

Scaling of Electrode-Electrolyte Interface Model Parameters In Phosphate Buffered Saline

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Abstract—We report how the impedance presented by a platinum electrode scales with the concentration of phosphate buffered saline (PBS). We measure the response in various dilutions of PBS with an electrode array as is commonly used in spinal cord stimulator (SCS) implants. We match the parameters of a non-linear electrode-electrolyte interface model to these measurements. We find that the constant phase element of the model scales with approximately the log of concentration, whereas the resistivity is inversely proportional. Using a novel DC measurement technique we show that the onset of Faradaic conduction for a platinum electrode, and thus the safe exposure limit, does not scale with concentration. We compare objective measurements made in saline to those made in the spinal cavity of live sheep. We comment upon the appropriateness of using PBS as a substitute for *in-vivo* measurements.

Index Terms—Bioelectric phenomena, bioimpedance, biomedical electrodes, biomedical measurements, biophysics, electrical stimulation, implantable biomedical devices.

I. INTRODUCTION

THERE is considerable interest in the electrical modelling of electrodes. [1]–[4] In [5] a linearised model was presented, and in [6] a non-linear model suitable for use with the SPICE circuit simulator was presented. The SPICE model usefully characterises an electrode in a given electrolyte with a small number of parameters. One major reason for the interest in the electrical impedance of electrodes concerns the design of electronics intended for integration into pacemakers, cochlear implants, spinal cord stimulators, etc. The design of successful circuits depends upon a good understanding of external load impedance, while maximisation of battery life is linked to the use of electrodes whose impedance is well understood.

Having a compact model of an electrode in the appropriate electrolyte allows circuit designers to simulate their designs with valid loads. For example, the authors of [7] are concerned with the impedance presented by their electrode loads, and calculate power efficiency of their circuits. This value is affected by the load impedance assumed. Nevertheless, it is clear that they do not have a good idea of, nor a good way to simulate with, the load truly presented by an electrode *in vivo* or *in*

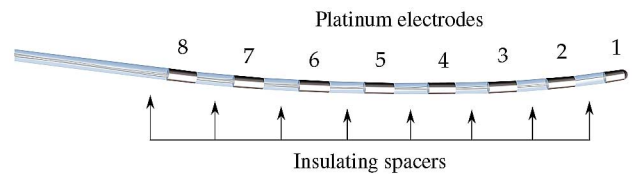


Fig. 1. St. Jude Medical Octrode lead comprised of eight coaxial cylindrical platinum electrodes (1 mm in diameter, 3 mm long) separated by 4 mm insulating spacers.

vitro. Knowing the effect upon the model of changes in the electrode or electrolyte would allow designers to anticipate circumstances that will affect the load seen by their circuits.

Another appeal of a compact model stems from the fact that electrode characteristics are succinctly and objectively represented by the model parameters. These parameters enable direct comparison of different electrodes, or the change in electrode properties over time. For example, in [8] changes in chronically-implanted electrodes are observed but there is no standard quantitative way of presenting the changes. Similarly, while there is a good understanding of what represents safe exposure, a compact model with parameters permits the prediction of the safety of any given stimulus regime. [9]

Electrodes in the laboratory are typically tested in a saline solution selected to mimic the circumstances in which they will operate when implanted. In an implanted setting there is no reference electrode and many of the well established electrochemical measurement regimes do not apply. Here we measure the behaviour in an electrode-electrolyte situation as it would be seen from an implant device.

The electrode used in this work is a commercial linear array of eight platinum electrodes, called an “Octrode”, intended for Spinal Cord Stimulator (SCS) implantation. [10] A picture of an Octrode is presented in Fig. 1. A solution of phosphate buffered saline (PBS) diluted to one-tenth concentration ($0.1 \times$) is commonly used as a crude phantom in the case of SCS, while full concentration ($1.0 \times$) PBS is considered more representative of the situation in blood.

Six solutions ranging from $1.0 \times$ to $0.025 \times$ the concentration (by mass) of a stock PBS solution have been examined. The ingredients of that stock solution are given in Table I.

In this manuscript we use an adapted version of the compact non-linear model presented in [6]. The model is compact in that it is suitable for entry into freely available electrical circuit simulation software. The schematic of the electrode-electrolyte interface model that we have used is presented as Fig. 2.

The electrical interface models of [5] and [6] have two parts, a displacement branch and a Faradaic branch. Scott and Single

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TABLE I
PBS STOCK SOLUTION INGREDIENTS

Ingredient	Quantity
$NaCl$	8.0 g
KCl	0.2 g
Na_2HPO_4	1.44 g
KH_2PO_4	0.24 g
Distilled Water	1.0 L

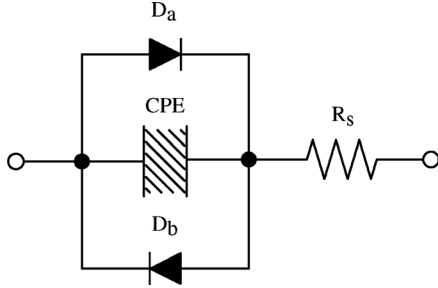


Fig. 2. Electrical schematic of the electrode-interface impedance model used in this work. D_a and D_b represent diodes; CPE and R_s represent the constant phase element and series resistance respectively.

incorporate memristors into their model, used to mimic a diffusion limited current conduction mode, i.e., their purpose is to throttle electrical current associated with Faradaic reactions. In an implanted setting an electrode should never be operated in a Faradaic conduction mode and certainly never to the point of becoming diffusion limited; for this reason, memristors have been omitted.

