

# **CPT\_S 260 Intro to Computer Architecture**

## **Lecture 24**

**Digital Design IV**  
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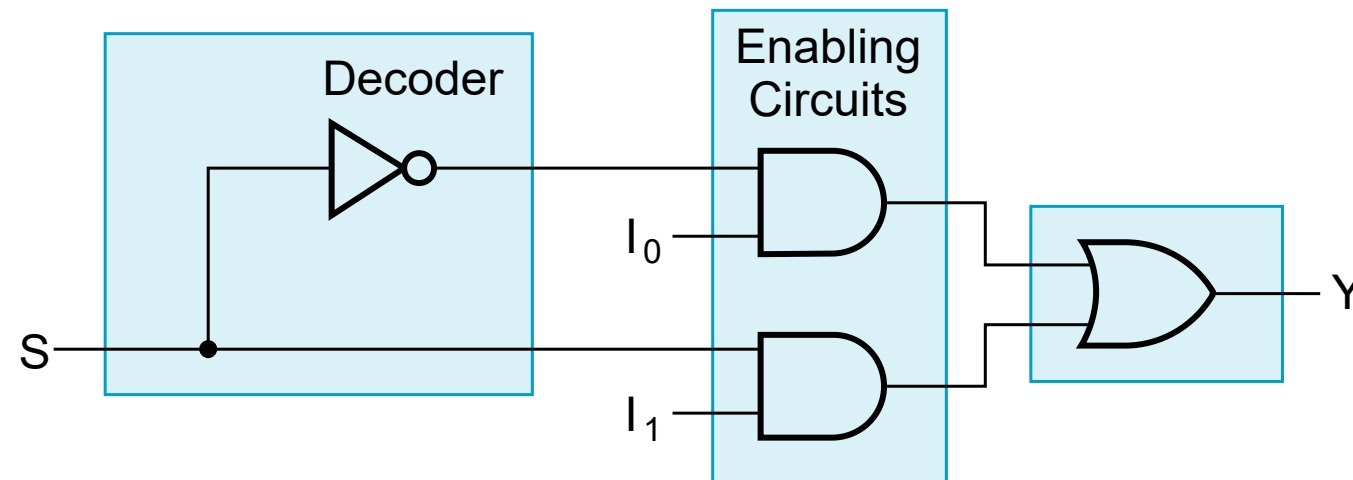
# Recap: 2-to-1-Line Multiplexer

- Since  $2 = 2^1$ ,  $n = 1$
- The single selection variable  $S$  has two values:
  - $S = 0$  selects input  $I_0$
  - $S = 1$  selects input  $I_1$

- The equation:

$$Y = \bar{S}I_0 + SI_1$$

- The circuit:

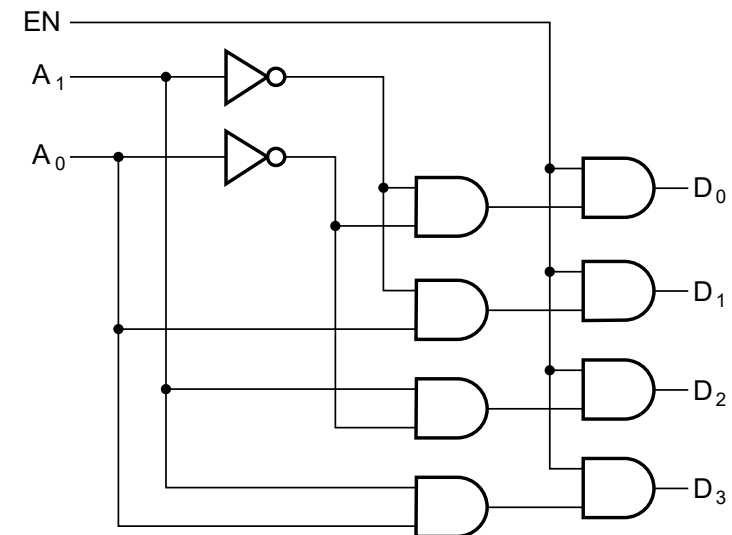


# Recap: Decoders

- **Decoding** - the conversion of an  $n$ -bit input code to an  $m$ -bit output code with  $n \leq m \leq 2^n$  such that each valid code word produces a unique output code
- Circuits that perform decoding are called *decoders*
- Here, functional blocks for decoding are
  - called  $n$ -to- $m$  line decoders, where  $m \leq 2^n$ , and
- **Example: 2-to-4 decoder:**

