
Solution Manual

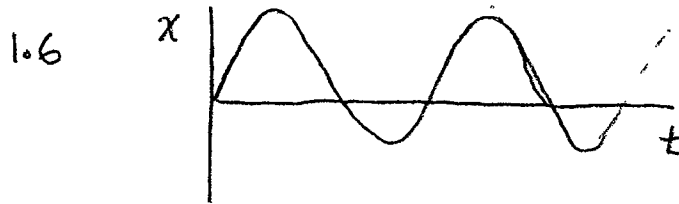
System Dynamics

**Modeling, Simulation and Control of
Mechatronic Systems**

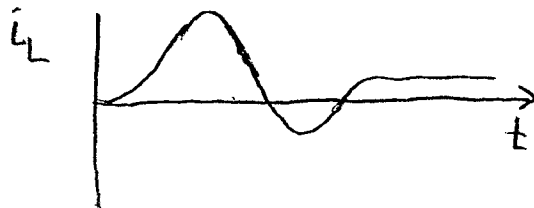
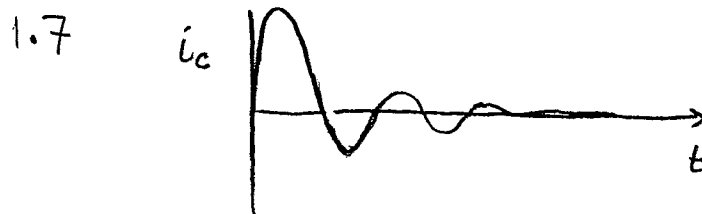
5th Edition

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Problems 1.1 to 1.5 are mainly discussion questions.



steady state deflection, $x_0 = \frac{(m+M)g}{k}$



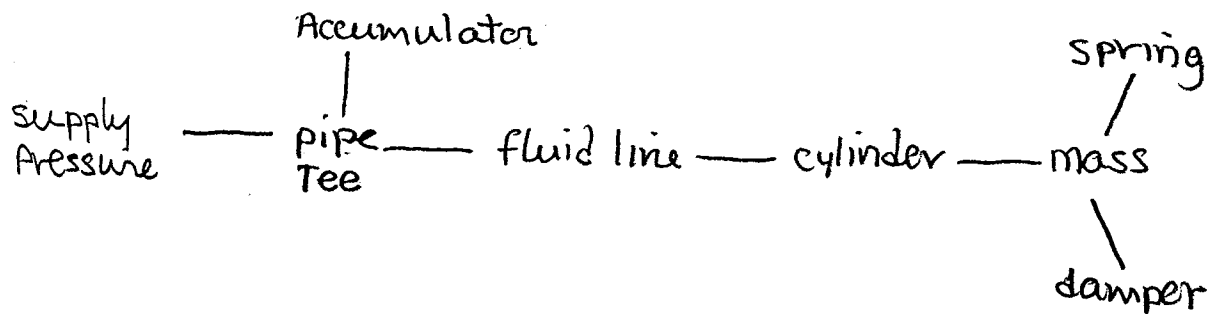
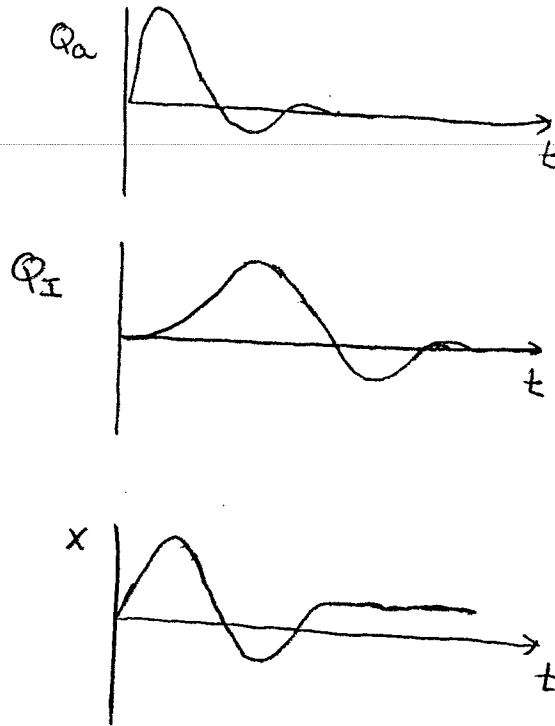
$$\dot{I}_{Lss} = \frac{E_0}{R}$$

1.8 $\left. \begin{array}{l} Q_a \text{ similar to } \dot{I}_c \\ Q_I \text{ similar to } \dot{I}_L \end{array} \right\} \text{ in prob. 1.7}$

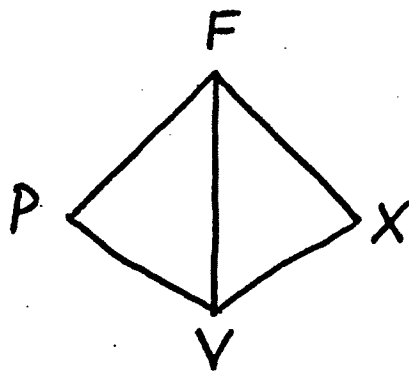
$$Q_{Iss} = P_s / R_f$$

1.9

1-2



2-1

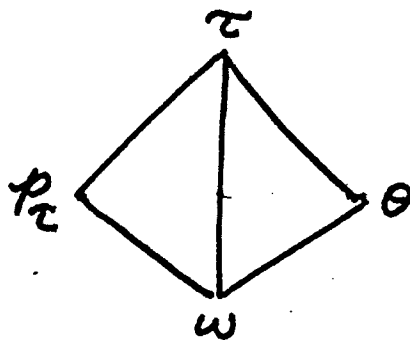


$$[F] - N$$

$$[V] - m/s$$

$$[P] - N \cdot s$$

$$[X] - m$$

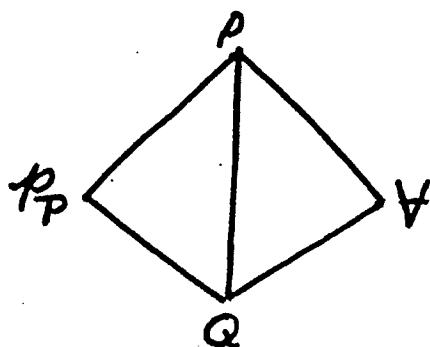


$$[\tau] - N \cdot m$$

$$[\omega] - rad/s$$

$$[p_\tau] - N \cdot m \cdot s$$

$$[\theta] - rad$$

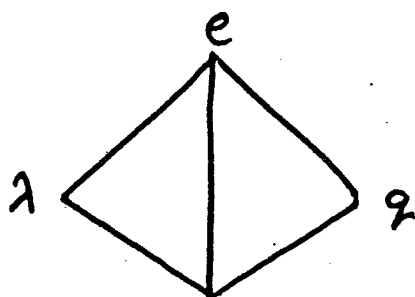


$$[P] - N/m^2$$

$$[Q] - m^3/s$$

$$[p_P] - N \cdot s/m^2$$

$$[V] - m^3$$



$$[e] - V$$

$$[i] - A$$

$$[\lambda] - V \cdot s$$

$$[q] - A \cdot s = C$$

$$2-2 \quad \frac{\tau}{\omega} \text{ Electric Motor } \frac{e}{i}$$

(a)

2-2

$$\frac{\tau}{\omega} \text{ Hydraulic Pump } \frac{P}{Q}$$

(b)

$$\frac{\tau}{\omega_1} \text{ Shaft } \frac{\tau}{\omega_2}$$

(c)

$$\frac{F}{V_1} \text{ Shock Absorber } \frac{F}{V_2}$$

(d)

$$\frac{e_1}{i_1} \text{ Transistor } \frac{e_2}{i_2}$$

(e)

$$\frac{e}{i} \text{ Speaker}$$

(f)

$$\frac{\tau}{\omega} \text{ Crank } \frac{F}{V}$$

(g)

$$\frac{F}{V} \text{ Wheel } \frac{\tau}{\omega}$$

(h)

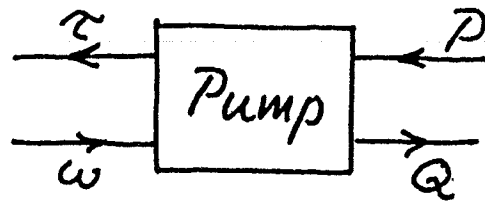
$$\frac{e_f}{i_f} \text{ Motor } \frac{e_a}{i_a}$$

(i)

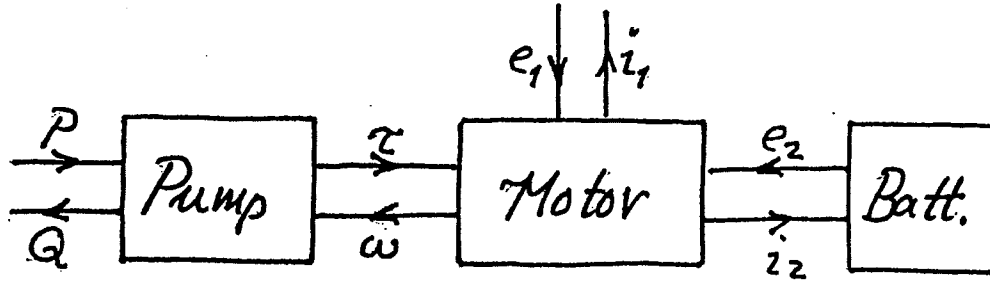
2-3



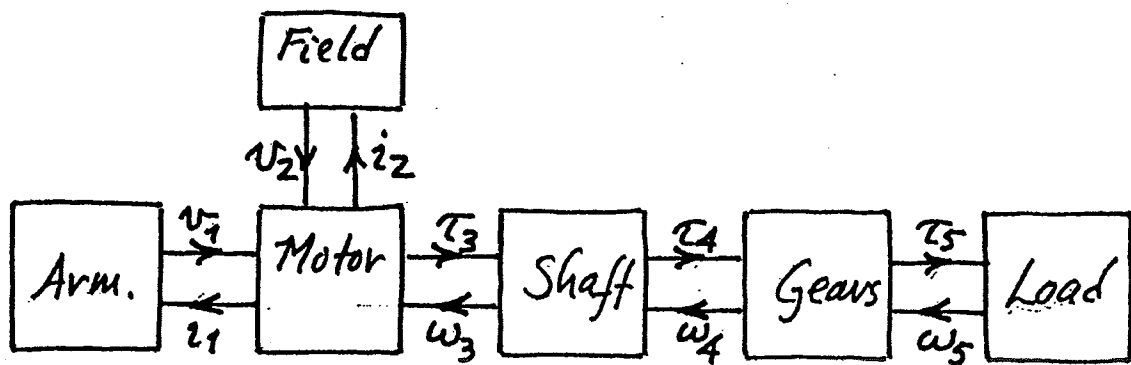
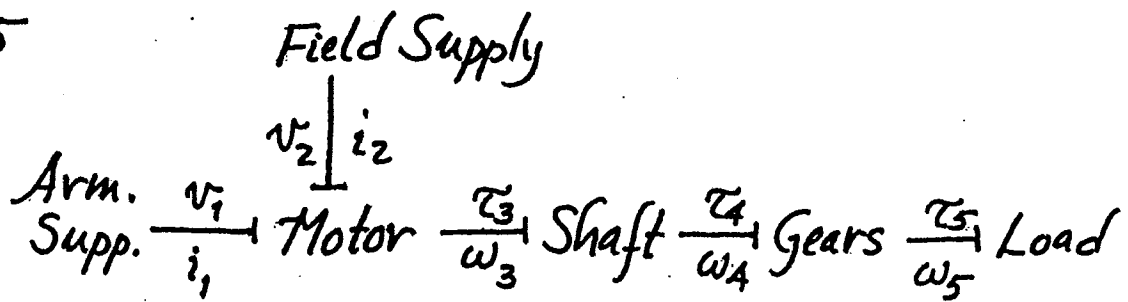
2-3



2-4

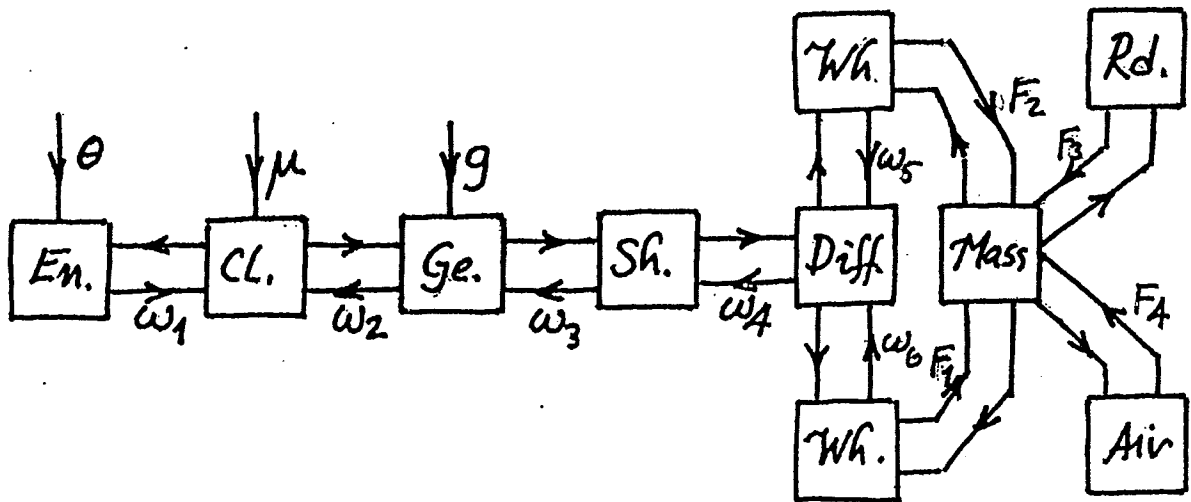
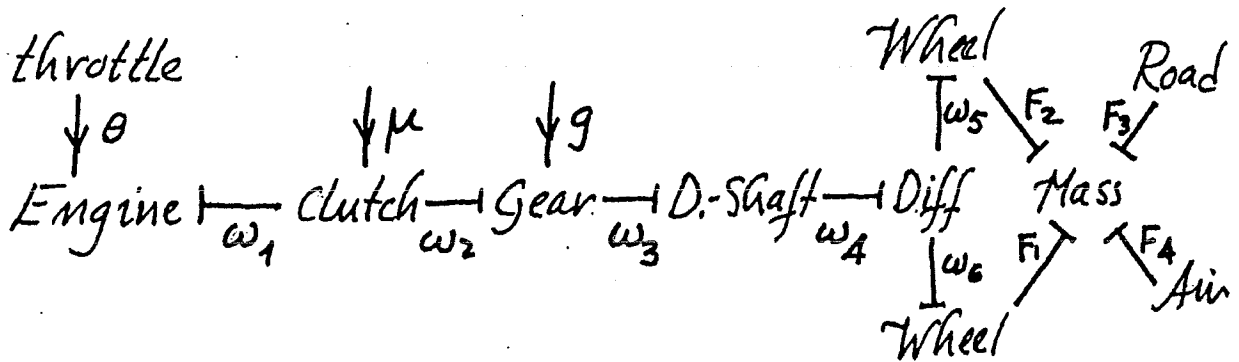


2-5



2-6

2-4



2-7 Inputs: P, e_1, e_2

Outputs: Q, i_1, i_2

2-8

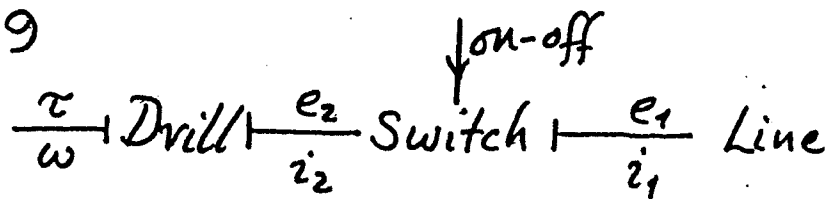
$$\text{Power} \times \text{Time} = \text{Energy}$$

$$P \cdot t = mgh$$

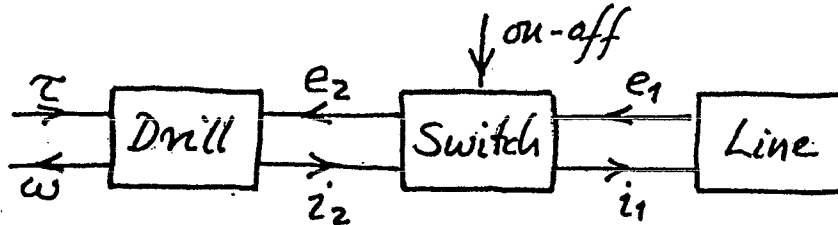
$$100 \cdot t = 10 \cdot (9.81) \cdot 30$$

$$t = 29.43 \text{ s}$$

2-9



2-5



2-10

$$\tau \omega = P Q$$

$$\omega = \frac{P}{\tau} \cdot Q = \frac{7.0 \times 10^6}{5} Q$$

$$\omega = 1.4 \times 10^6 Q$$

$$\left[\frac{\text{rad}}{\text{s}} \right] = \left[\frac{1}{\text{m}^3} \right] \cdot \left[\frac{\text{m}^3}{\text{s}} \right]$$

2.11

using kinematics:

$$x = R \cos \theta + l \cos \alpha$$

$$l \sin \alpha = R \sin \theta$$

$$\cos \alpha = \sqrt{1 - \sin^2 \alpha} = \sqrt{1 - \frac{R^2}{l^2} \sin^2 \theta}$$

 \therefore

$$x = R \cos \theta + l \sqrt{1 - \frac{R^2}{l^2} \sin^2 \theta}$$

Then

$$\dot{x} = -v = -R \sin \theta \dot{\theta} + l \frac{1}{2} \left[1 - \frac{R^2}{l^2} \sin^2 \theta \right]^{-1/2} \left(-\frac{R^2}{l^2} 2 \sin \theta \cos \theta \right) \dot{\theta}$$

or

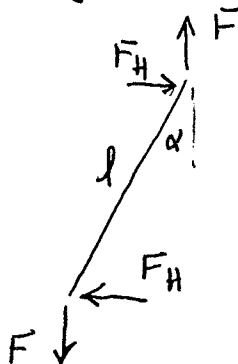
$$v = \left[R \sin \theta + l \frac{\left(\frac{R}{l} \right)^2 \sin \theta \cos \theta}{\sqrt{1 - \left(\frac{R}{l} \right)^2 \sin^2 \theta}} \right] \dot{\theta}$$

$m(\theta)$

$$\therefore v = m(\theta) \omega$$

$$\tau = m(\theta) F$$

using Forces and moments:

moment equilibrium
for rod:

$$F l \sin \alpha = F_H l \cos \alpha$$

or

$$F_H = F \sin \alpha / \cos \alpha$$

moment equilibrium
for crank:

$$\tau = F_H R \cos \theta + F R \sin \theta$$

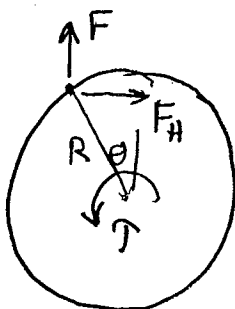
$$= F \frac{\sin \alpha}{\cos \alpha} R \cos \theta + F R \sin \theta$$

$$\text{use } \sin \alpha = \frac{R}{l} \sin \theta$$

$$\cos \alpha = \sqrt{1 - \left(\frac{R}{l} \right)^2 \sin^2 \theta}$$

substitute and you will
end up with:

$$\tau = m(\theta) F$$

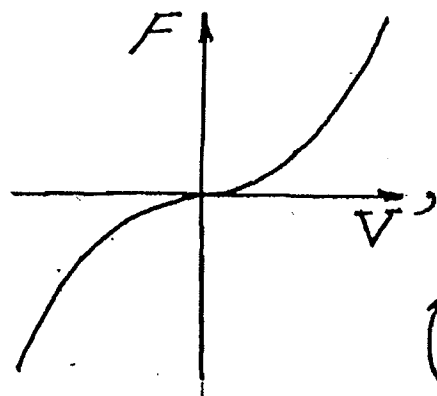


2.12

Z-7

element	inputs	outputs
supply pressure	Q_s	P_s
Tube	P_s, P_a, Q_e	Q_s, Q_a, P_c
Accumulator	Q_a	P_a
Cylinder	P_c, v_m	Q_e, F_m
Mass	F_m, F_s, F_d	v_m, v_s, v_d
Spring	v_s	F_s
Damper	v_d	F_d

3-1



$$\frac{F}{V} \rightarrow R,$$

3-1

$$V = \sqrt{\frac{|F|}{A}} \cdot \operatorname{sgn} F$$

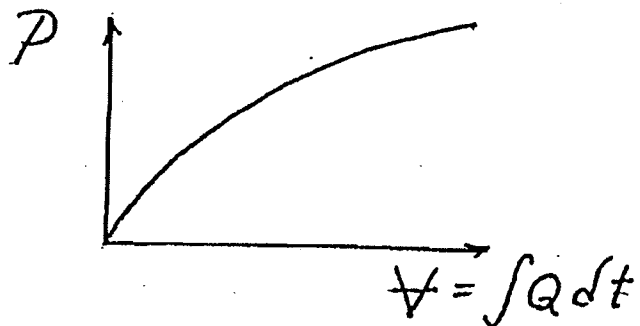
$$\frac{F}{V} \rightarrow R$$

$$3-2 \quad hA = V = \int Q dt,$$

$$P = \rho g h = \frac{\rho g}{A} \int Q dt$$

$$\therefore C = A / \rho g$$

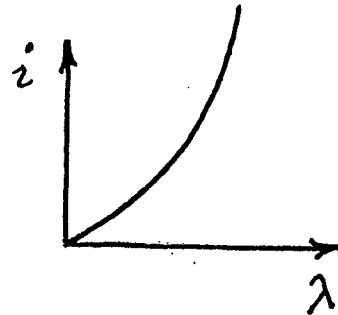
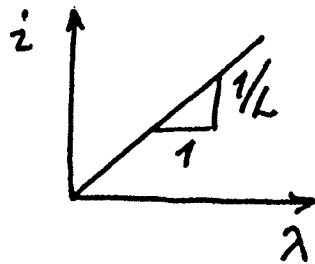
3-3



3-4

$$\frac{F}{\dot{x}} \rightarrow C, \quad F = \left(\frac{3EI}{L^3} \right) x.$$

3-5 $Li = \lambda = \int e dt$



3-2

$$i = i(\lambda)$$

$$\text{or } \lambda = \lambda(i)$$

$$\frac{di}{dt} = \frac{di(\lambda)}{d\lambda} \cdot \frac{d\lambda}{dt}$$

$$\frac{d\lambda}{dt} = e = \frac{d\lambda(i)}{di} \cdot \frac{di}{dt}$$

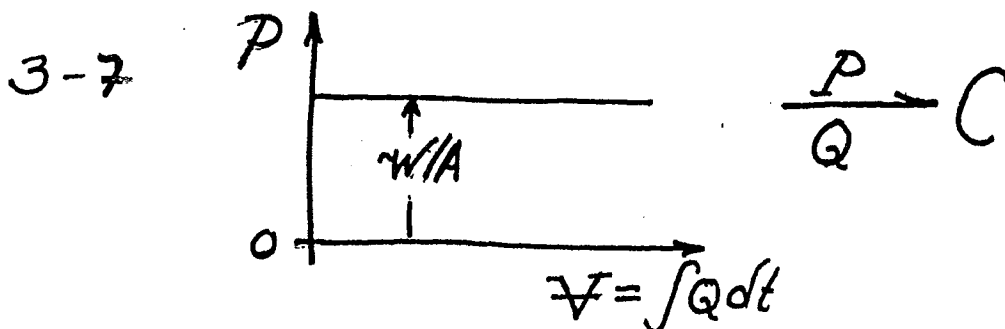
$$= \frac{di(\lambda)}{d\lambda} \cdot e$$

3-6 $\bar{F} = m\bar{a} = m\dot{\bar{v}}$

$$P_1 A - P_3 A = \rho A L \cdot \frac{dQ_2}{dt}$$

$$\text{or } P_1 - P_3 = \left(\frac{\rho L}{A}\right) \cdot \frac{dQ_2}{dt} = P_2$$

$$P_2 = \int (P_1 - P_3) dt = \frac{\rho L}{A} \cdot Q_2$$



3-8

$$\tau = I \alpha, \quad p_{\tau} = \int \tau dt = I \omega$$

$$I = \frac{1}{2} m R^2 = \frac{1}{2} \rho \cdot \pi R^2 t \cdot R^2$$

$$= \frac{0.28 \pi \cdot 1 \cdot (5)^4}{2 \cdot 386} = 0.712 \text{ lb s}^2/\text{in}$$

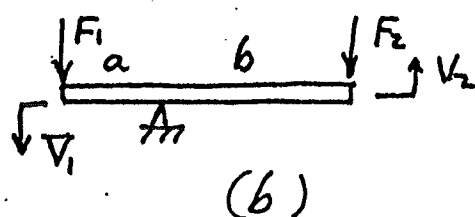
3-9

$$\frac{F}{V} \quad TF \quad \frac{P}{Q}, \quad \text{Area} = A$$

$$F = A P$$

$$A V = Q$$

3-10



$$\frac{F_1}{V_1} \quad TF \quad \frac{F_2}{V_2}$$

$$a F_1 = b F_2$$

$$\frac{V_1}{a} = \frac{V_2}{b}$$

$$(c) \quad \frac{\tau_1}{\omega_1} \quad TF \quad \frac{\tau_2}{\omega_2}$$

$$\left. \begin{aligned} \omega_1 r_1 &= \omega_2 r_2, \\ \tau_1 / r_1 &= \tau_2 / r_2, \end{aligned} \right\} r_1, r_2 \text{ radii of gears}$$

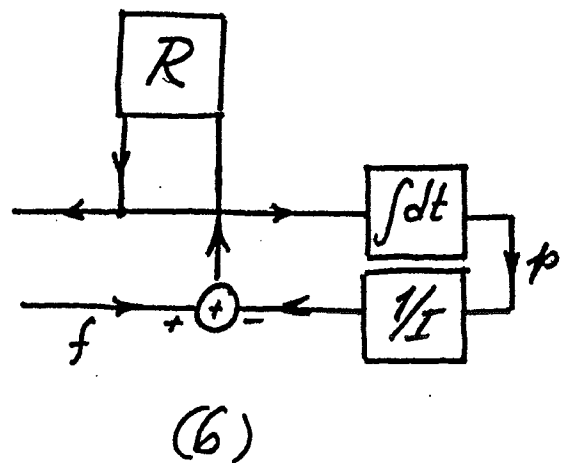
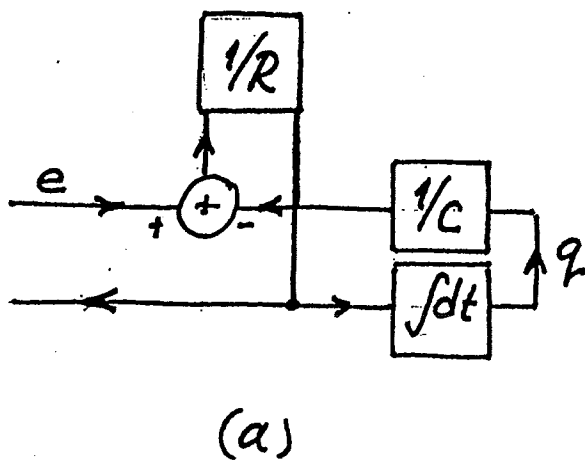
3-11 Ang. Momentum = \bar{H} 3-4

$|\bar{H}| \approx J\Omega$, torques associated with F_1 and F_2 cause change in direction of \bar{H} not magnitude. Consider F_1 first; let shaft length be L . Torque is then $F_1 L$, $\dot{\bar{H}} = \frac{F_1 L}{J}$ means that tip of \bar{H} vector must move up with angular rate V_2/L . $|\dot{\bar{H}}| = J\Omega \cdot V_2/L = F_1 L$

so

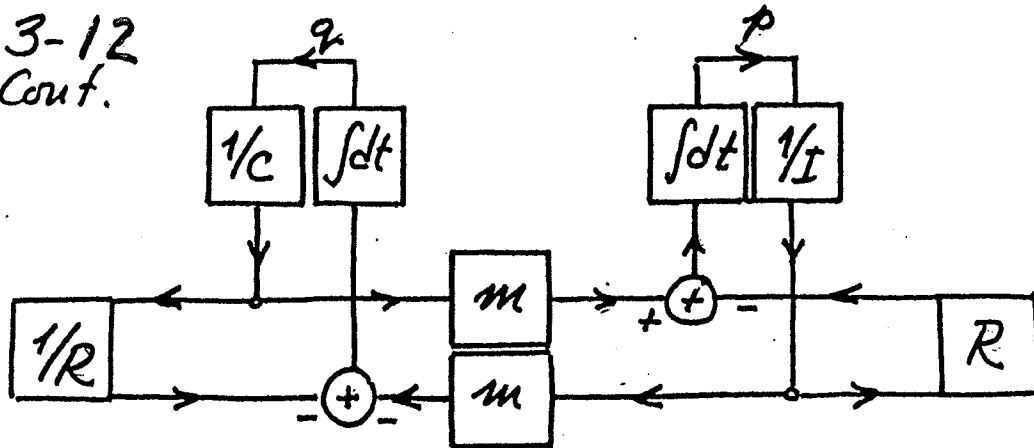
$$\left. \begin{aligned} F_1 &= \left(\frac{J\Omega}{L^2} \right) V_2 \\ \text{Similarly } F_2 &= \left(\frac{J\Omega}{L^2} \right) V_1 \end{aligned} \right\} \frac{F_1}{V_1} GY \frac{F_2}{V_2}$$

3-12

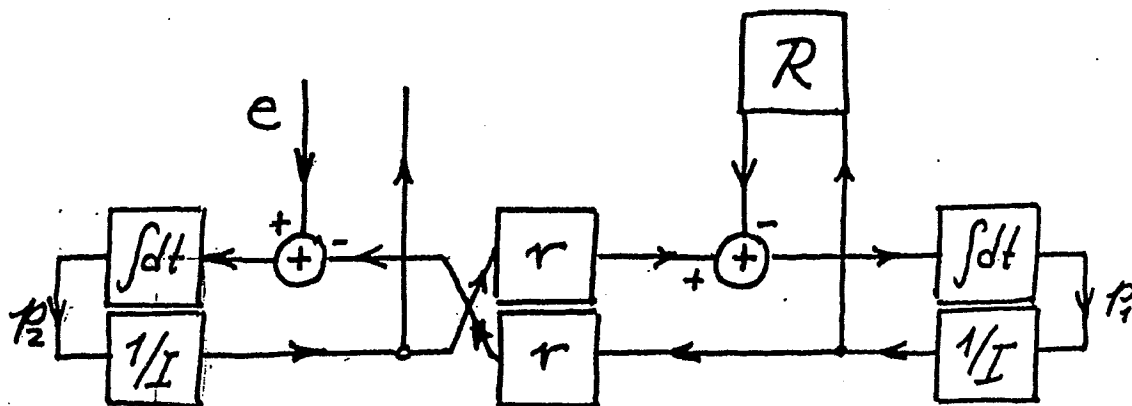


3-12
Cont.

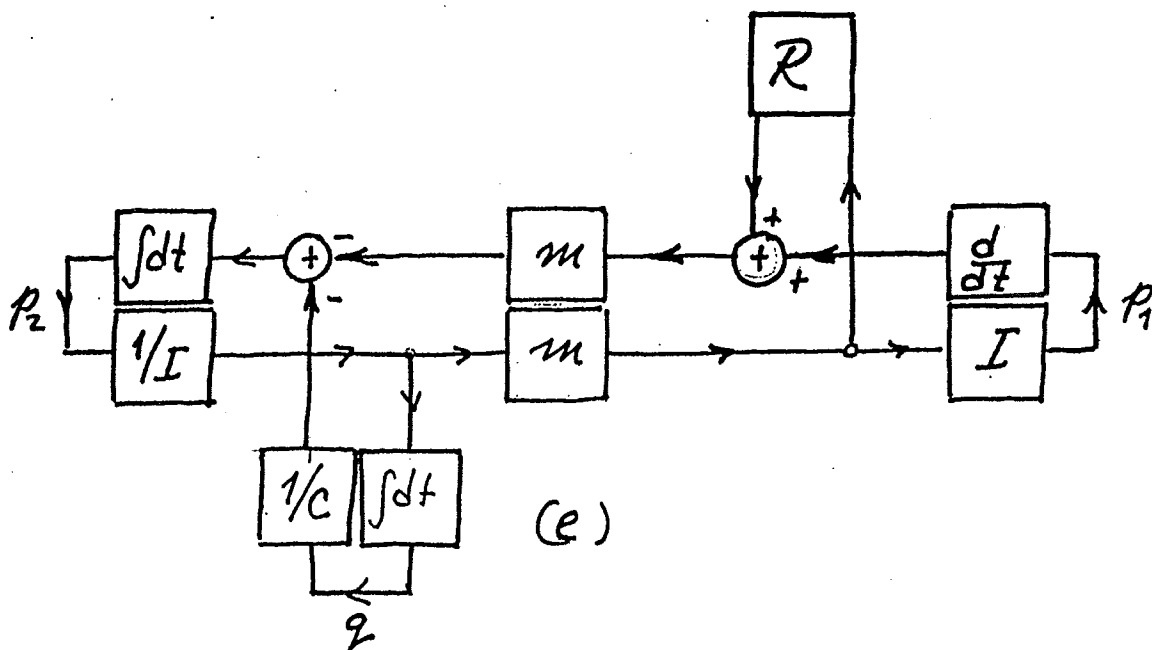
3-5



(c)

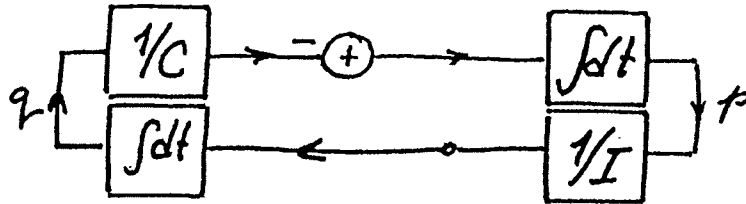


(d)



(e)

3-13



3-6

3-14

$$\frac{\tau}{\omega} \rightarrow TF \frac{F}{V}$$

$$\tau = rF$$

$$r\omega = V$$

3-15 $\frac{e}{i} \rightarrow G \ddot{y} \frac{F}{V} \rightarrow I$

$$e = TV = T \frac{p}{I} = \frac{T \int F dt}{I}$$

$$= \frac{T}{I} \int T i dt = \frac{T^2}{I} \int i dt = \frac{T^2}{I} q$$

so $\frac{1}{C} \leftrightarrow \frac{T^2}{I}$

3.16

$$P = P_0 \frac{V_0^r}{V^r} = \frac{P_0 V_0^r}{(V_0 - A_p x)^r} = \frac{P_0}{\left[1 - \frac{A_p x}{V_0}\right]^r}$$

3-7

P is absolute pressure:

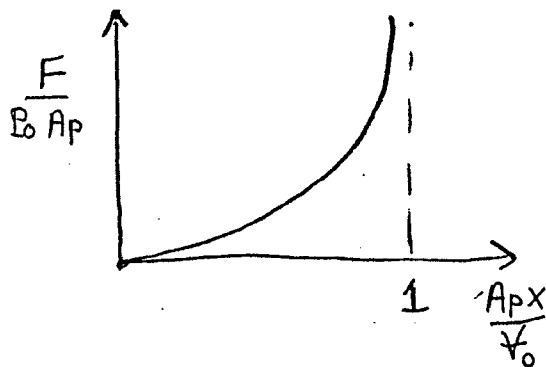
$P - P_0$ is gage pressure in cylinder,

$$(P - P_0) A_p = F$$

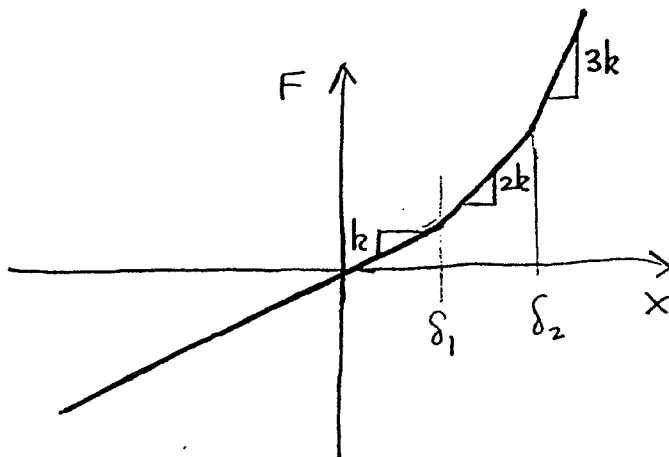
or

$$F = \left\{ \frac{P_0}{\left[1 - \frac{A_p x}{V_0}\right]^r} - P_0 \right\} A_p$$

$$F = P_0 A_p \left[\frac{1}{\left(1 - \frac{A_p x}{V_0}\right)^r} - 1 \right]$$

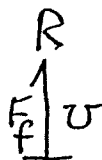


3-17



3-18

a)

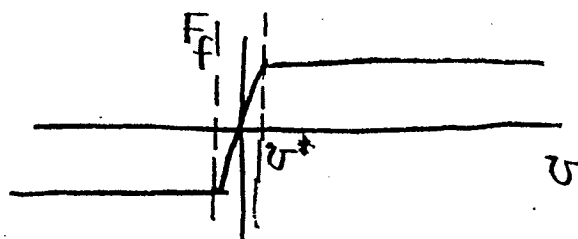


The only possible causality for the friction model shown in the problem is "effort" out, "flow" in.

For any specified velocity, F_f can be computed, but if F_f is specified, v is indeterminate.

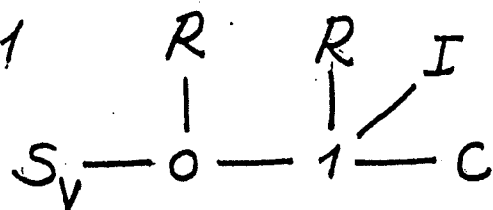
- (b) If used simply as $F_f = \mu N$, then F_f will be applied when $v=0$, which is not correct. When $v=0$, the mass "sticks", and the friction force exactly balances all other forces on the mass. When the other forces exceed the "stick" force, then F_f returns to $F_f = \mu N$.

A possible change in the constitutive law might be,

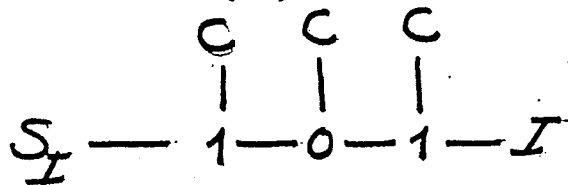


For v^* very small, the fundamental character of friction is maintained.