It has been suggested that the primary coupling mechanism between platinum electrodes and physiological saline is the reversible evolution of gas, probably oxygen, at the platinum surface, which dissolves without bubble formation. [11] Understanding the exact mechanism of displacement or Faradaic charge transfer at the electrode surface is beyond the scope of this work. Rather, we wish to characterise the electrode-electrolyte interface using a small number of parameters that can be fed into an electrical model.

The remainder of this manuscript is broken into five sections. Section II details the measurement of the inter-electrode resistances (arising from the bulk resistance of the solution) and the generation of a resistor network to replicate these resistances. Sections III and IV are concerned with the measurement and parameter fitting of displacement currents and Faradaic currents, respectively, across the electrolyte interface. Section V compares the results obtained for PBS with measurements made *in-vivo*. Finally, in Section VI we summarise our findings and give concluding comments.

II. BULK RESISTIVITY/RESISTOR NETWORK

The electrical impedance between two electrodes in an electrolyte arises from two interface impedances in series with the resistance due to the bulk of the electrolyte itself. To quantify the impedance presented by a single interface we first need to determine inter-electrode resistance presented by the electrolyte bulk. Finding these resistances allows us to quantify the series resistance of the interface (R_s) via subtraction. As there are always two interface impedances between any two electrodes, inter-electrode resistances cannot be measured in isolation.

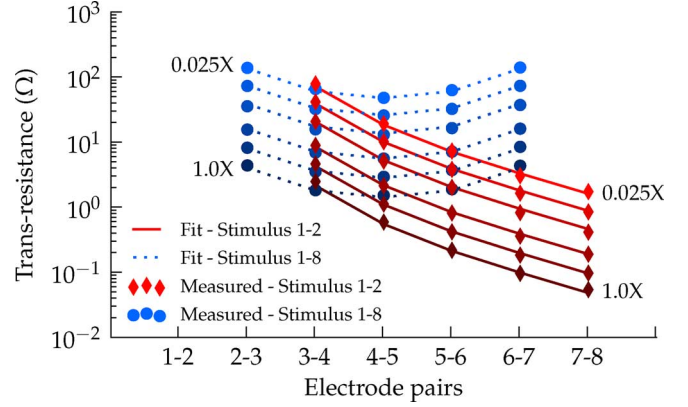


Fig. 3. Transresistance measurements (markers) used to generate the resistor mesh and the corresponding fit (lines). Red diamonds indicate results where a stimulus is placed across electrodes 1 and 2. Blue circles represent measurements where the stimulus is placed across electrodes 1 and 8. Each trace represents one of six concentrations of PBS, increasing in transresistance magnitude monotonically from $1.0 \times$ to $0.025 \times$.

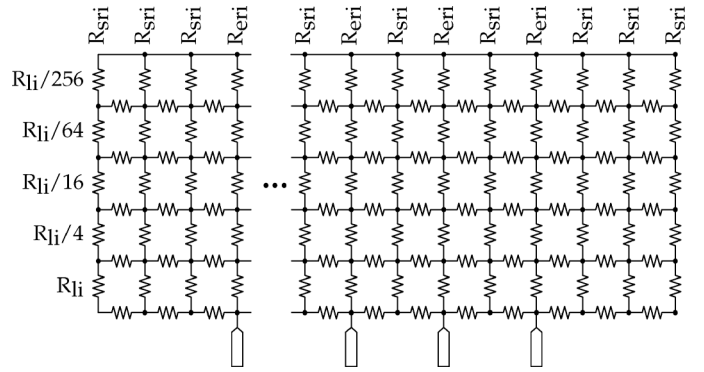


Fig. 4. Layout of the generated resistor mesh used to connect each of the interface models. The final mesh for an 8 electrode array contains 205 resistors.

In [6], a series of transresistance measurements were used in conjunction with a physical description of the electrode geometry to build up a representative network of resistors. That resistor network was defined using five parameters: R_{eri} , R_{sri} , R_{li} , mesh depth, and edge depth. Readers are referred to [6] for the interpretation of these model parameters. Repeating that work, we created a network of resistors that mimic the resistances due to the solution's bulk conductivity.

The mesh depth and padding values of five columns and three rows, respectively, has been taken directly from [6]. These parameters, together with an array of 8 electrodes, leads to a generated network of 205 resistors, the layout of which is presented in Fig. 4.

A numerical fit has been made to the two independent scaling parameters R_{eri} and R_{sri} .¹ The resulting fit and parameter values are presented in Fig. 3 and Table II.

Measurements of the inter-electrode impedance were made using a Tektronix TPS2014 DSO with four fully isolated and floating channels; an Agilent 33220A Function Generator; and a desktop PC running Python 2.7 for instrumentation control and

¹The values of R_{eri} and R_{sri} are defined by the lengths of the electrode and inter-electrode spacers respectively as described in [6]. The two parameters are proportional to each other by the ratio of their lengths. Hence, we define R_{sri} to be 3/4 that of R_{eri} making it a dependant parameter.