EN	A <sub>1</sub>	A <sub>0</sub>	D <sub>0</sub>	D <sub>1</sub>	D <sub>2</sub>	D <sub>3</sub>
0	X	X	0	0	0	0
1	0	0	1	0	0	0
1	0	1	0	1	0	0
1	1	0	0	0	1	0
1	1	1	0	0	0	1

(a)



(b)

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# Logic Simplification

# Rules of Boolean Algebra

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- **Associative Law of multiplication**

$$A \cdot (B \cdot C) = (A \cdot B) \cdot C$$

- **Distributive Law of multiplication**

$$A + BC = (A + B) \cdot (A + C)$$

- **Annulment law:**

$$A \cdot 0 = 0$$

$$A + 1 = 1$$

- **Identity law:**

$$A \cdot 1 = A$$

$$A + 0 = A$$

# Rules of Boolean Algebra

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- **Complement law:**

$$\begin{aligned}A + \bar{A} &= 1 \\A \cdot \bar{A} &= 0\end{aligned}$$

- **Double negation law:**

$$\bar{\bar{A}} = A$$

- **Absorption law:**

$$\begin{aligned}A \cdot (A + B) &= A \\A + AB &= A \\A + \bar{A}B &= A + B\end{aligned}$$

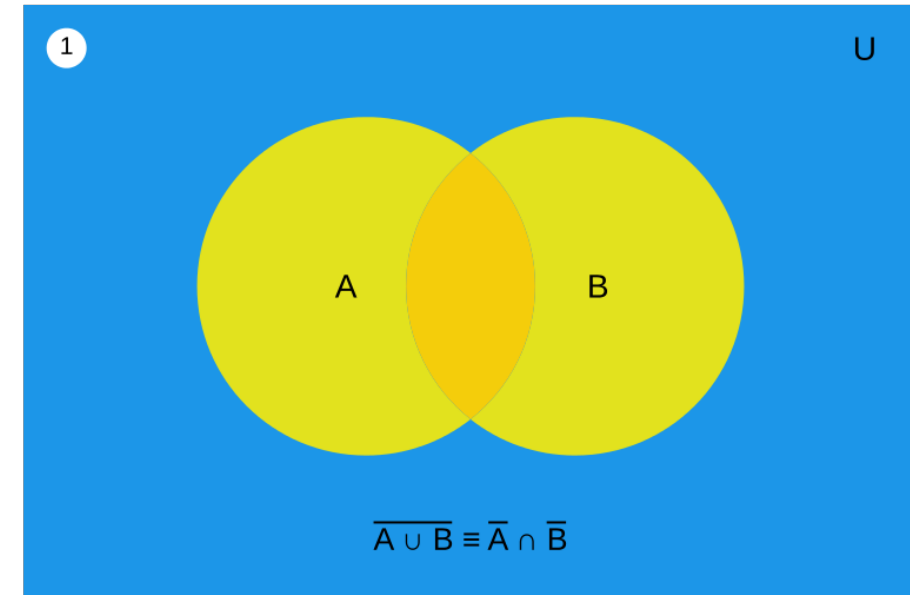
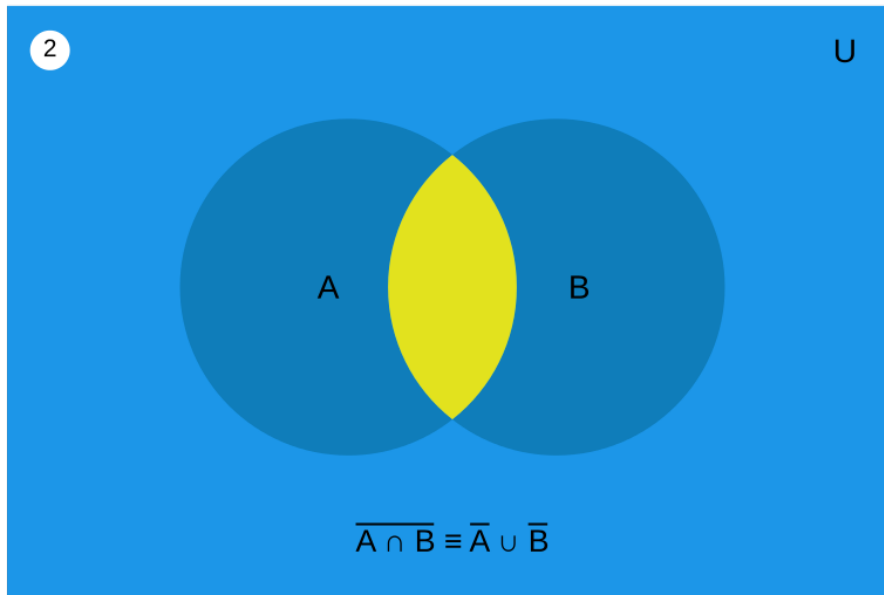
- **Idempotent law:**