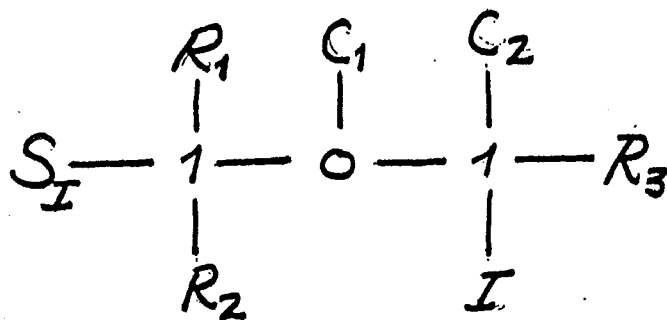
4-1



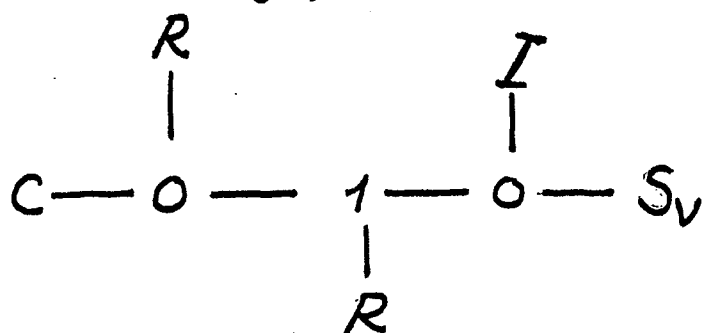
(a)



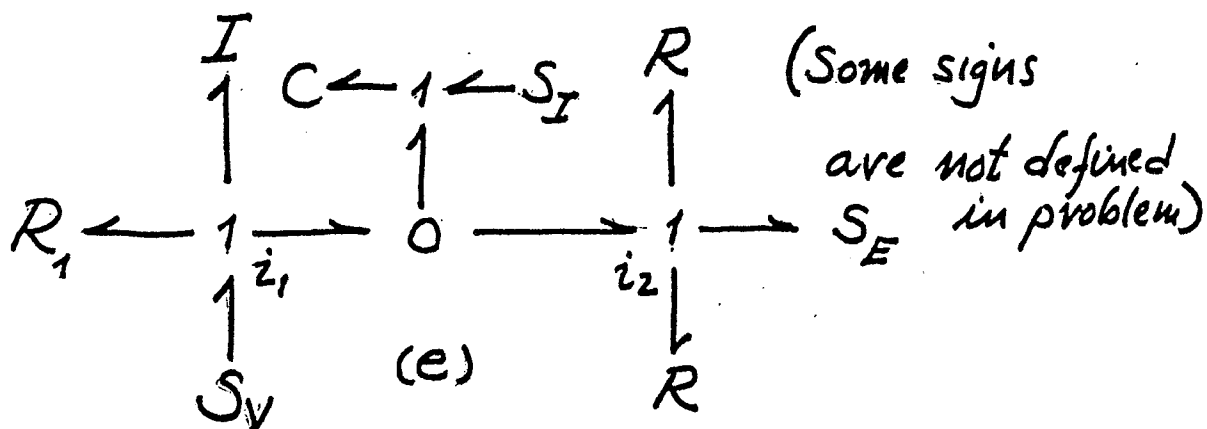
(b)



(c)

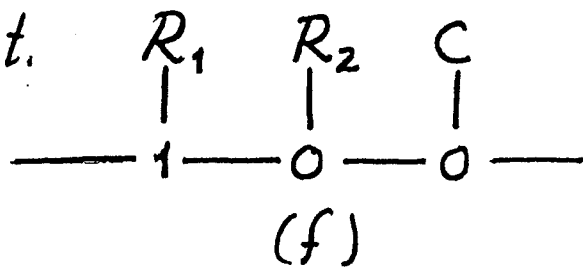


(d)

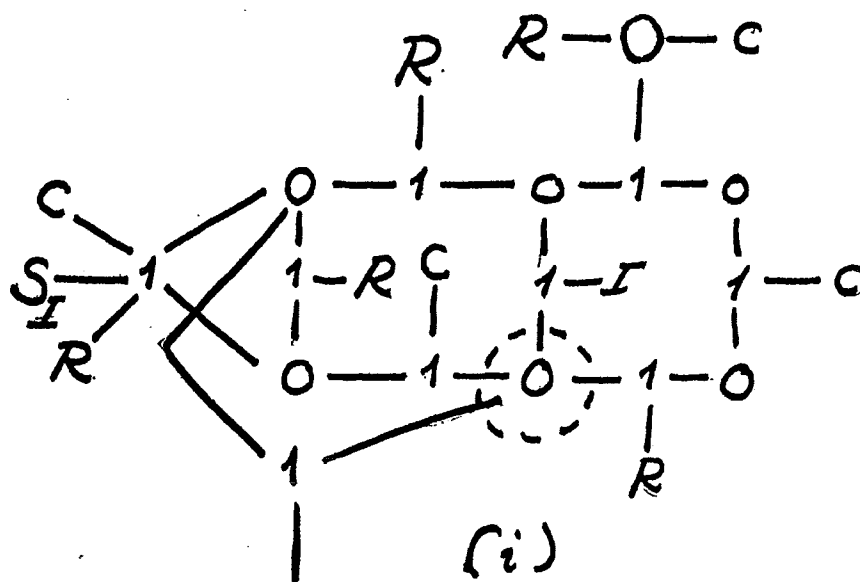
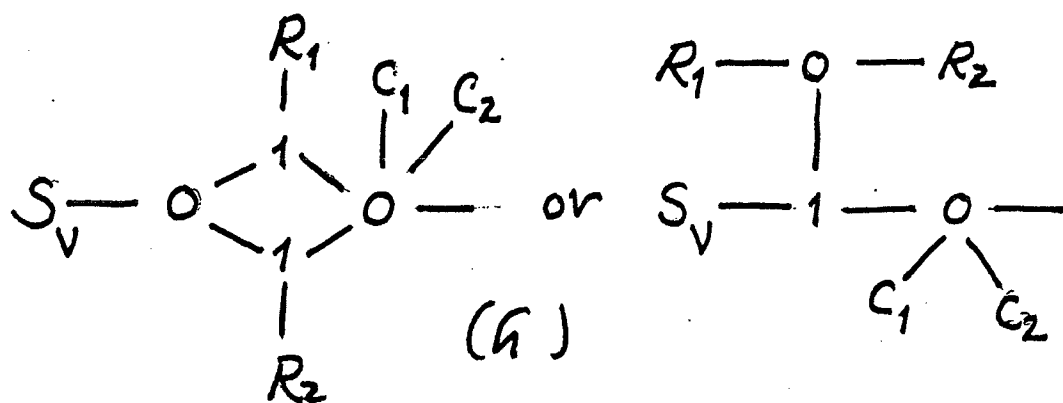
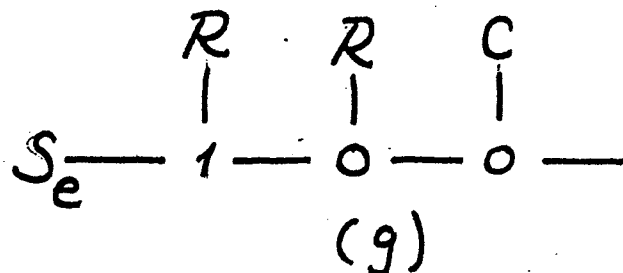


(e)

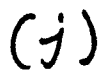
4-1 Cont.



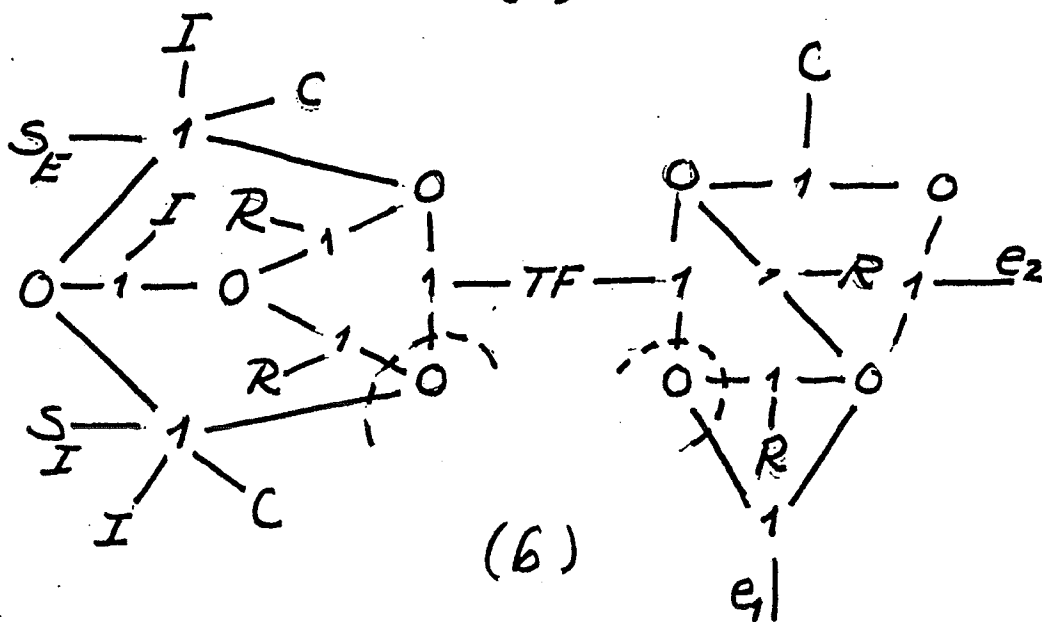
4-2



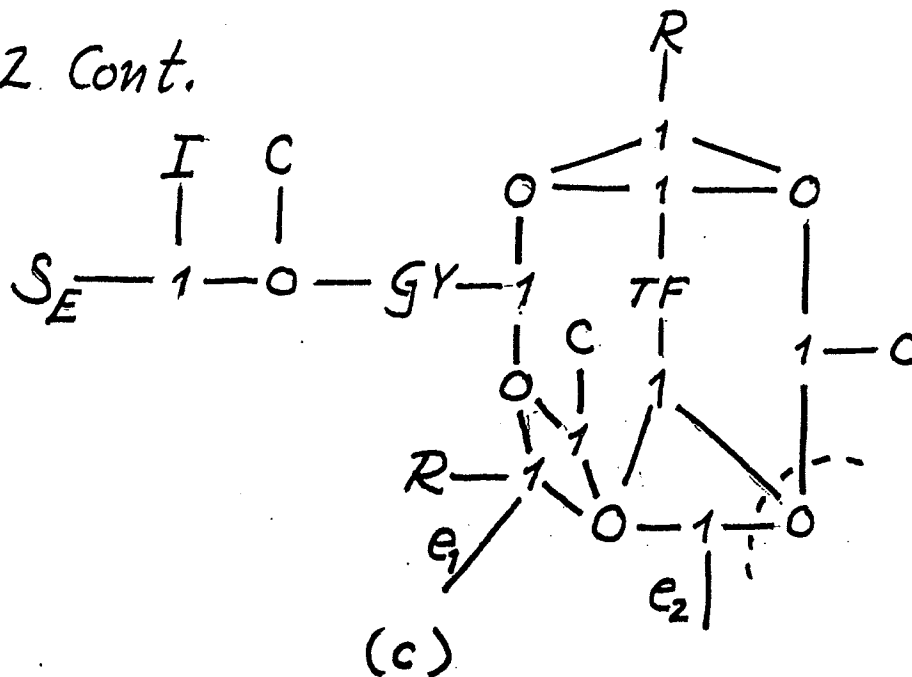
4-3


$$\begin{array}{ccccccc} Z & & R & C & & & R & I \\ & & | & | & & & | & / \\ S_E - & 1 & - & 0 & - & TF & - & 1 & - & C \end{array}$$

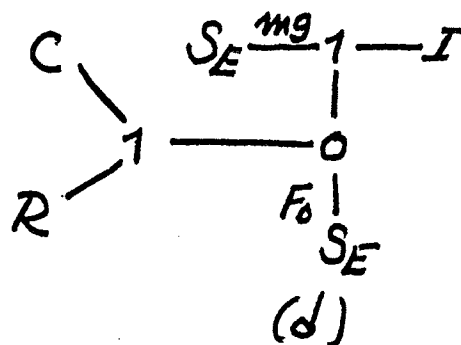
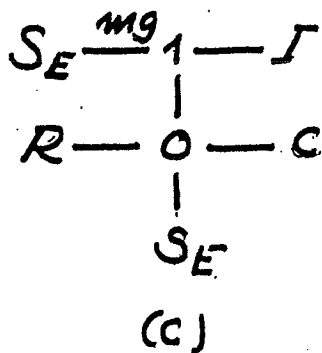
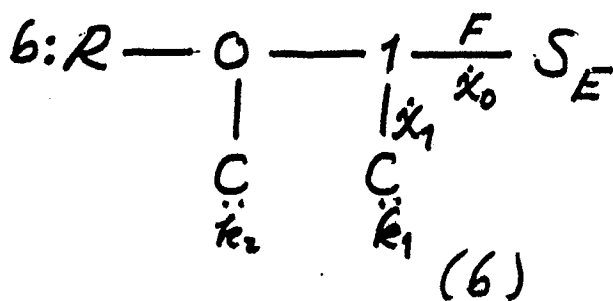
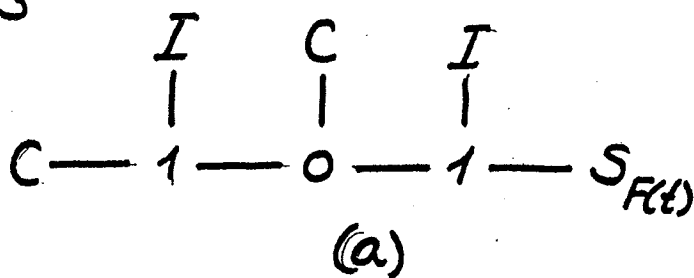
(a)



4-2 Cont.

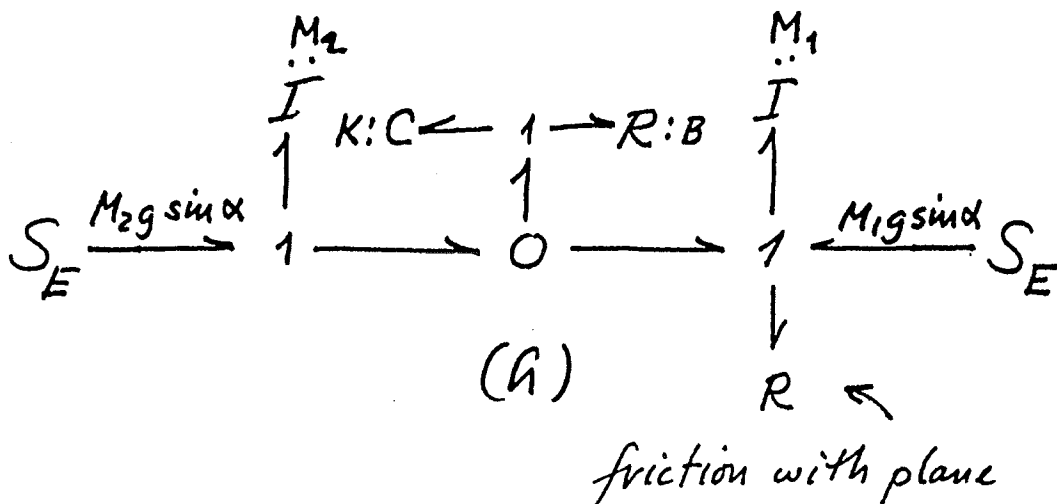
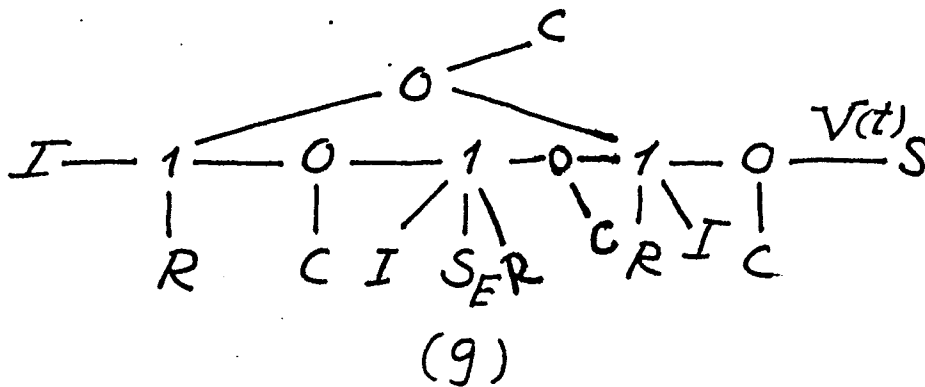
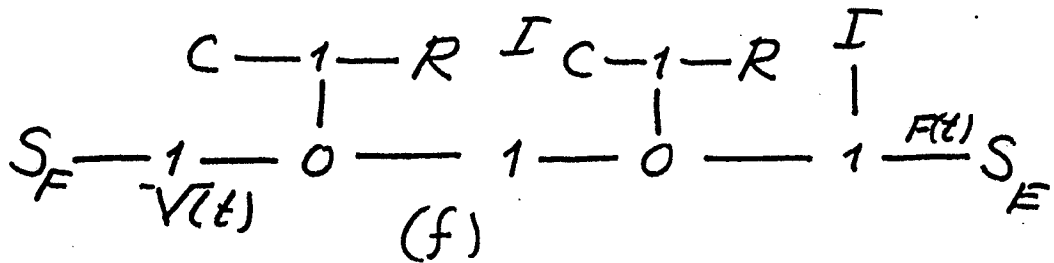
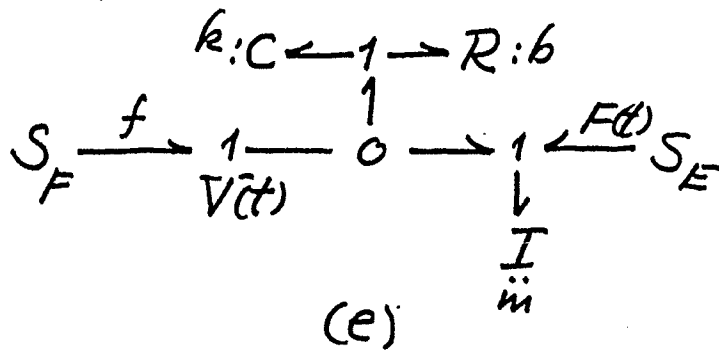


4-3

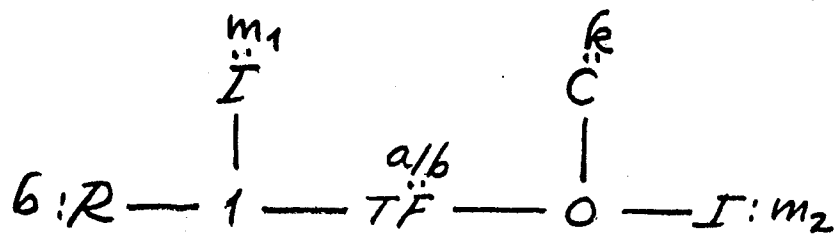


4-3 Cont.

4-5

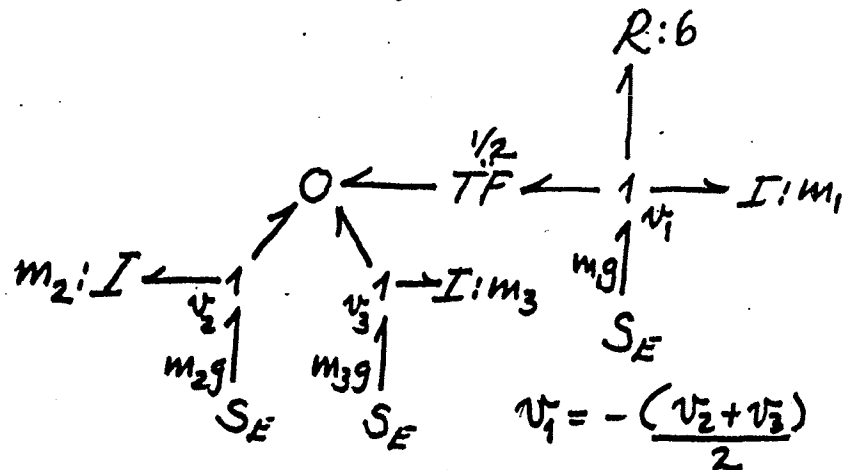


4-4

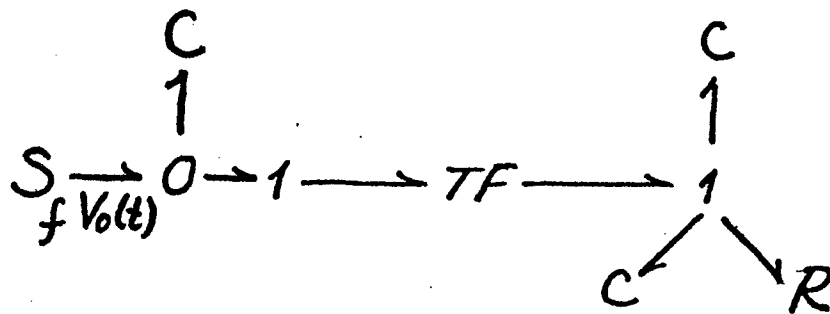


4-6

(a)

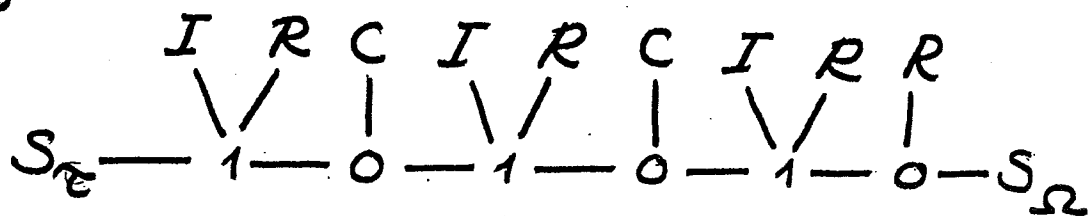


(6)



(c)

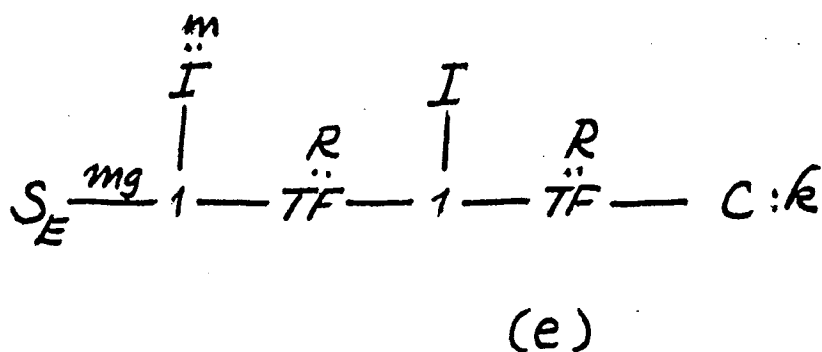
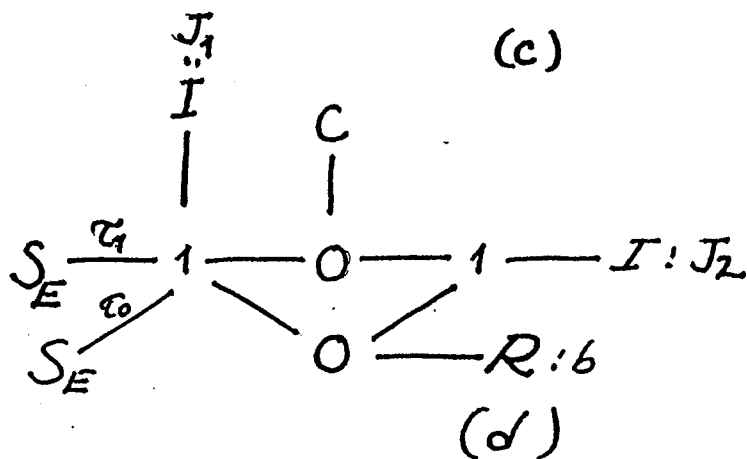
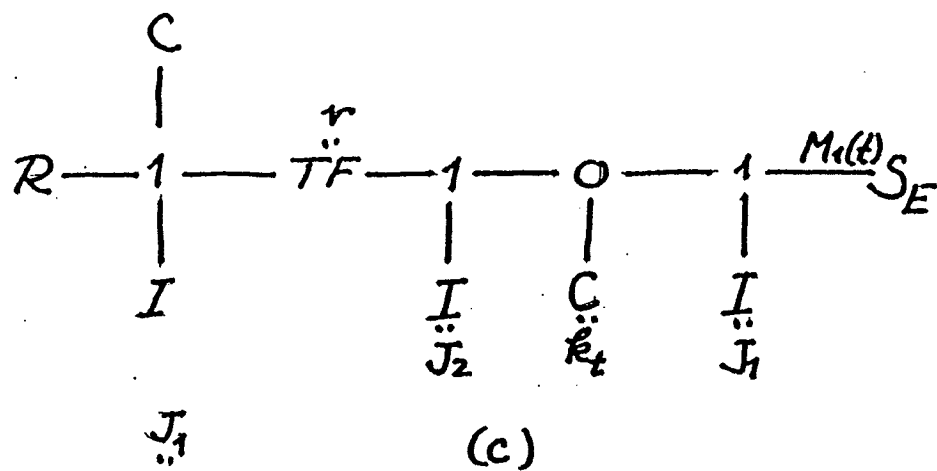
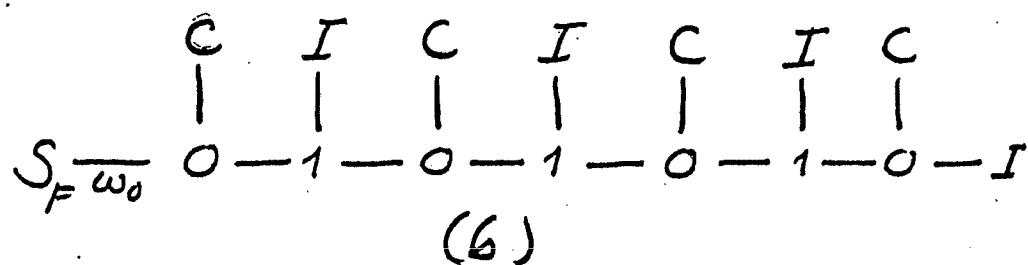
4-5

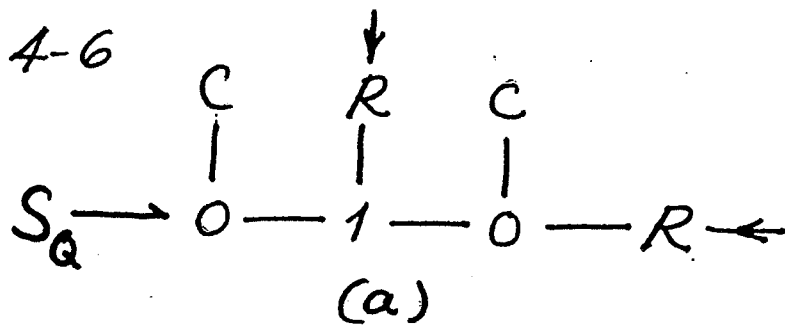


(a)

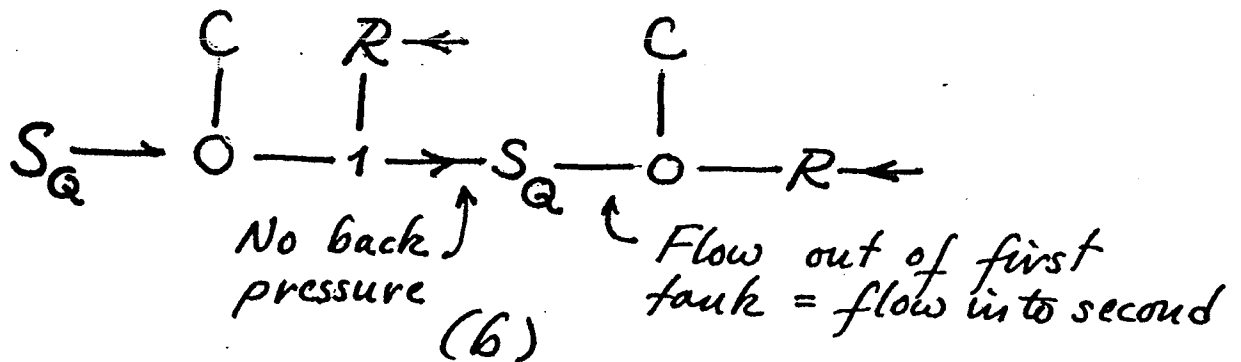
4-5 Cont.

4-7

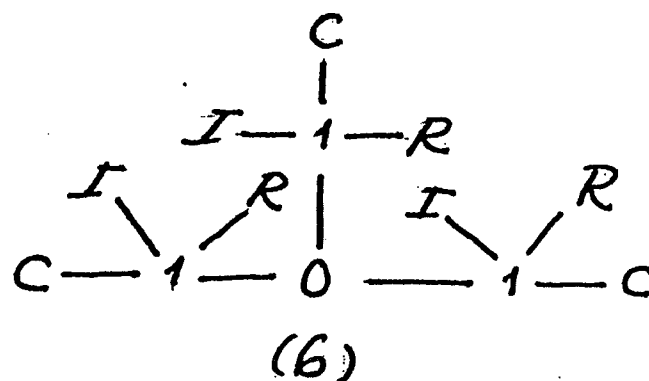
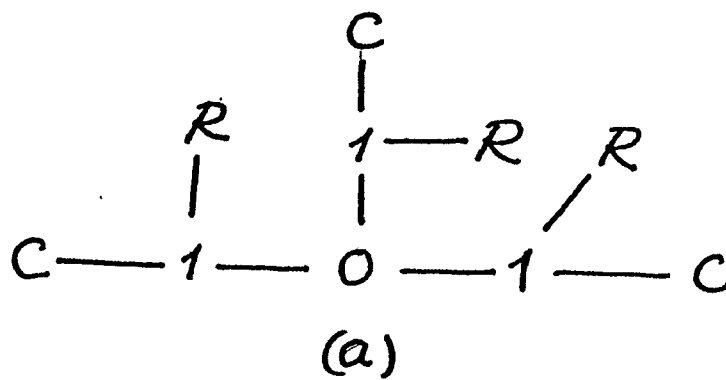




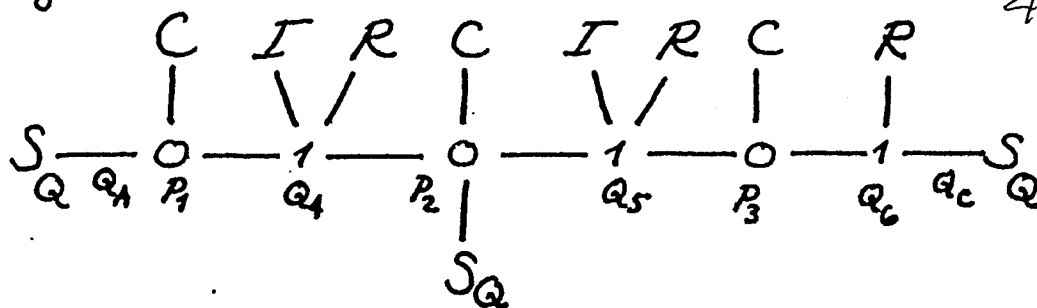
4-8



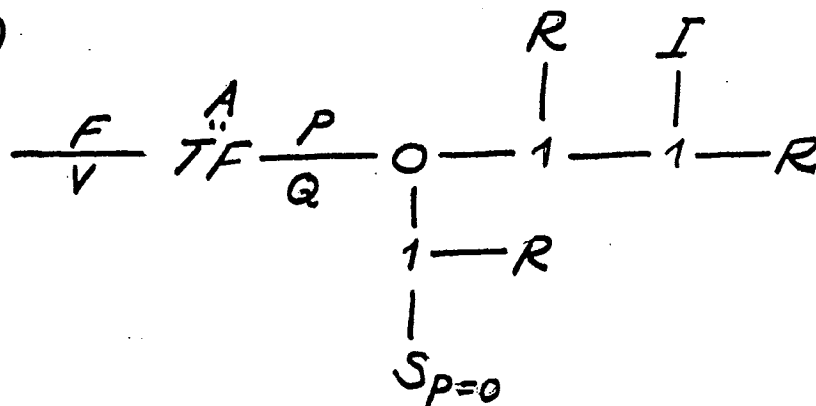
4-7



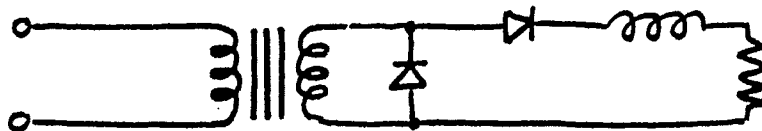
4-8



4-9



(a)



(b)

4-10

$$\frac{\tau}{\omega} \rightarrow GY \rightarrow \frac{e}{i} R$$

(a)

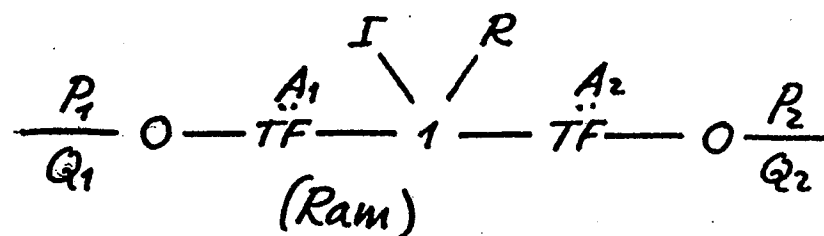
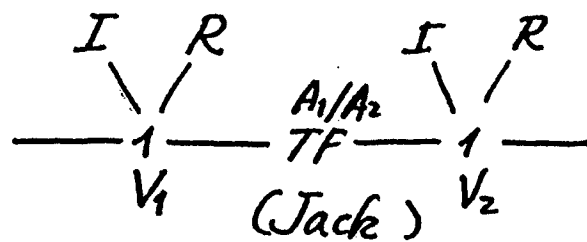
(b) If $R \rightarrow 0$, $e \rightarrow 0$ but $e = T\omega$ so $\omega \rightarrow 0$

If bulb shorts, there still is coil resistance

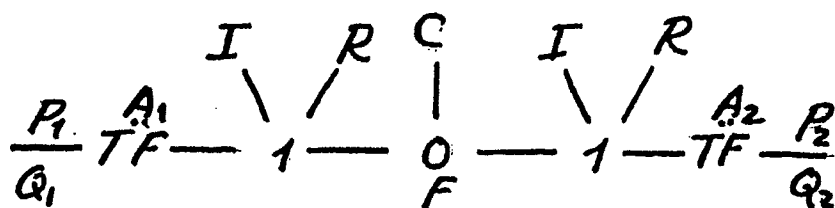
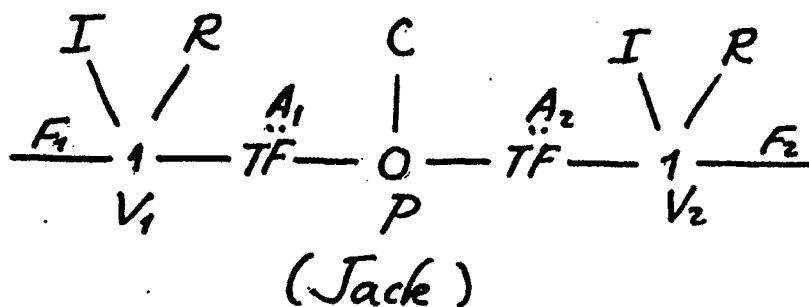
$$\frac{\tau}{\omega} \rightarrow GY \rightarrow \overset{R_{coil}}{\underset{1}{\rightarrow}} R_{bulb}$$

4-11

4-10

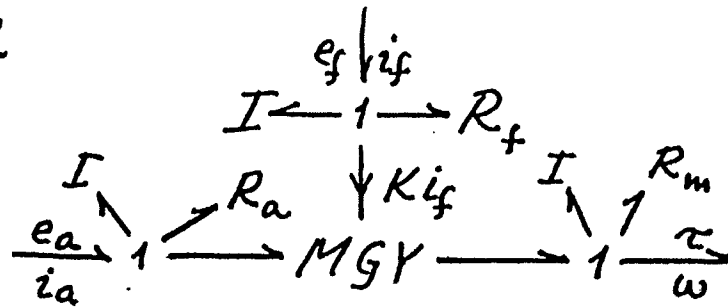


(a)



(b)

4-12



4-11

(a, b)

(c) No change.

$$(d) \quad \eta_{\text{motor}} = \frac{\tau \omega}{e_a i_a + e_f i_f}$$

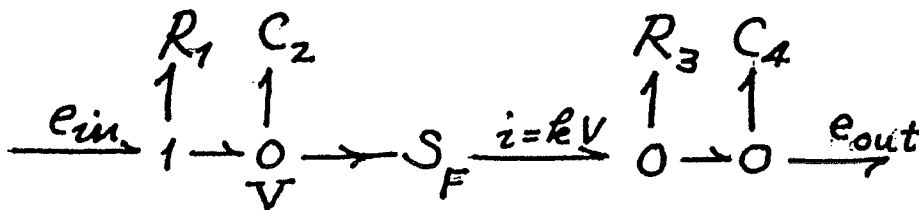
$$\eta_{\text{generator}} = \frac{e_a i_a}{\tau \omega + e_f i_f}$$

For steady operation, I 's play no role

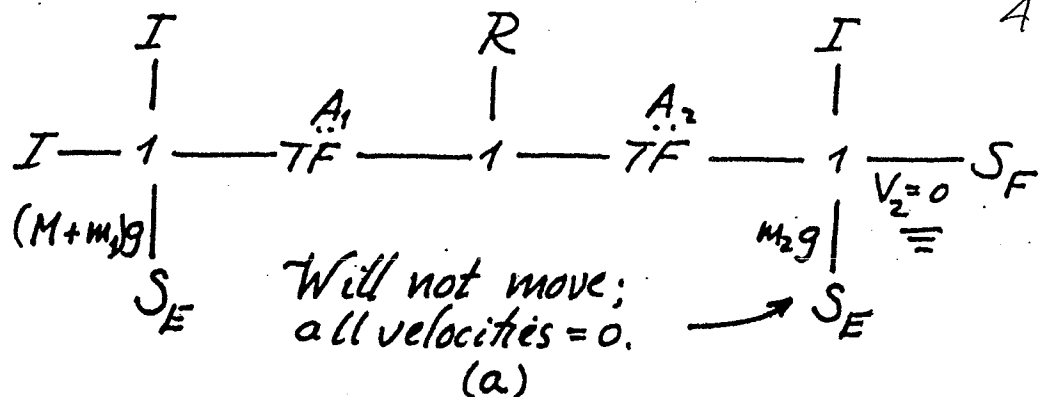
$$i_a = \frac{e_a - K i_f \omega}{R_a}, \quad i_f = e_f / R_f$$

$$\tau = K i_f i_a - R_m \omega$$

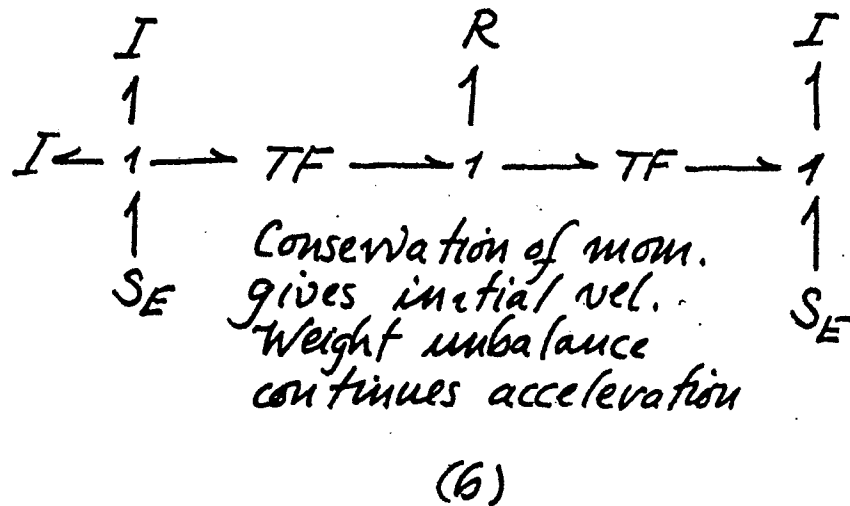
4-13



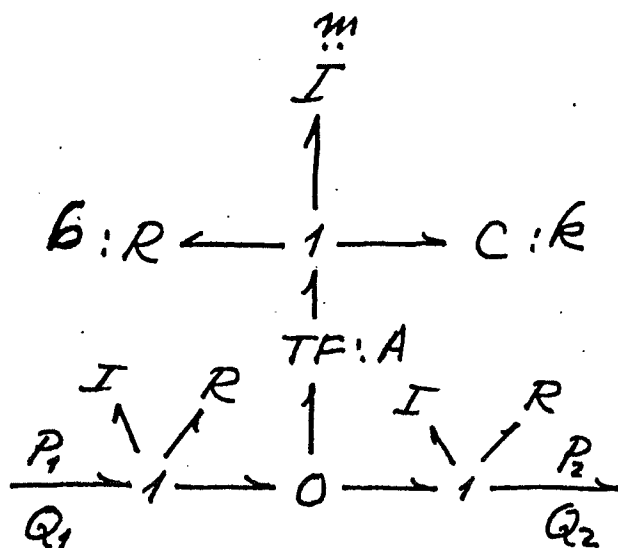
4-14



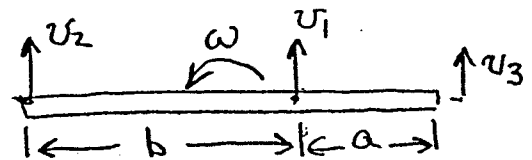
4-12



4-15



4-16 look at the lever,

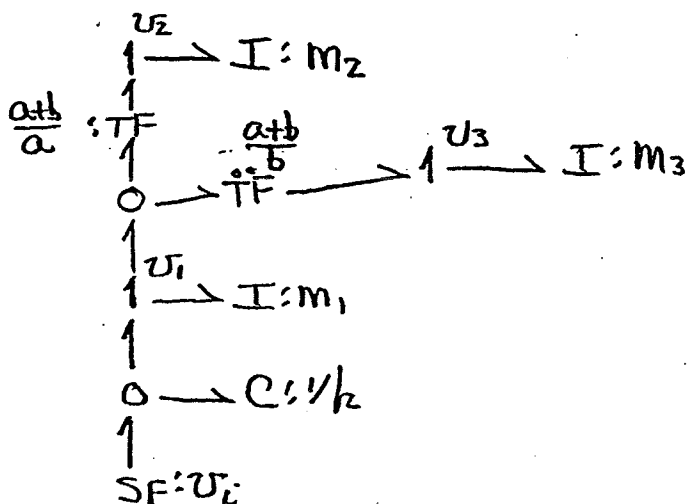


kinematics:

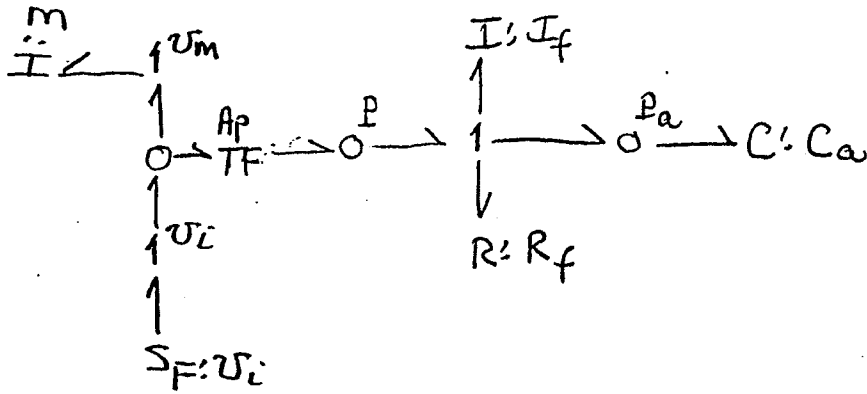
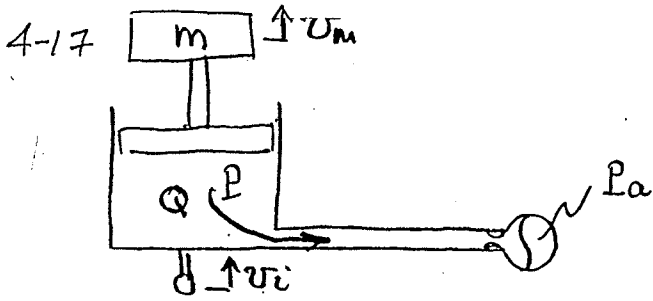
$$\omega = \frac{v_3 - v_2}{(a+b)}$$

$$v_3 = v_1 + a \left[\frac{v_3 - v_2}{(a+b)} \right]$$

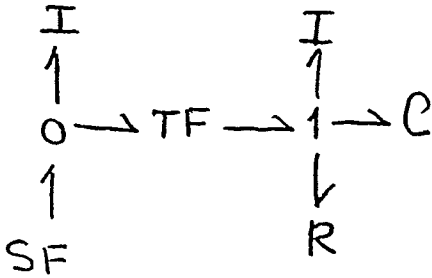
$$v_1 = \frac{b}{a+b} v_3 + \frac{a}{a+b} v_2 \quad \text{kinematic constraint}$$



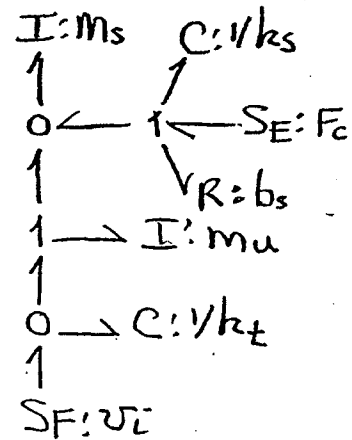
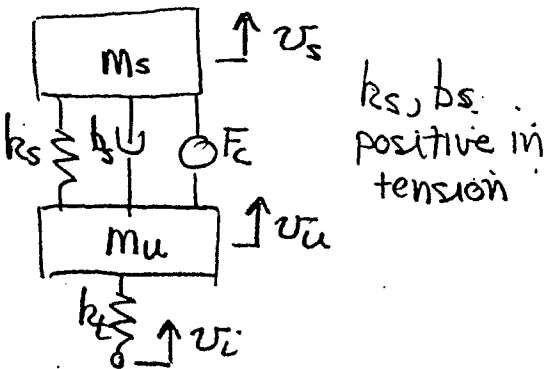
4-14



can be simplified to:



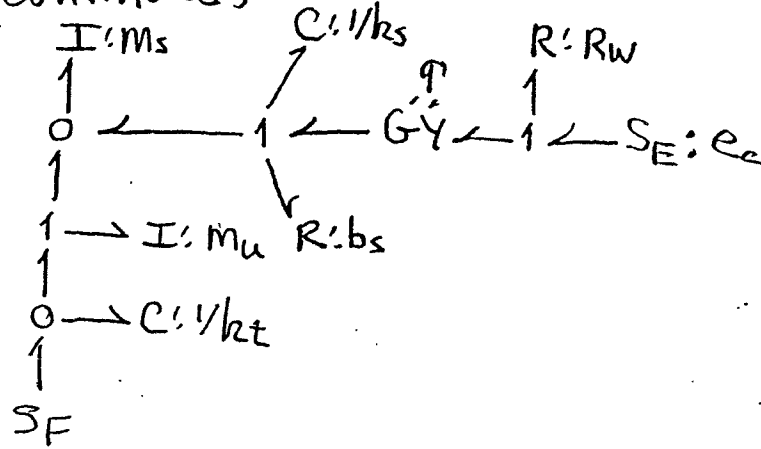
4-18



(a)

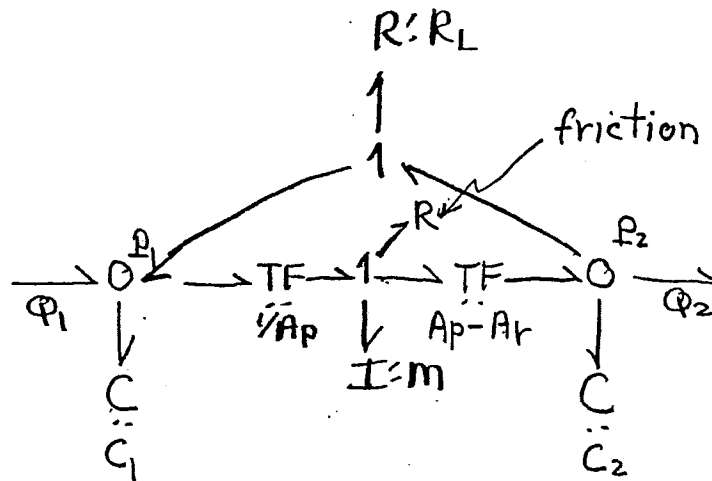
4-18
(b)

(continued)

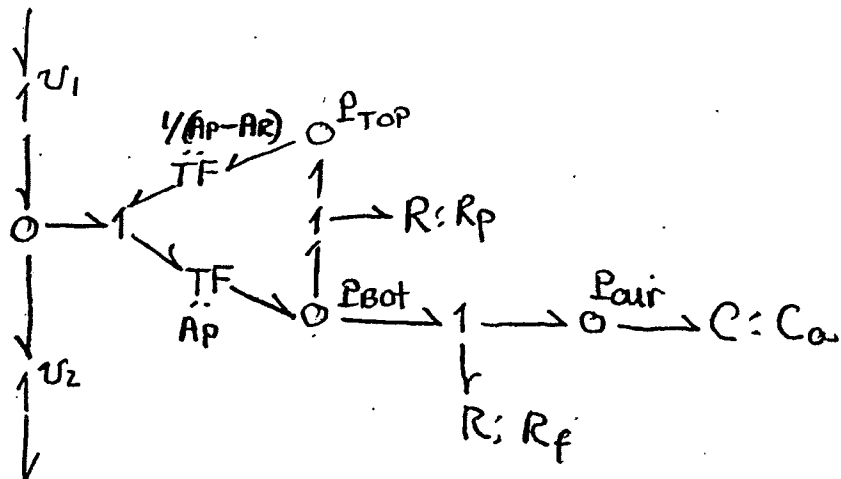


4-15

4-19

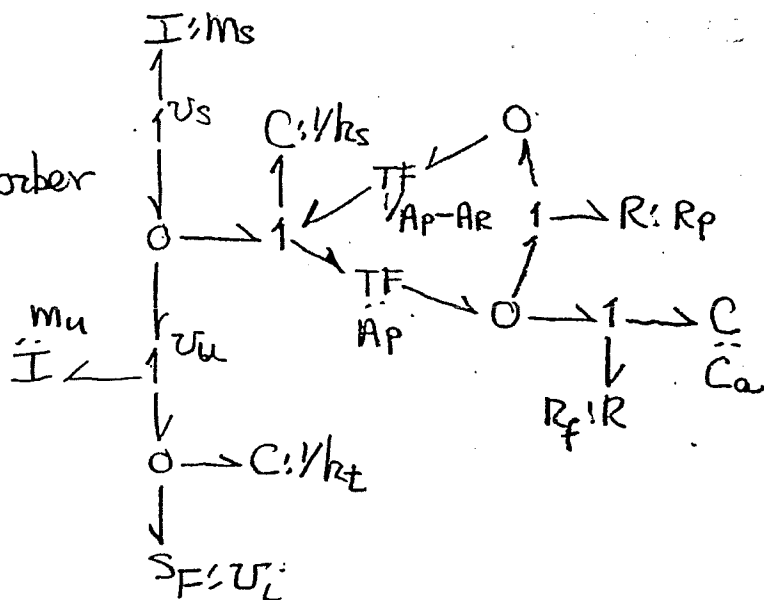
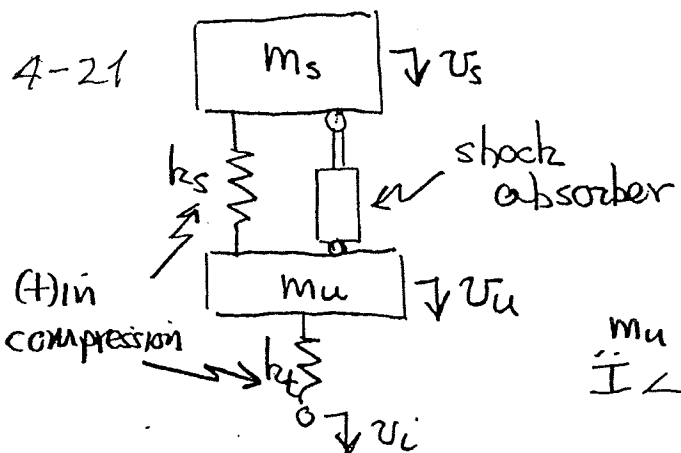


4-20



Realistic that might be included:

- (1) compliance of the top + bottom chambers
- (2) nonlinear resistances for the piston and foot valve
- (3) seal friction
- (4) leakage

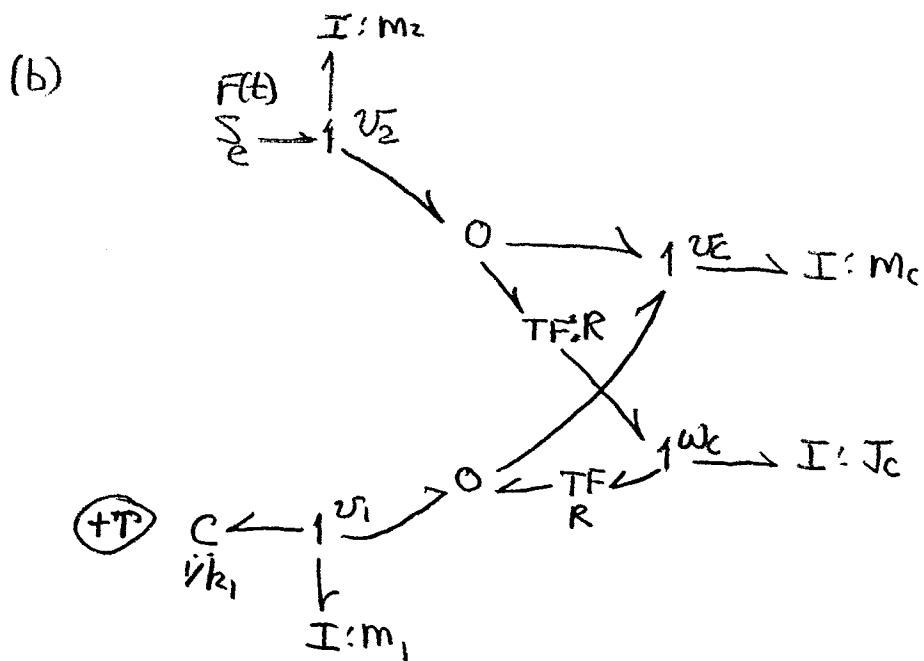


4-23

(a) kinematic constraints using $\underline{v}_P = \underline{v}_O + \underline{\omega} \times \underline{r}_{P/O}$

$$v_2 = v_C + R\omega$$

$$v_1 = v_C - R\omega$$



36
4-24

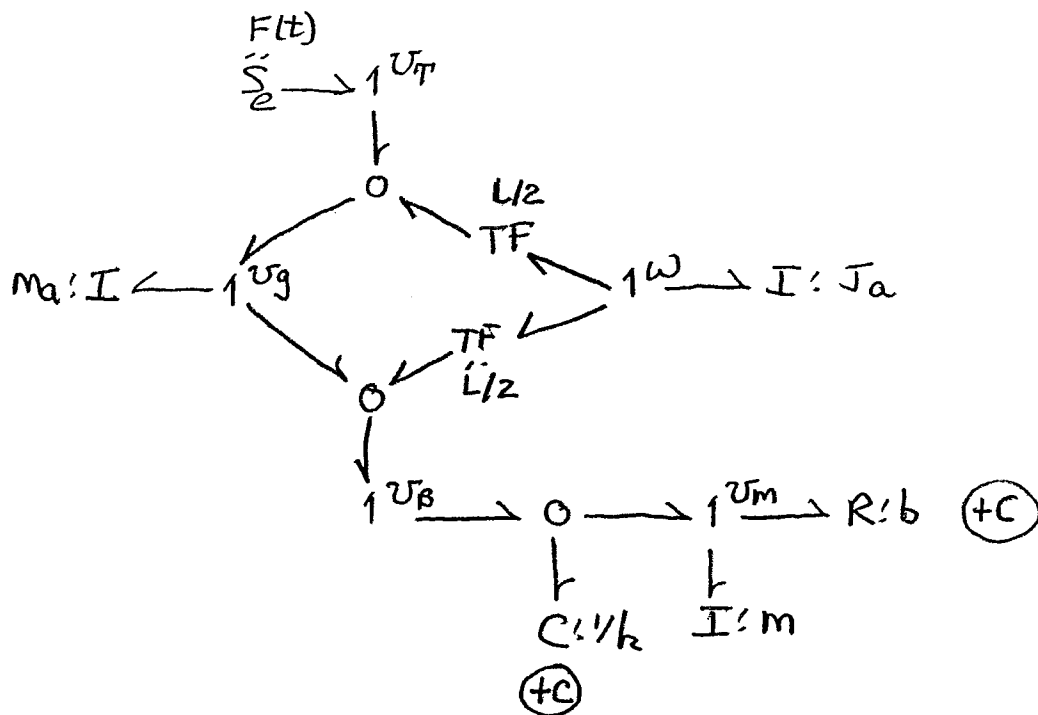
4-19

(a) kinematic constraints using $\underline{v}_P = \underline{v}_O + \underline{\omega} \times \underline{r}_{P/O}$

$$v_T = v_g - \frac{L}{2} \omega$$

$$v_B = v_g + \frac{L}{2} \omega$$

(b)



4-20

(a) $\omega_n = dc / D(IL)^{1/2}$, $f_n = dc / 2\pi D(IL)^{1/2}$

4-25

(b) $f_n = 343 / 2\pi = 54.6 \text{ Hz}$ This frequency is near the lower limit of audibility for humans.

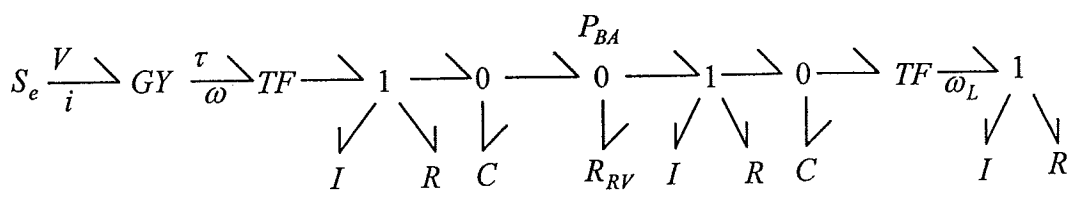
$C = V_o (1/B + 2r_o / t_w E)$; $B = 1.52 \times 10^9$; $E = 2.3 \times 10^9$;

4-26

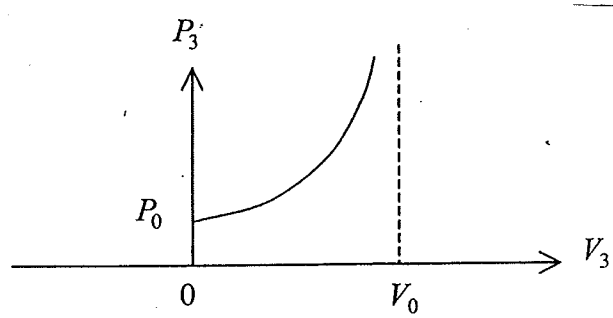
$V_o / B = 6.58 \times 10^{-10} V_o [m^5 / N]$; $V_o (2r_o / t_w E) = 2.61 \times 10^{-9} V_o [m^5 / N]$

In this example, the hose flexibility is the major contributor to the hydraulic compliance.

4-27



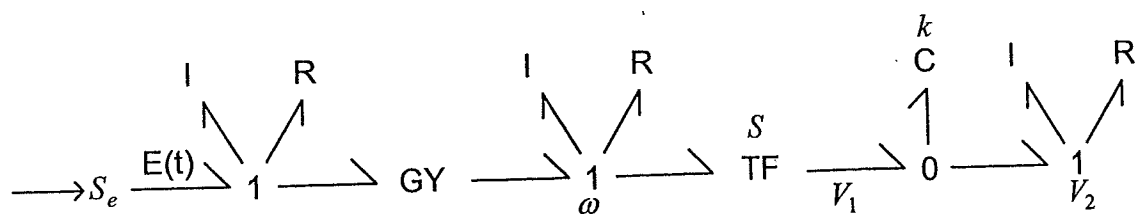
4-28

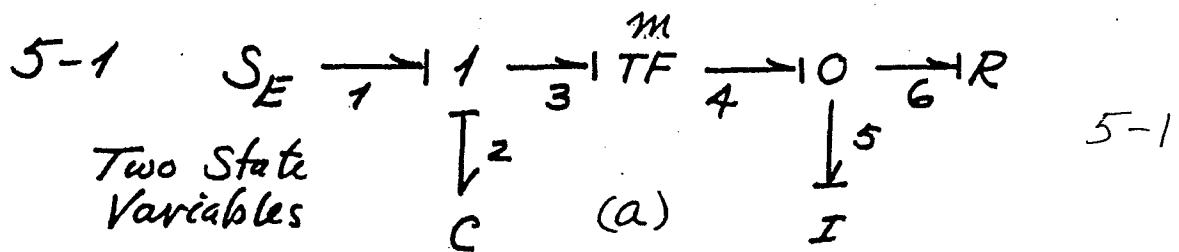


4-29

(a) $I_{eq} = (m + \rho LA^2 / a)$, (b) $a = \rho LA^2 / m$, $r = 15 [mm]$.

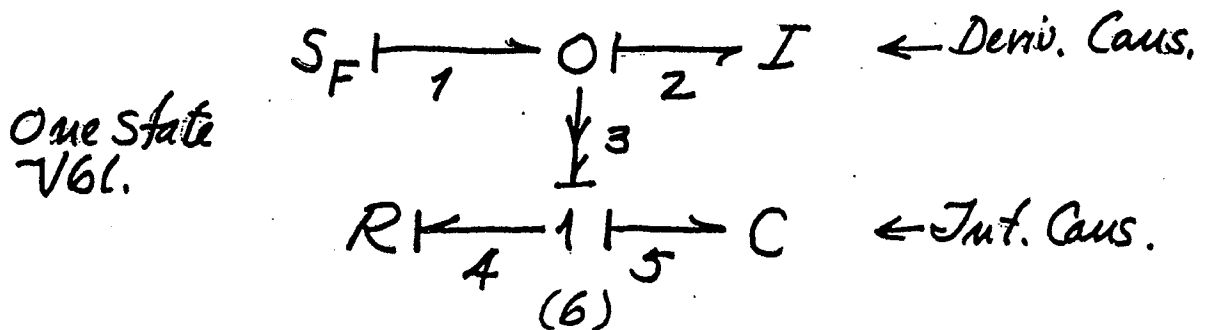
4-30





$$\dot{p}_5 = e_4 = m e_3 = m (E_1(t) - e_2) = \underline{m (E_1(t) - q_2/C_2)}$$

$$\begin{aligned} \dot{q}_2 &= f_3 = m f_4 = m (f_5 + f_6) = m (p_5/I_5 + e_6/R_6) \\ &= m (p_5/I_5 + m e_3/R_6) = m \left[\frac{p_5}{I_5} + m \frac{(E_1(t) - e_2)}{R_6} \right] \\ &= \underline{m \left[\frac{p_5}{I_5} + \frac{m (E_1(t) - q_2/C_2)}{R_6} \right]} \end{aligned}$$

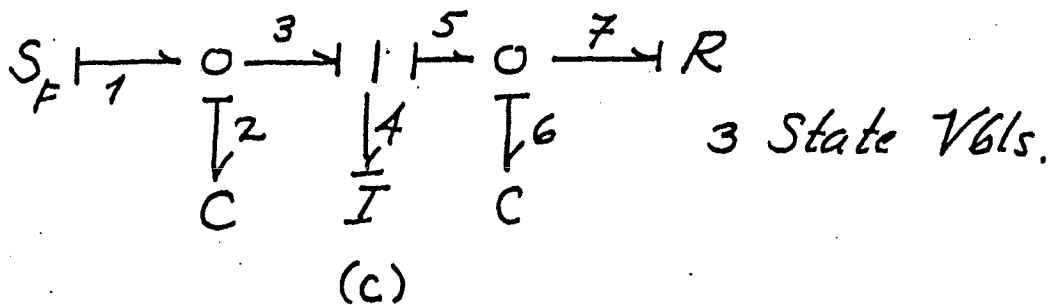


$$\begin{aligned} \dot{q}_5 &= f_4 = e_4/R_4 = (e_3 - e_5)/R_4 \\ &= \frac{e_2 - q_5/C_5}{R_4} = \frac{dp_2/dt - q_5/C_5}{R_4} \\ &= \frac{d(I_2 f_2)/dt - q_5/C_5}{R_4} \end{aligned}$$

$$\dot{q}_5 = \frac{I_2 d F_1(t)/dt - q_5/C_5}{R_4}$$

5-1 Cont.

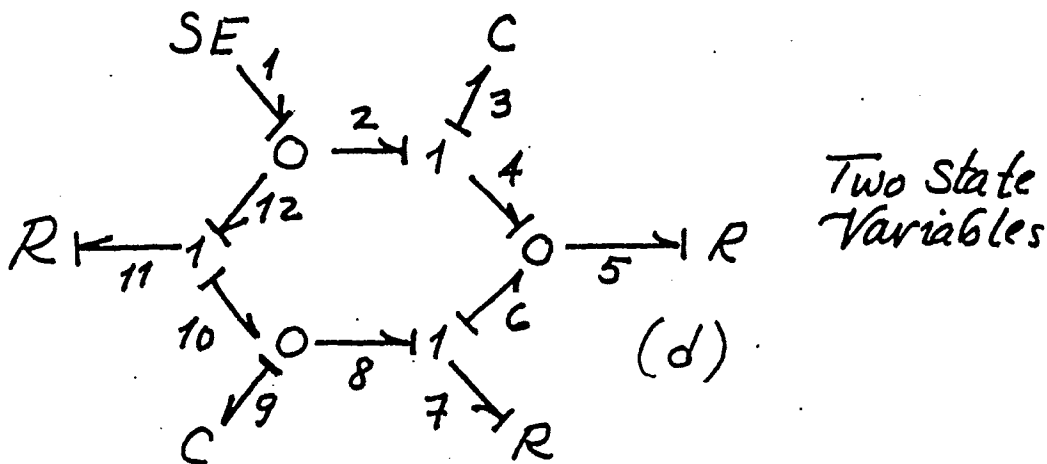
5-2



$$\dot{q}_2 = f_1(t) - f_3 = f_1(t) - f_4 = \frac{f_1(t) - p_4 / I_4}{C_2}$$

$$\dot{p}_4 = e_3 - e_5 = e_2 - e_6 = \frac{q_2 / C_2 - q_6 / C_6}{R_7}$$

$$\dot{q}_6 = f_5 - f_7 = f_4 - e_7 / R_7 = \frac{p_4 / I_4 - e_6}{R_7} = \frac{p_4 / I_4 - q_6 / C_6}{R_7}$$



$$\begin{aligned} \dot{q}_3 &= f_4 = f_5 - f_6 = \frac{e_5}{R_5} - f_7 = \frac{e_4}{R_5} - \frac{e_7}{R_7} = \frac{e_2 - e_3}{R_5} - \left(\frac{e_8 - e_6}{R_7} \right) \\ &= \frac{E_1(t) - q_3 / C_3}{R_5} - \left(\frac{e_9 - e_4}{R_7} \right) = \frac{E_1 - q_3 / C_3}{R_5} - \left(\frac{q_9 / C_9 - e_2 + e_3}{R_7} \right) \\ &= \frac{E_1(t) - q_3 / C_3}{R_5} - \left(\frac{q_9 / C_9 - E_1(t) + q_3 / C_3}{R_7} \right) \end{aligned}$$

$$\dot{q}_9 = \frac{E_1(t) - q_9 / C_9}{R_{11}} - \left(\frac{q_9 / C_9 - E_1(t) + q_3 / C_3}{R_7} \right)$$

$$5-2 \quad \begin{array}{c} 1 \xrightarrow{\quad} 1 \xrightarrow{\quad 3 \quad} R \\ \quad \quad \quad \downarrow 2 \\ \quad \quad \quad R \end{array} \quad e_1 = e_2 + e_3 = R_2 f_2 + R_3 f_3$$

$$(a) \quad \quad \quad = (R_2 + R_3) f_1$$

$$R_{eq} = R_2 + R_3$$

5-3

$$\text{or } e_1 = \Phi_{eq}(f_1), \quad \Phi_{eq}(f_1) = \Phi_{R_1}(f_1) + \Phi_{R_2}(f_1)$$

$$(6) \quad \begin{array}{c} 1 \xrightarrow{\quad} 0 \xrightarrow{\quad 3 \quad} 1 \xrightarrow{\quad 5 \quad} C \\ \quad \quad \quad \downarrow 2 \quad \quad \quad \downarrow 4 \\ \quad \quad \quad C \quad \quad \quad C \end{array}$$

$$e_1 = e_3 = e_4 + e_5 = q_4/C_4 + q_5/C_5 \quad (1)$$

now find q_4, q_5 in terms of q_1

$$\begin{aligned} \dot{q}_4 &= \dot{q}_3 = \dot{q}_1 - \dot{q}_2 = \dot{q}_1 - \frac{d}{dt} C_2 e_2 = \dot{q}_1 - \frac{d}{dt} C_2 e_3 \\ &= \dot{q}_1 - C_2 \frac{d}{dt} \left(\frac{q_4}{C_4} + \frac{q_5}{C_5} \right) \end{aligned}$$

$$\dot{q}_5 = \dot{q}_4; \quad \text{integrating}$$

$$q_4 = q_1 - C_2 \left(\frac{q_4}{C_4} + \frac{q_5}{C_5} \right) + \text{const.}^0$$

$$q_5 = q_1 - C_2 \left(\frac{q_4}{C_4} + \frac{q_5}{C_5} \right) + \text{const.}^0$$

solving,

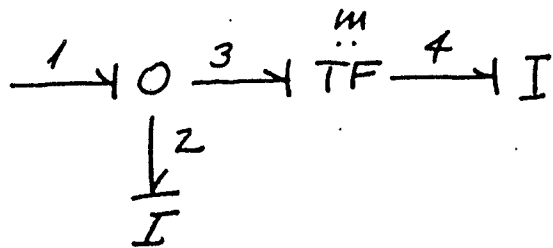
$$q_4 = q_5 = \frac{C_4 C_5}{C_4 C_5 + C_2 C_4 + C_2 C_5} q_1$$

then, from (1) we eventually conclude

$$e_1 = \underbrace{\frac{C_4 + C_5}{C_4 C_5 + C_4 C_2 + C_5 C_2}}_{1/C_{eq}} \cdot q_1 \quad \text{Wow!}$$

5-2

Continued



5-4

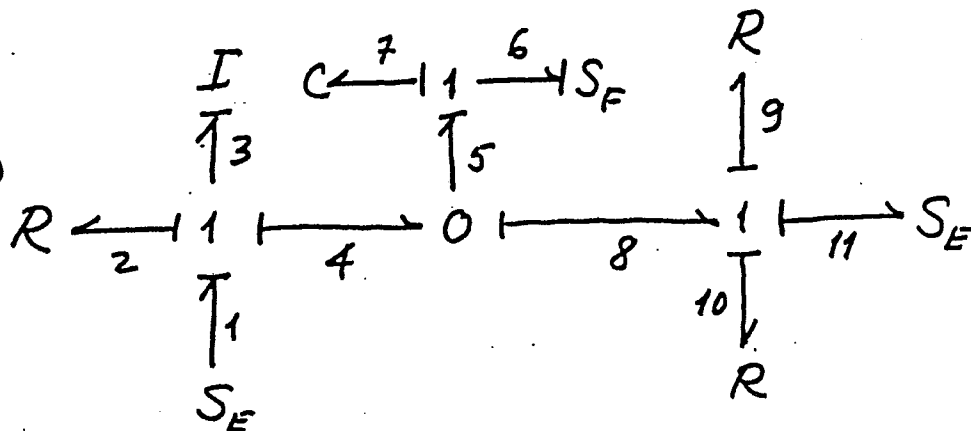
$$f_1 = f_3 + f_2 = \frac{p_2}{I_2} + m \frac{p_4}{I_4}$$

$$\left. \begin{array}{l} \dot{p}_2 = e_1 \\ \dot{p}_4 = m e_1 \end{array} \right\} \begin{array}{l} p_2 = p_1 \\ p_4 = m p_1 \end{array} \quad \text{so } f_1 = \frac{p_1}{I_2} + m m \frac{p_1}{I_4}$$

$$I_{eq} = \left[\frac{1}{I_2} + \frac{m^2}{I_4} \right]^{-1}$$

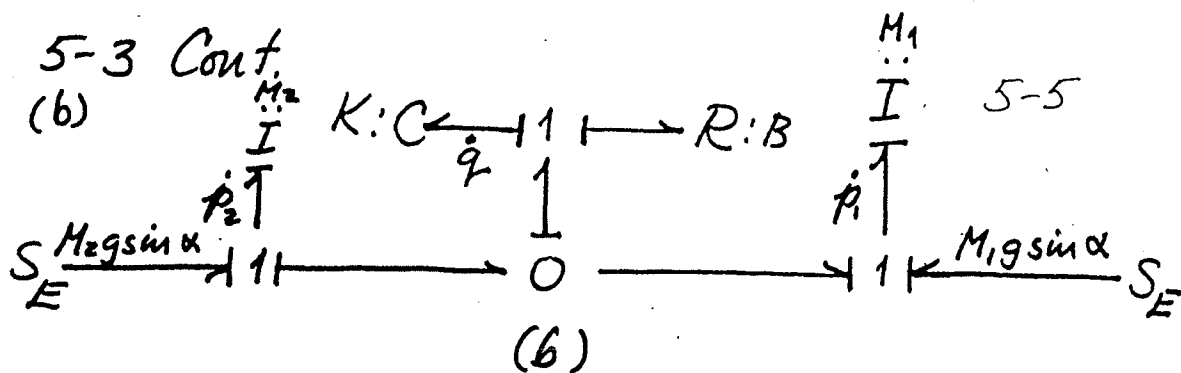
5-3

(a)



$$\dot{p}_3 = -\frac{R_2}{I_3} p_3 + E_1(t) - E_{11}(t) + (R_9 + R_{10})(F_6(t) - \frac{p_3}{I_3})$$

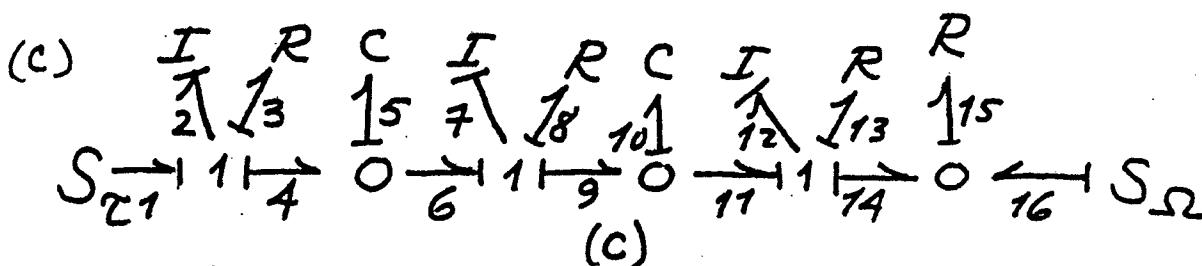
$$\dot{q}_7 = F_6(t)$$



$$\dot{p}_1 = M_1 g \sin \alpha + Kq + B\left(\frac{p_2}{M_2} - \frac{p_1}{M_1}\right)$$

$$\dot{q} = \frac{p_2}{M_2} - \frac{p_1}{M_1}$$

$$\dot{p}_2 = M_1 g \sin \alpha - Kq - B\left(\frac{p_2}{M_2} - \frac{p_1}{M_1}\right)$$



$$\dot{p}_2 = \tau_1(t) - R_3 \frac{p_2}{J_2} - K_5 q_5$$

$$\dot{q}_5 = \frac{p_2}{J_2} - \frac{p_7}{J_7}$$

$$\dot{p}_7 = K_5 q_5 - R_8 \frac{p_7}{J_7} - K_{10} q_{10}$$

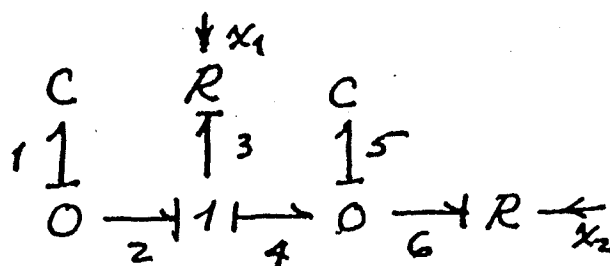
$$\dot{q}_{10} = \frac{p_7}{J_7} - \frac{p_{12}}{J_{12}}$$

$$\dot{p}_{12} = K_{10} q_{10} - R_{13} \frac{p_{12}}{J_{12}} - R_{15} \left(\frac{p_{12}}{J_{12}} + \Omega_{16}(t) \right)$$

5-3 Cont.

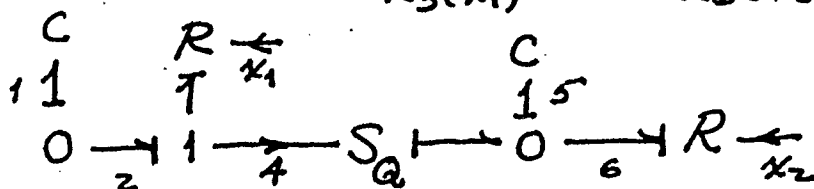
5-6

(d)



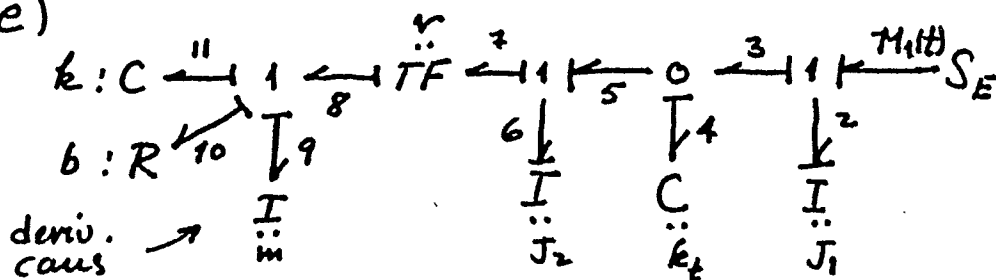
$$\dot{q}_1 = - \frac{(q_1/C_1 - q_5/C_5)}{R_3(x_1)}$$

$$\dot{q}_5 = \frac{q_1/C_1 - q_5/C_5}{R_3(x_1)} - \frac{q_5/C_5}{R_6(x_2)}$$



$$\dot{q}_1 = - \frac{q_1/C_1}{R_3(x_1)} ; \quad \dot{q}_5 = \frac{q_1/C_1}{R_3(x_1)} - \frac{q_5/C_5}{R_6(x_2)}$$

(e)



$$\dot{p}_2 = \frac{M_1(t) - k_t q_4}{J_2} ; \quad \dot{q}_4 = \frac{p_2/J_1 - p_6/J_2}{J_2}$$

$$\dot{q}_{11} = \frac{r p_6/J_2}{J_2}$$

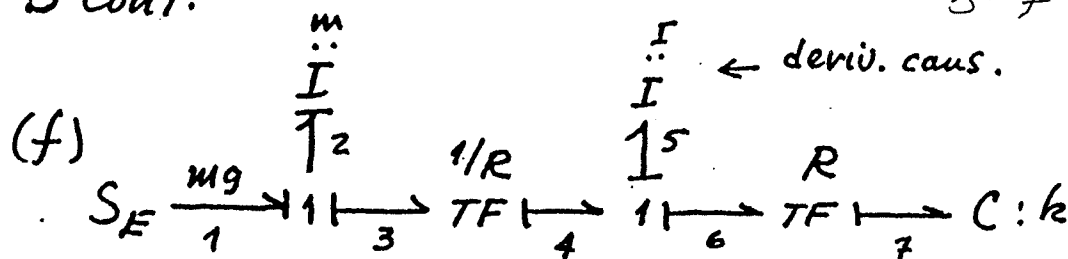
$$\dot{p}_6 = k_t q_4 - r(e_{11} + e_{10} + e_9) = k_t q_4 - r[k q_{11} + b r p_6/J_2 + \dot{p}_9]$$

but $p_9 = m f_9 = m f_8 = m r f_7 = m r f_6 = m r p_6/J_2$
sub. in last eq., solve for \dot{p}_6

$$\dot{p}_6 = \left[\frac{1}{1 + m r^2/J_2} \right] \left[k_t q_4 - r k q_{11} - \frac{r^2 b}{J_2} p_6 \right]$$

5-3 Cont.