TABLE II
RESISTOR MESH PARAMETERS. ELECTROLYTE CONDUCTIVITY (σ) IS
EXPRESSED IN UNITS OF S/cm

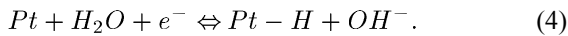
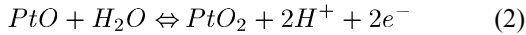
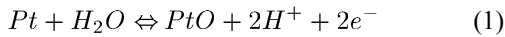
Parameter	Value
R_{eri} (Ω)	$0.407 / \sigma$
R_{sri} (Ω)	$R_{eri} \cdot 3/4$
R_{li} (Ω)	$3.71 / \sigma$
Depth (layers)	5
Padding (layers)	3

result processing. The pH level of the stock solution was measured using a calibrated EDT Instruments BA-350 pH meter. It had a pH of 7.4 and no pH adjustments were made to the derived solutions. The conductivity of solutions was measured using an EDT Instruments RE 388 Tx conductivity meter.

To be clear, the resistive mesh created in this step models *only* the resistivity of the solution bulk. No contribution from the interface itself appears in these measurements. This measurement is possible because voltages are always measured across pairs of non-driven electrodes and the measurements are of suitably high impedance. We determine the resistive component of the interface, R_S , in the following section.

III. DISPLACEMENT PARAMETERS/CPE

Displacement currents are responsible for capacitive behaviour at interfacial boundaries. They are brought about by the redistribution of charge in response to applied fields, e.g., a charged electrode will attract or repel Cl^- from its surface and reorientate polar molecules in the surrounding solution. [9] This capacitive behaviour may also be the result of electrode polarisation in the form of reversible Faradaic reactions at the surface of the electrode. Such reactions involving water and platinum, as identified in [9], [12], [13], include



Displacement currents are capable of transferring charge between electrical and biological systems without damaging electrodes or tissue. [13] This behaviour is modelled by way of a constant phase element (CPE), also known as a fractional capacitor. For simulation purposes this element is realised as an array of R-C branches [6], [14], [15] where the value of each R-C pair is chosen to set the slope of impedance relative to frequency. The placement of R-C pairs within the CPE is determined by Morrison's parameters m and k , where k determines the spacing, or density, of R-C branches and m determines the resulting slope. R-C pairs were generated using a Python script to span 1 nHz – 1 GHz with a spacing of 3 pairs per decade. The CPE contains 55 branches, i.e., 110 electrical elements are used to represent the CPE.

We measure displacement currents by sweeping the frequency of a sinusoidal stimulus across electrodes 2 and 7 of the Octrode while measuring the voltage between electrodes 2 and 3 (numbering shown in Fig. 1). This allows for the

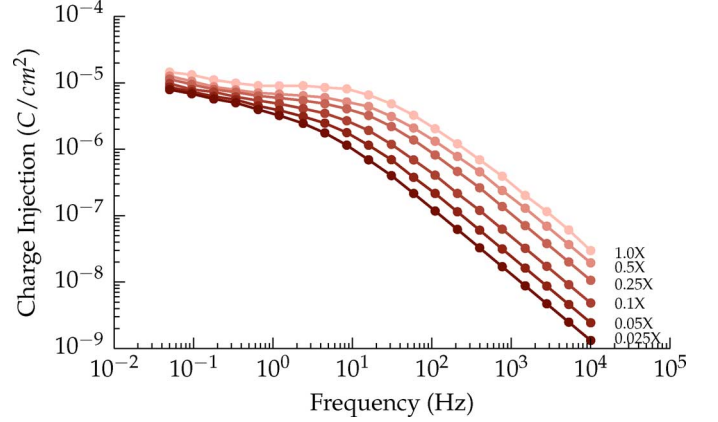


Fig. 5. Per phase charge injected during CPE measurement in each concentrations of PBS, calculated with an effective surface area of 14 mm^2 .

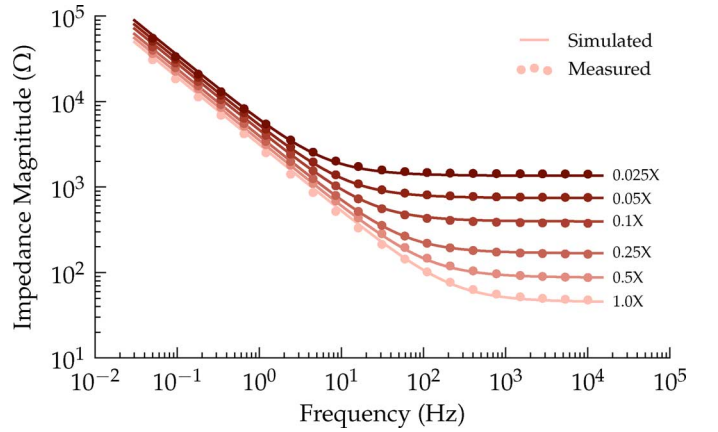


Fig. 6. Magnitude of interfacial impedance between electrodes 2 and 7 against the frequency of a sine wave stimulus for six concentrations of PBS. The slope of the CPE is visible the left, whereas the resistance due to R_{S+bulk} appears on the right. Traces represent simulated data while markers represent measured values.

measurement of a single interface impedance in series with the resistance presented by the bulk solution.