$$\begin{aligned}A + A &= A \\A \cdot A &= A\end{aligned}$$

# De Morgan's Laws

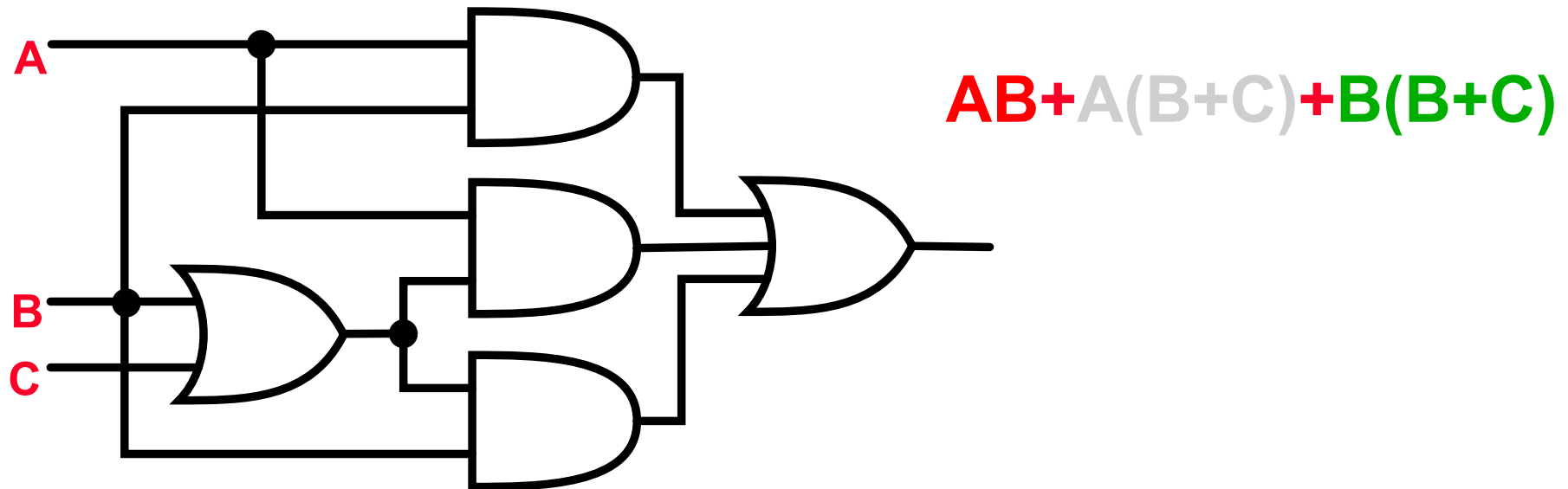
- Transformation rules that help simplification of negations
- Statement:

$$\overline{AB} = \bar{A} + \bar{B}$$
$$\overline{(A + B)} = \bar{A} \cdot \bar{B}$$



# Simplification Using Boolean Algebra

- A simplified Boolean expression uses the fewest gates possible to implement a given expression.