5-7



$$\dot{q}_7 = R \cdot \frac{1}{R} \cdot p_2/m = \underline{p_2/m}$$

$$\dot{p}_2 = mg - e_3 = mg - \frac{1}{R}(e_5 + e_6) = mg - \frac{1}{R}(\dot{p}_5 + Rkq_7)$$

but $p_5 = I f_5 = I \cdot \frac{1}{R} \cdot \frac{p_2}{m}$ or $\dot{p}_5 = \frac{I}{Rm} \dot{p}_2$

so

$$\dot{p}_2 = [1 + I/R^2m]^{-1} [mg - kq_7]$$

5.4 $\dot{q}_1 = I(t) - \lambda_1/L_1$

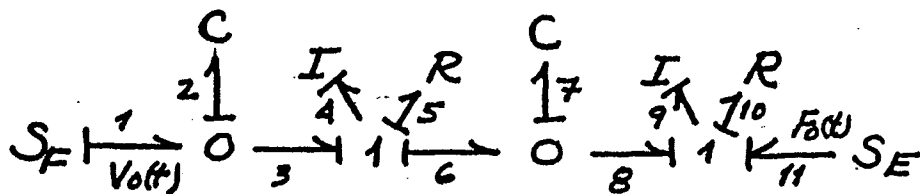
$$\dot{\lambda}_1 = -R_1 \lambda_1/L_1 + q_1/C_1 - q_2/C_2$$

$$\dot{q}_2 = \lambda_1/L_1 - \lambda_2/L_2$$

$$\dot{\lambda}_2 = -R_2 \lambda_2/L_2 + q_2/C_2 - R_L \lambda_2/L_2$$

$$e = R_L \lambda_2/L_2$$

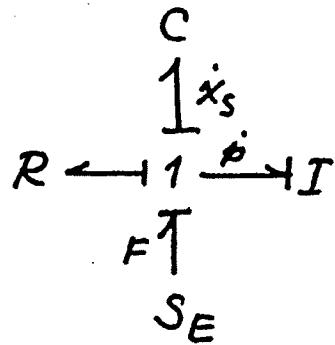
5-5



$$\dot{q}_2 = V_0(t) - p_4/m ; \dot{p}_4 = Aq_2^3 - B\text{sgn}(\frac{p_4}{m}) - Aq_7^3 ;$$

$$\dot{q}_7 = p_4/m - p_9/m ; \dot{p}_9 = Aq_7^3 - B\text{sgn}(\frac{p_9}{m}) + F_0(t).$$

5-7



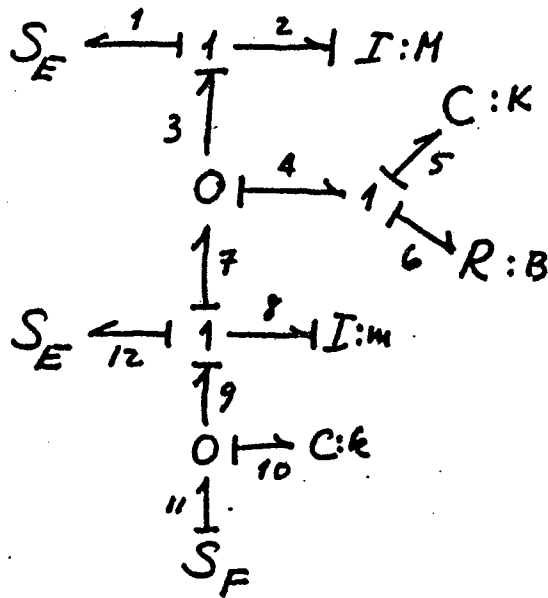
or $\dot{x}_s = v$

$$\dot{x}_s = p/m \quad 5-9$$

$$\dot{p} = -\phi_s(x_s) - \frac{b}{m^2} |p|p + F(t)$$

$$\dot{v} = -\frac{\phi(x_s)}{m} - \frac{b}{m} |v|v + \frac{F(t)}{m}$$

5-8

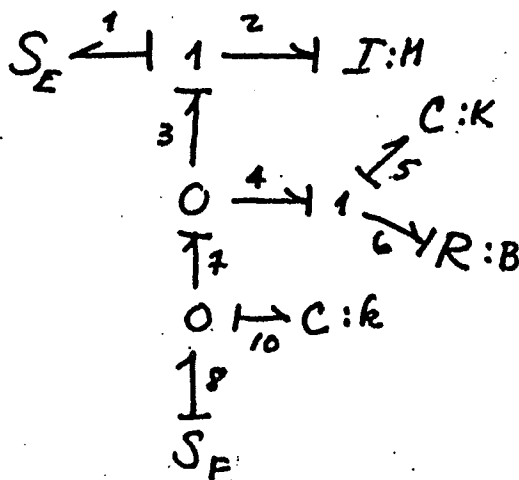


$$\dot{p}_2 = -Mg + Kq_5 + B\left(\frac{p_8}{m} - \frac{p_2}{M}\right)$$

$$\dot{q}_5 = p_8/m - p_2/M$$

$$\dot{p}_8 = -mg + kq_{10} - Kq_5 - B\left(\frac{p_8}{m} - \frac{p_2}{M}\right)$$

$$\dot{q}_{10} = V(t) - p_8/m$$



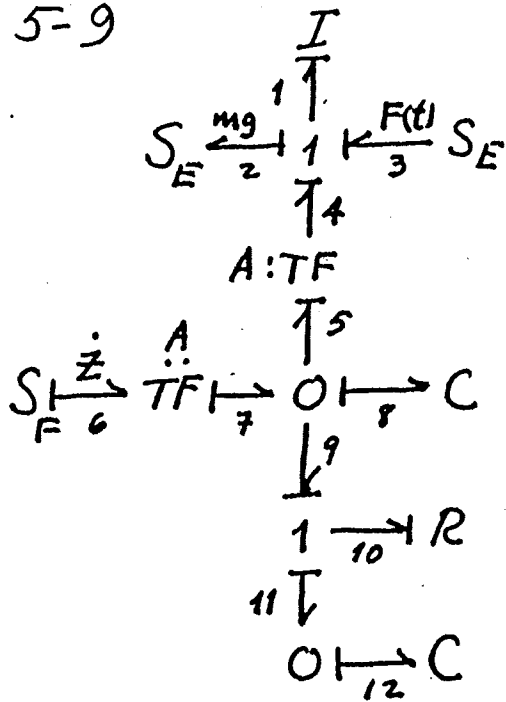
$$\dot{p}_2 = -Mg + kq_{10}$$

$$\dot{q}_5 = \frac{1}{B}(-Kq_5 + kq_{10})$$

$$\dot{q}_{10} = V_0(t) - \frac{1}{B}(-Kq_5 + kq_{10}) - p_2/M$$

5-10

5-9

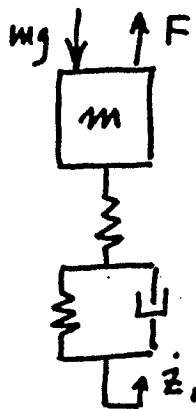


$$\dot{p}_1 = F(t) - mg + A \Phi_{c8}^{-1}(q_8),$$

$$\dot{q}_8 = A \dot{z}_1(t) - A p/m$$

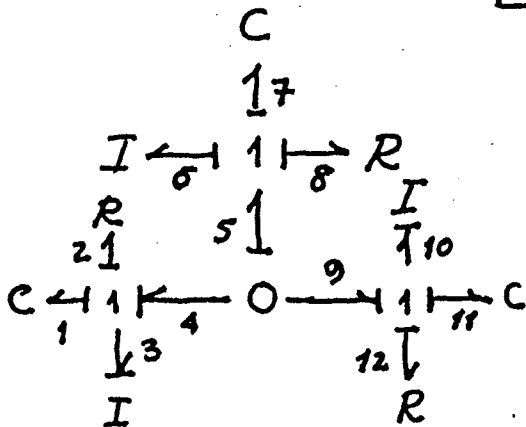
$$+ \Phi_{R10}^{-1} (\Phi_{c8}^{-1}(q_8) - \Phi_{c12}^{-1}(q_{12})),$$

$$\dot{q}_{12} = + \Phi_{R10}^{-1} (\Phi_{c8}^{-1}(q_8) - \Phi_{c12}^{-1}(q_{12})).$$



Transformers not shown here.

5-10



Note: These equations are for linear capacitors.

$$\dot{p}_3 = -q_1/C_1 - R_2 p_3/I_3 + (\dot{p}_6 + q_7/C_7 + R_8 (-p_3/I_3 - p_{10}/I_{10}))$$

$$\dot{p}_{10} = -q_{11}/C_1 - R_{12} p_{10}/I_{10} + (\dot{p}_6 + q_7/C_7 + R_8 (-p_3/I_3 - p_{10}/I_{10}))$$

$$\dot{q}_1 = p_3/I_3 ; \quad \dot{q}_{11} = p_{10}/I_{10} ; \quad \dot{q}_7 = -p_3/I_3 - p_{10}/I_{10}$$

$$\text{Diff. Caus. } p_6 = I_6 f_6 = I_6 (-p_3/I_3 - p_{10}/I_{10})$$

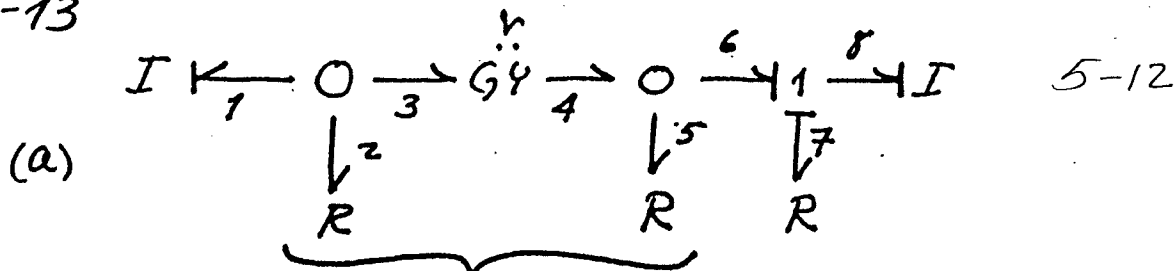
$$\dot{p}_6 = \frac{-I_6}{I_3} \dot{p}_3 - \frac{I_6}{I_{10}} \dot{p}_{10} \leftarrow \text{Sub. into state eq.}$$

2x2 matrix inversion is req'd.

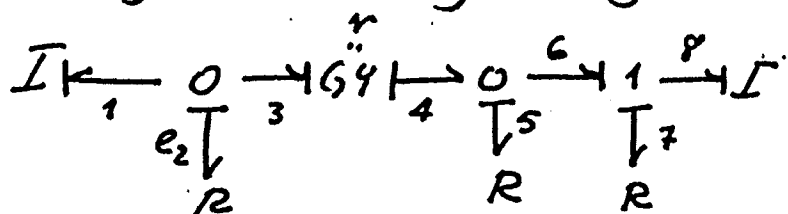
5-11

With coil resistance $\frac{R}{s} \rightarrow \frac{R}{s} + 1 \rightarrow s^2 + \frac{R}{L}s + \frac{1}{LC} = 0$, second order again

5-13



Causality in this R -field not determined by I_1 and I_8 . Some bond must be assigned causality arbitrarily. Try bond 2:



$$\dot{p}_1 = e_2 ; \quad \dot{p}_8 = -R_7 \frac{p_8}{I_8} + R_5 \left(-\frac{p_8}{I_8} + \frac{1}{r} e_2 \right)$$

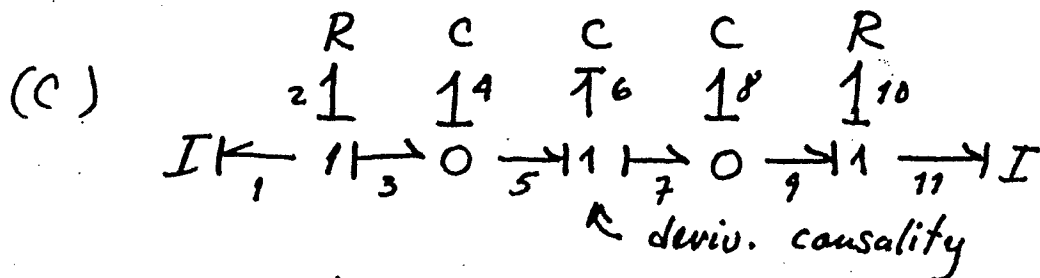
now find e_2 in terms of itself, p_1 and p_8

$$e_2 = R_2 \left(-\frac{p_1}{I_1} - \frac{R_5}{r} \left(-\frac{p_8}{I_8} + \frac{e_2}{r} \right) \right)$$

Solve for e_2 and use in state equations.

(b) System has a causal conflict of type

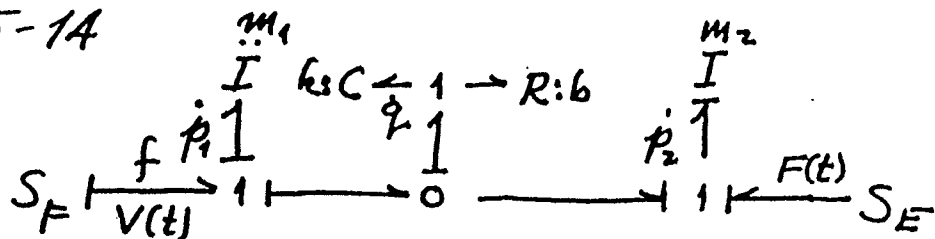
$S_E - G_4 - S_F$ or $S_E - S_F$ i.e. a nonsense system.



$$q_6 = C_6 e_6 = C_6 \left(\frac{q_4}{C_4} - \frac{q_8}{C_8} \right)$$

\dot{q}_6 appears in state eqn's. - may be replaced by expression in \dot{q}_4, \dot{q}_8 , 2x2 matrix inversion req'd.

5-14

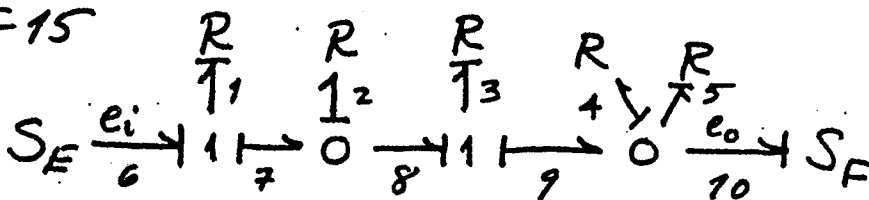


5-13

$$\left. \begin{aligned} (a) \quad \dot{p}_2 &= F(t) - k_2 q - b(V(t) - p_2/m_2) \\ \dot{q} &= V(t) - p_2/m_2 \end{aligned} \right\} \begin{array}{l} \text{complete} \\ \text{state Eqs.} \end{array}$$

$$\begin{aligned} (b) \quad f &= \dot{p}_1 + k_1 q + b(V(t) - p_2/m_2) \\ &= m_1 \dot{V}_1(t) + b(V(t)) + k_1 q - b p_2/m_2 \end{aligned}$$

5-15



causality chosen arbitrarily on bonds 1, 2, 3, 4. We use f_1, e_2, f_3, e_4 as auxiliary variables, e_i and f_{10} as inputs. Eqs are

$$\left. \begin{aligned} e_o &= e_4 = R_4(f_3 - e_4/R_5 - f_{10}) \\ f_3 &= (e_2 - e_4)/R_3 \\ e_2 &= R_2(f_1 - f_3) \\ f_1 &= (e_i - e_2)/R_1 \end{aligned} \right\} \begin{array}{l} 4 \text{ eqns may} \\ \text{be solved} \\ \text{for aux. var} \\ \text{in terms of} \\ e_i \text{ and } f_{10} \end{array}$$

If there is a load resistance instead of $f_{10} S_F$, use $f_{10} = e_4/R_L$ instead of f_{10} as input quantity.

5.16 state variables $\phi_5, q_7, \phi_{10}, q_{12}$

5-14

$$\dot{\phi}_5 = \frac{\tau}{R_2} (e - \frac{\tau}{I_5} \phi_5) - \frac{q_7}{C_7}$$

$$\dot{q}_7 = \frac{\phi_5}{I_5} - \frac{1}{R} \frac{\phi_{10}}{I_{10}}$$

$$\dot{\phi}_{10} = -\frac{q_{12}}{C_{12}} - \frac{R_{11}}{I_{10}} \phi_{10} + \frac{1}{R} \frac{q_7}{C_7}$$

$$\dot{q}_{12} = \phi_{10} / I_{10}$$

$$\frac{d}{dt} \begin{bmatrix} \phi_5 \\ q_7 \\ \phi_{10} \\ q_{12} \end{bmatrix} = \begin{bmatrix} -\tau^2/R_2 I_5 & -1/C_7 & 0 & 0 \\ 1/I_5 & 0 & -1/R I_{10} & 0 \\ 0 & 1/R C_7 & -R_{11}/I_{10} & -1/C_{12} \\ 0 & 0 & 1/I_{10} & 0 \end{bmatrix} \begin{bmatrix} \phi_5 \\ q_7 \\ \phi_{10} \\ q_{12} \end{bmatrix} + \begin{bmatrix} \tau/R_2 \\ 0 \\ 0 \\ 0 \end{bmatrix} e$$

output eqns:

$$e_4 = \frac{\tau}{R_2} (e - \frac{\tau}{I_5} \phi_5)$$

$$f_7 = \frac{\phi_5}{I_5} - \frac{1}{R} \frac{\phi_{10}}{I_{10}}$$

$$e_2 = e - \frac{\tau}{I_5} \phi_5$$

5.17

state variables p_5, p_{10}, q_{12}

5-15

To handle derivative causality, $p_8 = I_8 \left[\frac{R}{I_5} p_5 + \frac{p_{10}}{I_{10}} \right]$

$$\dot{p}_8 = e_8 = R \frac{I_8}{I_5} \dot{p}_5 + \frac{I_8}{I_{10}} \dot{p}_{10}$$

state eqns:

$$\dot{p}_5 = \frac{\tau}{R_2} (e - \frac{\tau}{I_5} p_5) - R \dot{p}_8$$

$$\dot{p}_{10} = \frac{q_{12}}{C_{12}} - \dot{p}_8$$

$$\dot{q}_{12} = v_i - \frac{p_{10}}{I_{10}}$$

substitute for \dot{p}_8

$$\dot{p}_5 \left[1 + R^2 \frac{I_8}{I_5} \right] + R \frac{I_8}{I_{10}} \dot{p}_{10} = \frac{\tau}{R_2} e - \frac{\tau^2}{R_2 I_5} p_5$$

$$R \frac{I_8}{I_5} \dot{p}_5 + \dot{p}_{10} \left[1 + \frac{I_8}{I_{10}} \right] = q_{12}/C_{12}$$

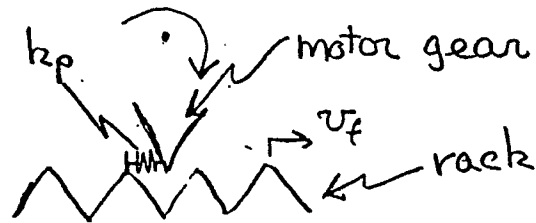
$$\dot{q}_{12} = v_i - \frac{p_{10}}{I_{10}}$$

can write as,

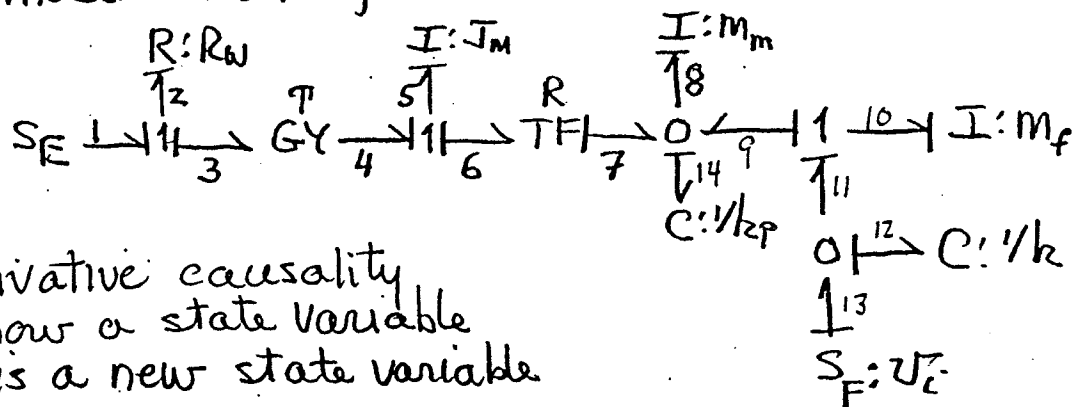
$$\begin{bmatrix} 1 + R^2 \frac{I_8}{I_5} & R \frac{I_8}{I_{10}} & 0 \\ R \frac{I_8}{I_5} & 1 + \frac{I_8}{I_{10}} & 0 \\ 0 & 0 & 1 \end{bmatrix} \begin{bmatrix} \dot{p}_5 \\ \dot{p}_{10} \\ \dot{q}_{12} \end{bmatrix} = \begin{bmatrix} -\frac{\tau^2}{R_2 I_5} & 0 & 0 \\ 0 & 0 & 1/C_{12} \\ 0 & -1/I_{10} & 0 \end{bmatrix} \begin{bmatrix} p_5 \\ p_{10} \\ q_{12} \end{bmatrix} + \begin{bmatrix} \tau/R_2 \\ 0 \\ 0 \end{bmatrix} e + \begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix} v_i$$

final eqns. requires inversion of this matrix

5.18 If we imagine that a very stiff spring is between the gear tooth of the motor and the tooth of the rack, such as



the model becomes,



No derivative causality
 p_8 now a state variable
 q_{14} is a new state variable

$$\dot{p}_5 = \frac{T}{R_2} (e - \frac{T}{I_5} p_5) - \frac{R}{C_{14}} q_{14}$$

$$\dot{q}_8 = q_{14} / C_{14}$$

$$\dot{p}_{10} = \frac{q_{12}}{C_{12}} - \frac{q_{14}}{C_{14}}$$

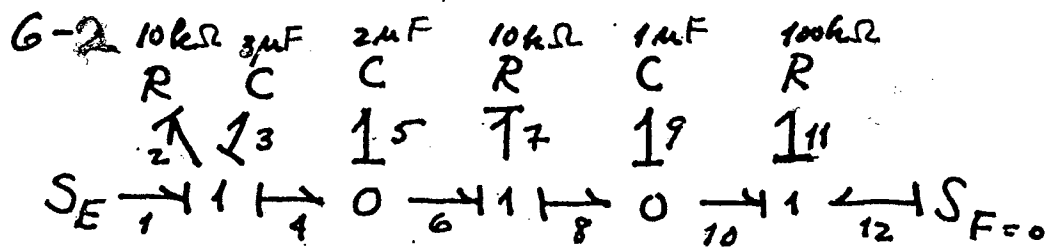
$$\dot{q}_{12} = v_f - p_{10} / I_{10}$$

$$\dot{q}_{14} = \frac{R}{I_5} p_5 - \frac{p_8}{I_8} + \frac{p_{10}}{I_{10}}$$

6-1. $x = \frac{V_0}{\omega_n} \sin \omega_n t$; $\omega_n = 10 \text{ rad/sec} = \sqrt{\frac{k}{m}}$

Period = $T = \frac{2\pi}{\omega_n}$, Δt should be a relatively small fraction of T say $1/10$ or less

$$\Delta t \approx \frac{1}{10} \cdot \frac{2\pi}{10} \text{ sec.}$$



$$\begin{bmatrix} \dot{q}_3 \\ \dot{q}_5 \\ \dot{q}_9 \end{bmatrix} = \begin{bmatrix} -\frac{1}{R_2 C_3} & \frac{1}{R_2 C_5} & 0 \\ -\frac{1}{R_2 C_3} & \frac{1}{R_2 C_5} - \frac{1}{R_7 C_5} & \frac{1}{R_7 C_9} \\ 0 & \frac{1}{R_7 C_5} & -\frac{1}{R_7 C_9} \end{bmatrix} \begin{bmatrix} q_3 \\ q_5 \\ q_9 \end{bmatrix} + \begin{bmatrix} 1/R_2 \\ 1/R_2 \\ 0 \end{bmatrix} E_1$$

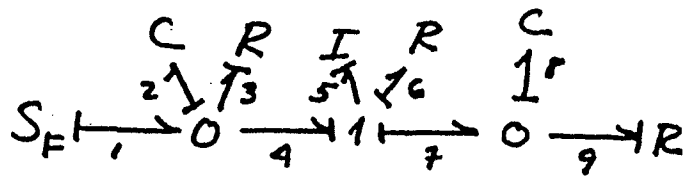
$$e_{12} = q_9 / C_9$$

$$R_2 C_3 = 10 \cdot 10^3 \cdot 3 \times 10^{-6} = 30 \times 10^{-3} \text{ s etc.}$$

Characteristic Eqn. $\det [sI - A] = 0$

Third order in this case.

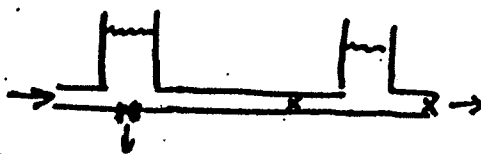
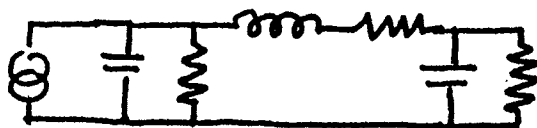
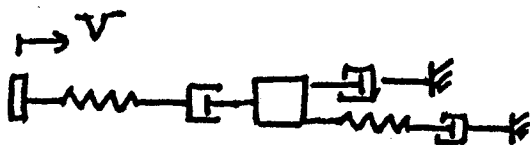
6-3



6-2

$$\begin{bmatrix} \dot{q}_2 \\ \dot{p}_5 \\ \dot{q}_8 \end{bmatrix} = \underbrace{\begin{bmatrix} -1/R_3 C_2 & -1/I_5 & 0 \\ 1/C_2 & -R_6/I_5 & -1/C_9 \\ 0 & 1/I_5 & -1/R_9 C_8 \end{bmatrix}}_A \begin{bmatrix} q_2 \\ p_5 \\ q_8 \end{bmatrix} + \underbrace{\begin{bmatrix} 1 \\ 0 \\ 0 \end{bmatrix}}_B F_1(t)$$

$$\begin{bmatrix} f_4 \\ e_7 \end{bmatrix} = \underbrace{\begin{bmatrix} 0 & 1/I_5 & 0 \\ 0 & 0 & 1/C_8 \end{bmatrix}}_C \begin{bmatrix} q_2 \\ p_5 \\ q_8 \end{bmatrix} + \underbrace{\begin{bmatrix} 0 \\ 0 \end{bmatrix}}_D F_1(t)$$



6-4 $\operatorname{Re} e^{j\omega t} = \cos \omega t$

$$\frac{d}{dt} \operatorname{Re} e^{j\omega t} = -\omega \sin \omega t$$

$$\frac{d}{dt} e^{j\omega t} = j\omega e^{j\omega t} = \omega e^{j(\omega t + \pi/2)}$$

$$\operatorname{Re} \frac{d}{dt} e^{j\omega t} = \omega \cos(\omega t + \pi/2) = -\omega \sin \omega t$$

6-3

6-5 for 1 d.o.f system, $\frac{b}{m} = 2 \zeta \omega_n$

for 4 dampers $b = \frac{1}{4} \cdot 2 m \zeta \omega_n$

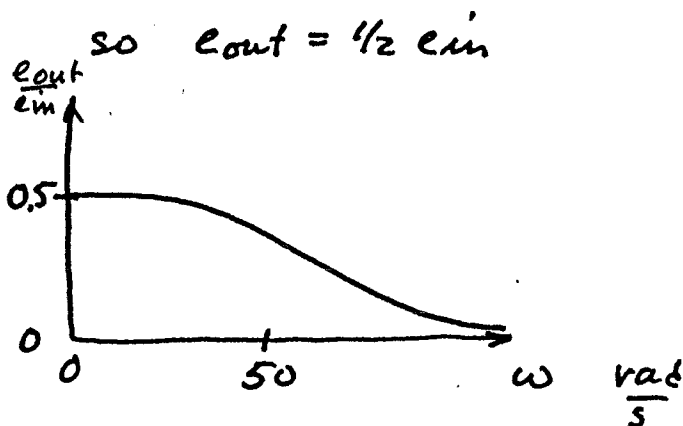
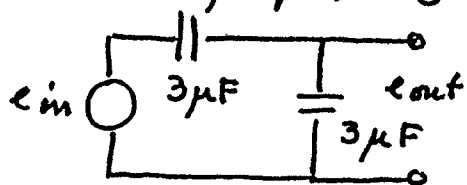
$$b = \frac{3000}{2 \cdot 386} \cdot 0.707 \cdot 2\pi \cdot 1 \frac{16}{\text{in/sec}}$$

6-6 System has three real roots -

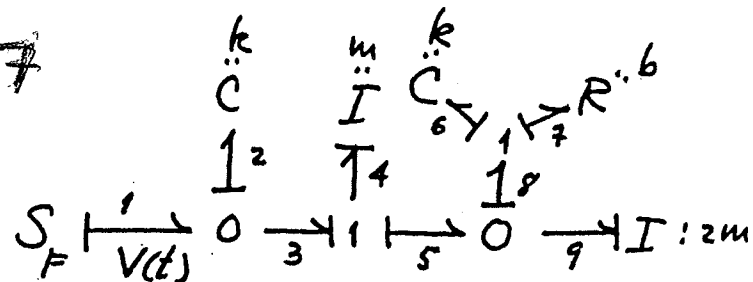
time constants are in the neighborhood of

$$(1-3) \times 10^{-2} \text{ s or } \omega = 1/\tau \approx 30 \text{ to } 100 \text{ rad/s}$$

at zero freq., system looks like



6-7



6-4

(Gravity forces left out so deflections are measured from equilibrium.)

State equations

$$\begin{bmatrix} \dot{q}_2 \\ \dot{p}_4 \\ \dot{q}_6 \\ \dot{p}_9 \end{bmatrix} = \underbrace{\begin{bmatrix} 0 & -1/m & 0 & 0 \\ k & -b/m & -k & b/2m \\ 0 & 1/m & 0 & -1/2m \\ 0 & b/m & k & -b/2m \end{bmatrix}}_A \begin{bmatrix} q_2 \\ p_4 \\ q_6 \\ p_9 \end{bmatrix} + \underbrace{\begin{bmatrix} 1 \\ 0 \\ 0 \\ 0 \end{bmatrix}}_B V(t)$$

Acceleration

$$a = \frac{1}{2m} \cdot \dot{p}_9 = \frac{1}{2m} \left(k q_6 + \frac{b}{m} p_4 - \frac{b}{2m} p_9 \right)$$

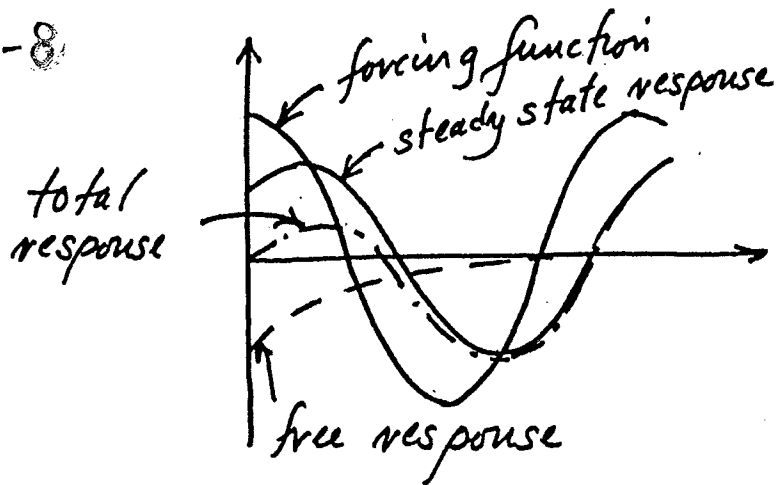
$$a = \underbrace{\begin{bmatrix} 0 & \frac{b}{2m^2} & \frac{k}{2m} & -\frac{b}{4m^2} \end{bmatrix}}_C \begin{bmatrix} q_2 \\ p_4 \\ q_6 \\ p_9 \end{bmatrix}$$

if $V(t) = 1 \cdot e^{i\omega t}$, Assume $q_2 = Q_2 e^{i\omega t}$ etc.
then

$$\begin{bmatrix} i\omega & 1/m & 0 & 0 \\ -k & i\omega + b/m & k & -b/2m \\ 0 & -1/m & i\omega & 1/2m \\ 0 & -b/m & -k & i\omega + b/2m \end{bmatrix} \begin{bmatrix} Q_2 \\ P_4 \\ Q_6 \\ P_9 \end{bmatrix} = \begin{bmatrix} 1 \\ 0 \\ 0 \\ 0 \end{bmatrix}$$

must be solved, for example, by Cramer's rule. Then a can be found from P_4 , Q_6 and P_9 . The algebra is a bit lengthy, it must be admitted

6-8



6-5

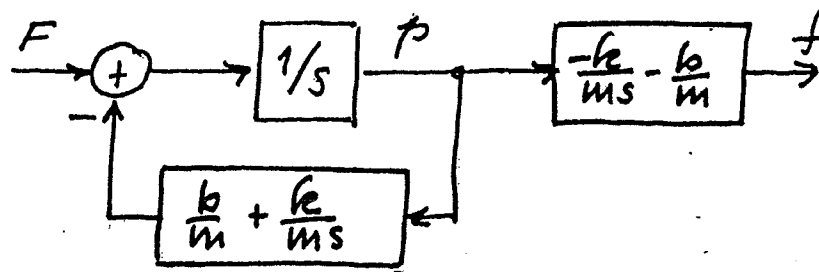
6-9 Both Fig. 6.24 (a) and (b) contain errors in sign. The transfer function in Fig. 6.24 (a) should be $\frac{-(bs+k)}{ms^2+bs+k}$ as in Eq. 6-117 and in Fig. 6.24 (b) f should be computed as $kx - \frac{b}{m}p$ as in Eq. 6.24.

From 6-24 (a) we can relate p and x to themselves as follows: $x = -\frac{1}{ms}p$,
 $p = \frac{1}{s}(F + kx - \frac{b}{m}p)$. substituting for x
 we find $p = \frac{msF}{ms^2+bs+k}$ and then $x = \frac{-F}{ms^2+bs+k}$

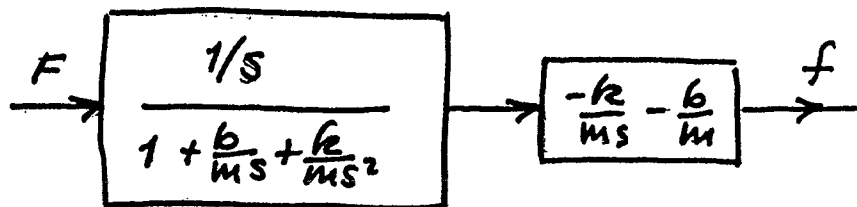
using the correct expression for f (above)
 the correct transfer function is found.

6-10

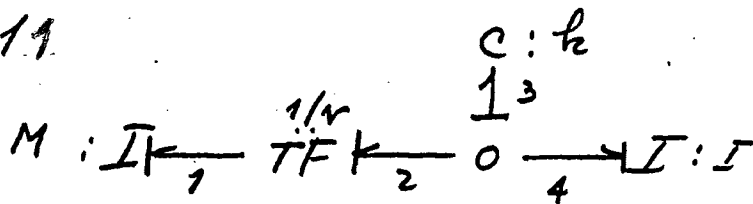
6-6



||



6-11



$$\begin{bmatrix} \dot{p}_1 \\ \dot{p}_2 \\ \dot{p}_4 \end{bmatrix} = \underbrace{\begin{bmatrix} 0 & k/r & 0 \\ -1/rm & 0 & -1/I \\ 0 & k & 0 \end{bmatrix}}_A \begin{bmatrix} p_1 \\ p_2 \\ p_4 \end{bmatrix}$$

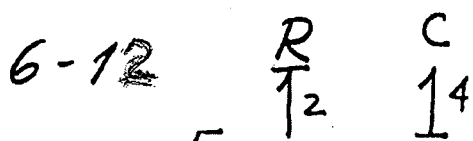
$$\det [sI - A] = s \left[s^2 + \frac{k}{I} \right] + \frac{k}{r} \cdot \frac{1}{rm} = 0$$

$$\text{or } s \left(s^2 + \frac{k}{I} + \frac{k}{r^2 M} \right) = 0$$

$s = 0 \leftarrow$ steady spin mode

$$s = \pm j\omega_n = \pm j \sqrt{k \left(\frac{1}{I} + \frac{1}{r^2 M} \right)}$$

undamped oscillator mode



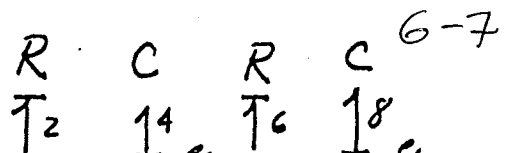
$$S_E \xrightarrow{1} 1 \xrightarrow{3} 0 \xrightarrow{e_a} S_{I=0}$$

(a)

$$\dot{q}_4 = -\frac{q_4}{RC} + \frac{E(t)}{R}$$

$$e_a = q_4/C$$

(b) $H_{e_a/E} = \frac{1}{RCs+1}$



$$S_E \xrightarrow{1} 1 \xrightarrow{3} 0 \xrightarrow{e_a} S_{I=0} ; S_E \xrightarrow{1} 1 \xrightarrow{3} 0 \xrightarrow{e_a} S_{I=0}$$

$$\dot{q}_4 = -\frac{2q_4}{RC} + \frac{q_8}{RC} + \frac{E(t)}{R}$$

$$\dot{q}_8 = \frac{q_4}{RC} - \frac{q_8}{RC}$$

$$e_a = q_4/C, e_b = q_8/C$$

$$H_{e_a/E} = \frac{RCs+1}{R^2C^2s^2+3RCs+1}$$

(c) Loading effects preclude affirmative answers.

$$H_{e_b/E} = \frac{1}{R^2C^2s^2+3RCs+1}$$

6-13



$$A = \begin{bmatrix} -R/L & -1/C \\ 1/L & C \end{bmatrix}$$

$$s^2 + s\frac{R}{L} + \frac{1}{LC} = 0$$

$R=0$, undamped

$R \rightarrow \infty$, overdamped



$$A = \begin{bmatrix} 0 & 1/C \\ -1/L & -1/RC \end{bmatrix}$$

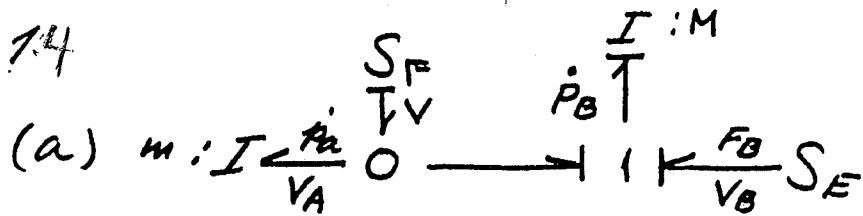
$$s^2 + \frac{s}{RC} + \frac{1}{LC} = 0$$

$R \rightarrow 0$, overdamped

$R \rightarrow \infty$, undamped

6-14

6-8



(b) $\dot{p}_B = F_B(t) + \dot{p}_A$, but

$$p_A = m \dot{x}_A = m (\dot{x}(t) - \dot{x}_B/M)$$

so $\dot{p}_A = F_B(t) + m (\dot{x}(t) - \dot{x}_B/M)$

or $\dot{p}_B = \frac{M}{m+M} (F_B(t) + m \dot{x}(t)) \leftarrow \text{state eqn.}$

$$\left. \begin{aligned} V_B &= p_B/M \\ V_A &= \dot{x}(t) - p_B/M \end{aligned} \right\} \text{output eqns.}$$

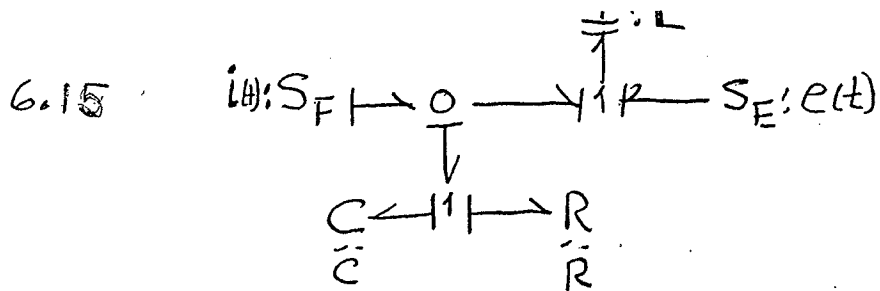
(c) i. $\frac{V_B}{F_B} = \frac{p_B}{M F_B} = \frac{M}{i\omega(m+M)} \frac{F_B}{F_B} \checkmark$

ii. $p_B \equiv 0 = F_B - m \dot{x}$, $F_B = m j\omega V \checkmark$

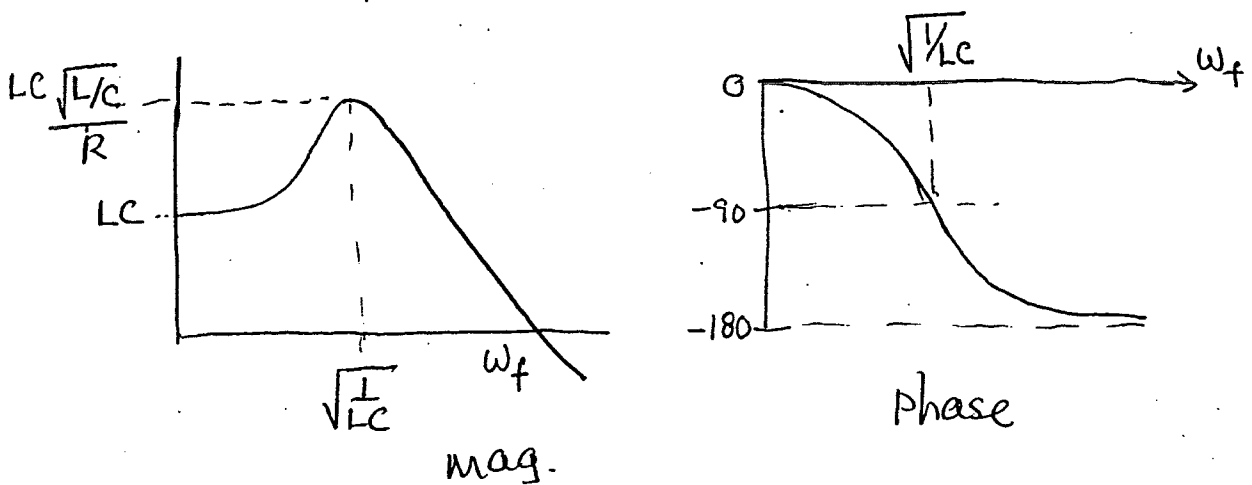
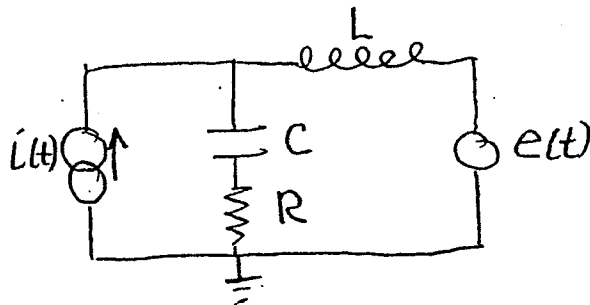
iii. $j\omega p_B = \frac{M}{m+M} (m j\omega V)$

$$V_B = \frac{1}{M} \cdot \frac{M m}{(m+M)} V \checkmark$$

iv. Son of a gun!



circuit:



magnitude and phase of $\frac{1}{\frac{1}{LC} - \omega_f^2 + j \frac{R}{L} \omega_f}$

6.16 (a) $\frac{Y}{X}(j\omega) = \frac{G(\omega_0^2 - \omega^2)}{\omega_h^2 - \omega^2 + j 2\zeta \omega_h \omega}$

(b) $\left| \frac{Y}{X} \right| = \frac{G[(\omega_0^2 - \omega^2)^2]^{1/2}}{[(\omega_h^2 - \omega^2)^2 + (2\zeta \omega_h \omega)^2]^{1/2}}$

6.16 (Continued)

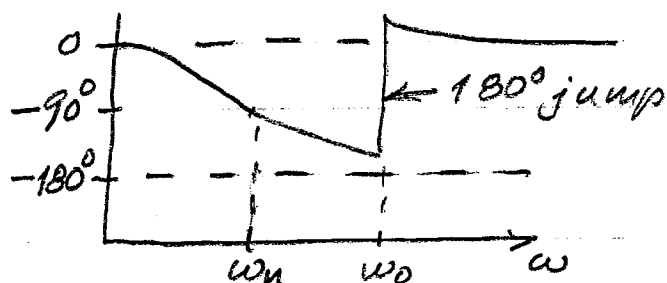
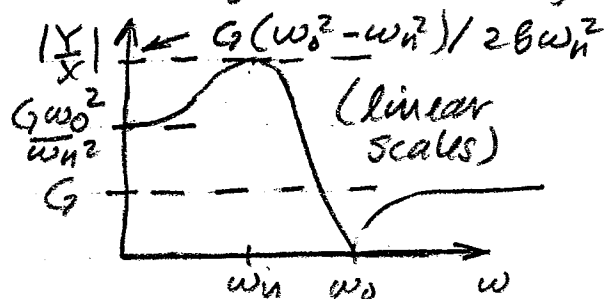
$$\omega \rightarrow 0 \quad |Y/X| \rightarrow \frac{G\omega^2}{\omega_n^2}; \quad \angle Y/X \rightarrow 0$$

$$\omega = \omega_n, \quad |Y/X| = \frac{G(\omega_0^2 - \omega_n^2)}{2\beta\omega_n}; \quad \angle Y/X = -\pi/2, -90^\circ$$

$$\omega = \omega_0, \quad |Y/X| = 0; \quad \angle Y/X \text{ jumps by } \pi, 180^\circ$$

for $\omega < \omega_0$ to $\omega > \omega_0$
(sign change for numerator)

$$\omega \rightarrow \infty, \quad |Y/X| \rightarrow G, \quad \angle Y/X \rightarrow 0$$



6.17

a) $P_f = I_f A_p (\dot{v}_i - \frac{p_m}{m})$ dealing with derivative causality.

$$\dot{p}_m = A_p \left[\frac{q_a}{C_a} + R_f A_p (\dot{v}_i - \frac{p_m}{m}) + \dot{p}_f \right]$$

$$= \frac{A_p}{C_a} q_a + R_f A_p^2 \dot{v}_i - R_f A_p^2 \frac{p_m}{m} + A_p \left[I_f A_p (\dot{v}_i - \frac{p_m}{m}) \right]$$

$$(1) \quad \dot{p}_m = \frac{A_p/C_a}{1 + \frac{m_e}{m}} q_a + \frac{R_f A_p^2}{1 + \frac{m_e}{m}} \dot{v}_i - \frac{R_f A_p^2}{1 + \frac{m_e}{m}} \frac{p_m}{m} + \frac{m_e}{1 + \frac{m_e}{m}} \dot{v}_i$$

$$(2) \quad \dot{q}_a = A_p (\dot{v}_i - \frac{p_m}{m})$$

where $m_e = A_p^2 I_f$

6.17 (continued)

6-11

(b) In the s-domain,

$$\begin{bmatrix} s + \frac{R_f A_p^2}{1 + m_e/m} & -\frac{A_p/Ca}{1 + m_e/m} \\ A_p/m & s \end{bmatrix} \begin{bmatrix} p_m \\ q_a \end{bmatrix} = \begin{bmatrix} \frac{R_f A_p^2 + m_e s}{1 + m_e/m} \\ A_p \end{bmatrix} \frac{v_i}{s}$$

$$D = s^2 + \frac{R_f A_p^2/m}{1 + m_e/m} s + \frac{A_p^2/mCa}{1 + m_e/m}$$

solve for p_m using Cramer's rule:

$$\frac{p_m}{v_i} = \frac{(R_f A_p^2 + m_e s)s + \frac{A_p^2/Ca}{1 + m_e/m}}{D}$$

and

$$\frac{v_m}{v_i} = \frac{\frac{m_e/m}{1 + m_e/m} s^2 + \frac{R_f A_p^2/m}{1 + m_e/m} s + \frac{A_p^2/mCa}{1 + m_e/m}}{s^2 + \frac{R_f A_p^2/m}{1 + m_e/m} s + \frac{A_p^2/mCa}{1 + m_e/m}}$$

or

$$\boxed{\frac{v_m}{v_i} = G \frac{s^2 + 2\zeta_0 \omega_0 s + \omega_0^2}{s^2 + 2\zeta \omega_n s + \omega_n^2}}$$

where

$$G = \frac{m_e/m}{1 + m_e/m} \quad 2\zeta_0 \omega_0 = \frac{R_f A_p^2/m}{m_e/m} \quad \omega_0^2 = \frac{A_p^2/mCa}{m_e/m}$$

$$2\zeta \omega_n = \frac{R_f A_p^2/m}{1 + m_e/m} \quad \omega_n^2 = \frac{A_p^2/mCa}{1 + m_e/m}$$

(c) Comparing to Prob 6.17 we see that the transfer functions are the same except for the damping, ζ_0 , in the numerator of (b). This damping will compromise the notch in the frequency response at $\omega = \omega_0$. Still, this is an excellent isolator but only over a narrow frequency range near ω_0 . The high frequency isolation is poor.