Current density during measurements, calculated using an effective surface area of 14 mm^2 , peaked at $161 \mu A/cm^2$ at 10 kHz in $1.0 \times$ PBS and fell as low as 197 nA/cm^2 at 50 mHz in $0.025 \times$ PBS. Charge injection per phase is shown in Fig. 5 for each solution. The stimulus waveform was set at 300 mV (peak) at each measurement point. The measurement instruments used are the same as those used to measure the inter-electrode resistance in the previous section.

Fig. 6 shows the magnitude of the measured response for each concentration of PBS accompanied by simulation results. Fig. 7 shows the phase response for the same measurements, again with the simulated response at each concentration. The simulated responses are generated using fitted parameter values for m , k , $|Z|$ @ 1 Hz, and R_S , which are presented in Table III.

As shown in the interface model schematic (Fig. 2), the interface contains its own internal series resistance (R_S). The resistance seen in series with the CPE will therefore be the sum of both the resistance due to the bulk resistivity of the fluid and R_S , to which we will refer as R_{S+bulk} . In Fig. 6, the slope and magnitude of the CPE are visible below 1 Hz whereas R_{S+bulk}

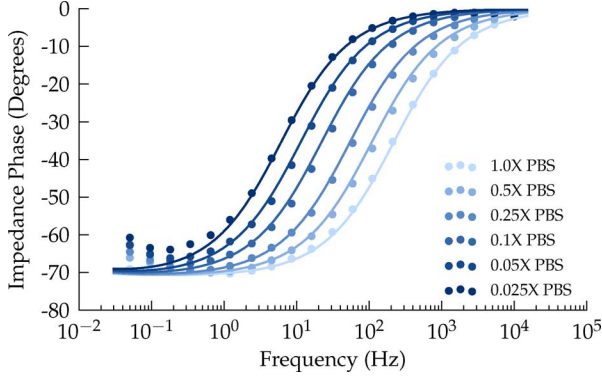


Fig. 7. Phase data of interfacial impedance across electrodes 2 and 7 against frequency of a sine wave stimulus. Traces represent simulated data while markers represent measured values.

TABLE III
DISPLACEMENT PARAMETER SCALING. *conc* REFERS TO THE MASS CONCENTRATION OF PBS

Parameter	Value
m	1.34
k	1.773
$ Z @ 1Hz (\Omega)$	$3284 \times conc^{-0.158}$
$R_S (\Omega)$	$13.38 \times conc^{-0.8397}$

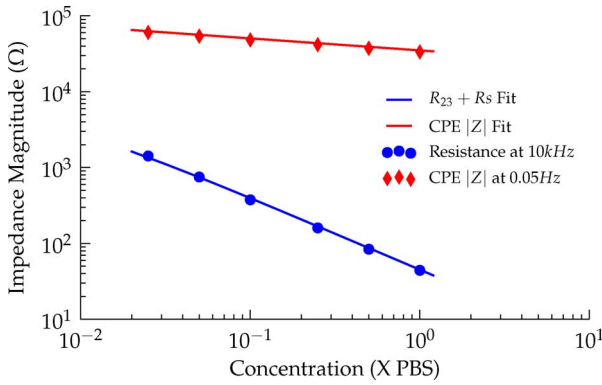


Fig. 8. Simulated and measured values of CPE parameters versus concentration of PBS. $|Z|$ is compared at 0.05 Hz as measured data is unaffected by R_{S+bulk} at this frequency.

dominates above 1 kHz; it is clear that the two do not scale similarly. As the impedance of the CPE and R_{S+bulk} are separable with frequency, these measurements can be used to determine the CPE parameters and R_S .

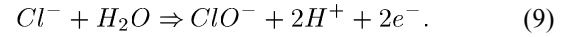
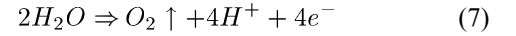
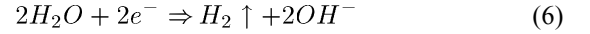
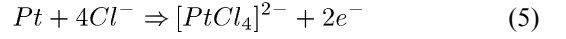
Fig. 8 shows measured data and the corresponding fits for both the CPE magnitude at 0.05 Hz and R_{S+bulk} . Here we express the CPE offset parameter as $|Z| @ 1 Hz$, but the fits used $|Z| @ 0.05 Hz$, as data points at this frequency are outside the transitional zone where the impedance of the CPE and that of R_{S+bulk} overlap.

The data presented in Fig. 6 shows a large, linear, relative shift of $|Z|$ with salinity at higher frequencies, where the resistive component of the solution dominates. Conversely, at lower frequencies where $|Z|$ changes with frequency, the relative change is closer to being logarithmic. It is this region that is modelled by the CPE. This suggests that the CPE does not chiefly rely on the added salt ions, but they do have some effect. Instead,

the constant-phase effect must arise from phenomena not associated with the salt ions.

