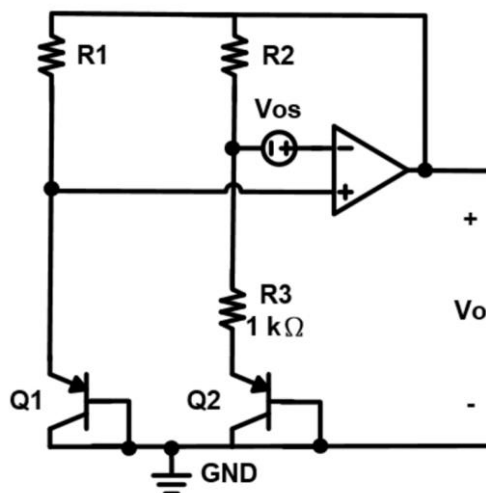


Fig.2