(d) The high frequency asymptote is,

$$G = \frac{m_e/m}{1 + m_e/m} \quad \text{where } m_e = A_p^2 I_f$$

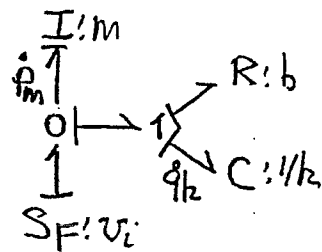
I_f is the fluid inertia and would be calculated from,

$$I_f = \frac{\rho L}{A_t}$$

and
$$m_e = A_p^2 \frac{\rho L}{A_t}$$

To improve the high frequency isolation we need to reduce m_e . This could be done by increasing A_t .

6-18



$$\ddot{p}_m = kq_k + b(v_i - \dot{p}_m/m)$$

$$\dot{q}_k = v_i - \dot{p}_m/m$$

s-domain:

$$\begin{bmatrix} s + \frac{b}{m} & -k \\ \frac{1}{m} & s \end{bmatrix} \begin{bmatrix} p_m \\ q_k \end{bmatrix} = \begin{bmatrix} b \\ 1 \end{bmatrix} v_i$$

$$\begin{bmatrix} s + \frac{b}{m} & -k \\ \frac{1}{m} & s \end{bmatrix} \begin{bmatrix} p_m \\ q_k \end{bmatrix} = \begin{bmatrix} 0 \\ 1 \end{bmatrix} v_i$$

use Cramer's rule: solve for p_m , then

$$\frac{v_m}{v_i} = \frac{\frac{b}{m}s + \frac{k}{m}}{s^2 + \frac{b}{m}s + \frac{k}{m}}$$

$$\frac{v_m}{v_i} = \frac{k/m}{s^2 + \frac{b}{m}s + \frac{k}{m}}$$

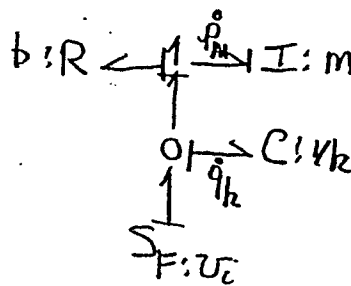
For frequency response

$$\left| \frac{v_m(\omega)}{v_i} \right| = \frac{[(\omega_n^2)^2 + (2\zeta\omega_n\omega)^2]^{1/2}}{[(\omega_n^2 - \omega^2)^2 + (2\zeta\omega_n\omega)^2]^{1/2}}$$

$$\left| \frac{v_m}{v_i} \right| = \frac{\omega_n^2}{[(\omega_n^2 - \omega^2)^2 + (2\zeta\omega_n\omega)^2]^{1/2}}$$

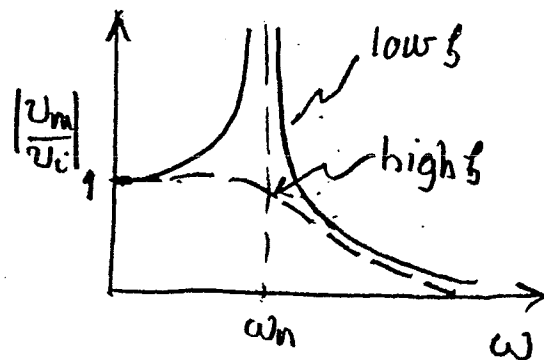
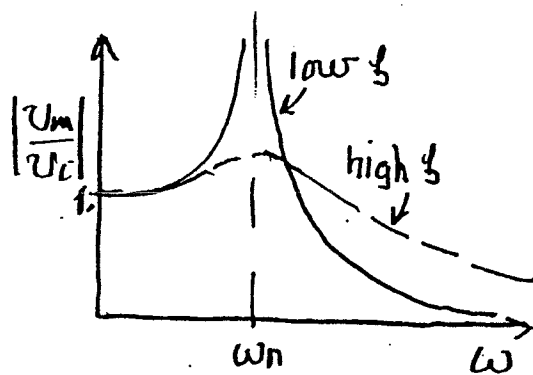
$$\omega_n^2 = \frac{k}{m} \quad 2\zeta\omega_n = \frac{b}{m}$$

6-13

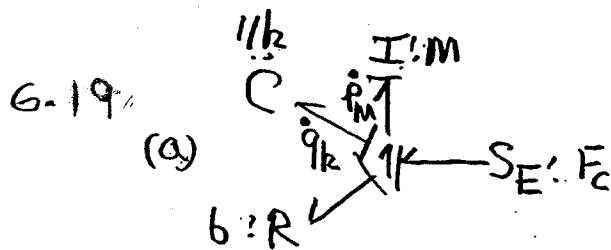


$$\ddot{p}_m = -\frac{b}{m}\dot{p}_m + kq_k$$

$$\dot{q}_k = v_i - \dot{p}_m/m$$



The system with the damper attached to inertial ground has much better high frequency isolation without large response at the resonant frequency, ω_n .



s-domain;

use Cramer's rule, solve for $q_k = x$

$$\frac{x}{F_c} = \frac{1/m}{s^2 + \frac{b}{m}s + \frac{k}{m}} = G_p(s)$$

$$\begin{aligned} \dot{p}_m &= -k q_k - \frac{b}{m} p_m + F_c \\ \dot{q}_k &= p_m / m \end{aligned}$$

$$\begin{bmatrix} s + \frac{b}{m} & k \\ -1/m & s \end{bmatrix} \begin{bmatrix} p_m \\ q_k \end{bmatrix} = \begin{bmatrix} 1 \\ 0 \end{bmatrix} F_c$$

$$(b) \quad \frac{x}{x_{des}} = \frac{G_c G_p}{1 + G_c G_p}$$

$$G_c = K_p + K_D s + \frac{K_I}{s} \equiv \frac{K_D s^2 + K_p s + K_I}{s}$$

$$\begin{aligned} &= \frac{(K_D s^2 + K_p s + K_I) 1/m}{s(s^2 + \frac{b}{m}s + \frac{k}{m}) + (K_D s^2 + K_p s + K_I) 1/m} 1/m \end{aligned}$$

6.19 (continued)

6-15

$$\frac{x}{x_{des}} = \frac{\frac{k_D}{m} s^2 + \frac{k_P}{m} s + \frac{k_I}{m}}{s^3 + \left(\frac{b}{m} + \frac{k_D}{m}\right) s^2 + \left(\frac{k}{m} + \frac{k_P}{m}\right) s + \frac{k_I}{m}}$$

(c) The eigenvalues are solutions of,

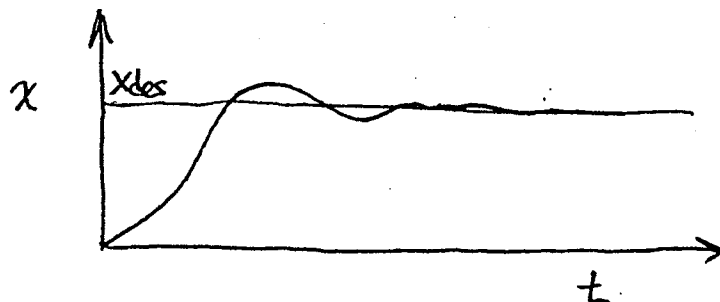
$$s^3 + \left(\frac{b}{m} + \frac{k_D}{m}\right) s^2 + \left(\frac{k}{m} + \frac{k_P}{m}\right) s + \frac{k_I}{m} = 0$$

without control

$$s^2 + \frac{b}{m} s + \frac{k}{m} = 0$$

The control allows independent adjustment of all coefficients of s in the characteristic eqn. Thus we can place the closed loop eigenvalues anywhere we desire.

(d)



Final value theorem shows that,

$$\left. \frac{x}{x_{des}} \right|_{s \rightarrow 0} \rightarrow 1$$

so the controller will move the mass to the proper final location. And with gain adjust we can probably get the response to be acceptable.

Notice that without integral control, $k_I = 0$, the system will not reach x_{des} .

6-20

$$\text{a. } \begin{bmatrix} \dot{p}_4 \\ \dot{q}_5 \end{bmatrix} = \begin{bmatrix} 0 & 1/C_5 \\ -1/I_4 & -(1/R_2 + 1/R_6)/C_5 \end{bmatrix} \begin{bmatrix} p_4 \\ q_5 \end{bmatrix} + \begin{bmatrix} 0 \\ 1/R_2 \end{bmatrix} E_1$$

b.

$$\begin{vmatrix} s & -1/C_5 \\ 1/I_4 & s + (1/R_2 + 1/R_6)/C_5 \end{vmatrix} = s^2 + \frac{1}{C_5} \left(\frac{1}{R_2} + \frac{1}{R_6} \right) s + \frac{1}{I_4 C_5} = s^2 + \frac{1}{C} \left(\frac{1}{R_a} + \frac{1}{R_b} \right) s + \frac{1}{LC} = 0$$

$$\text{c. } \omega_n = (1/LC)^{1/2}, \quad 2\zeta\omega_n = \frac{1}{C} \left(\frac{1}{R_a} + \frac{1}{R_b} \right), \quad \zeta = \frac{1}{2} \left(\frac{L}{C} \right)^{1/2} \left(\frac{1}{R_a} + \frac{1}{R_b} \right).$$

6-21.

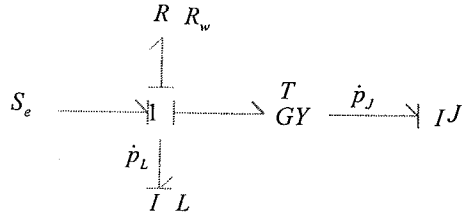
$$\text{a. } \dot{p}_1 = \frac{-(R_2 + R_3 + R_5)p_1}{I_1} + R_5 F_6(t), \quad \tau = \frac{I_1}{(R_2 + R_3 + R_5)} = \frac{M}{(B_2 + B_3 + B_5)}$$

$$\text{b. } \dot{p}_1 = \frac{-(R_2 + R_3)p_1}{I_1} + E_6(t), \quad \tau = \frac{I_1}{(R_2 + R_3)} = \frac{M}{(B_2 + B_3)}$$

Problem 6-22

(a) The bond graph is shown below.

The state variables are the momentum variables p_L and p_J



The state equations are:

$$\frac{d}{dt} \begin{bmatrix} p_L \\ p_J \end{bmatrix} = \begin{bmatrix} -\frac{R_w}{L} & -\frac{T}{J} \\ \frac{T}{L} & 0 \end{bmatrix} \begin{bmatrix} p_L \\ p_J \end{bmatrix} + \begin{bmatrix} 1 \\ 0 \end{bmatrix} e_c \quad (\text{SP6-22-1})$$

(b) In the s -domain these become,

$$\begin{bmatrix} s + \frac{R_w}{L} & \frac{T}{J} \\ -\frac{T}{L} & s \end{bmatrix} \begin{bmatrix} p_L(s) \\ p_J(s) \end{bmatrix} = \begin{bmatrix} 1 \\ 0 \end{bmatrix} e_c(s) \quad (\text{SP6-22-2})$$

The characteristic equation is $\det[sI - A] = 0$ or

$$s^2 + \frac{R_w}{L}s + \frac{T^2}{JL} = 0 \quad (\text{SP6-22-3})$$

The eigenvalues are,

$$s_{1,2} = -\frac{1}{2} \frac{R_w}{L} \pm \sqrt{\left(\frac{1}{2} \frac{R_w}{L}\right)^2 - \frac{T^2}{JL}} \quad (\text{SP6-22-4})$$

If the resistance is set to zero, the undamped eigenvalues are,

$$s_{1,2} = \pm j \sqrt{\frac{T^2}{JL}} = \pm j \omega_n$$

thus the natural frequency is,

$$\omega_n = \sqrt{\frac{T^2}{JL}} \quad (\text{SP6-22-5})$$

To obtain the damping ratio, set,

$$\zeta \omega_n = \frac{1}{2} \frac{R_w}{L}$$

thus,

$$\zeta = \frac{1}{2} \frac{R_w}{L} \frac{1}{\omega_n} \quad (\text{SP6-22-6})$$

(c) Using Cramer's Rule we obtain the transfer function relating the angular momentum $p_J(s)$ to the input voltage $e_c(s)$, thus,

$$\frac{p_J}{e_c}(s) = \frac{\begin{vmatrix} s + \frac{R_w}{L} & 1 \\ -\frac{T}{L} & 0 \end{vmatrix}}{s^2 + \frac{R_w}{L}s + \frac{T^2}{JL}} = -\frac{\frac{T}{L}}{s^2 + \frac{R_w}{L}s + \frac{T^2}{JL}} \quad (\text{SP6-22-7})$$

and we obtain the angular velocity ω by dividing the angular momentum by the moment of inertia, thus,

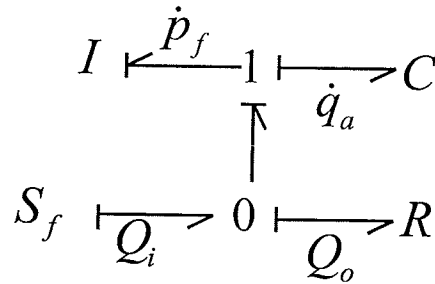
$$\frac{\omega}{e_c}(s) = -\frac{\frac{T}{JL}}{s^2 + \frac{R_w}{L}s + \frac{T^2}{JL}} \quad (\text{SP6-22-8})$$

(d) The transfer function for angular position is simply obtained from the transfer function for angular velocity by dividing by s (i. e. integration in the s -domain)

$$\frac{\theta}{e_c}(s) = -\frac{\frac{T}{JL}}{s \left(s^2 + \frac{R_w}{L}s + \frac{T^2}{JL} \right)} \quad (\text{SP6-22-9})$$

Problem 6.23

The bond graph is given in the problem statement and repeated here.



(a) The state equations come directly from the bond graph by following the causality indicated, thus,

$$\begin{aligned} \frac{d}{dt} p_f &= -\frac{q_a}{C_a} + R_f(Q_i - \frac{p_f}{I_f}) \\ \frac{d}{dt} q_a &= \frac{p_f}{I_f} \end{aligned} \quad (\text{SP6-23-1})$$

or

$$\frac{d}{dt} \begin{bmatrix} p_f \\ q_a \end{bmatrix} = \begin{bmatrix} -\frac{R_f}{I_f} & -\frac{1}{C_a} \\ \frac{1}{I_f} & 0 \end{bmatrix} \begin{bmatrix} p_f \\ q_a \end{bmatrix} + \begin{bmatrix} R_f \\ 0 \end{bmatrix} Q_i \quad (\text{SP6-23-2})$$

(b) In the s -domain,

$$\begin{bmatrix} s + \frac{R_f}{I_f} & \frac{1}{C_a} \\ -\frac{1}{I_f} & s \end{bmatrix} \begin{bmatrix} p_f(s) \\ q_a(s) \end{bmatrix} = \begin{bmatrix} R_f \\ 0 \end{bmatrix} Q_i(s) \quad (\text{SP6-23-3})$$

The characteristic equation is the determinant of the left side,

$$s^2 + \frac{R_f}{I_f}s + \frac{1}{I_f C_a} = 0 \quad (\text{SP6-23-4})$$

and the eigenvalues are the solutions to the characteristic equation,

$$s_{1,2} = -\frac{1}{2} \frac{R_f}{I_f} \pm \sqrt{\left(\frac{1}{2} \frac{R_f}{I_f}\right)^2 - \frac{1}{I_f C_a}} \quad (\text{SP6-23-5})$$

It is convenient to let

$$\omega_0^2 = \frac{1}{I_f C_a}$$

$$\zeta \omega_0 = \frac{1}{2} \frac{R_f}{I_f}$$

such that the eigenvalues become,

$$s_{1,2} = -\zeta \omega_0 \pm \sqrt{(\zeta \omega_0)^2 - \omega_0^2} \quad (\text{SP6-23-6})$$

(c) The output we are after is Q_0 and we can write the output equation for Q_0 by following the causality on the bond graph with the result,

$$Q_0 = Q_i - \frac{p_f}{I_f}$$

Thus,

$$\frac{Q_0}{Q_i} = 1 - \frac{1}{I_f} \frac{p_f}{Q_i} \quad (\text{SP6-23-7})$$

We obtain the desired transfer function by using Cramer's rule and deriving first the transfer function for $\frac{p_f}{Q_i}(s)$ and then follow the operations above to obtain the desired transfer function.

$$\frac{p_f}{Q_i}(s) = \frac{R_f s}{s^2 + \frac{R_f}{I_f} s + \frac{1}{I_f C_a}} \quad (\text{SP6-23-8})$$

and

$$\frac{Q_0}{Q_i}(s) = 1 - \frac{\frac{R_f}{I_f} s}{s^2 + \frac{R_f}{I_f} s + \frac{1}{I_f C_a}} = \frac{s^2 + \frac{1}{I_f C_a}}{s^2 + \frac{R_f}{I_f} s + \frac{1}{I_f C_a}} \quad (\text{SP6-23-9})$$

or using the parameter definitions above,

$$\frac{Q_0}{Q_i}(s) = \frac{s^2 + \omega_0^2}{s^2 + 2\zeta \omega_0 s + \omega_0^2} \quad (\text{SP6-23-10})$$

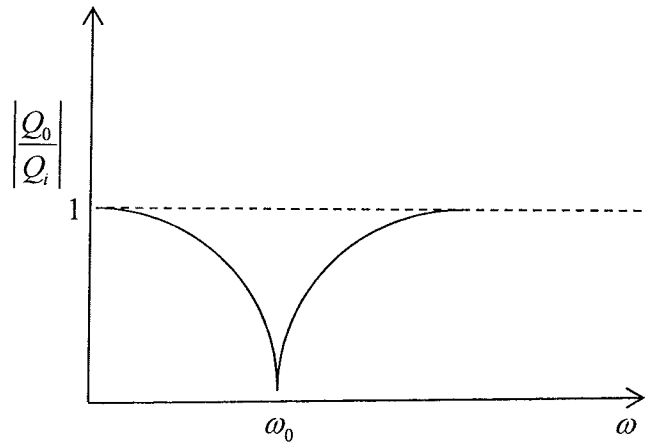
(d) The complex frequency response function comes from letting $s = j\omega$ in the transfer function, thus

$$\frac{Q_0}{Q_i}(j\omega) = \frac{\omega_0^2 - \omega^2}{\omega_0^2 - \omega^2 + j2\zeta \omega_0 \omega} \quad (\text{SP6-23-11})$$

and the magnitude of the frequency response is the magnitude of the complex frequency response function, thus

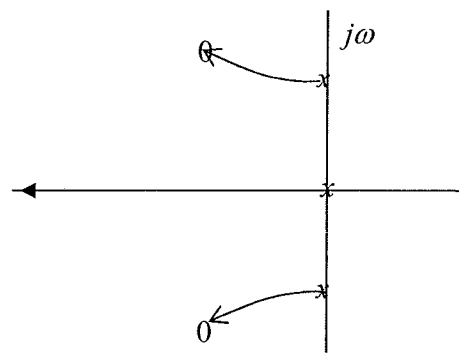
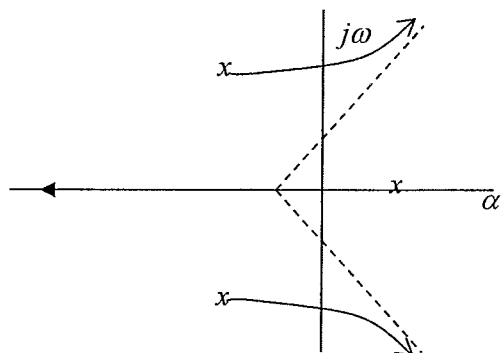
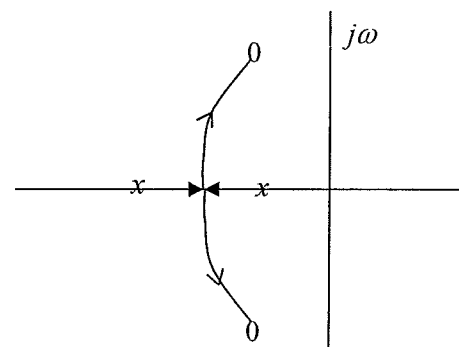
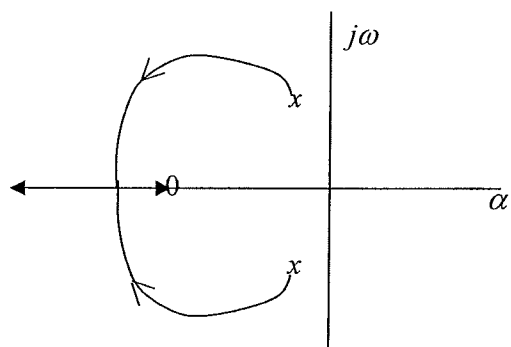
$$\left| \frac{Q_o}{Q_i} \right| = \frac{[(\omega_0^2 - \omega^2)^2]^{1/2}}{[(\omega_0^2 - \omega^2)^2 + (2\zeta\omega_0\omega)^2]^{1/2}} \quad (\text{SP6-23-12})$$

This frequency response is interesting in that at very low input frequencies the response is unity (i. e. set $\omega = 0$ and evaluate) and at very high input frequencies the response is unity (i. e. imagine that the input frequency is much larger than ω_0). However when the input frequency equals ω_0 the response is ZERO! The frequency response would look something like that shown below.

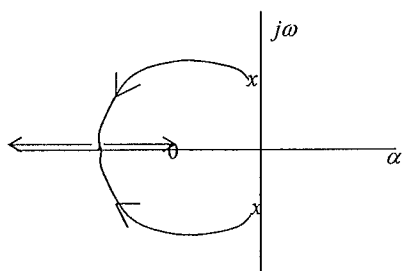


The side branch accumulator shown in the problem statement for Problem 6-24 is very useful in hydraulic systems where the system pressure is generated by a multi-piston positive displacement pump. For such pumps the output flow has a steady state component and an oscillatory component due to the motion of the pistons. The side branch accumulator removes much of the oscillation and the downstream flow is very smooth.

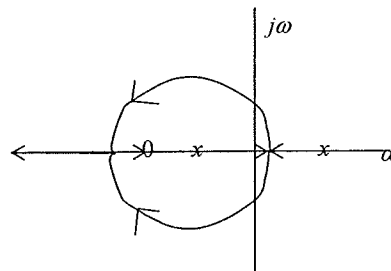
Problem 6-24



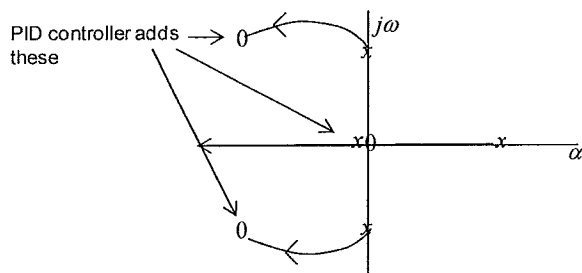
Problem 6.25



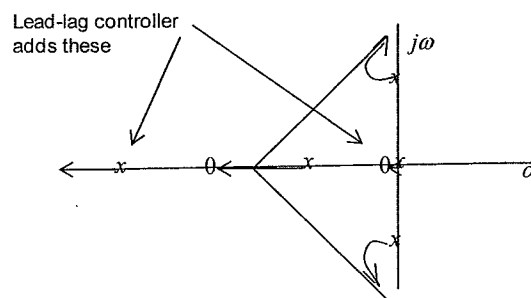
PD control yields stable closed loop behavior and will produce nicely located closed loop eigenvalues



PD control stabilizes the system and potentially produces nicely located closed loop eigenvalues



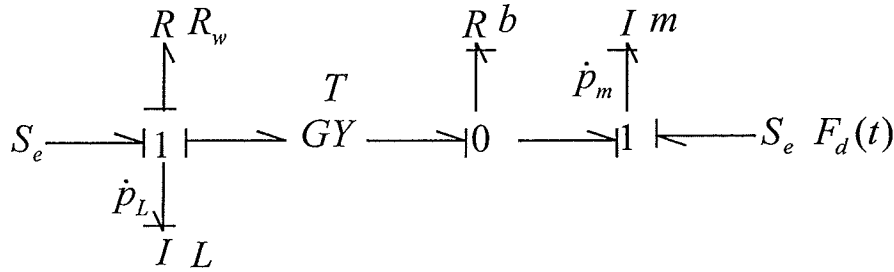
PID adds the 2 arbitrary zeros and the pole at the origin. This pole effectively cancels the system zero and yields a stable closed loop system



Lead-lag control pulls the centroid to the left allowing stable closed loop behavior for some range of system gain. The performance needs to be tested to determine if acceptable response can be achieved

Problem 6-26

The bond graph from the problem statement is repeated here with the state variables identified



(a) The state equations come directly from the bond graph as,

$$\begin{aligned} \frac{d}{dt} p_L &= e_c - \frac{R_w}{L} p_L - T \left[\frac{p_m}{m} + \frac{1}{b} T \frac{p_L}{L} \right] \\ \frac{d}{dt} p_m &= F_d + T \frac{p_L}{L} \end{aligned} \quad (\text{SP6-26-1})$$

In matrix form these become,

$$\frac{d}{dt} \begin{bmatrix} p_L \\ p_m \end{bmatrix} = \begin{bmatrix} -\left(\frac{R_w}{L} + \frac{T^2}{bL}\right) & -\frac{T}{m} \\ \frac{T}{L} & 0 \end{bmatrix} \begin{bmatrix} p_L \\ p_m \end{bmatrix} + \begin{bmatrix} 1 \\ 0 \end{bmatrix} e_c + \begin{bmatrix} 0 \\ 1 \end{bmatrix} F_d \quad (\text{SP6-26-2})$$

(b)

In the s -domain we get,

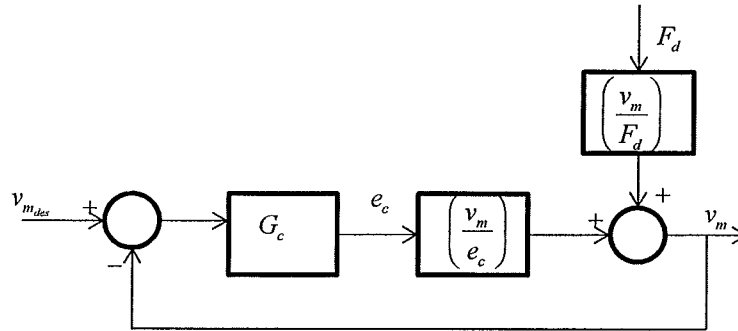
$$\begin{bmatrix} s + \left(\frac{R_w}{L} + \frac{T^2}{bL}\right) & \frac{T}{m} \\ -\frac{T}{L} & s \end{bmatrix} \begin{bmatrix} p_L(s) \\ p_m(s) \end{bmatrix} = \begin{bmatrix} 1 \\ 0 \end{bmatrix} e_c + \begin{bmatrix} 0 \\ 1 \end{bmatrix} F_d \quad (\text{SP6-26-3})$$

Use Cramer's Rule to obtain the transfer functions. The easiest way to obtain the velocity transfer functions $\frac{v_m}{e_c}(s)$ and $\frac{v_m}{F_d}(s)$ is to derive the transfer functions for the momentum p_m and divide the result by the mass m . Thus,

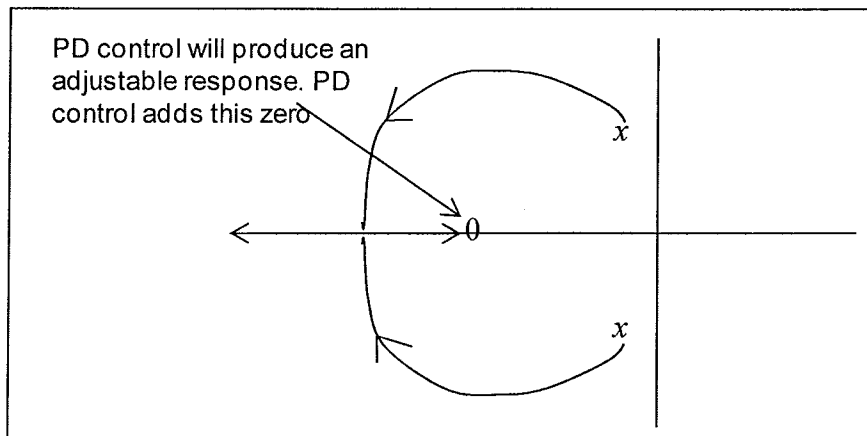
$$\frac{v_m}{e_c}(s) = \frac{\frac{1}{m} \text{Det} \begin{bmatrix} s + \left(\frac{R_w}{L} + \frac{T^2}{bL}\right) & 1 \\ -\frac{T}{L} & 0 \end{bmatrix}}{s^2 + \left(\frac{R_w}{L} + \frac{T^2}{bL}\right)s + \frac{T^2}{mL}} = \frac{\frac{T}{mL}}{s^2 + \left(\frac{R_w}{L} + \frac{T^2}{bL}\right)s + \frac{T^2}{mL}} \quad (\text{SP6-26-4})$$

$$\frac{v_m}{F_d}(s) = \frac{\frac{1}{m} \text{Det} \begin{bmatrix} s + (\frac{R_w}{L} + \frac{T^2}{bL}) & 0 \\ -\frac{T}{L} & 1 \end{bmatrix}}{s^2 + (\frac{R_w}{L} + \frac{T^2}{bL})s + \frac{T^2}{mL}} = \frac{\frac{1}{m} \left[s + (\frac{R_w}{L} + \frac{T^2}{bL}) \right]}{s^2 + (\frac{R_w}{L} + \frac{T^2}{bL})s + \frac{T^2}{mL}} \quad (\text{SP6-26-5})$$

(c)



(d) PD control produces a stable closed loop system with potential for good overall response. The control would be as shown in the figure below.



(e) Representing PD control as $G_c(s) = K(s + z)$, the closed loop response of the velocity to the disturbance force is derived as,

$$\frac{v_m}{F_d} \Big|_{cl} = \frac{\left(\frac{v_m}{F_d}\right)}{1 + G_c \left(\frac{v_m}{e_c}\right)} = \frac{\frac{1}{m} \left[s + (\frac{R_w}{L} + \frac{T^2}{bL}) \right]}{s^2 + (\frac{R_w}{L} + \frac{T^2}{bL})s + \frac{T^2}{mL} + K(s + z) \frac{T}{mL}} \quad (\text{SP6-26-6})$$

or

$$\frac{v_m}{F_d} \Big|_{cl} = \frac{\frac{1}{m} \left[s + \left(\frac{R_w}{L} + \frac{T^2}{bL} \right) \right]}{s^2 + \left(\frac{R_w}{L} + \frac{T^2}{bL} + \frac{KT}{mL} \right) s + \frac{T^2}{mL} + Kz \frac{T}{mL}} \quad (\text{SP6-26-7})$$

If we let $s = 0$ in this transfer function to test for the final value to a step input, we obtain,

$$\frac{v_m}{F_d} \Big|_{cl, s \rightarrow 0} = \frac{\frac{1}{m} \left(\frac{R_w}{L} + \frac{T^2}{bL} \right)}{\frac{T^2}{mL} + Kz \frac{T}{mL}} \quad (\text{SP6-26-8})$$

In order to keep the response to the disturbance force small, it is necessary to have a large system gain K and/or have the location of the controller zero far out on the negative real axis of the s -plane. This will affect the response of the system to a desired velocity. Thus there will exist a tradeoff between desired response and disturbance rejection.

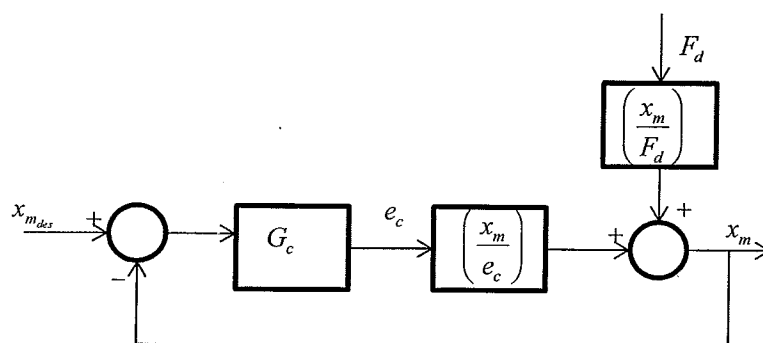
Problem 6.27

The transfer functions from Problem 6-26 are repeated here only for position as output rather than velocity as output.

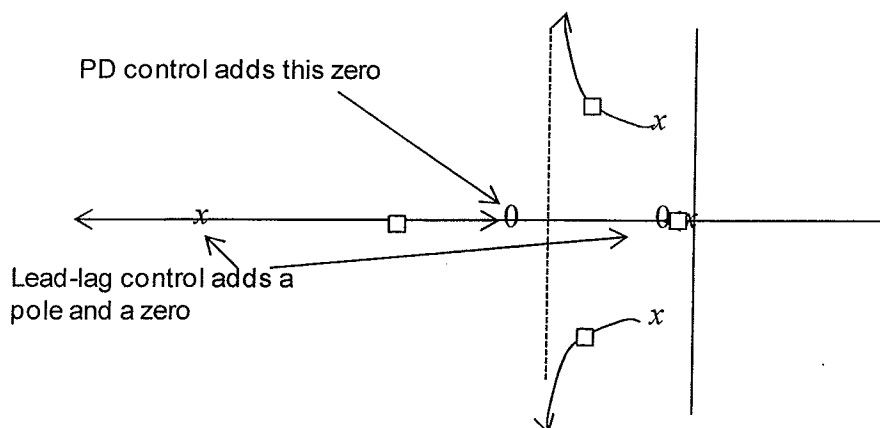
$$\frac{x_m}{e_c}(s) = \frac{\frac{T}{mL}}{s(s^2 + (\frac{R_w}{L} + \frac{T^2}{bL})s + \frac{T^2}{mL})} \quad (\text{SP6-27-1})$$

$$\frac{x_m}{F_d}(s) = \frac{\frac{1}{m} \left[s + (\frac{R_w}{L} + \frac{T^2}{bL}) \right]}{s(s^2 + (\frac{R_w}{L} + \frac{T^2}{bL})s + \frac{T^2}{mL})} \quad (\text{SP6-27-2})$$

(c) The general block diagram for the control system is shown in the figure below.



(d) We now have 3 open loop zeros and PD control alone produces a pole-zero excess of 2 thus creating 2, 90 degree asymptotes that very likely will be in the unstable region of the s-plane. A lead-lag compensator is added to the PD controller and this is shown in the Root Locus below. The lead-lag compensator allows us to “pull” the asymptotes to the left leading to the possibility of a stable, well behaved closed loop system.



(e) Representing PD plus lead-lag control as $G_c(s) = K(s+z)\frac{(s+z_1)}{(s+p_1)}$, the closed loop response of the velocity to the disturbance force is derived as,

$$\frac{x_m}{F_d} \Big|_{cl} = \frac{\frac{1}{m} \left[s + \left(\frac{R_w}{L} + \frac{T^2}{bL} \right) \right] (s+p_1)}{s^4 + \left[p_1 + \frac{R_w}{L} + \frac{T^2}{bL} \right] s^3 + \left[\frac{T^2}{mL} + \left(\frac{R_w}{L} + \frac{T^2}{bL} \right) p_1 + K \frac{T}{mL} \right] s^2 + \left[\frac{T^2}{mL} p_1 + K \frac{T}{mL} (z+z_1) \right] s + K \frac{T}{mL} z z_1} \quad (\text{SP6-27-3})$$

If we let $s=0$ in this transfer function to test for the final value to a step input, we obtain,

$$\frac{x_m}{F_d} \Big|_{cl, s \rightarrow 0} = \frac{\frac{1}{m} \left(\frac{R_w}{L} + \frac{T^2}{bL} \right) p_1}{K \frac{T}{mL} z z_1} \quad (\text{SP6-27-4})$$

This result indicates that it will be a problem having good disturbance rejection because the basic structure of the lead-lag controller is such that $\frac{p_1}{z_1}$ is a large value perhaps 10 or 100. Thus a very large overall gain K is required to keep the response small. This needs to be investigated further.