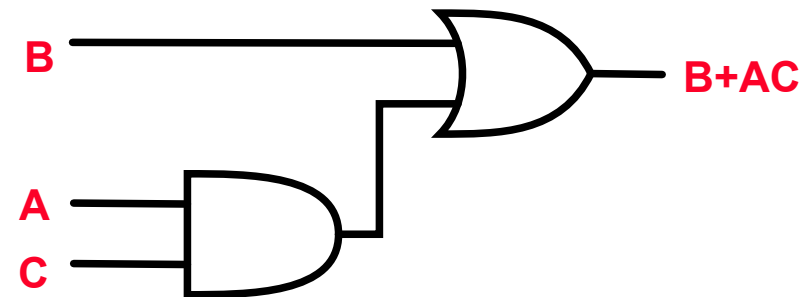
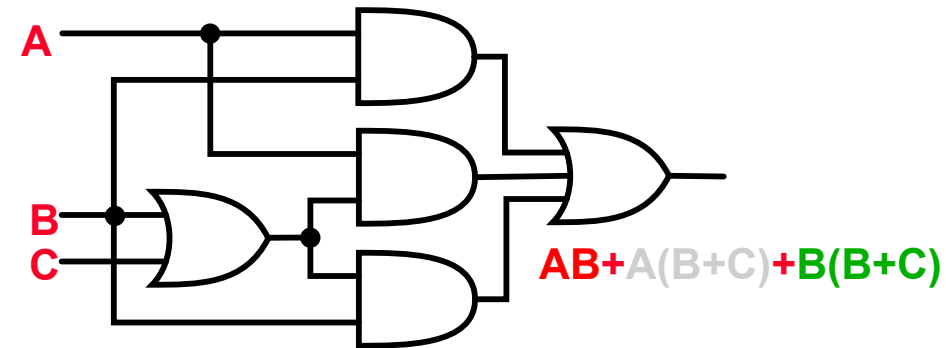




# Simplification Using Boolean Algebra

## ▪ $AB + A(B+C) + B(B+C)$

- (distributive law)
  - »  $AB + AB + AC + BB + BC$
- ( $BB=B$ )
  - »  $AB + AB + AC + \textcolor{red}{B} + BC$
- ( $AB + AB = AB$ )
  - »  $\textcolor{red}{AB} + AC + B + BC$
- ( $B + BC = B$ )
  - »  $AB + AC + \textcolor{red}{B}$
- ( $AB + B = B$ )
  - »  $\textcolor{red}{B} + AC$



# Examples

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- $[A\bar{B}(C + BD) + \bar{A}\bar{B}]C$
- $\bar{A}BC + A\bar{B}\bar{C} + \bar{A}\bar{B}\bar{C} + A\bar{B}C + ABC$
- $\overline{AB + AC} + \bar{A}\bar{B}C$

# Examples

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$$\begin{aligned} & [A\bar{B}(C + BD) + \bar{A}\bar{B}]C \\ &= [A\bar{B}C + A\bar{B}BD + \bar{A}\bar{B}]C \text{ (distributive law)} \\ &= A\bar{B}CC + \bar{A}\bar{B}C \text{ } (\bar{B}B = 0 \text{ using complement law)} \\ &= \bar{B}C(A + \bar{A}) \\ &= \bar{B}C \text{ (Complement law)} \end{aligned}$$