IV. FARADAIC PARAMETERS/DIODES

With parameters fitted for the scaling of both the CPE and R_S with PBS concentration, we turn to Faradaic conduction. Our model uses two reverse-connected diodes to represent Faradaic current conduction between electrode and electrolyte. Specifically, we use the diodes to model irreversible Faradaic reactions. (Any reversible reactions, those producing reactants bound to the surface of the electrode, are encapsulated by the displacement branch of the model.) Possible irreversible Faradaic reactions for platinum electrodes (each with differing reaction potentials) in saline are



We use a diode to model the electrical current conduction caused by one reaction proceeding in one direction, i.e., a diode is the circuit representation of the Butler-Volmer equation for a single reaction proceeding in a single direction. We wish to model the onset of Faradaic current as a way of knowing that an electrode has been pushed “too far”. It is expected that the situation *in-vivo* will be much more complex and that fitted diode parameters may encapsulate multiple reactions if they occur close together.

To measure Faradaic conduction we must separate its contribution from any displacement behaviour. By using a DC measurement technique, as opposed to cyclic measurements, we are able to separate the effects in time. Like a conventional capacitor, the CPE responds to a change in voltage by absorbing or ejecting charge, causing a spike in current. The CPE will draw negligible current once it has settled and in the case of our model we assume the remaining current conduction to be the result of Faradaic processes.

The equation governing current conduction in a diode where the forward voltage across the diode, V_D , is small is

$$I = i_0 e^{V_D/nV_T} \quad (10)$$

where V_T is the thermal voltage and is approximately 25 mV at a temperature of 300 K. We wish to find how the saturation current (i_0) and the ideality factor (n) scale as the solution salinity is varied.

Faradaic measurements were made using an Agilent E5270B Precision Measurement Mainframe producing a stepped DC voltage stimulus whilst continuously measuring the output current.

Measurements commenced with a 4.5 litre solution of $1.0 \times$ PBS which was progressively diluted in factors of two between measurements. The solution was mixed continuously using a standard laboratory grade magnetic stirrer. The solution was in

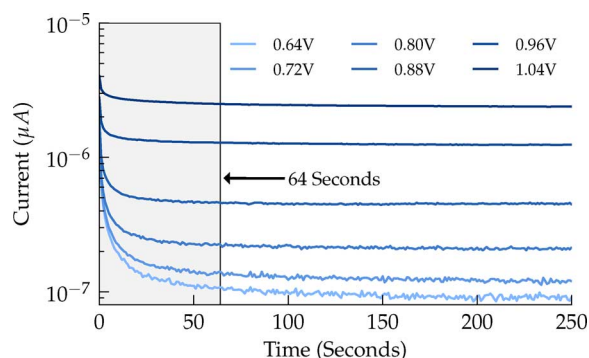


Fig. 9. Current versus time after a stepped voltage increment across a pair of interfaces. Time before 64 seconds is highlighted in grey. Measurements were taken with a wait time of 10,000 seconds between increments.

equilibrium with air prior to and during measurement and the electrodes remained submerged at all times. A single measurement run involved stepping the voltage across the electrodes from 0.5 V to 1.2 V in increments of 0.05 V; this range was previously determined to capture the onset of Faradaic conduction.

In [11] it is shown that the stirring of an air-saturated solution of physiological saline reduces the settling time of platinum electrodes to stimulus transients; with both situations eventually settling to the same impedance.

In [1] it has been noted that results obtained using cyclic voltammetry are dependant on, among other factors, the measurement sweep rate. This indicates a lack of isolation between displacement and Faradaic mechanisms. Our measurements were repeated using wait-times of 4, 16, 32 and 64 seconds between voltage increments, allowing us to determine the effect of sweep rate. From these measurements and those of a separate investigation² we concluded that a wait-time of 64 seconds is sufficient.

In Fig. 10, electrical current measurements (solid lines) are plotted against time. Each spike marks the point at which the voltage is incremented by 50 mV. Note that the amount of charge absorbed by the CPE increases with concentration of PBS, but the current for each concentration converges to approximately the same value. Dotted lines have been added that link the electrical current measurements taken 10 seconds after each increment for 0.125 × PBS (lower trace) and 1.0 × PBS (upper trace). These dotted traces show results typical of cyclic measurement methods, where the electrode overpotential is never constant. This illustrates the benefit of using DC measurements when measuring Faradaic currents.

Fig. 11 shows the settled currents plotted on a log scale, versus the voltage applied across the electrode pair. This uses the same data as the previous figure but has been processed so that each point represents the average of the final 24 current

²We took measurements of the interface's response to voltage steps when left to settle for 10 000 seconds, of which the first 250 seconds are shown in Fig. 9. Those measurements show a trend of decreasing current stabilisation time as the overpotential steps increase. Sixty four seconds after stepping from 0.64 V to 0.72 V, the current has stabilised to 95% of its value at 10000 seconds. At overpotential steps for higher voltages, the settling time is further reduced. These measurements were made in still solution and we expect stirring will further reduce settling time.