Problem 6.28

(a) The state equations are put into the s-domain with the result,

$$\begin{bmatrix} s + \frac{b_s}{m_s} & -k & 0 \\ \frac{1}{m_s} & s & 0 \\ -\frac{1}{m_s} & 0 & s \end{bmatrix} \begin{bmatrix} P_s \\ Q_s \\ X \end{bmatrix} = \begin{bmatrix} 0 \\ 1 \\ 0 \end{bmatrix} V_c + \begin{bmatrix} 1 \\ 0 \\ 0 \end{bmatrix} F_d \quad (\text{SP6-28-1})$$

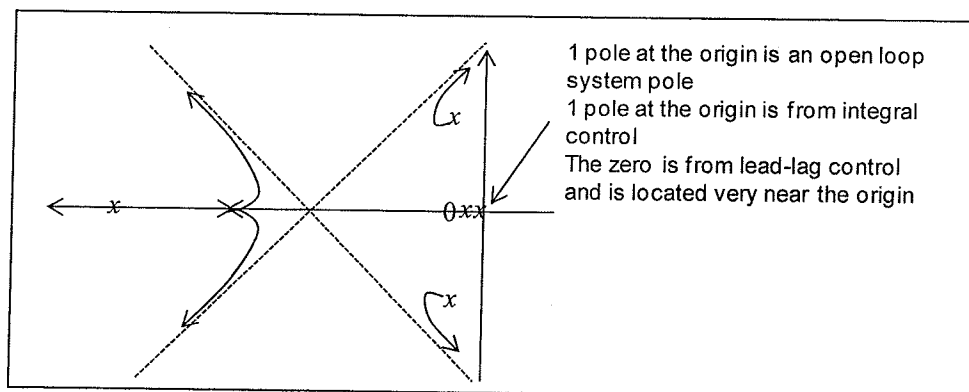
To derive the transfer functions $\frac{X}{V_c}(s)$ and $\frac{X}{F_d}(s)$, use Cramer's Rule and substitute the appropriate forcing vector into the 3rd column of the matrix $[sI - A]$. Take the appropriate determinants with the result,

$$\frac{X}{V_c}(s) = \frac{\frac{k_s}{m_s}}{s(s^2 + \frac{b_s}{m_s}s + \frac{k_s}{m_s})} \quad (\text{SP6-28-2})$$

$$\frac{X}{F_d}(s) = \frac{\frac{1}{m_s}s}{s(s^2 + \frac{b_s}{m_s}s + \frac{k_s}{m_s})}$$

(b) In the Root Locus below the 3 open loop poles are shown along with poles and zeros from a proposed controller. For a load leveling application we desire that the response to a disturbance force return to the original desired position. To insure that this occurs it is proposed to use integral control. In the Root Locus below, the second pole at the origin is from the integral control action, $G_c(s) = \frac{K_I}{s}$. Integral control alone would yield 4 open

loop poles with 4 asymptotes at $\pm 45^\circ$ and $\pm 135^\circ$. The centroid of such a configuration would be on the real axis somewhere near the 4 poles. Such a control system would be unstable for virtually any value of gain, depending on the amount of passive damping in the system. It is further proposed to add a lead-lag controller which can "pull" the centroid to the left and yield the closed loop system shown in the Root Locus below.



This control action would be,

$$G_c(s) = \frac{K (s + z)}{s (s + p)}$$

and it is possible to obtain stable closed loop behavior.

(c) Block diagram reduction yields the closed loop transfer function for response to a disturbance as,

$$\frac{X}{F_d} \Big|_{cl} = \frac{\left(\frac{X}{F_d} \right)}{1 + G_c \left(\frac{X}{V_c} \right)} = \frac{\frac{1}{m_s} s^2 (s + p)}{s^5 + \left(\frac{b_s}{m_s} + p \right) s^4 + \left(\frac{b_s}{m_s} p + \frac{k_s}{m_s} \right) s^3 + \left(\frac{k_s}{m_s} p \right) s^2 + K \frac{k_s}{m_s} s + K z \frac{k_s}{m_s}} \quad (\text{SP6-28-3})$$

Letting $s \rightarrow 0$ shows that the final value of response to a step change in disturbance force is zero indicating that the system returns to the original position.

Problem 6.29

The s-domain representation from problem 6.28 is repeated here for convenience.

$$\begin{bmatrix} s + \frac{b_s}{m_s} & -k & 0 \\ \frac{1}{m_s} & s & 0 \\ -\frac{1}{m_s} & 0 & s \end{bmatrix} \begin{bmatrix} P_s \\ Q_s \\ X \end{bmatrix} = \begin{bmatrix} 0 \\ 1 \\ 0 \end{bmatrix} V_c + \begin{bmatrix} 1 \\ 0 \\ 0 \end{bmatrix} F_d \quad (\text{SP6-29-1})$$

This time the control is state variable feedback with the form,

$$V_c = K \left(x_{ref} - [k_{p_s} \ k_{q_s} \ k_x] \begin{bmatrix} P_s \\ Q_s \\ X \end{bmatrix} \right) \quad (\text{SP6-29-2})$$

Substituting into the matrix equation,

$$\begin{bmatrix} s + \frac{b_s}{m_s} & -k_s & 0 \\ \frac{1}{m_s} + Kk_{p_s} & s + Kk_{q_s} & Kk_x \\ -\frac{1}{m_s} & 0 & s \end{bmatrix} \begin{bmatrix} P_s \\ Q_s \\ X \end{bmatrix} = \begin{bmatrix} 0 \\ 1 \\ 0 \end{bmatrix} Kx_{ref} + \begin{bmatrix} 1 \\ 0 \\ 0 \end{bmatrix} F_d \quad (\text{SP6-29-3})$$

From this representation we use Cramer's rule to derive the closed loop transfer function for $\frac{X}{F_d}(s)$,

$$\frac{X}{F_d}(s) = \frac{\frac{1}{m_s}(s + Kk_{q_s})}{s^3 + \left(\frac{b_s}{m_s} + Kk_{q_s}\right)s^2 + \left(\frac{b_s}{m_s}Kk_{q_s} + \frac{k_s}{m_s} + k_sKk_{p_s}\right)s + \frac{k_s}{m_s}Kk_x} \quad (\text{SP6-29-4})$$

As guaranteed with state variable feedback, the closed loop eigenvalues can be arbitrarily placed. There are sufficient gains to affect each coefficient of s in the closed loop denominator.

The final value of the response to a step change in disturbance force is not zero as can be seen by letting $s \rightarrow 0$ in the transfer function. The result is,

$$\frac{X}{F_d} \Big|_{s \rightarrow 0} = \frac{Kk_{q_s}}{k_sKk_x} \quad (\text{SP6-29-5})$$

It may not be practical to let k_{q_s} be zero due to its effect on the closed loop eigenvalues. So what is done in this situation is add an additional integral control action which will drive the error to zero. One must be careful when doing this to not make the integral control too fast as this will drive the system unstable.

Problem 6.30

The equations of motion from the bond graph and put into matrix format are,

$$(a) \quad \frac{d}{dt} \begin{bmatrix} p_s \\ p_a \\ q_s \\ q_a \end{bmatrix} = \begin{bmatrix} -\frac{b}{m} - \frac{b_a}{m} & \frac{b_a}{m_a} & k & -k_a \\ \frac{b_a}{m} & -\frac{b_a}{m_a} & 0 & k_a \\ -\frac{1}{m} & 0 & 0 & 0 \\ \frac{1}{m} & -\frac{1}{m_a} & 0 & 0 \end{bmatrix} \begin{bmatrix} p_s \\ p_a \\ q_s \\ q_a \end{bmatrix} + \begin{bmatrix} b \\ 0 \\ 1 \\ 0 \end{bmatrix} v_{in} \begin{bmatrix} -1 \\ 1 \\ 0 \\ 0 \end{bmatrix} F_c \quad (SP-30-1)$$

(b) In the s-domain these become,

$$\begin{bmatrix} s + \frac{b}{m} + \frac{b_a}{m} & -\frac{b_a}{m_a} & -k & k_a \\ -\frac{b_a}{m} & s + \frac{b_a}{m_a} & 0 & -k_a \\ \frac{1}{m} & 0 & s & 0 \\ -\frac{1}{m} & \frac{1}{m_a} & 0 & s \end{bmatrix} \begin{bmatrix} p_s \\ p_a \\ q_s \\ q_a \end{bmatrix} = \begin{bmatrix} b \\ 0 \\ 1 \\ 0 \end{bmatrix} v_{in} \begin{bmatrix} -1 \\ 1 \\ 0 \\ 0 \end{bmatrix} F_c \quad (SP-30-2)$$

It is bit tedious to carry out the steps to derive the transfer functions but with a little patience these become,

$$\frac{v_m}{v_{in}}(s) = \frac{(\frac{b}{m}s + \frac{k}{m})(s^2 + \frac{b_a}{m_a}s + \frac{k_a}{m_a})}{D(s)} \quad (SP-30-3)$$

$$\frac{v_m}{F_c}(s) = \frac{-\frac{1}{m}s^3}{D(s)}$$

where,

$$D(s) = s^4 + [\frac{b}{m} + \frac{b_a}{m_a} + \frac{b_a}{m}]s^3 + [\frac{k_a}{m_a} + \frac{k}{m} + \frac{k_a}{m} + \frac{b}{m} \frac{b_a}{m_a}]s^2 + [\frac{b}{m} \frac{k_a}{m_a} + \frac{b_a}{m_a} \frac{k}{m}]s + \frac{k}{m} \frac{k_a}{m_a} \quad (SP-30-4)$$

(c) The complex frequency response comes from substituting $s = j\omega$ in the appropriate transfer function with the result,

$$\frac{v_m(j\omega)}{v_{in}} = \frac{(\frac{k}{m} + j\frac{b}{m}\omega)(\frac{k_a}{m_a} - \omega^2 + j\frac{b_a}{m_a}\omega)}{\omega^4 - [\frac{k_a}{m_a} + \frac{k}{m} + \frac{k_a}{m} + \frac{b}{m}\frac{b_a}{m_a}]\omega^2 + \frac{k}{m}\frac{k_a}{m_a} + j\omega[(\frac{b}{m}\frac{k_a}{m_a} + \frac{b_a}{m_a}\frac{k}{m}) - (\frac{b}{m} + \frac{b_a}{m_a} + \frac{b_a}{m})\omega^2]} \quad (\text{SP-30-5})$$

The magnitude response comes from the taking the magnitude of the complex frequency response function with the result,

$$\left| \frac{v_m(j\omega)}{v_{in}} \right| = \frac{[(\frac{k}{m})^2 + (\frac{b}{m}\omega)^2]^{1/2} [(\frac{k_a}{m_a} - \omega^2)^2 + (\frac{b_a}{m_a}\omega)^2]^{1/2}}{[\omega^4 - (\frac{k_a}{m_a} + \frac{k}{m} + \frac{k_a}{m} + \frac{b}{m}\frac{b_a}{m_a})\omega^2 + \frac{k}{m}\frac{k_a}{m_a}]^2 + \omega^2 \{ \frac{b}{m}\frac{k_a}{m_a} + \frac{b_a}{m_a}\frac{k}{m} - (\frac{b}{m} + \frac{b_a}{m_a} + \frac{b_a}{m})\omega^2 \}^2]^{1/2}} \quad (\text{SP-30-6})$$

Notice that if the damper b_a is very small then the numerator will be near zero at the operating frequency $\omega^2 = \frac{k_a}{m_a}$.

(d) In order to use control to cancel the effect of the actuator damping it is proposed to use,

$$F_c = -Kb_a(\frac{p_s}{m} - \frac{p_a}{m_a}) \quad (\text{SP-30-7})$$

The gain K is included to facilitate testing the effect of the control force being too large or too small. If $K=1$ then the control force perfectly cancels the actuator damping force. If we substitute Eq. (SP-30-7) into (SP-30-1), we can carry out the steps and derive the closed loop transfer function relating the structure mass velocity to the input velocity. It is convenient to write the control force as shown below before substituting into (SP-30-1).

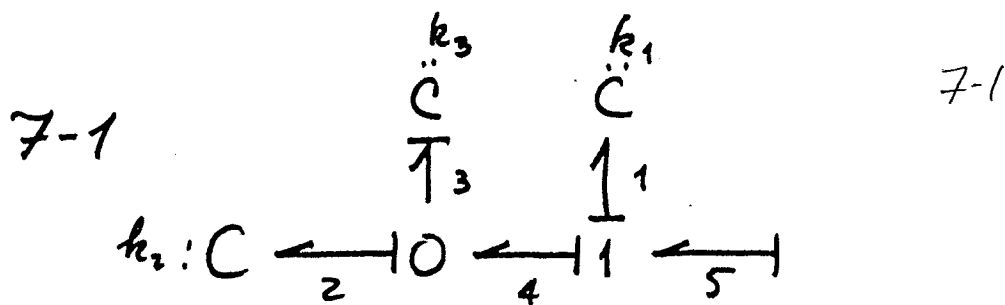
$$F_c = \begin{bmatrix} -K\frac{b_a}{m} & K\frac{b_a}{m_a} & 0 & 0 \end{bmatrix} \begin{bmatrix} p_s \\ p_a \\ q_s \\ q_a \end{bmatrix} \quad (\text{SP-30-8})$$

The result of the algebra is the closed loop transfer function,

$$\frac{v_m}{v_{in}}(s) = \frac{\left(\frac{b}{m}s + \frac{k}{m}\right)\left(s^2 + \frac{b_a}{m_a}(1-K)s + \frac{k_a}{m_a}\right)}{s^4 + \left[\frac{b}{m} + \frac{b_a}{m_a}(1-K) + \frac{b_a}{m}(1-K)\right]s^3 + \left[\frac{k_a}{m_a} + \frac{k}{m} + \frac{k_a}{m} + \frac{b}{m}\frac{b_a}{m_a}(1-K)\right]s^2 + \left[\frac{b}{m}\frac{k_a}{m_a} + \frac{b_a}{m_a}\frac{k}{m}(1-K)\right]s + \frac{k}{m}\frac{k_a}{m_a}} \quad (\text{SP-30-9})$$

You can turn this into a complex frequency response and then into a magnitude response if you wish. However, from the transfer function itself we see that if $K = 1$, the actuator damping term in the numerator will vanish and we will obtain a perfect zero response at the operating frequency $\omega^2 = \frac{k_a}{m_a}$.

We also can observe that if K is positive and made too large, then some terms in the denominator could become negative and this would yield unstable closed loop eigenvalues.



State Eqs: $\dot{q}_1 = f_5 \rightarrow \underline{q_1 = q_5 + \text{const}}$

$\dot{q}_2 = -f_3 + f_4 = -\dot{q}_3 + f_5 \rightarrow \underline{q_2 = -q_3 + q_5 + \text{const}}$

Deriv. Caus.

$$q_3 = \frac{e_3}{k_3} = \frac{e_2}{k_2} = \frac{k_2 q_2}{k_3}$$

so $q_2 = -\frac{k_2}{k_3} q_2 + q_5$

or $\underline{q_2 = \frac{k_3}{k_2 + k_3} q_5}$

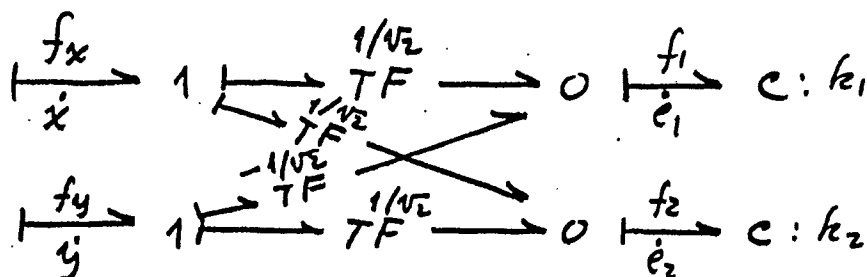
Output Eqn: $e_5 = e_1 + e_4 = k_1 q_1 + k_2 q_2$

$$e_5 = k_1 q_5 + \frac{k_2 k_3}{k_2 + k_3} q_5$$

$$e_5 = \underbrace{\left[k_1 + \frac{k_2 k_3}{k_2 + k_3} \right]}_{k_{eq}} q_5$$

7-2 (in this case, f 's are forces, e 's deflections)

$$e_1 = \frac{1}{\sqrt{2}} x - \frac{1}{\sqrt{2}} y ; e_2 = \frac{1}{\sqrt{2}} x + \frac{1}{\sqrt{2}} y \quad 7-2$$



$$f_x = \frac{1}{\sqrt{2}} f_1 + \frac{1}{\sqrt{2}} f_2 = \frac{k_1 e_1}{\sqrt{2}} + \frac{k_2 e_2}{\sqrt{2}}$$

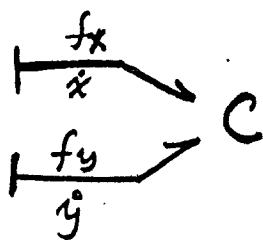
$$= \frac{k_1}{\sqrt{2}} \left(\frac{x}{\sqrt{2}} - \frac{y}{\sqrt{2}} \right) + \frac{k_2}{\sqrt{2}} \left(\frac{x}{\sqrt{2}} + \frac{y}{\sqrt{2}} \right)$$

$$= \left(\frac{k_1 + k_2}{2} \right) x + \left(\frac{k_2 - k_1}{2} \right) y$$

$$f_y = -\frac{1}{\sqrt{2}} f_1 + \frac{1}{\sqrt{2}} f_2 = -\frac{k_1 e_1}{\sqrt{2}} + \frac{k_2 e_2}{\sqrt{2}}$$

$$= -\frac{k_1}{\sqrt{2}} \left(\frac{1}{\sqrt{2}} x - \frac{1}{\sqrt{2}} y \right) + \frac{k_2}{\sqrt{2}} \left(\frac{1}{\sqrt{2}} x + \frac{1}{\sqrt{2}} y \right)$$

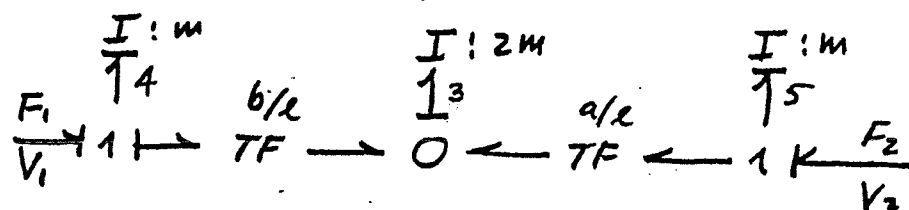
$$= \left(\frac{k_2 - k_1}{2} \right) x + \left(\frac{k_1 + k_2}{2} \right) y$$



7-3 $l = a + b$, velocity of middle mass $= V_3$

$$V_3 = \frac{b}{l} V_1 + \frac{a}{l} V_2$$

7-3



$$V_1 = p_4/m, \quad V_2 = p_5/m$$

$$\dot{p}_4 = F_1 - \frac{b}{l} \dot{p}_3, \quad \dot{p}_5 = F_2 - \frac{a}{l} \dot{p}_3$$

but $p_3 = 2m V_3 = 2m \left(\frac{b}{l} V_1 + \frac{a}{l} V_2 \right)$

$$= 2m \left(\frac{b}{l} \frac{p_4}{m} + \frac{a}{l} \frac{p_5}{m} \right)$$

combining

$$\left(1 + \frac{2b^2}{l^2} \right) \dot{p}_4 + \frac{2ab}{l^2} \dot{p}_5 = F_1$$

$$\frac{2ab}{l^2} \dot{p}_4 + \left(1 + \frac{2a^2}{l^2} \right) \dot{p}_5 = F_2$$

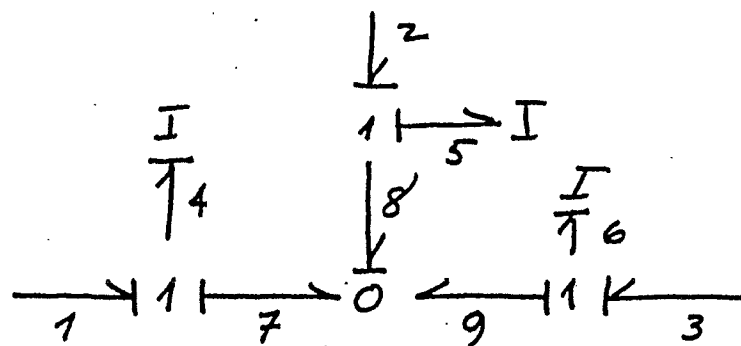
integrate in time and use first eq's above

$$\left(1 + \frac{2b^2}{l^2} \right) m V_1 + \frac{2ab}{l^2} m V_2 = P_1$$

$$\frac{2ab}{l^2} m V_1 + \left(1 + \frac{2a^2}{l^2} \right) m V_2 = P_2$$

I-field in mass matrix form

7-4



$$\frac{p_2}{A} = I$$

7-4

$$\dot{p}_4 = e_1 - e_7 = e_1 - e_8 = e_1 - (e_2 - \dot{p}_5)$$

$$\dot{p}_6 = e_3 - e_9 = e_3 - e_8 = e_3 - (e_2 - \dot{p}_5)$$

$$\text{but } p_5 = I f_5 = I f_8 = -I(f_7 + f_9) = -I\left(\frac{p_4}{I} + \frac{p_6}{I}\right)$$

$$\dot{p}_5 = -\dot{p}_4 - \dot{p}_6$$

Combining and solving, the result is

$$\left. \begin{aligned} \dot{p}_6 &= \frac{2}{3} \dot{p}_3 - \frac{1}{3} \dot{p}_1 - \frac{1}{3} \dot{p}_2 \\ \dot{p}_4 &= \frac{2}{3} \dot{p}_1 - \frac{1}{3} \dot{p}_2 - \frac{1}{3} \dot{p}_3 \end{aligned} \right\} \begin{aligned} p_6 &= \frac{2}{3} p_3 - \frac{1}{3} p_1 - \frac{1}{3} p_2 \\ p_4 &= \frac{2}{3} p_1 - \frac{1}{3} p_2 - \frac{1}{3} p_3 \end{aligned}$$

Finally

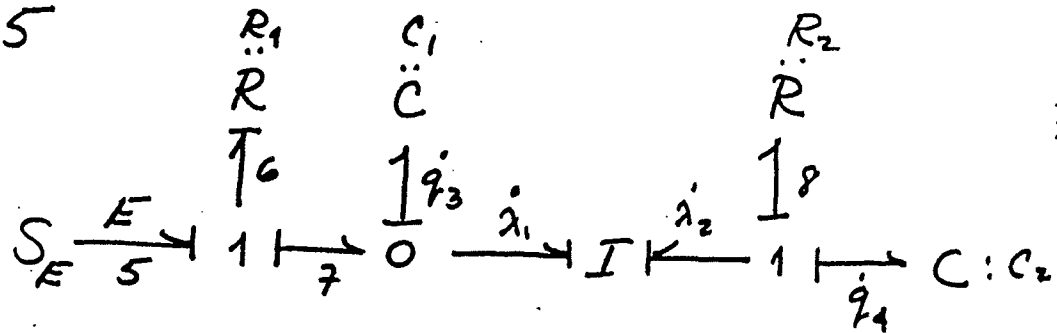
$$f_1 = \frac{p_4}{I} = \frac{1}{I} \left(\frac{2}{3} p_1 - \frac{1}{3} p_2 - \frac{1}{3} p_3 \right)$$

$$f_3 = \frac{p_6}{I} = \frac{1}{I} \left(\frac{2}{3} p_3 - \frac{1}{3} p_1 - \frac{1}{3} p_2 \right)$$

$$f_2 = -\frac{p_4}{I} - \frac{p_6}{I} = \frac{1}{I} \left(\frac{2}{3} p_2 - \frac{1}{3} p_1 - \frac{1}{3} p_3 \right)$$

Final I-fied relations in inverse mass matrix form

7-5



7-5

$$\begin{bmatrix} \lambda_1 \\ \lambda_2 \end{bmatrix} = \begin{bmatrix} L_1 & M_{12} \\ M_{12} & L_2 \end{bmatrix} \begin{bmatrix} \dot{i}_1 \\ \dot{i}_2 \end{bmatrix} \text{ or } \begin{bmatrix} \dot{i}_1 \\ \dot{i}_2 \end{bmatrix} = \frac{1}{L_1 L_2 - M_{12}^2} \begin{bmatrix} L_2 & -M_{12} \\ -M_{12} & L_1 \end{bmatrix}$$

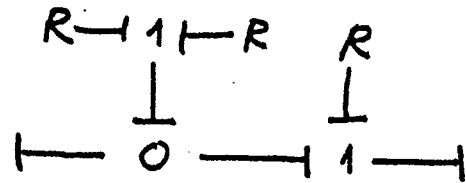
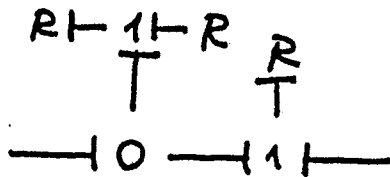
$$\dot{q}_3 = \left(E - \frac{q_3}{C_1} \right) \frac{1}{R_1} - \frac{L_2 \lambda_1 - M_{12} \lambda_2}{L_1 L_2 - M_{12}^2}$$

$$\dot{\lambda}_1 = \dot{q}_3 / C_3$$

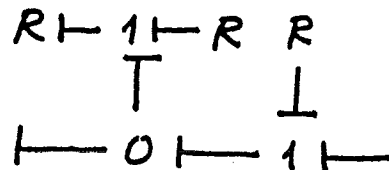
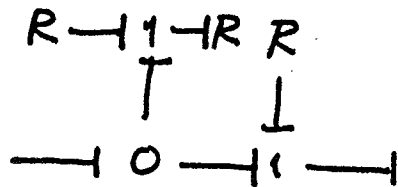
$$\dot{\lambda}_2 = -\dot{q}_4 / C_2 - R_2 \left(\frac{-M_{12} \lambda_1 + L_1 \lambda_2}{L_1 L_2 - M_{12}^2} \right)$$

$$\dot{q}_4 = -\frac{M_{12} \lambda_1 + L_1 \lambda_2}{L_1 L_2 - M_{12}^2}$$

7-6

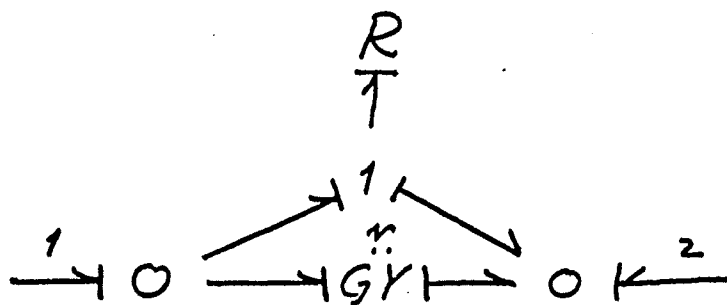


Ousager Formus



Casimir Formus

7-7



7-6

$$\begin{bmatrix} f_1 \\ f_2 \end{bmatrix} = \begin{bmatrix} 1/R & 1/r - 1/R \\ -1/R - 1/r & 1/R \end{bmatrix} \begin{bmatrix} e_1 \\ e_2 \end{bmatrix}$$

7-8

$$[e] = [R][f] ; \text{Power} = P = [f]^T [e]$$

$$P = [f]^T [R][f]$$

but $[e]^T = [f]^T [R]^T$ so another expression for P is $P = [e]^T [f] = [f]^T [R]^T [f]$

if R is antisymmetric, $[R]^T = -[R]$

so $P = -P$ which means $P = 0$

7-9

$$e_1 = m_1 e_3 = m_1 C_3^{-1}(\dot{q}_3) = m_1 C_3^{-1}(m_1 \dot{q}_1 + m_2 \dot{q}_2) \quad 7-7$$

$$e_2 = m_2 e_3 = m_2 C_3^{-1}(\dot{q}_3) = m_2 C_3^{-1}(m_1 \dot{q}_1 + m_2 \dot{q}_2)$$

$$\frac{\partial e_1}{\partial \dot{q}_2} = m_1 \frac{\partial C_3^{-1}}{\partial \dot{q}_3} \cdot \frac{\partial \dot{q}_3}{\partial \dot{q}_2} = m_1 \frac{\partial C_3^{-1}}{\partial \dot{q}_3} \cdot m_2$$

$$\frac{\partial e_2}{\partial \dot{q}_1} = m_2 \frac{\partial C_3^{-1}}{\partial \dot{q}_3} \cdot \frac{\partial \dot{q}_3}{\partial \dot{q}_1} = m_2 \frac{\partial C_3^{-1}}{\partial \dot{q}_3} \cdot m_1 = \frac{\partial e_1}{\partial \dot{q}_2}$$

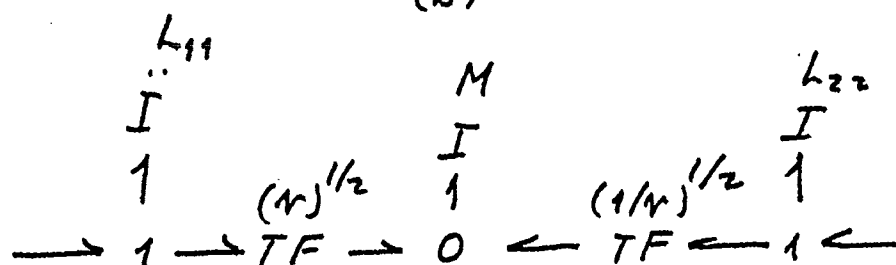
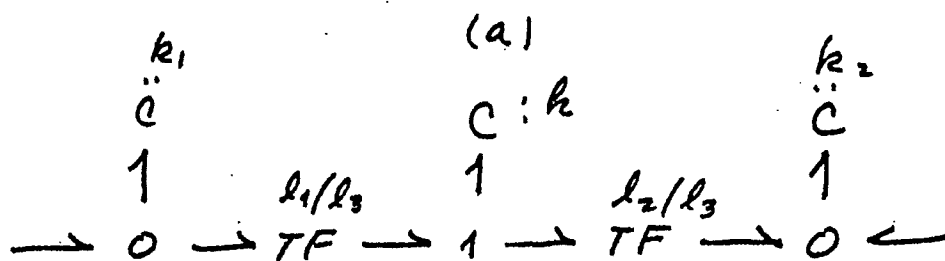
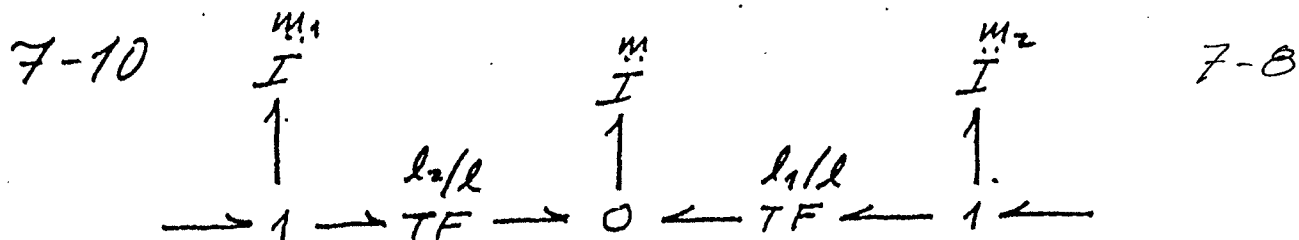
$$\text{if } \dot{q}_3 = \dot{q}_3(\dot{q}_1, \dot{q}_2), \quad \dot{q}_3 = \frac{\partial \dot{q}_3}{\partial \dot{q}_1} \dot{q}_1 + \frac{\partial \dot{q}_3}{\partial \dot{q}_2} \dot{q}_2$$

$$\begin{array}{ccccc} & & C & & \\ & \frac{\partial \dot{q}_3}{\partial \dot{q}_1} & \dot{q}_3 & \frac{\partial \dot{q}_3}{\partial \dot{q}_2} & \\ \dot{q}_1 & \xrightarrow{MTF} & 0 & \xleftarrow{MTF} & \dot{q}_2 \end{array}$$

$$e_1 = \frac{\partial \dot{q}_3}{\partial \dot{q}_1} C_3^{-1}(\dot{q}_3); \quad \frac{\partial e_1}{\partial \dot{q}_2} = \frac{\partial^2 \dot{q}_3}{\partial \dot{q}_2 \partial \dot{q}_1} C_3^{-1}(\dot{q}_3) + \frac{\partial \dot{q}_3}{\partial \dot{q}_1} \frac{\partial C_3^{-1}}{\partial \dot{q}_3} \frac{\partial \dot{q}_3}{\partial \dot{q}_2}$$

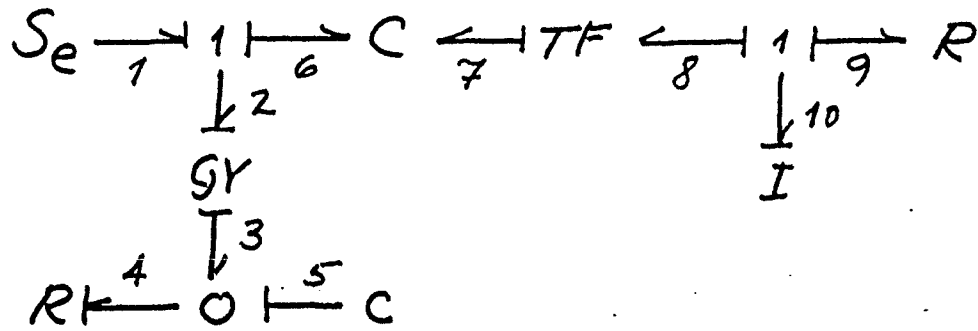
$$e_2 = \frac{\partial \dot{q}_3}{\partial \dot{q}_2} C_3^{-1}(\dot{q}_3); \quad \frac{\partial e_2}{\partial \dot{q}_1} = \frac{\partial^2 \dot{q}_3}{\partial \dot{q}_1 \partial \dot{q}_2} C_3^{-1}(\dot{q}_3) + \frac{\partial \dot{q}_3}{\partial \dot{q}_2} \frac{\partial C_3^{-1}}{\partial \dot{q}_3} \frac{\partial \dot{q}_3}{\partial \dot{q}_1}$$

$$\therefore \frac{\partial e_1}{\partial \dot{q}_2} = \frac{\partial e_2}{\partial \dot{q}_1}$$



Note: in part (3) some signs do not correspond to the figure. For part (a) you can show that $V_1 = -\frac{(l_1/l_2)V_2 + (l^2/ml_2^2)p_2}{mm}$. For part (b), $(-l_2/l_1)x_1 = x_2$. It might be nicer to show V_2 to be positive upwards in both Fig.(a) and Fig.(b) so that in the limiting cases, a simple $\rightarrow TF \rightarrow$ with through power convention results.

7-11



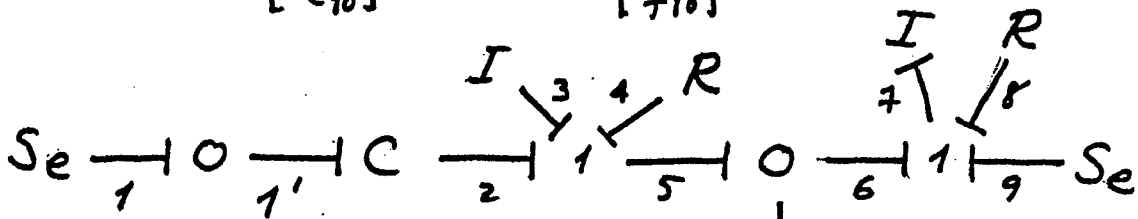
7-9

external, 1, 4, 5, 6, 7, 9, 10; internal, 2, 3, 8;
Storage, 5, 6, 7, 10; Source, 1; diss., 4, 9.

$$U = [e_1] \quad V = [f_1]$$

$$D_i = \begin{bmatrix} e_4 \\ f_9 \end{bmatrix} \quad D_o = \begin{bmatrix} f_4 \\ e_9 \end{bmatrix}$$

$$\dot{X}_i = \begin{bmatrix} f_5 \\ f_6 \\ f_7 \\ e_{10} \end{bmatrix} \quad Z_i = \begin{bmatrix} e_5 \\ e_6 \\ e_7 \\ f_{10} \end{bmatrix}$$



$$U = \begin{bmatrix} e_1 \\ e_9 \end{bmatrix}, \quad V = \begin{bmatrix} f_1 \\ f_9 \end{bmatrix}$$

$$D_i = \begin{bmatrix} f_4 \\ f_8 \\ f_{12} \\ e_{15} \\ f_{14} \end{bmatrix} \quad D_o = \begin{bmatrix} e_4 \\ e_8 \\ e_{12} \\ f_{15} \\ e_{14} \end{bmatrix}$$

$$\dot{X}_i = \begin{bmatrix} f_2 \\ e_7 \\ e_{11} \\ f_{15} \\ e_{14} \end{bmatrix}$$

$$\dot{X}_d = \begin{bmatrix} f_1' \\ e_3 \end{bmatrix} \quad Z_d = \begin{bmatrix} e_1' \\ f_3 \end{bmatrix}$$

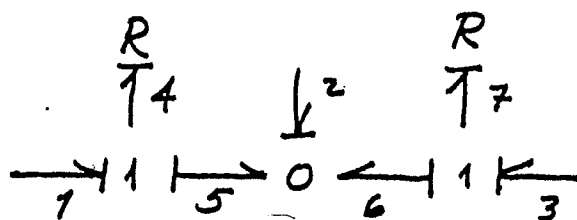
internal 5, 6, 10.

7-13

7-10

$$K = \begin{bmatrix} R_4 & 1 & 0 \\ -1 & 0 & -1 \\ 0 & 1 & R_7 \end{bmatrix}$$

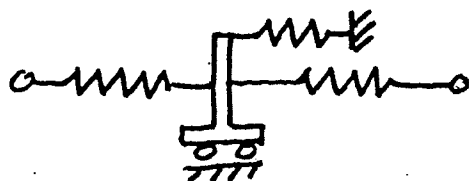
Cascinir form



$$\begin{bmatrix} f_1 \\ f_2 \\ f_3 \end{bmatrix} = \begin{bmatrix} 1/R_4 & -1/R_4 & 0 \\ -1/R_4 & 1/R_4 + 1/R_7 & -1/R_7 \\ 0 & -1/R_7 & 1/R_7 \end{bmatrix} \begin{bmatrix} e_1 \\ e_2 \\ e_3 \end{bmatrix}$$

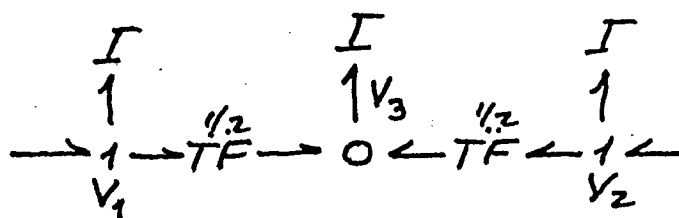
Ousager form

7-14



7-15

$$\uparrow V_1 \quad \uparrow V_3 \quad \uparrow V_2 \quad V_3 = \frac{V_1 + V_2}{2}$$

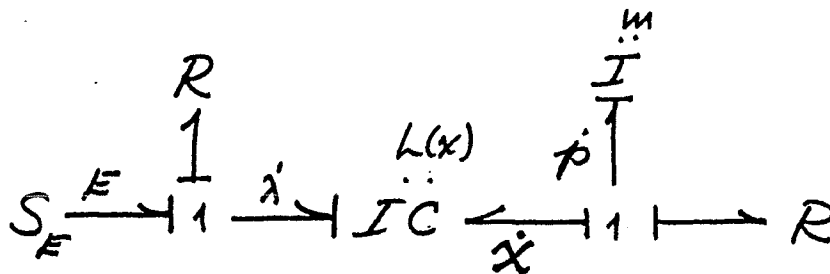


8-1 See Problem 5-11.

8-2 See Problem 5-12.