# Examples

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$$\begin{aligned} & \overline{AB + AC} + \bar{A}\bar{B}C \\ &= (\overline{AB} \cdot \overline{AC}) + \bar{A}\bar{B}C \text{ (DeMorgan Law)} \\ &= (\bar{A} + \bar{B})(\bar{A} + \bar{C}) + \bar{A}\bar{B}C \text{ (DeMorgan Law)} \\ &= \bar{A}\bar{A} + \bar{A}\bar{C} + \bar{A}\bar{B} + \bar{B}\bar{C} + \bar{A}\bar{B}C \\ &= \bar{A} + \bar{B}\bar{C} \text{ (Take } \bar{A} \text{ out and use annulment law)} \end{aligned}$$

# Standard Forms of Boolean Expressions

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- **All Boolean expressions, regardless of their form, can be converted into either of two standard forms:**
  - The sum-of-products (SOP) form (minterms)
  - The product-of-sums (POS) form (maxterms)
- **Standardization makes the evaluation, simplification, and implementation of Boolean expressions much more systematic and easier**

# Sum of Products

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- **Minterm Expressions**
- If input is 0 we take the complement of the variable
- If input is 1 we take the variable as is
- To get the desired canonical SOP expression we will add the minterms (product terms) for which the output is 1

$$F = \bar{A}B + A\bar{B} + AB$$

A	B	F	Minterm
0	0	0	A'B'
0	1	1	A'B
1	0	1	AB'
1	1	1	AB

# Product of Sums

- **Maxterm Expressions**
- If input is 1, we take the complement
- If input is 0, we take the variable
- To get the desired canonical POS expression we will multiply the maxterms (sum terms) for which the output is 0

A	B	F	Minterm
0	0	0	A'B'
0	1	1	A'B
1	0	1	AB'
1	1	1	AB

$$F = (A + B)$$

A	B	F	Maxterm
0	0	0	A + B
0	1	1	A + $\bar{B}$
1	0	1	$\bar{A}$ + B
1	1	1	$\bar{A}$ + $\bar{B}$

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# **Designing a Simple ALU**

## **(Arithmetic Logic Unit)**



# Adder Algorithm

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	1	0	0	1
	0	1	0	1
Sum (S)	1	1	1	0
Carry (C)	0	0	0	1

Truth Table for the above operations:

A	B	Cin	Sum	Cout
0	0	0		
0	0	1		
0	1	0		
0	1	1		
1	0	0		
1	0	1		
1	1	0		
1	1	1		

# Adder Algorithm

	1	0	0	1
	0	1	0	1
Sum	1	1	1	0
Carry	0	0	0	1

Truth Table for the above operations:

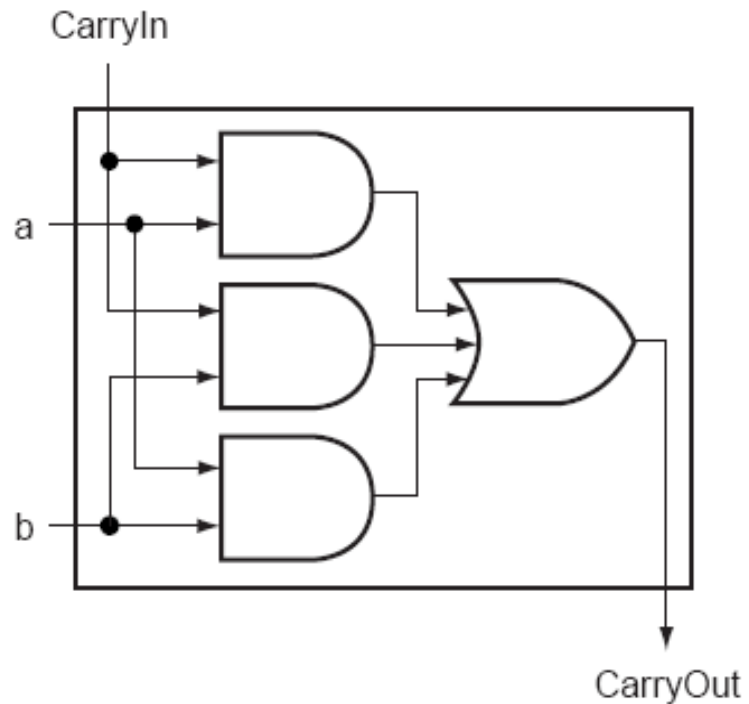
A	B	Cin	Sum	Cout
0	0	0	0	0
0	0	1	1	0
0	1	0	1	0
0	1	1	0	1
1	0	0	1	0
1	0	1	0	1
1	1	0	0	1
1	1	1	1	1