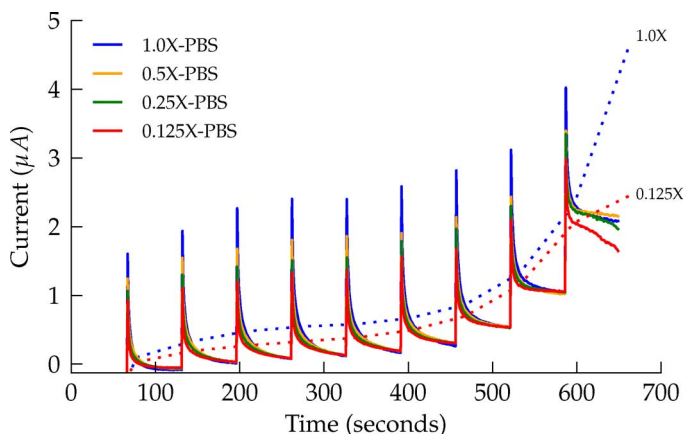


Fig. 10. Current conduction with 0.05 V stepped voltage increments from 0.5 V to 0.9 V. Each voltage is held for 64 seconds before being further incremented. Dotted lines connect current measurements occurring 10 seconds after an increment.

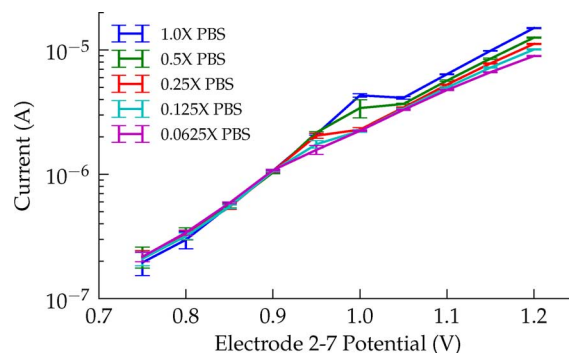


Fig. 11. Faradaic conduction as a function of voltage applied across electrodes 2 and 7. Samples shown were taken between 40 and 64 seconds after each voltage increment. Error bars indicate spread in measurement results, where 95% of samples lie within the bars.

measurements at each step. Below 0.9 V it appears that the concentration of the solution had no identifiable effect on Faradaic conduction as there is overlap between each trace and the mean values are not monotonic with concentration in this region. Between 0.9 V and 1.05 V a transition occurs which results in current conduction becoming dependent on solution concentration. Above 1.05 V the traces diverge showing clear dependence upon concentration. The point at which this transition occurs is determined by the concentration of saline, with lower concentrations transitioning earlier. A concentration of 0.0625 × PBS causes a transition at around 0.95 V whereas electrodes in the 1.0 × PBS solution transition somewhere between 1.0 V and 1.05 V. The effect of transition can be seen in the green trace at 1.0 V where the error bar is wide due to the current shift being captured in the measurement at that voltage. Part of a transition with respect to time is visible in Fig. 10 where the red trace drops uncharacteristically after the increment at around 600 seconds.

We attribute the change in behaviour between 0.9 V and 1.05 V to a transition to diffusion-controlled conduction. We hypothesise that the charging of the CPE draws available ions to the electrode, creating a layer of high ionic concentration at the surface irrespective of that of the solution bulk. It is this layer that is subsequently consumed by the Faradaic reactions at a rate that increases exponentially with electrode overpotential. The effect of bulk solution concentration while this

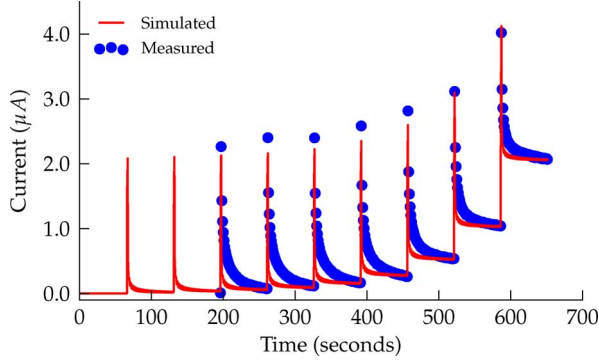


Fig. 12. Current versus time for $1.0 \times$ PBS overlaid with simulated results of the interface model's response to increasing voltage steps. The model used for simulation includes the CPE, diodes and R_s .

TABLE IV
FARADAIC PARAMETERS

Parameter	Value
i_0	2.757 pA
n	1.36

layer has formed is negligible until the point at which the layer is consumed faster than it can be replenished. At this stage, and with increasing overpotential, the Faradaic conduction is governed by diffusion of ions from the bulk into the layer, the rate of which increases with the ion concentration in the bulk. We believe this explains the divergence of conduction with concentration between 0.9 V and 1.05 V and why there is no observable dependence on bulk ion concentration beforehand. For the purposes of our model, we are content with placing a limit of 0.9 V across the pair of interfaces and using the assumption that Faradaic conduction does not change with concentration. As we are primarily interested with capturing only the onset of Faradaic conduction, this limitation does not degrade the usefulness of the model for implant modelling purposes.

Fig. 12 shows a subset of the data used in Fig. 10 (the $1.0 \times$ PBS trace) to show the fit between simulated and measured data in the time domain. Here the simulation contains fitted diode parameters, shown in Table IV, which give rise to the exponentially increasing steady state current. It is apparent from this trace that the temporal response of the CPE does not match the transient decay curves seen in the Faradaic measurements, something the CPE should predict. These transient decay curves appear to have a time-constant³ that decreases with each subsequent measurement. This variable time-constant phenomena is especially clear when looking at Fig. 9. Although the Faradaic response transients appear to approach the time-constant predicted by the CPE, there is still behaviour not completely described by the CPE.

The match between fitted diode parameters (n and i_0) and the measurement data 64 seconds after each voltage increment has been plotted in Fig. 13.