8-1

8-3

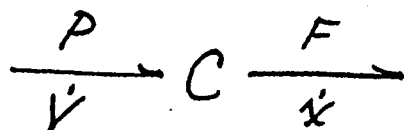


$$\dot{\lambda} = E(t) - R\lambda/L(x)$$

$$\dot{p} = -\left(-\frac{\lambda^2}{2} \frac{L'(x)}{L^2(x)}\right) - F_0 \operatorname{sgn}(p/m)$$

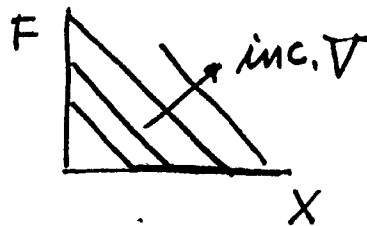
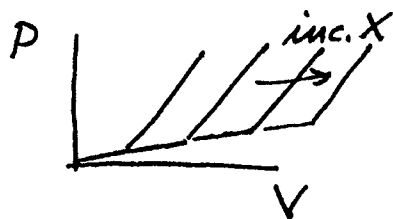
$$\dot{x} = p/m$$

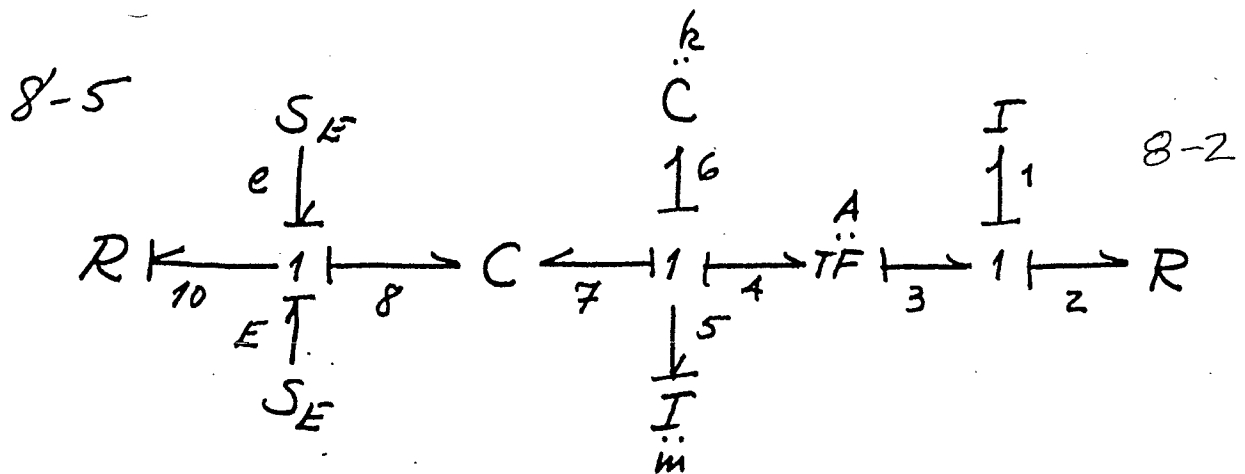
8-4



$$U = \int P dV - F dx$$

$$P = \frac{\partial U}{\partial V}, \quad -F = \frac{\partial U}{\partial x} \quad \text{so} \quad \frac{\partial P}{\partial x} = -\frac{\partial F}{\partial V}$$





$$C(q_7) = \frac{EA}{q_7}; U = \frac{q_8^2 q_7}{2EA}$$

$$\therefore e_8 = \frac{q_8 q_7}{EA}, \quad e_7 = \frac{q_8^2}{2EA}$$

$$\dot{q}_8 = \frac{1}{R_{10}} \left(E + e(t) - \frac{q_8 q_7}{EA} \right) \quad \checkmark$$

$$\left. \begin{aligned} \dot{q}_7 &= \frac{p_5}{m} \\ \dot{q}_6 &= \frac{p_5}{m} \end{aligned} \right\} q_7 = q_6 + \text{const}$$

$$\dot{p}_5 = -\frac{q_8^2}{2EA} - k q_6 - A \left(\dot{p}_1 + R_2 \left(A \frac{p_5}{m} \right) \right)$$

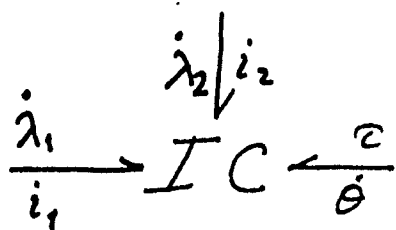
but $p_1 = I, f_1 = I, A \frac{p_5}{m} \rightarrow \dot{p}_1 = \frac{I_1 A}{m} \dot{p}_5$

so

$$\left(1 + \frac{I_1 A^2}{m} \right) \dot{p}_5 = -\frac{q_8^2}{2EA} - k q_6 - \frac{R A^2}{m} p_5$$

↑ divide through to get final form.

8-6



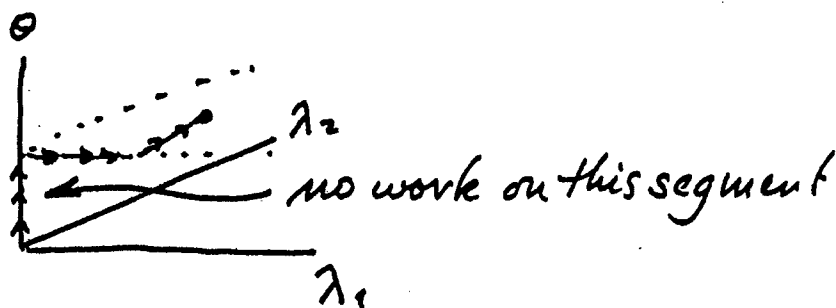
8-3

$$[L(\theta)] = \begin{bmatrix} L_1 & L_0 \cos \theta \\ L_0 \cos \theta & L_2 \end{bmatrix}$$

$$[\Gamma(\theta)] = [L(\theta)]^{-1} = \frac{\begin{bmatrix} L_2 & -L_0 \cos \theta \\ -L_0 \cos \theta & L_1 \end{bmatrix}}{L_1 L_2 - L_0^2 \cos^2 \theta}$$

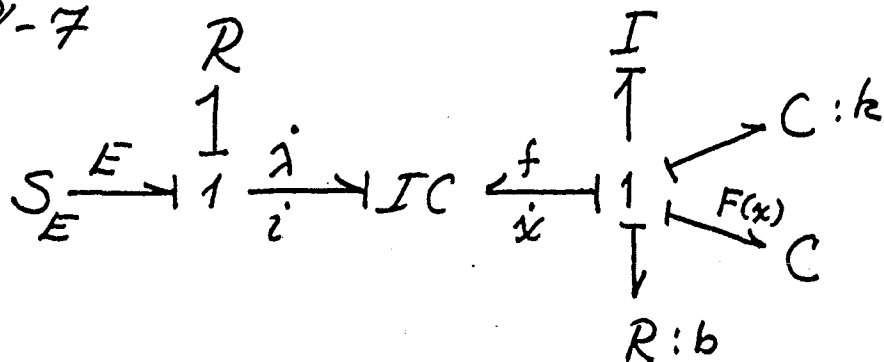
$$\text{so } \begin{bmatrix} \dot{i}_1 \\ \dot{i}_2 \end{bmatrix} = [\Gamma(\theta)] \begin{bmatrix} \lambda_1 \\ \lambda_2 \end{bmatrix}$$

then $W_m(\lambda_1, \lambda_2, \theta) = \frac{1}{2} [\lambda_1, \lambda_2] [\Gamma(\theta)] \begin{bmatrix} \lambda_1 \\ \lambda_2 \end{bmatrix}$
 where W_m is the stored energy evaluated
 along this special path:



$$\tau = \frac{\partial W_m(\lambda_1, \lambda_2, \theta)}{\partial \theta}$$

8-7



8-4

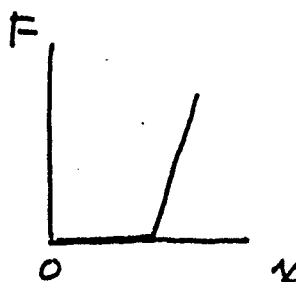
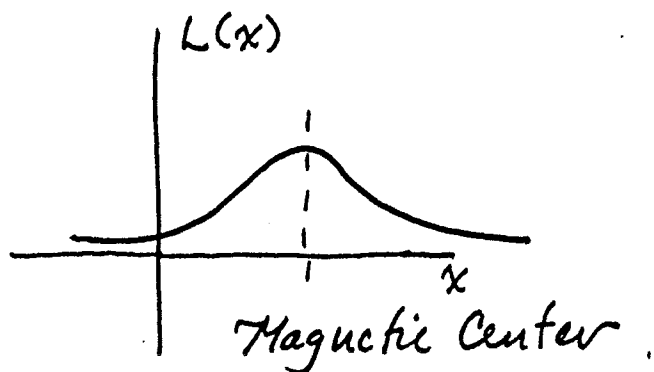
$x=0$ is no-current rest position

$$U = \frac{\lambda^2}{2} L(x); \quad \dot{\lambda} = \frac{\lambda}{L(x)}, \quad f = -\frac{\lambda^2}{2} \frac{L'(x)}{L^2(x)}$$

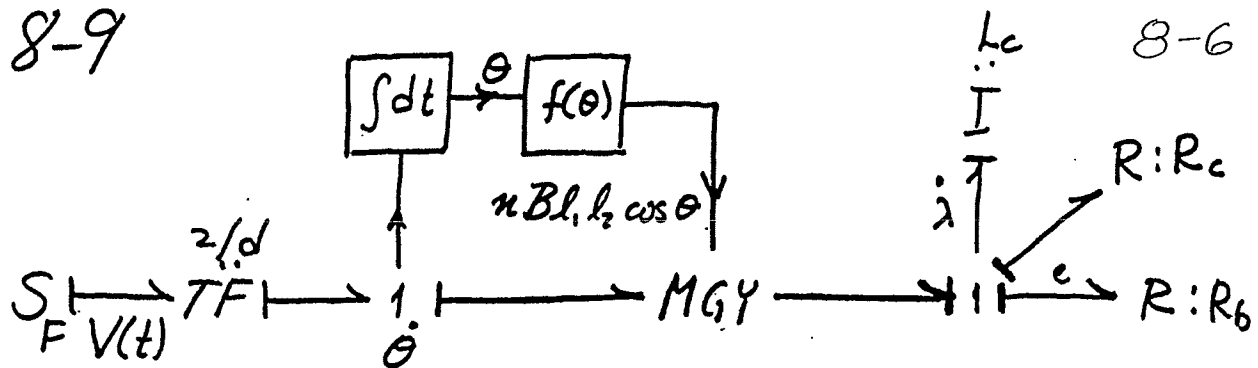
$$\dot{\lambda} = E(t) - \frac{R\lambda}{L(x)}$$

$$\dot{x} = p/m$$

$$\dot{p} = \frac{\lambda^2}{2} \frac{L'(x)}{L^2(x)} - \frac{b}{m} p - kx - F(x)$$



8-9



8-6

$$\dot{\theta} = \frac{z}{d} V(t)$$

$$\dot{\lambda} = nBl_1l_2 \cos \theta \cdot \frac{z}{d} V(t) - \left(\frac{R_c + R_b}{L_c} \right) \lambda$$

$$e = \frac{R_b}{L_c} \lambda$$

if $V = \text{const}$, $\theta = \frac{zV}{d} t$

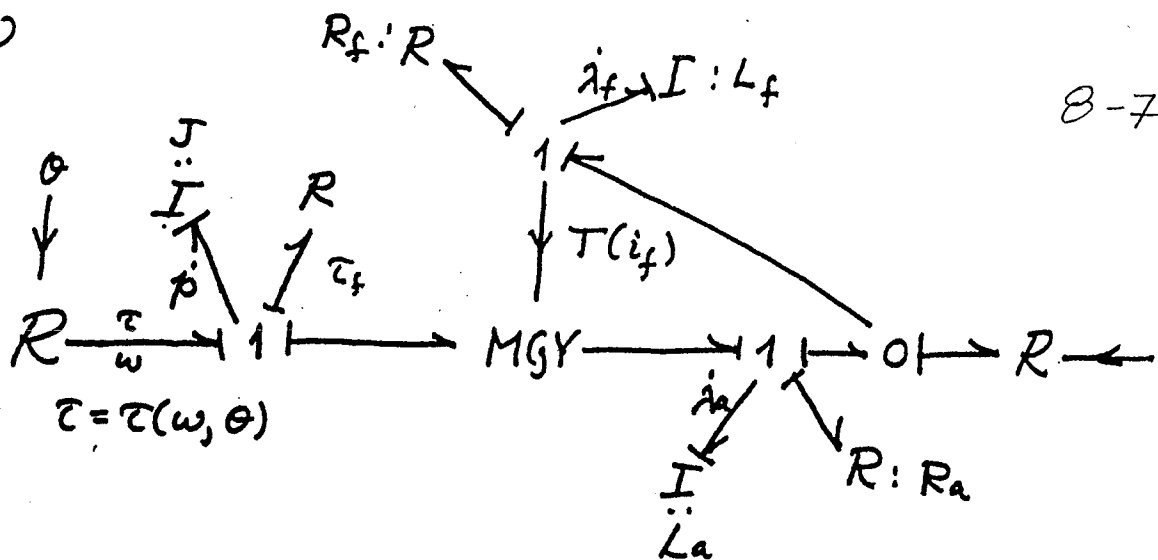
$$\dot{\lambda} + \frac{R_c + R_b}{L_c} \lambda = \frac{zV}{d} \cdot nBl_1l_2 \cos\left(\frac{zV}{d} t\right)$$

↑ amplitude
increases with V ↑

but frequency does also
so λ (and e) do
not increase so fast
with V when

$$\frac{zV}{d} > \frac{R_c + R_b}{L_c}$$

8-7

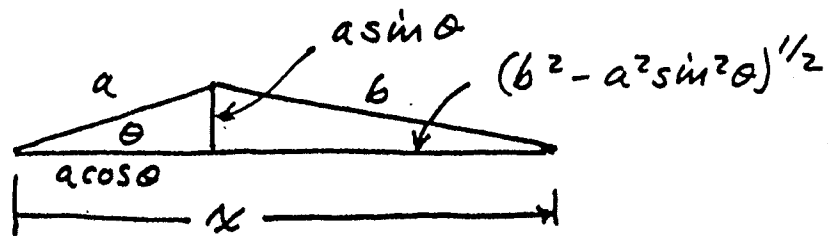


$$\dot{p} = \tau\left(\frac{p}{J}, \theta(t)\right) - \tau_f\left(\frac{p}{J}\right) - T\left(\frac{\lambda_f}{L_f}\right) \frac{\lambda_a}{L_a}$$

$$\dot{\lambda}_a = T\left(\frac{\lambda_f}{L_f}\right) \frac{p}{J} - \frac{R_a \lambda_a}{L_a} - R(t) \left(\frac{\lambda_a}{L_a} - \frac{\lambda_f}{L_f} \right)$$

$$\dot{\lambda}_f = R(t) \left(\frac{\lambda_a}{L_a} - \frac{\lambda_f}{L_f} \right) - R_f \frac{\lambda_f}{L_f}$$

8-11



8-8

$$x = a \cos \theta + (b^2 - a^2 \sin^2 \theta)^{1/2}$$

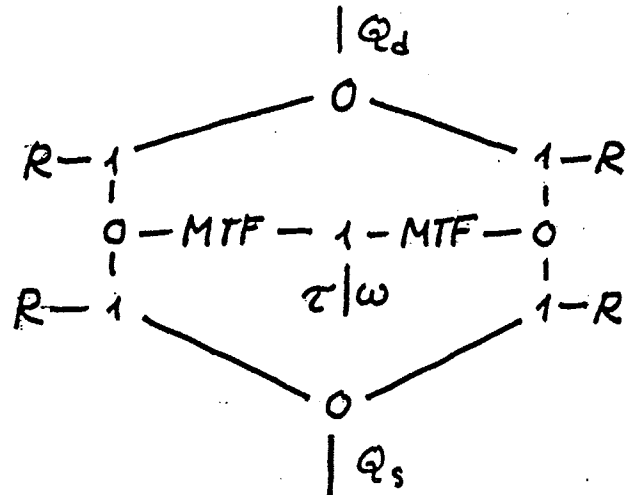
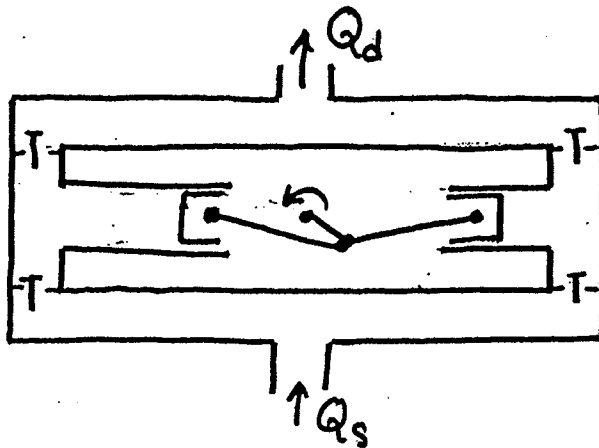
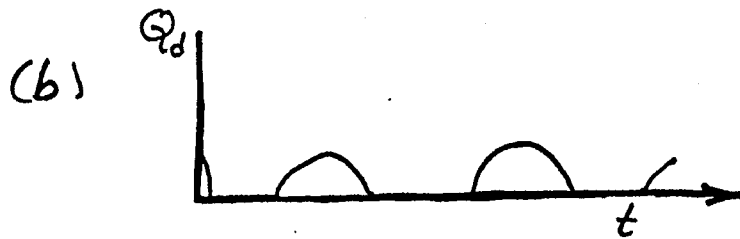
$$\dot{x} = \left[-a \sin \theta + \frac{1}{2} (b^2 - a^2 \sin^2 \theta)^{-1/2} (-a^2 \sin \theta \cdot 2 \cos \theta) \right] \dot{\theta}$$

$$Q = A \dot{x}$$

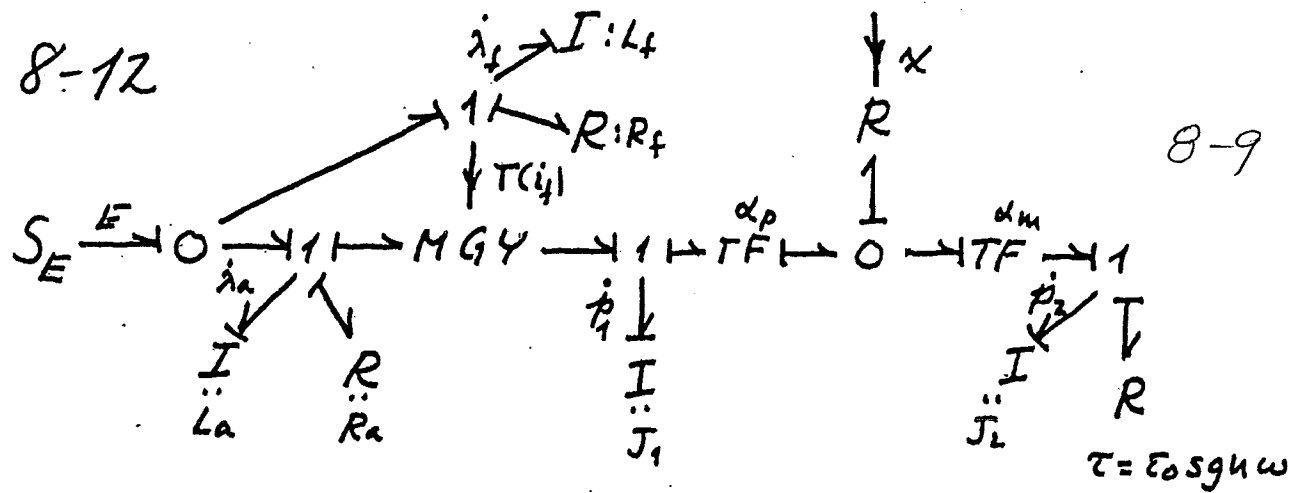
area ↑

$$\dot{\theta} \xrightarrow{f(\theta)} MTF \xrightarrow{Q}$$

$$(a) \quad f(\theta) = A \left[-a \sin \theta - (b^2 - a^2 \sin^2 \theta)^{1/2} (a^2 \sin \theta \cos \theta) \right]$$



8-12



8-9

$$\dot{\lambda}_f = E - \frac{R_f \lambda_f}{L_f}$$

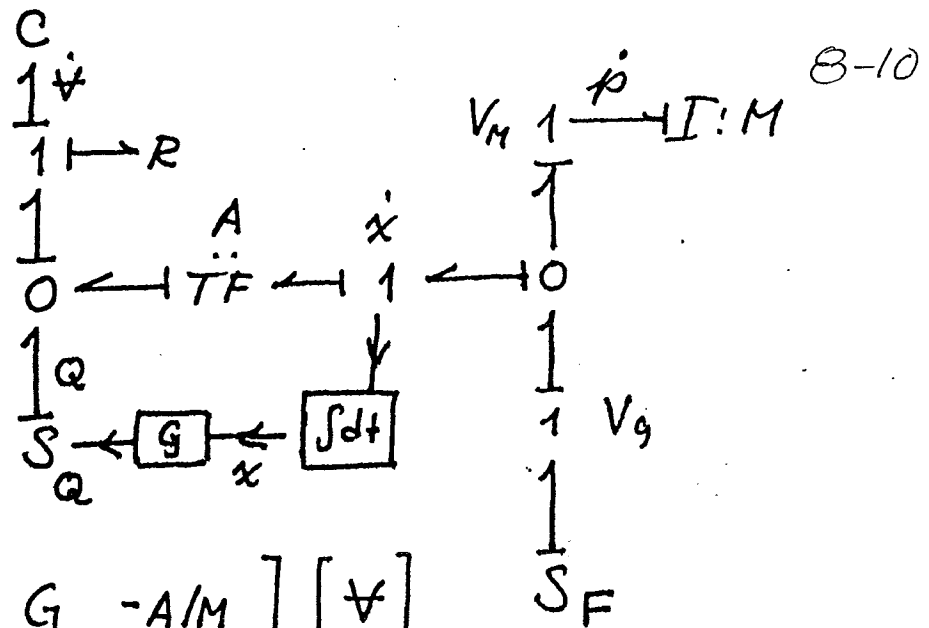
$$\dot{\lambda}_a = E - \frac{R_a \lambda_a}{L_a} - T\left(\frac{\lambda_f}{L_f}\right) \cdot \frac{p_1}{J_1}$$

$$\dot{p}_1 = T\left(\frac{\lambda_f}{L_f}\right) \cdot \frac{\lambda_a}{L_a} - \alpha_p A(x) \left(\alpha_p \frac{p_1}{J_1} - \alpha_m \frac{p_2}{J_2} \right) \left| \alpha_p \frac{p_1}{J_1} - \alpha_m \frac{p_2}{J_2} \right|$$

$$\dot{p}_2 = \alpha_m A(x) \left(\alpha_p \frac{p_1}{J_1} - \alpha_m \frac{p_2}{J_2} \right) \left| \alpha_p \frac{p_1}{J_1} - \alpha_m \frac{p_2}{J_2} \right|$$

$$- \tau_0 \text{sgn} \frac{p_2}{J_2}$$

8-13



$$\begin{bmatrix} \dot{V} \\ \dot{x} \\ \dot{p} \end{bmatrix} = \begin{bmatrix} 0 & G & -A/M \\ 0 & 0 & -1/M \\ A/C & 0 & -RA^2/M \end{bmatrix} \begin{bmatrix} V \\ x \\ p \end{bmatrix}$$

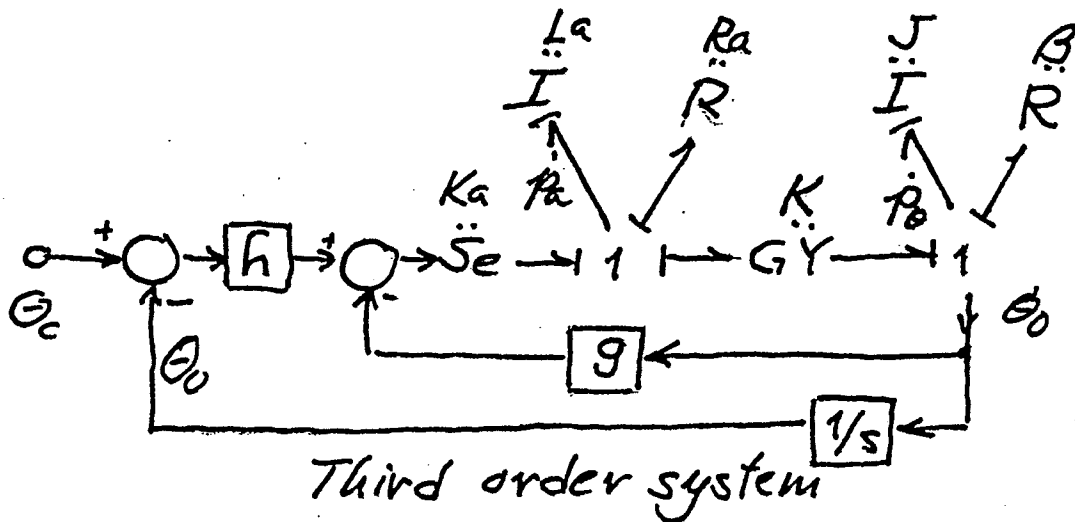
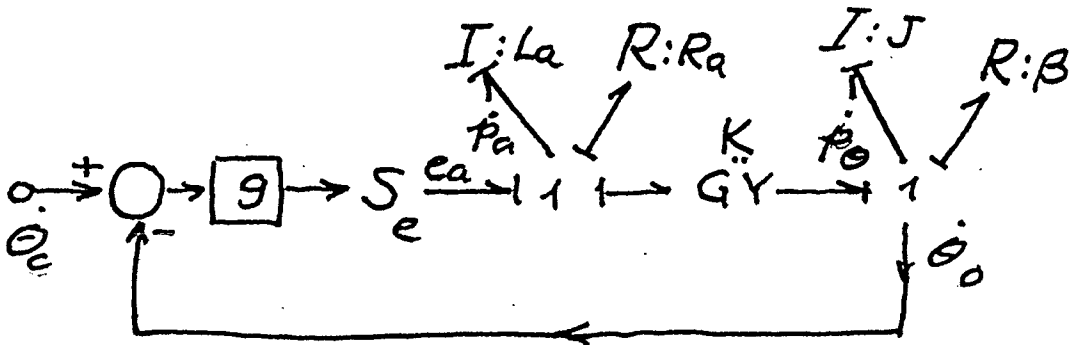
$$+ \begin{bmatrix} A \\ 1 \\ RA^2 \end{bmatrix} V_g(t)$$

$$\det \begin{bmatrix} s & -G & A/M \\ 0 & s & 1/M \\ -A/C & 0 & s + RA^2/M \end{bmatrix} = 0$$

(for eigenvalues)

8-15

8-12



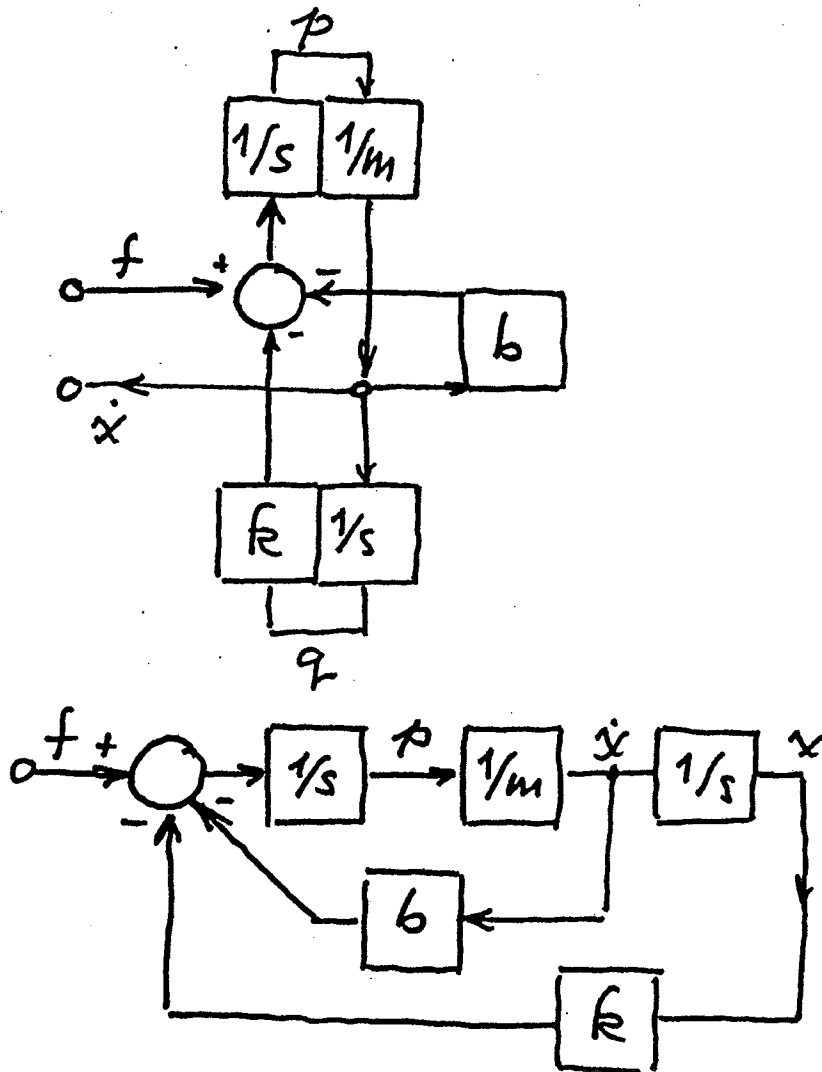
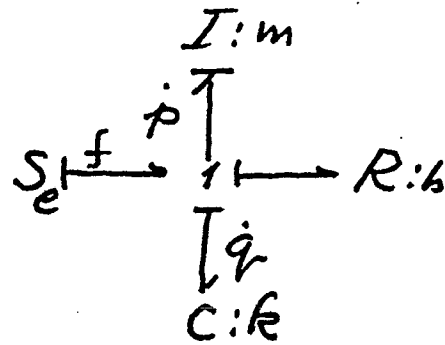
$$\dot{p}_a = K_a [-g p_o / J + h(\theta_c - \theta_o)] - R_a [p_a / L_a - K p_o / J]$$

$$\dot{p}_o = K p_a / L_a - \beta p_o / J$$

$$\dot{\theta}_o = p_o / J$$

8-16

8-13



If linear operators $1/s$ and $1/m$ are interchanged, original diagram results

8-17

8-14

$$F(q, X) = \frac{q^2}{2} \cdot \frac{d}{dX} \left(\frac{1}{C} \right) = - \frac{q^2}{2C^2} \cdot \frac{dC}{dX}$$

a) if $e_0 = \frac{q}{C(X)}$, $F = - \frac{e_0^2 C(X)}{2 C^2(X)} \cdot \frac{dC}{dX}$

$$F(e_0, X) = - \frac{e_0^2}{2} \frac{dC}{dX}$$

b) if $C(X) = \epsilon A / X$, $1/C = \frac{X}{\epsilon A}$

$$F(q, X) = + \frac{q^2}{2} \cdot \frac{d}{dX} \left(\frac{X}{\epsilon A} \right) = \frac{q^2}{2 \epsilon A}$$

$$F(e_0, X) = - \frac{e_0^2}{2} \frac{d}{dX} \left(\frac{\epsilon A}{X} \right) = + \frac{\epsilon A e_0^2}{2 X^2}$$

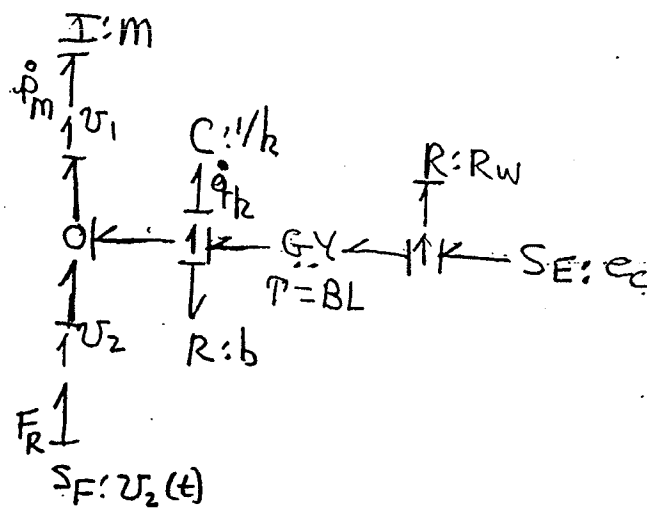
8-18

$$\tau(e_0, \theta) = - \frac{e_0^2}{2} \cdot \frac{dC}{d\theta}; \quad C = C_0 + C_1 \cos 2\theta$$

$$\begin{aligned} \tau(e_0, \theta) &= - \frac{e_0^2}{2} (-C_1 (\sin 2\theta) \cdot 2) \\ &= e_0^2 C_1 \sin 2\theta \end{aligned}$$

$$\begin{aligned} \tau_{\max} &= (1000)^2 \cdot 10 \times 10^{-12} \cdot 1 \\ &= 10^{-5} \text{ Nm} \end{aligned}$$

8-15



$$\dot{q}_h = \frac{F_m}{M} - v_2(t)$$

$$F_R = -k q_k - b \left(\frac{f_m}{m} - v_2(t) \right) + \frac{T}{R_w} \left[e_c - T \left(\frac{f_m}{m} - v_2(t) \right) \right]$$

8.20 In the s -domain

$$\begin{bmatrix} \left[\frac{s+b+\frac{q^2}{m}}{R_{Wm}} \right] & \left[\frac{1}{R_{Wm}} \right] \\ -\frac{1}{m} & 0 \end{bmatrix} \begin{bmatrix} p_m \\ q_h \end{bmatrix} = \begin{bmatrix} \left[\frac{b+\frac{q^2}{R_W}}{R_W} \right] v_2 \\ -1 \end{bmatrix} + \begin{bmatrix} \left[\frac{1}{R_W} \right] e_c \\ 0 \end{bmatrix}$$

$$F_R = -k q_k - \left[\frac{b}{m} + \frac{T^3}{R_w m} \right] p_m + \left[b + \frac{T^2}{R_w} \right] v_z + \frac{T}{R_w} e_c$$

Derive

$$\frac{F_m}{v_2} = \frac{(b + \frac{P^2}{R_w})s + k}{D}$$

$$\frac{q_k}{U_2} = \frac{-(s + \frac{b}{m} + \frac{q^2}{R_{wm}}) + \frac{b}{m} + \frac{q^2}{R_{wm}}}{0}$$

$$\frac{P_m}{P_c} = \frac{\frac{T}{R_W} S}{D}$$

$$\frac{q_k}{e_c} = \frac{\frac{P}{R_{wm}}}{D}$$

$$D = S^2 + \left(\frac{b}{m} + \frac{q^2}{R_0 m}\right) S + \frac{k}{m}$$

8.20 (continued)

8-16

Get output eqn:

$$\begin{aligned}
 \frac{F_R(s)}{U_2} &= -k \frac{q_k(s)}{U_2} - \left[\frac{b}{m} + \frac{P^2}{R_W m} \right] \frac{F_m}{U_2} + \left[b + \frac{P^2}{R_W} \right] \\
 &= + \frac{k}{D} - \frac{\left[\frac{b}{m} + \frac{P^2}{R_W m} \right] \left[\left(b + \frac{P^2}{R_W} \right) s + k \right]}{D} + \left(b + \frac{P^2}{R_W} \right) \\
 &= \frac{\left[k - \left(\frac{b}{m} + \frac{P^2}{R_W m} \right) \left(b + \frac{P^2}{R_W} \right) \right] s - \left(\frac{b}{m} + \frac{P^2}{R_W m} \right) k + \left(b + \frac{P^2}{R_W} \right) \left(s^2 + \left(\frac{b}{m} + \frac{P^2}{R_W m} \right) s + \frac{k}{m} \right)}{D} \\
 &= \frac{\left(b + \frac{P^2}{R_W} \right) s^2 + k s}{D}
 \end{aligned}$$

and

$$\boxed{\frac{F_R(s)}{U_2} = \frac{s \left[\left(b + \frac{P^2}{R_W} \right) s + k \right]}{s^2 + \left(\frac{b}{m} + \frac{P^2}{R_W m} \right) s + \frac{k}{m}} = G_{FV}(s)}$$

$$\begin{aligned}
 \frac{\bar{F}_R(s)}{E_c} &= -k \frac{q_k(s)}{E_c} - \left[\frac{b}{m} + \frac{P^2}{R_W m} \right] \frac{F_m}{E_c} + \frac{P}{R_W} \\
 &= - \frac{k \frac{P}{R_W m}}{D} - \frac{\left[\frac{b}{m} + \frac{P^2}{R_W m} \right] \frac{P}{R_W} s}{D} + \frac{P}{R_W} \\
 &= \frac{- \frac{k P}{R_W m} - \left[\frac{b}{m} + \frac{P^2}{R_W m} \right] \frac{P}{R_W} s + \frac{P}{R_W} \left(s^2 + \left(\frac{b}{m} + \frac{P^2}{R_W m} \right) s + \frac{k}{m} \right)}{D}
 \end{aligned}$$

$$\boxed{\frac{\bar{F}_R(s)}{E_c} = \frac{\frac{P}{R_W} s^2}{s^2 + \left(\frac{b}{m} + \frac{P^2}{R_W m} \right) s + \frac{k}{m}} = G_{F_c}(s)}$$

8.21

8-17

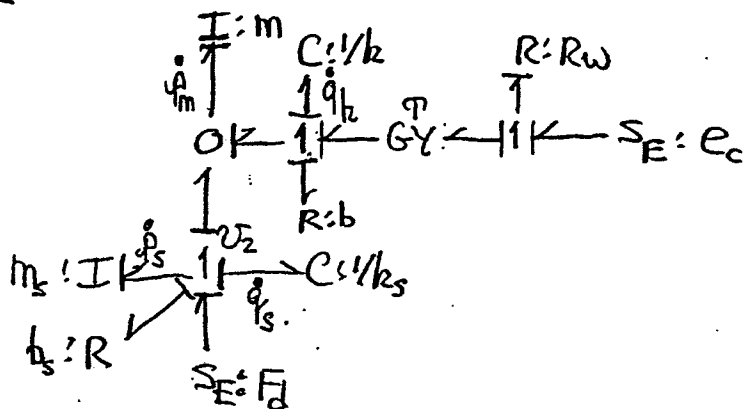
$$F_R = G_{FV} v_2 + G_{FC} e_c = b_c v_2$$

$$e_c = \frac{(b_c - G_{FV}) v_2}{G_{FC}}$$

$$= \frac{(b_c D - N F V)}{N F C} v_2 = \frac{\left[b_c \left(s^2 + \left(\frac{b}{m} + \frac{\tau^2}{R_w m} \right) s + \frac{k}{m} \right) - s \left[\left(b + \frac{\tau^2}{R_w} \right) s + k \right] \right]}{\frac{\tau}{R_w} s^2} v_2$$

$$e_c = \left\{ \frac{\left[b_c - \left(b + \frac{\tau^2}{R_w} \right) \right] s^2 + \left[b_c \left(\frac{b}{m} + \frac{\tau^2}{R_w m} \right) - k \right] s + b_c \frac{k}{m}}{\frac{\tau}{R_w} s^2} \right\} v_2$$

8.22



$$\ddot{q}_m = -k q_k - b \left(\frac{\dot{q}_m}{m} - \frac{\dot{q}_s}{m_s} \right) + \frac{\tau}{R_w} (e_c - \tau \left(\frac{\dot{q}_m}{m} - \frac{\dot{q}_s}{m_s} \right))$$

$$\ddot{q}_s = -\frac{b_s}{m_s} \dot{q}_s - k_s q_s + F_d - \ddot{q}_m \leftarrow \text{little trick to save some time.}$$

$$\dot{q}_k = \frac{\dot{q}_m}{m} - \frac{\dot{q}_s}{m_s}$$

$$\dot{q}_s = \frac{\dot{q}_s}{m_s}$$

8.22 (continued)

8-18

$$\begin{bmatrix} s + \frac{b}{m} + \frac{\tau^2}{mR_w} & -(\frac{b}{m_s} + \frac{\tau^2}{R_w m_s}) \\ s & s + \frac{b_s}{m_s} \\ -1/m & 1/m_s \\ 0 & -1/m_s \end{bmatrix} \begin{bmatrix} k \\ 0 \\ s \\ 0 \end{bmatrix} = \begin{bmatrix} 0 \\ k_s \\ 0 \\ s \end{bmatrix} \begin{bmatrix} p_m \\ p_s \\ q_k \\ q_s \end{bmatrix} = \begin{bmatrix} 0 \\ 1 \\ 0 \\ 0 \end{bmatrix} F_d + \begin{bmatrix} \tau/R_w \\ 0 \\ 0 \\ 0 \end{bmatrix} e_c$$

$$D = \left(s + \frac{b}{m} + \frac{\tau^2}{mR_w}\right)s \left(s^2 + \frac{b_s}{m_s}s + \frac{k_s}{m_s}\right) + \left(\frac{b}{m_s} + \frac{\tau^2}{R_w m_s}\right)s^3 + k \left\{ \frac{s^2}{m_s} + \frac{1}{m} \left(s^2 + \frac{b_s}{m_s}s + \frac{k_s}{m_s}\right) \right\}$$

$$D = s^4 + \left[\frac{b_s}{m_s} + \frac{b}{m} + \frac{\tau^2}{R_w m} + \frac{b}{m_s} + \frac{\tau^2}{R_w m_s} \right] s^3 + \left[\frac{k_s}{m_s} + \frac{b_s}{m_s} \left(\frac{b}{m} + \frac{\tau^2}{R_w m} \right) + \frac{k}{m_s} + \frac{k}{m} \right] s^2 + \left[\frac{k_s}{m_s} \left(\frac{b}{m} + \frac{\tau^2}{R_w m} \right) + \frac{k}{m} \frac{b_s}{m_s} \right] s + \frac{k}{m} \frac{k_s}{m_s}$$

solve for p_s using Cramer's rule.

$$\frac{p_s}{F_d}(s) = \frac{s \left(s^2 + \left(\frac{b}{m} + \frac{\tau^2}{mR_w} \right) s + \frac{k}{m} \right)}{D}$$

$$\frac{v_z}{F_d} = \frac{\frac{1}{m_s} s \left[s^2 + \left(\frac{b}{m} + \frac{\tau^2}{R_w m} \right) s + \frac{k}{m} \right]}{D} = G_{vd} \text{ open loop TF.}$$

8.22 (continued)

Also need v_2/e_c ,

8-19

$$\frac{P_3(s)}{e_c} = \frac{-\tau/R_w s^3}{D}$$

$$\frac{v_2(s)}{e_c} = \frac{-\frac{\tau}{R_w m s} s^3}{D} = G_{vc}$$

Then

$$v_2 = G_{vd} F_d + G_{vc} e_c \quad \text{but } e_c = G_c v_2 \leftarrow \text{Prob 8.21}$$

$$\left. \frac{v_2}{F_d} \right|_{c.l.} = \frac{G_{vd}}{1 - G_{vc} G_c} = \frac{N_{vd}}{D - N_{vc} G_c}$$

$$= \frac{1}{m_s} s \left[s^2 + \left(\frac{b}{m} + \frac{\tau^2}{R_w m} \right) s + \frac{k}{m} \right]$$

$$D + \frac{\tau}{R_w m s} s^3 \left\{ \frac{[b_c - (b + \frac{\tau^2}{R_w})] s^2 + [b_c (\frac{b}{m} + \frac{\tau^2}{R_w m}) - k] s + b_c \frac{k}{m}}{\frac{\tau}{R_w} s^2} \right\}$$

$$\left. \frac{v_2}{F_d} \right|_{c.l.} = \frac{1}{m_s} s \left[s^2 + \left(\frac{b}{m} + \frac{\tau^2}{R_w m} \right) s + \frac{k}{m} \right]$$

closed
loop
TF.

$$s^4 + \left[\frac{b_c}{m_s} + \frac{b_s}{m_s} + \frac{b}{m} + \frac{\tau^2}{R_w m} \right] s^3 + \left[\left(\frac{b_c}{m_s} + \frac{b_s}{m_s} \right) \left(\frac{b}{m} + \frac{\tau^2}{R_w m} \right) + \frac{k_s}{m_s} + \frac{k}{m} \right] s^2$$

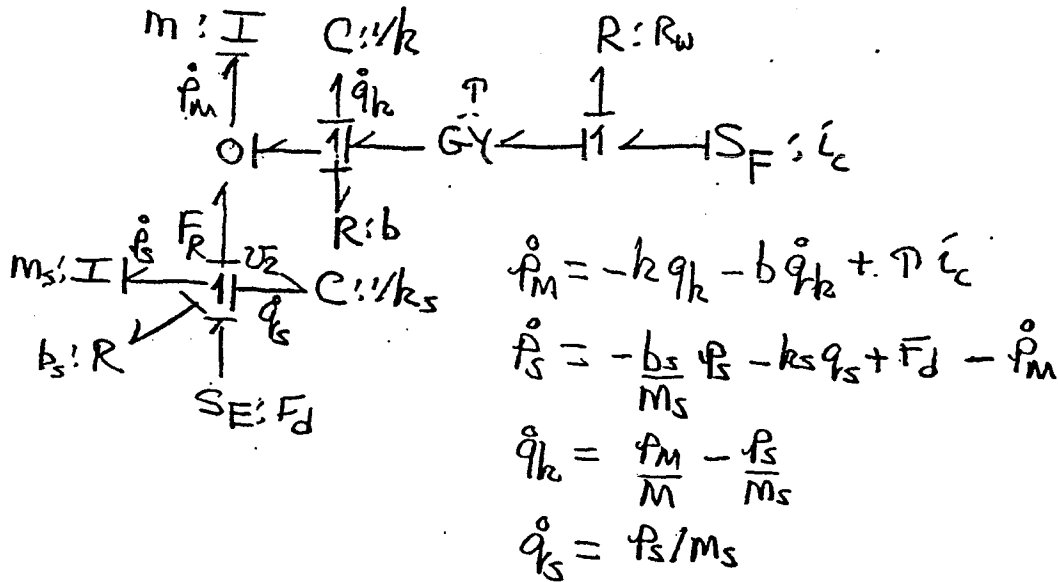
$$+ \left[\left(\frac{b_c}{m_s} + \frac{b_s}{m_s} \right) \frac{k}{m} + \frac{k_s}{m_s} \left(\frac{b}{m} + \frac{\tau^2}{m R_w} \right) \right] s + \frac{k}{m} \frac{k_s}{m_s}$$

From the Figure for Prob 8.22, we see that b_s is a damper to ground. From the c.l. transfer function we see that b_c ended up functioning identically to b_s . Thus the control did create a damper to ground.