Equations:

$$\begin{aligned} \text{Sum} = & \text{Cin} \cdot \bar{A} \cdot \bar{B} + \\ & B \cdot \bar{\text{Cin}} \cdot \bar{A} + \\ & A \cdot \bar{\text{Cin}} \cdot \bar{B} + \\ & A \cdot B \cdot \text{Cin} \end{aligned}$$

$$\begin{aligned} \text{Cout} = & A \cdot B \cdot \text{Cin} + \\ & A \cdot B \cdot \bar{\text{Cin}} + \\ & A \cdot \text{Cin} \cdot \bar{B} + \\ & B \cdot \text{Cin} \cdot \bar{A} \\ = & A \cdot B + \\ & A \cdot \text{Cin} + \\ & B \cdot \text{Cin} \end{aligned}$$

# Carry Out Logic

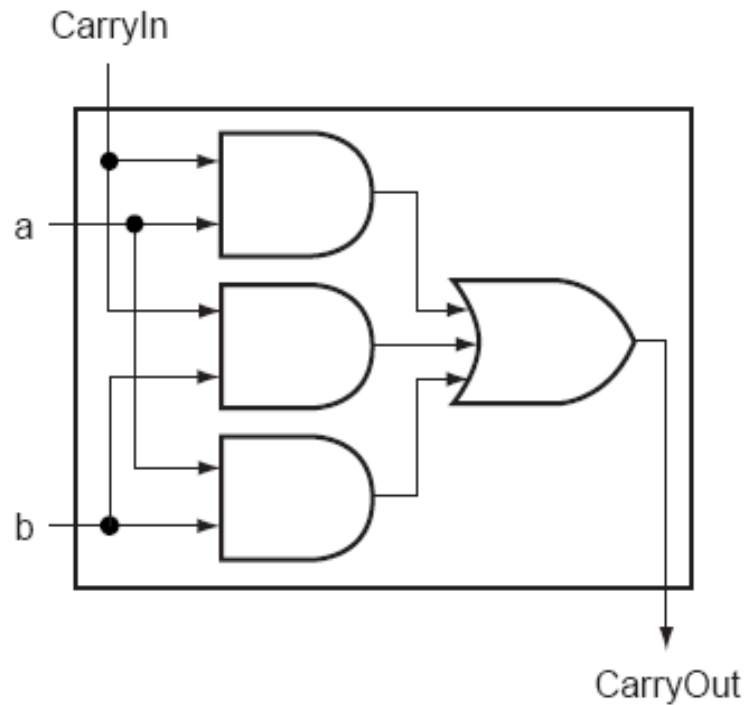


Equations:

$$\begin{aligned} \text{Sum} = & \text{Cin} \cdot \bar{A} \cdot \bar{B} + \\ & B \cdot \bar{\text{Cin}} \cdot \bar{A} + \\ & A \cdot \bar{\text{Cin}} \cdot \bar{B} + \\ & A \cdot B \cdot \text{Cin} \end{aligned}$$

$$\begin{aligned} \text{Cout} = & A \cdot B \cdot \text{Cin} + \\ & A \cdot B \cdot \bar{\text{Cin}} + \\ & A \cdot \text{Cin} \cdot \bar{B} + \\ & B \cdot \text{Cin} \cdot \bar{A} \\ = & A \cdot B + \\ & A \cdot \text{Cin} + \\ & B \cdot \text{Cin} \end{aligned}$$