³We use the term "time-constant" although a CPE does not have a simple exponential time-domain response.

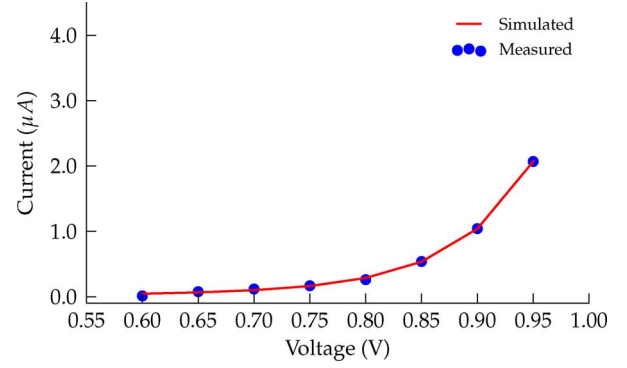


Fig. 13. Simulated response of the interface model compared to measurement from the electrodes with a stepped voltage overpotential. Measurement points show current values taken at the 64th second after each voltage increment.

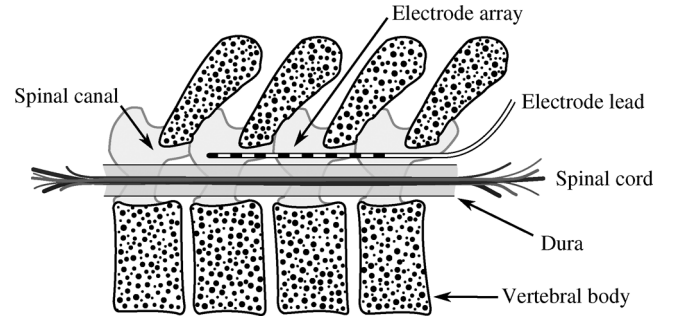


Fig. 14. Cross section of a sheep spine showing the position of the electrode array relative to the dura. Spotted regions represent cross sectioned vertebrae.

TABLE V
MODEL PARAMETERS FITTED TO SHEEP DATA

Parameter	Value
R_{eri}	500Ω
R_{sri}	375Ω
R_{li}	176Ω
m	1.34
k	1.77
$ Z @ 1Hz$	11300Ω
R_s	126Ω

In future we hope to extend this model into a diffusion controlled conduction scenario and extend the CPE element to better reproduce the time domain response.

V. In-Vivo MEASUREMENTS OF SHEEP

Cogan states the main differences between *in-vivo* and *in-vitro* responses come down to temperature, the presence of organic species, tortuous diffusion path for charge carriers, physiological responses of the host to a foreign body, and uncertain concentrations of electrolytes and buffers near the electrode surface.[1] Nevertheless, electrical engineers use baths of saline as a substitute for *in-vivo* testing (a "phantom") due to its convenience. We now test the claim that $0.1 \times$ PBS is a reasonable representation of the spine of a living sheep.

Each of the model characterisation measurements were repeated inside the spinal canal of living sheep, just outside the dura as shown in Fig. 14. Two sheep were prepared and anaesthetised using the procedures described in [16] under the Animal Care and Ethics Committee approval of the Royal North Shore

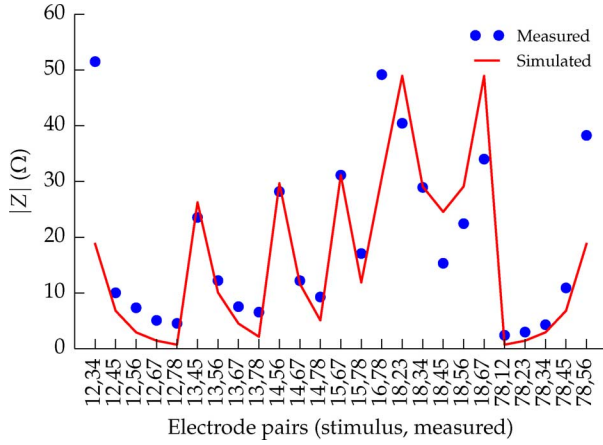


Fig. 15. Comparison between the magnitude of *in-vivo* transresistance measurements (blue dots) and simulation results using fitted mesh parameters (red trace).

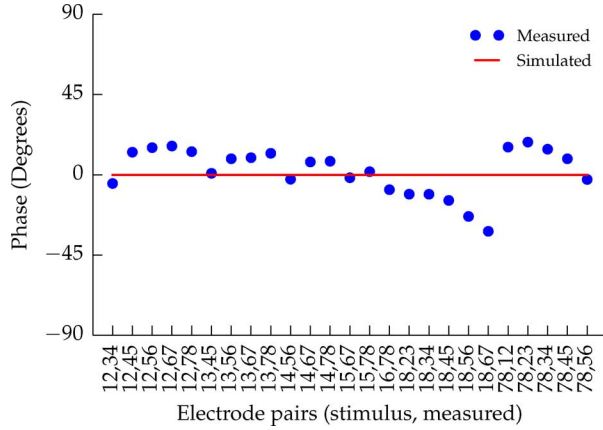


Fig. 16. Comparison between the phase of *in-vivo* transresistance measurements (blue dots) and simulation results using fitted mesh parameters (red trace). The simulation mesh does not capture any phase relationship due to being purely resistive.