# Carry Out Logic



$$\begin{aligned}C_{out} &= A \cdot B \cdot Cin + A \cdot B \cdot \overline{Cin} + A \cdot Cin \cdot \bar{B} + B \cdot Cin \cdot \bar{A} \\&= A \cdot B \cdot (Cin + \overline{Cin}) + A \cdot Cin \cdot \bar{B} + A \cdot B \cdot Cin + B \cdot Cin \cdot \bar{A} + A \cdot B \cdot Cin \\&= A \cdot B + A \cdot Cin \cdot (B + \bar{B}) + B \cdot Cin \cdot (A + \bar{A})\end{aligned}$$

# Full-Adder (3 inputs, 2 outputs)

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- Full-Adder Truth Table:

X	Y	Z	C	S
0	0	0	0	0
0	0	1	0	1
0	1	0	0	1
0	1	1	1	0
1	0	0	0	1
1	0	1	1	0
1	1	0	1	0
1	1	1	1	1

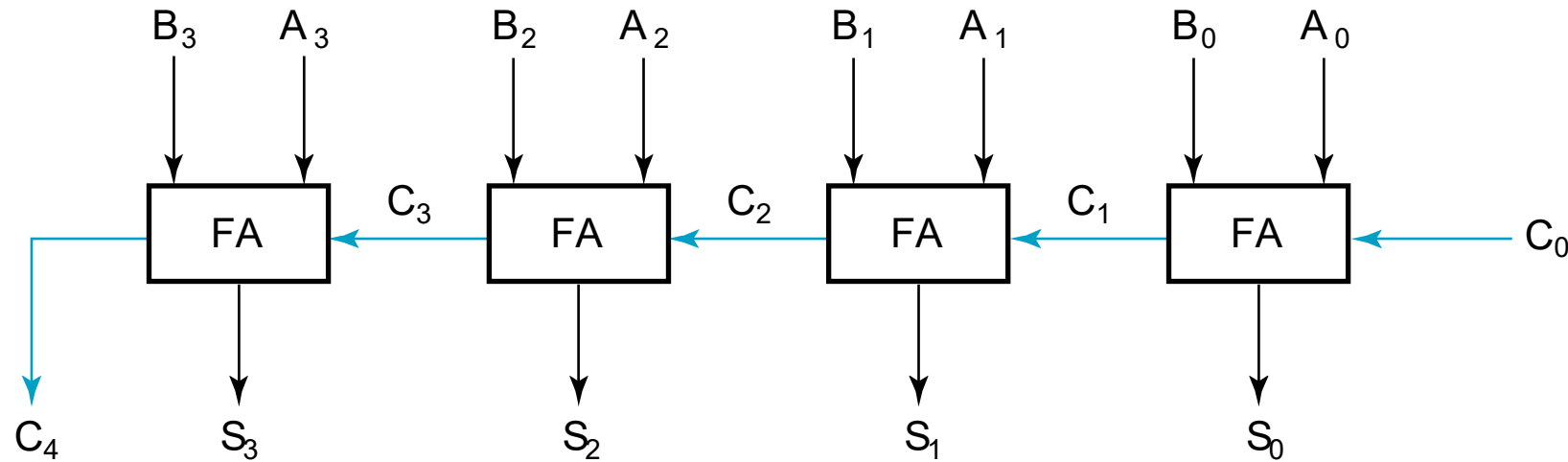
- Same as the previous example:

$$S = X \bar{Y} \bar{Z} + \bar{X} Y \bar{Z} + \bar{X} \bar{Y} Z + X Y Z$$
$$C = X Y + X Z + Y Z$$

# 4-bit Ripple-Carry Binary Adder

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- A four-bit Ripple Carry Adder made from four 1-bit Full Adders:



# 32-bit Ripple Carry Adder

**1-bit ALUs are connected  
“in series” with the  
carry-out of 1 box  
going into the carry-in  
of the next box**