Hospital, Sydney. The study complied with the Australian Code of Practice for the Care and Use of Animals for Scientific Purposes.

Transimpedance measurements *in-vivo* were conducted with additional configurations than were carried out in Section II. The same measurements were captured except this time using a larger number of stimulus-measure configurations in order to minimise the effect of nearby bone. Time constraints with the live animal meant only a single dataset was obtained for transimpedance. This data is shown in Figs. 17 and 18 along with results obtained from simulation with fitted parameters.

In these measurements a phase deviation of approximately 30 degrees is observed, visible in Fig. 16. There was no measurable phase deviation for these measurements in PBS. These results indicate that the sheep itself presents a significant reactive component.

Four measurements of displacement current (CPE element in the model) were taken at regular intervals over a period of 30 minutes, the mean response is presented in Figs. 17 and 18.

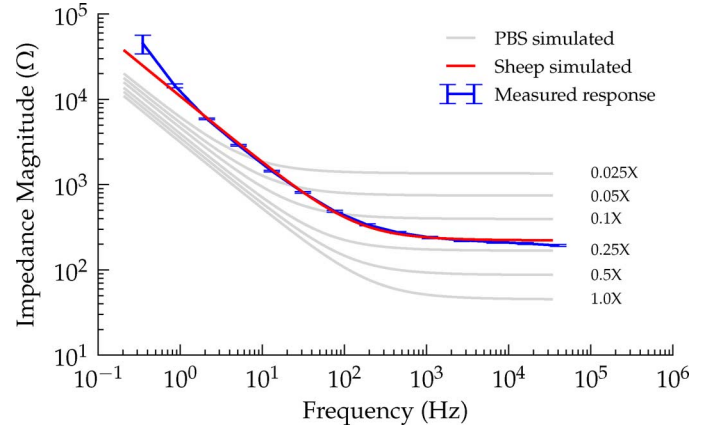


Fig. 17. Comparison between the mean magnitude of measured displacement currents in four live sheep (blue trace), each of the PBS traces from Fig. 6 (grey lines), and a simulated fit to the sheep data (red trace). Error bars have been placed one standard deviation either side of the mean for measured data.

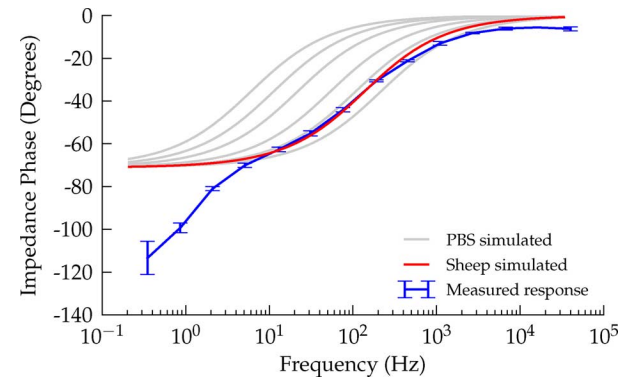


Fig. 18. Comparison of the mean phase response of measured displacement currents in live sheep (blue trace), each of the PBS traces taken from Fig. 7 (grey lines), and a simulated fit to the sheep data (red trace). Error bars have been placed one standard deviation either side of the mean for measured data.

These measurements appear on top of the traces taken from Figs. 6 and 7 for comparison.

Although it is clear that there are complications *in vivo*, a $0.25 \times$ PBS solution offers a good approximation to the resistive part of the trace. The reactive part (that of the CPE) would be better modelled by a solution of less than $0.025 \times$ concentration.

Faradaic measurements on the live animal were abandoned as they were likely to cause potentially violent muscle contraction. Attempts to measure the response at low stimulus potentials yielded unsatisfactory results and were subsequently discarded.

VI. CONCLUSION

We have used the parameters of a compact electrical model of an implantable electrode as a way of objectively comparing scenarios. We have measured how changing the concentration of PBS affects the parameters of this model. We have drawn direct relationships with concentration and conductivity sufficient to predict interface characteristics at arbitrary dilutions of PBS. We found that the magnitude of the CPE element moves much more slowly with concentration whereas the resistance moved almost linearly. Using DC measurements we conclude that the

concentration of PBS has no effect on the onset of Faradaic conduction for the range of concentrations measured. We hypothesise that this is because the double-layer sets the concentration of species involved in the Faradaic reactions within a volume surrounding each electrode.

In-vivo measurements using platinum electrodes in sheep spine show that no single concentration of PBS matches both the CPE and the resistive characteristics at once. These measurements also reveal a considerable amount of reactance (see transimpedance measurements *in vivo*) that is captured neither by PBS nor the model presented when using a purely resistive spreading resistance. Nevertheless, if a single solution of PBS is to be used as a human phantom, a one-tenth concentration solution (by mass) is a good compromise.

While the model used is limited to predicting only the onset of Faradaic current conduction, it usefully captures the transition into this biologically destructive stimulus regime sufficient for it to be avoided.

The final model, including 205 lines of resistive mesh, is represented by a SPICE netlist comprised of 348 lines. A simulation of this model, using the freely available ngSpice circuit simulator, at 1000 frequencies takes approximately 1 second on a modern CPU.

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