8.23

8-20

complete system:



system eqns

Actuator eqns:

$$\left. \begin{aligned} \dot{p}_m &= -k q_k - b \dot{q}_k + \tau \dot{i}_c \\ \dot{q}_k &= \frac{p_m}{m} - v_z \end{aligned} \right\} \begin{bmatrix} s & bs+k \\ -1/m & s \end{bmatrix} \begin{bmatrix} p_m \\ q_k \end{bmatrix} = \begin{bmatrix} 0 \\ -1 \end{bmatrix} v_z + \begin{bmatrix} \tau \\ 0 \end{bmatrix} \dot{i}_c$$

$$F_r = \dot{p}_m$$

$$\text{derive } \frac{p_m}{v_z} = \frac{bs+k}{s^2 + \frac{b}{m}s + \frac{k}{m}}$$

$$\therefore \frac{F_r(s)}{v_z} = \frac{s(bs+k)}{s^2 + \frac{b}{m}s + \frac{k}{m}}$$

$$\text{derive } \frac{p_m}{i_c} = \frac{\tau s}{s^2 + \frac{b}{m}s + \frac{k}{m}}$$

$$\therefore \frac{F_r(s)}{i_c} = \frac{\tau s^2}{s^2 + \frac{b}{m}s + \frac{k}{m}}$$

8-23 (continued)

derive control filter that yields $F_r = b_c v_z$ 8-21

$$\text{i.e. } F_r = \left(\frac{F_r}{v_z}\right) v_z + \left(\frac{F_r}{\dot{v}_c}\right) \dot{v}_c = b_c v_z$$

$$\therefore \dot{v}_c = \frac{\left[b_c - \left(\frac{F_r}{v_z}\right)\right]}{\left(\frac{F_r}{\dot{v}_c}\right)} v_z = \frac{b_c \left[s^2 + \frac{b}{m}s + \frac{k}{m}\right] - s(bs+k)}{\tau s^2} v_z$$

$$\dot{v}_c = \frac{[(b_c - b)]s^2 + \left[b_c \frac{b}{m} - k\right]s + b_c \frac{k}{m}}{\tau s^2} v_z$$

$G_c(s)$ needed below

Return to system eqns:

$$\begin{bmatrix} s & 0 & bs+k & 0 \\ s & s+\frac{bs}{m_s} & 0 & k_s \\ -1/m & 1/m_s & s & 0 \\ 0 & -1/m_s & 0 & s \end{bmatrix} \begin{bmatrix} p_m \\ p_s \\ q/k \\ q_s \end{bmatrix} = \begin{bmatrix} 0 \\ 1 \\ 0 \\ 0 \end{bmatrix} F_d + \begin{bmatrix} \tau \\ 0 \\ 0 \\ 0 \end{bmatrix} \dot{v}_c$$

$$D = s^2 \left(s^2 + \frac{bs}{m_s} s + \frac{k_s}{m_s}\right) + (bs+k) \left\{ \frac{s^3}{m_s} + \frac{1}{m} \left(s^2 + \frac{bs}{m_s} s + \frac{k_s}{m_s}\right) \right\}$$

$$D = s^4 + \left[\frac{bs}{m_s} + \frac{b}{m_s} + \frac{b}{m}\right] s^3 + \left[\frac{k_s}{m_s} + \frac{b}{m} \frac{bs}{m_s} + \frac{k}{m_s} + \frac{k}{m}\right] s^2 + \left[\frac{b}{m} \frac{k_s}{m_s} + \frac{bs}{m_s} \frac{k}{m} + \frac{k}{m} \frac{k_s}{m_s}\right] s$$

8.23 (continued)

8-22

solve for $\frac{P_s(s)}{F_d}$ using Cramers rule:

$$\frac{P_s(s)}{F_d} = \frac{s(s^2 + \frac{b}{M}s + \frac{k}{M})}{D} \quad \therefore \quad \frac{v_z}{F_d} = \frac{\frac{1}{M_s} s(s^2 + \frac{b}{M}s + \frac{k}{M})}{D}$$

solve for $\frac{P_s}{\dot{L}_c}(s)$:

$$\frac{P_s}{\dot{L}_c} = \frac{-\frac{1}{M_s} s^3}{D} \quad \therefore \quad \frac{v_z(s)}{\dot{L}_c} = \frac{-\frac{1}{M_s} s^3}{D}$$

Derive closed loop response:

$$v_z = \left(\frac{v_z}{F_d}\right) F_d + \left(\frac{v_z}{\dot{L}_c}\right) \dot{L}_c \quad \dot{L}_c = G_c v_z$$

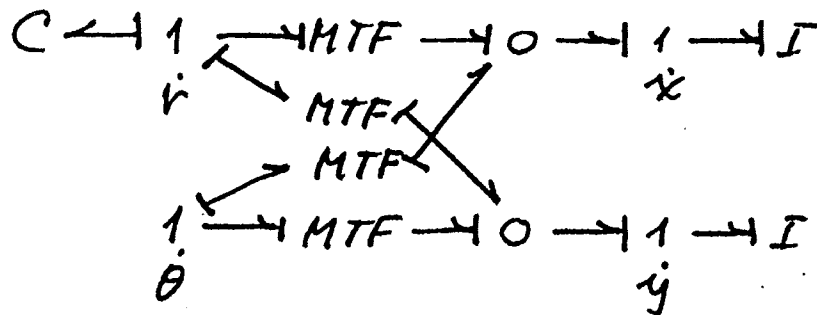
$$\frac{v_z}{F_d} \bigg|_{cl} = \frac{\left(\frac{v_z}{F_d}\right)}{1 - \left(\frac{v_z}{\dot{L}_c}\right) G_c} = \frac{\frac{1}{M_s} s(s^2 + \frac{b}{M}s + \frac{k}{M})}{D + \frac{1}{M_s} s^3 \left[(b_c - b)s^2 + (b_c \frac{b}{M} - k)s + b_c \frac{k}{M} \right]}$$

$\cancel{1/s^2}$

$$\frac{v_z}{F_d} \bigg|_{cl} = \frac{\frac{1}{M_s} s(s^2 + \frac{b}{M}s + \frac{k}{M})}{\dots}$$

$$s^4 + \left[\frac{b_c}{M_s} + \frac{b_s}{M_s} + \frac{b}{M} \right] s^3 + \left[\frac{b_c}{M_s} \frac{b}{M} + \frac{k_s}{M_s} + \frac{b}{M} \frac{b_s}{M_s} + \frac{k}{M} \right] s^2 + \left[\frac{b_c}{M_s} \frac{k}{M} + \frac{b_s}{M_s} \frac{k}{M} + \frac{b}{M} \frac{k_s}{M_s} \right] s + \frac{k}{M} \frac{k_s}{M_s}$$

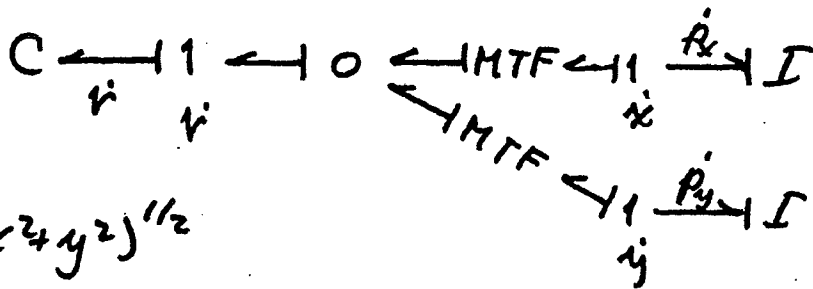
9-1



9-1

$$x = r \cos \theta \rightarrow \dot{x} = (\cos \theta) \dot{r} - (r \sin \theta) \dot{\theta}$$

$$y = r \sin \theta \rightarrow \dot{y} = (\sin \theta) \dot{r} + (r \cos \theta) \dot{\theta}$$



$$r = (x^2 + y^2)^{1/2}$$

$$\dot{r} = \frac{1}{2} (x^2 + y^2)^{-1/2} (2x\dot{x} + 2y\dot{y})$$

$$= \left[(x^2 + y^2)^{-1/2} x \right] \dot{x} + \left[(x^2 + y^2)^{-1/2} y \right] \dot{y}$$

$$\dot{x} = p_x / m, \quad \dot{y} = p_y / m, \quad \dot{r} = \left[(x^2 + y^2)^{-1/2} x \right] p_x / m + \left[(x^2 + y^2)^{-1/2} y \right] p_y / m$$

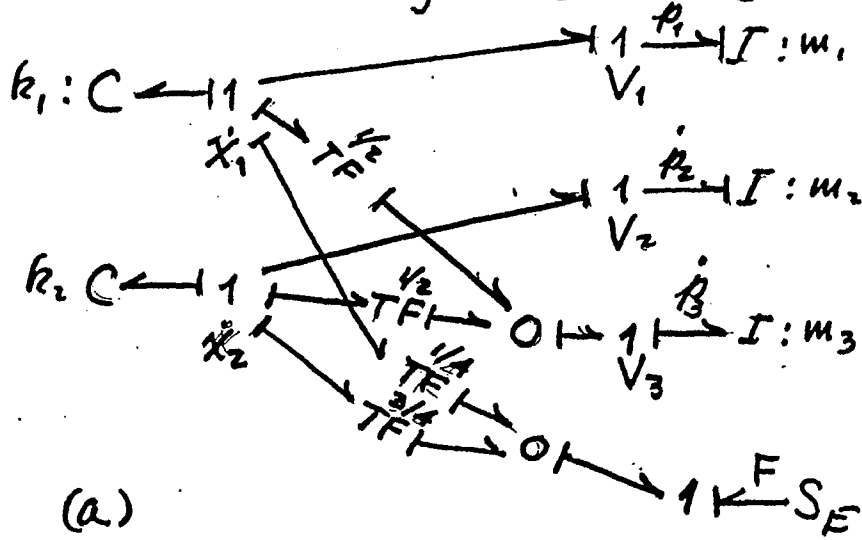
$$\dot{p}_x = - \left[(x^2 + y^2)^{-1/2} x \right] k(r - l_0)$$

$$\dot{p}_y = - \left[(x^2 + y^2)^{-1/2} y \right] k(r - l_0)$$

five state variables - this could be reduced to four by using $r = (x^2 + y^2)^{1/2}$.

9-2 Interchange m_2 and m_3 in Figure.

9-2



$$\dot{x}_1 = p_1/m_1$$

diff. eqns

$$\dot{x}_2 = p_2/m_2$$

$$p_3 = m_3 \left(\frac{p_1}{m_1} + \frac{p_2}{m_2} \right)$$

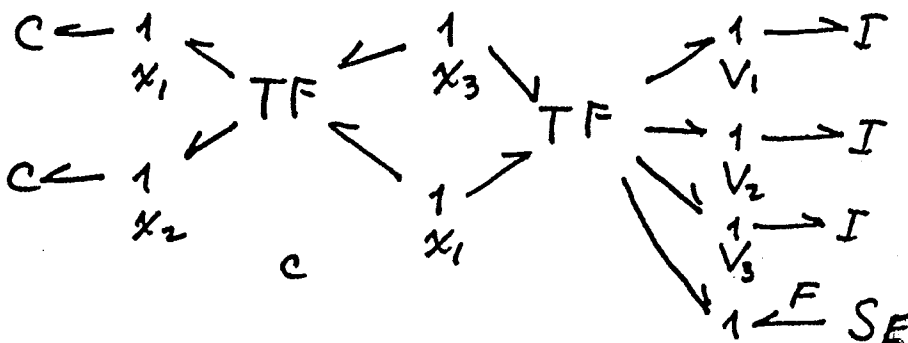
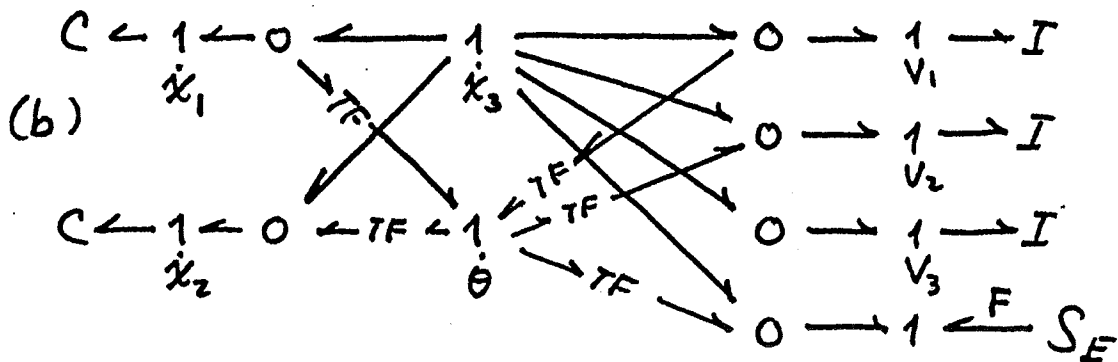
$$\dot{p}_1 = \frac{F(t)}{4} - \frac{p_3}{2} - k_1 x_1$$

so

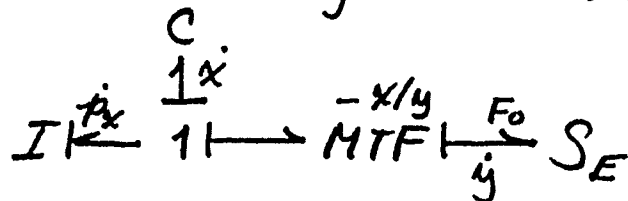
$$\dot{p}_3 = \frac{m_3}{m_1} \dot{p}_1 + \frac{m_3}{m_2} \dot{p}_2$$

$$\dot{p}_2 = \frac{3F(t)}{4} - \frac{p_3}{2} - k_2 x_2$$

Substitute in last two equations.



9-3 1. $x^2 + y^2 = l^2 \rightarrow x\dot{x} + y\dot{y} = 0$



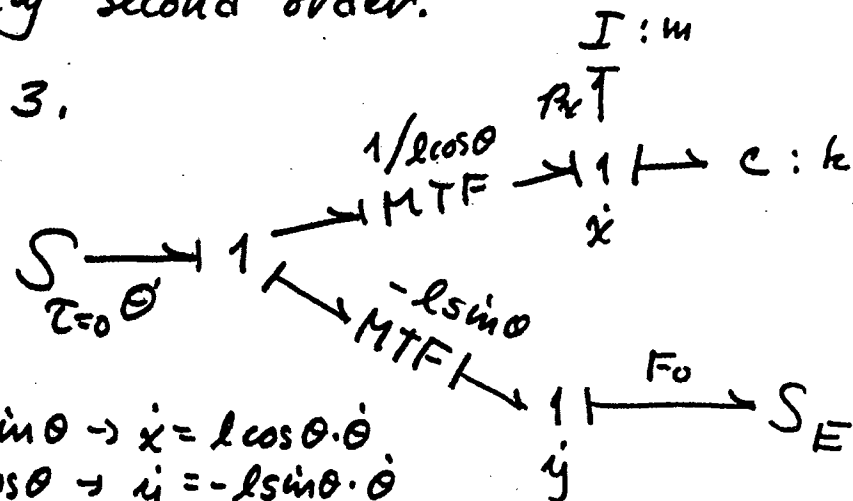
$$\dot{x} = p_x / m$$

$$\dot{p}_x = -kx - \left(-\frac{x}{y}\right) F_0$$

$$\dot{y} = -\frac{x}{y} \cdot \frac{p_x}{m} \text{ for use } x^2 + y^2 = l^2 \text{ to eliminate } y \text{ in equation above}$$

2. Essentially same as above except y has been eliminated.

Only second order.



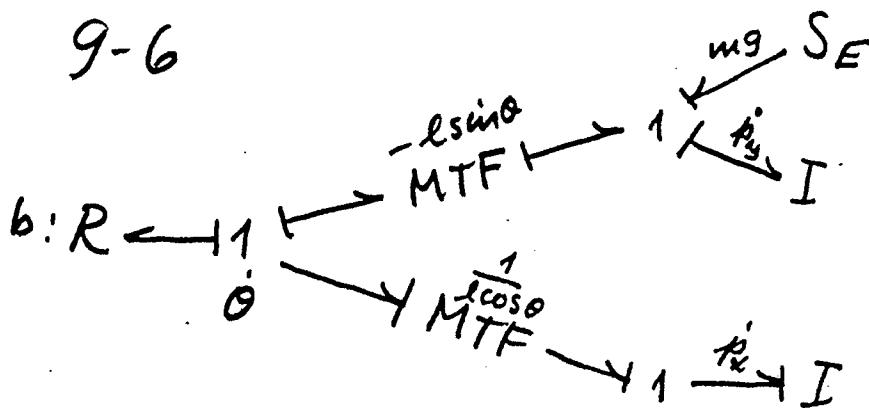
$$x = l \sin \theta \rightarrow \dot{x} = l \cos \theta \cdot \dot{\theta}$$

$$y = l \cos \theta \rightarrow \dot{y} = -l \sin \theta \cdot \dot{\theta}$$

$$\dot{x} = p_x / m ; \dot{p}_x = \frac{l \sin \theta F_0}{l \cos \theta} - kx$$

$$\dot{\theta} = \frac{1}{l \cos \theta} \cdot \frac{p_x}{m} \text{ or use } \theta = \sin^{-1} \frac{x}{l}$$

9-6



9-5

$$\boxed{\dot{\theta} = \frac{1}{l \cos \theta} \frac{p_x}{m}} \quad (1)$$

$$\dot{p}_x = \frac{1}{l \cos \theta} \left[-b \left(\frac{p_x}{m l \cos \theta} - (-l \sin \theta) [-mg + \dot{p}_y] \right) \right]$$

$$\text{but } p_y = m (-l \sin \theta \cdot \frac{1}{l \cos \theta} \cdot \frac{p_x}{m})$$

$$\dot{p}_y = \frac{d}{dt} \left(\frac{\sin \theta}{\cos \theta} \cdot p_x \right) = \frac{-1}{\cos^2 \theta} \dot{\theta} p_x - \frac{\sin \theta}{\cos \theta} \dot{p}_x$$

Substituting,

$$\dot{p}_x = \frac{-b p_x}{m l^2 \cos^2 \theta} - \frac{\sin \theta}{\cos \theta} mg + \frac{\sin \theta}{\cos \theta} \left(-\frac{\dot{\theta} p_x}{\cos^2 \theta} - \frac{\sin \theta}{\cos \theta} \dot{p}_x \right)$$

or

$$\left(1 + \frac{\sin^2 \theta}{\cos^2 \theta} \right) \dot{p}_x = \frac{-b p_x}{m l^2 \cos^2 \theta} - \frac{\sin \theta}{\cos \theta} mg - \frac{\sin \theta}{\cos^3 \theta} p_x \cdot \frac{p_x}{m l \cos \theta}$$

$$\boxed{\dot{p}_x = -\frac{b}{m l^2} p_x - \sin \theta \cos \theta mg - \frac{\sin \theta}{m l \cos^2 \theta} p_x^2} \quad (2)$$

$$\text{from (1), } p_x = m l \cos \theta \dot{\theta}, \quad p_x^2 = m^2 l^2 \cos^2 \theta \dot{\theta}^2$$

$$\dot{p}_x = \ddot{\theta} m l \cos \theta - m l \sin \theta \dot{\theta}^2$$

Substitute these into (2) to get standard eqn.

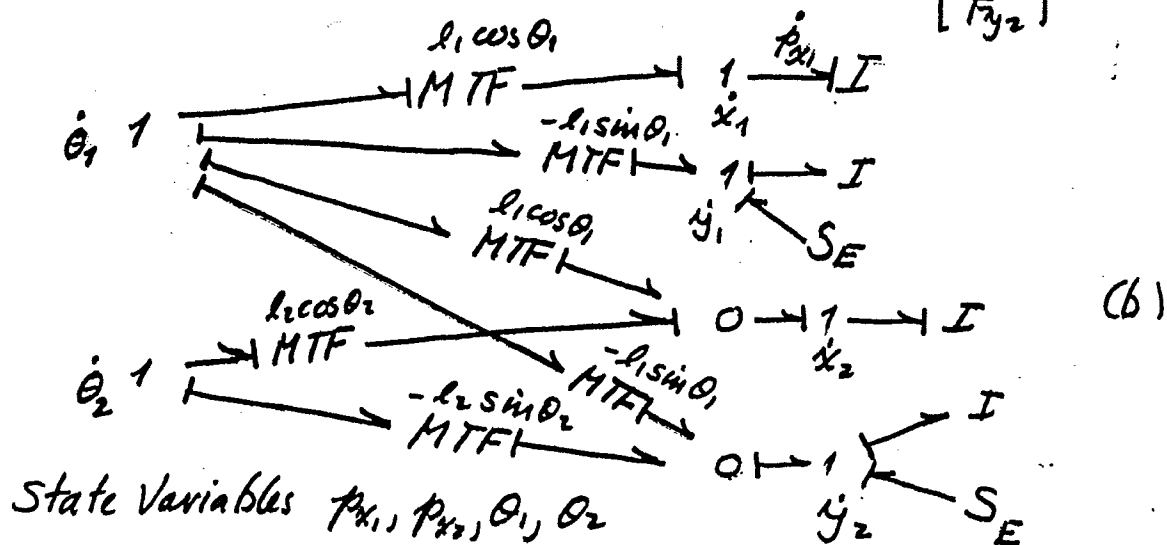
9-7 (a)

$$\begin{aligned}x_1 &= l_1 \sin \theta_1 \\y_1 &= l_1 \cos \theta_1 \\x_2 &= l_1 \sin \theta_1 + l_2 \sin \theta_2 \\y_2 &= l_1 \cos \theta_1 + l_2 \cos \theta_2\end{aligned}$$

9-6

$$\begin{bmatrix} \dot{x}_1 \\ \dot{y}_1 \\ \dot{x}_2 \\ \dot{y}_2 \end{bmatrix} = \begin{bmatrix} l_1 \cos \theta_1 & 0 \\ -l_1 \sin \theta_1 & 0 \\ l_1 \cos \theta_1 & l_2 \cos \theta_2 \\ -l_1 \sin \theta_1 & -l_2 \sin \theta_2 \end{bmatrix} \begin{bmatrix} \dot{\theta}_1 \\ \dot{\theta}_2 \end{bmatrix} \quad (a)$$

$$\begin{bmatrix} l_1 \cos \theta_1 & -l_1 \sin \theta_1 & l_1 \cos \theta_1 & -l_1 \sin \theta_1 \\ 0 & 0 & l_2 \cos \theta_2 & -l_2 \sin \theta_2 \end{bmatrix} \begin{bmatrix} F_{x1} \\ F_{y1} \\ F_{x2} \\ F_{y2} \end{bmatrix} = \begin{bmatrix} \tau_1 \\ \tau_2 \end{bmatrix}$$



$$\begin{aligned}T^* &= \frac{m_1}{2} [(l_1 \cos \theta_1 \dot{\theta}_1)^2 + (l_1 \sin \theta_1 \dot{\theta}_1)^2] \\ &+ \frac{m_2}{2} [(l_1 \cos \theta_1 \dot{\theta}_1 + l_2 \cos \theta_2 \dot{\theta}_2)^2 + (l_1 \sin \theta_1 \dot{\theta}_1 + l_2 \sin \theta_2 \dot{\theta}_2)^2]\end{aligned} \quad (c)$$

$$\begin{aligned}\delta W &= (m_1 g l_1 \sin \theta_1 + m_2 g l_1 \sin \theta_1) \delta \theta_1 \\ &+ (m_2 g l_2 \sin \theta_2) \delta \theta_2\end{aligned}$$

(T^* can be simplified using trigonometric identities.)

State Variables $\theta_1, \dot{\theta}_1, \theta_2, \dot{\theta}_2$

9-8

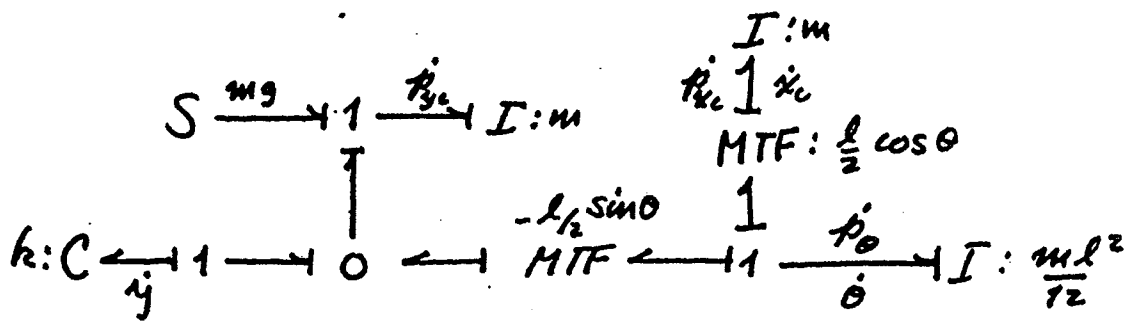
$$y_c = y + \frac{l}{2} \cos \theta$$

$$\dot{y}_c = \dot{y} - \frac{l}{2} \sin \theta \dot{\theta}$$

$$x_c = \frac{l}{2} \sin \theta$$

$$\dot{x}_c = \frac{l}{2} \cos \theta \dot{\theta}$$

9-7



$$\dot{\theta} = p_\theta \cdot \frac{12}{ml^2}$$

$$\dot{p}_{y_c} = mg - ky$$

$$\dot{y} = p_{y_c} / m + \frac{l}{2} \sin \theta p_\theta \cdot \frac{12}{ml^2}$$

$$\dot{p}_\theta = -\frac{l}{2} \sin \theta \cdot ky - \frac{l}{2} \cos \theta p_{x_c}$$

$$\begin{aligned} \text{but } p_{x_c} &= m \dot{x}_c = m \frac{l}{2} \cos \theta \cdot \dot{\theta} = m \frac{l}{2} \cos \theta p_\theta \cdot \frac{12}{ml^2} \\ &= \frac{6}{l} \cos \theta p_\theta \end{aligned}$$

so

$$\dot{p}_{x_c} = \frac{6}{l} \cos \theta \cdot \dot{p}_\theta + \frac{6}{l} p_\theta (-\sin \theta) \dot{\theta}$$

$$= \frac{6}{l} \cos \theta \dot{p}_\theta - \frac{6}{l} \sin \theta \cdot p_\theta^2 \cdot \frac{12}{ml^2}$$

After substitution in last state equation,

$$(1 + 3 \cos^2 \theta) \dot{p}_\theta = -\frac{l}{2} \sin \theta \cdot ky + \frac{36 \sin \theta \cos \theta}{ml^2} p_\theta^2$$

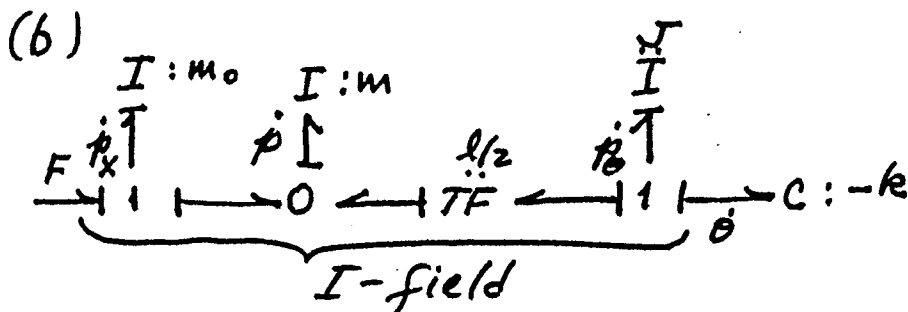
which is a short step from the final form.

$$9-9 \text{ (a)} \quad T^* = \frac{1}{2} J \dot{\theta}^2 + \frac{1}{2} m v^2 + \frac{1}{2} m_0 \dot{x}^2$$

9-8

$$v \approx \dot{x} + \frac{l}{2} \dot{\theta}; \quad V = mg \frac{l}{2} \cos \theta \approx mg \frac{l}{2} (1 - \frac{\theta^2}{2})$$

$$V = -\frac{k}{2} \theta^2 + \text{const}, \text{ so } k = mgl/2$$



$$\dot{\theta} = p_{\theta} / J$$

$$\dot{p}_x = F(t) - \dot{p}; \quad \dot{p}_{\theta} = -(-k)\theta - \frac{l}{2} \dot{p}$$

$$\text{but } p = m \left(\frac{p_x}{m_0} + \frac{l}{2} \frac{p_{\theta}}{J} \right), \text{ so } \dot{p} = \frac{m}{m_0} \dot{p}_x + \frac{ml}{2J} \dot{p}_{\theta}$$

then

$$\begin{bmatrix} \frac{1 + m/m_0}{ml/2m_0} & \frac{ml/2J}{1 + ml^2/4J} \end{bmatrix} \begin{bmatrix} \dot{p}_x \\ \dot{p}_{\theta} \end{bmatrix} = \begin{bmatrix} F(t) \\ k\theta \end{bmatrix}$$

(invert to get standard form.

(c) We can write T^* as follows

$$T^* = \frac{1}{2} [\dot{x} | \dot{\theta}] \underbrace{\begin{bmatrix} J + ml^2/4 & ml/2 \\ ml/2 & m_0 + m \end{bmatrix}}_{\text{Mass matrix of I-field}} \begin{bmatrix} \dot{x} \\ \dot{\theta} \end{bmatrix}$$

Mass matrix of I-field

$$[I]^{-1} = \frac{1}{(J + \frac{ml^2}{4})(m_0 + m) - \frac{m^2 l^2}{4}} \begin{bmatrix} m_0 + m & -ml/2 \\ -ml/2 & J + ml^2/4 \end{bmatrix}$$

$$= \begin{bmatrix} \Gamma_{11} & \Gamma_{12} \\ \Gamma_{21} & \Gamma_{22} \end{bmatrix} \text{ and } \begin{bmatrix} \dot{x} \\ \dot{\theta} \end{bmatrix} = \begin{bmatrix} \Gamma_{11} & \Gamma_{12} \\ \Gamma_{21} & \Gamma_{22} \end{bmatrix} \begin{bmatrix} p_x \\ p_{\theta} \end{bmatrix}$$

so state eqns are

$$\dot{p}_x = F; \quad \dot{p}_{\theta} = k\theta; \quad \dot{\theta} = \Gamma_{21} p_x + \Gamma_{22} p_{\theta}$$

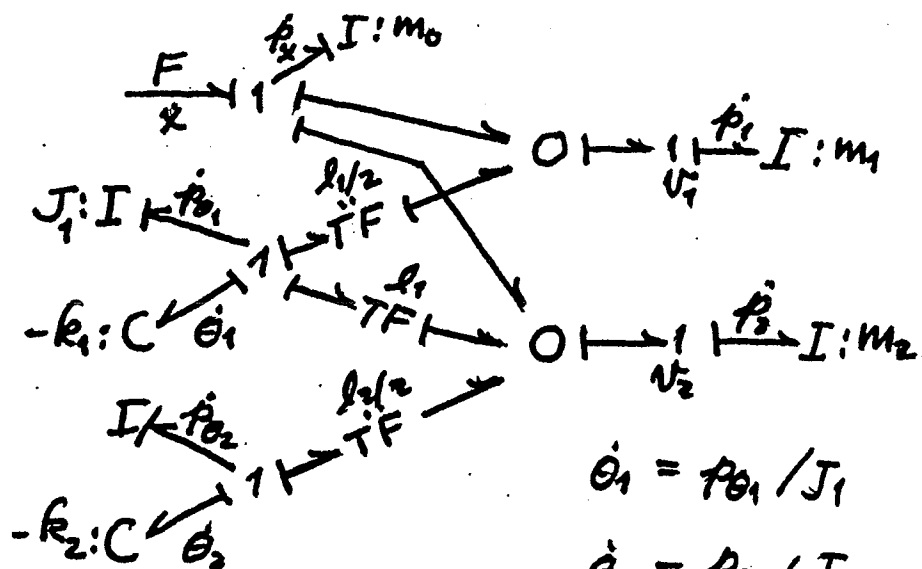
$$9-10 \quad T^* = \frac{1}{2} m_0 \dot{x}^2 + \frac{1}{2} J_1 \dot{\theta}_1^2 + \frac{1}{2} J_2 \dot{\theta}_2^2 + \frac{1}{2} m_1 v_1^2 + \frac{1}{2} m_2 v_2^2$$

$$v_1 \approx \dot{x} + \frac{l_1}{2} \dot{\theta}_1, \quad v_2 = \dot{x} + l_1 \dot{\theta}_1 + \frac{l_2}{2} \dot{\theta}_2$$

$$V = m_1 g \frac{l_1}{2} \cos \theta_1 + m_2 g (l_1 \cos \theta_1 + \frac{l_2}{2} \cos \theta_2) \quad 9-9$$

$$\approx (m_1 g \frac{l_1}{2} + m_2 g l_1) (-\frac{\theta_1^2}{2}) + m_2 g \frac{l_2}{2} (-\frac{\theta_2^2}{2}) + \text{const}$$

$$\approx -k_1 \frac{\theta_1^2}{2} - k_2 \frac{\theta_2^2}{2}; \quad k_1 = m_1 g \frac{l_1}{2} + m_2 g; \quad k_2 = m_2 g \frac{l_2}{2}.$$



$$\dot{\theta}_1 = \dot{\theta}_1 / J_1$$

$$\dot{\theta}_2 = \dot{\theta}_2 / J_2$$

$$\dot{p}_x = F(t) - \dot{p}_1 - \dot{p}_2$$

$$\dot{p}_{\theta_1} = k_1 \theta_1 - \frac{l_1}{2} \dot{p}_1 - l_1 \dot{p}_2$$

$$\dot{p}_{\theta_2} = k_2 \theta_2 - \frac{l_2}{2} \dot{p}_2$$

Basic
State
Equations

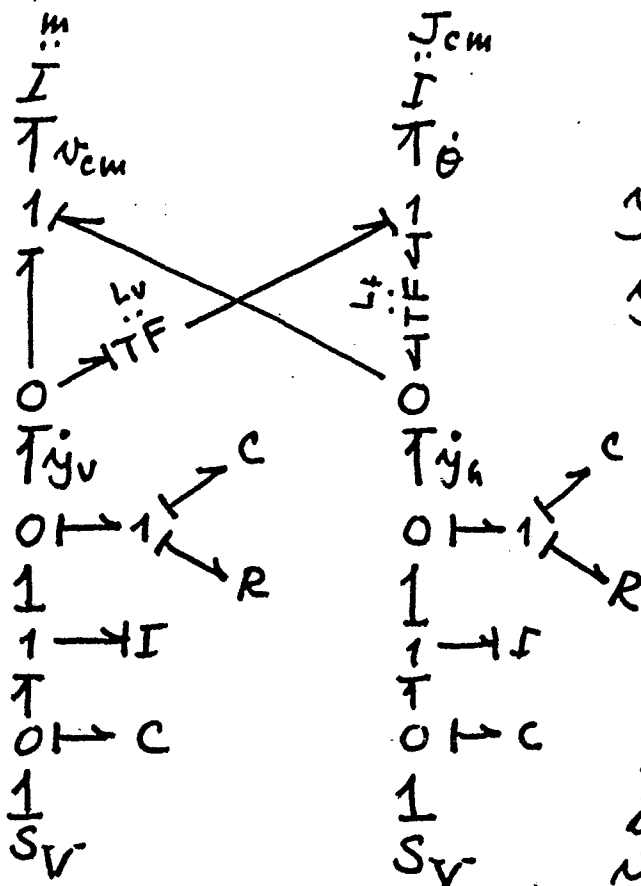
Derivative Causality Calculations

$$p_1 = m_1 v_1 = m_1 (\frac{p_x}{m_0} + \frac{l_1}{2} \frac{p_{\theta_1}}{J_1})$$

$$p_2 = m_2 v_2 = m_2 (\frac{p_x}{m_0} + l_1 \frac{p_{\theta_1}}{J_1} + \frac{l_2}{2} \frac{p_{\theta_2}}{J_2})$$

Substitute in for \dot{p}_1 and \dot{p}_2 and do a 3×3 matrix inversion to get standard form.

9-11



9-10

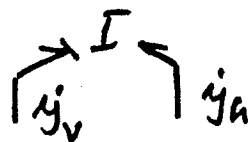
$$\dot{y}_v = \dot{y}_{cm} + L_v \dot{\theta}$$

$$\dot{y}_h = \dot{y}_{cm} - L_h \dot{\theta}$$

With y_{cm} and θ as components of the q_k vector, all integral causality is possible as shown.

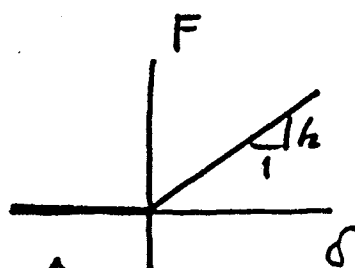
If the MTF structure is set up to compute \dot{y}_{cm} and $\dot{\theta}$ from y_v and y_h , then an algebraic loop results as the bond graph tries to invert the transformation.

The I-field form is just



where the I-matrix may be found directly from the transformer structure, or by evaluating T^* and then differentiating to get mass matrix elements.

Nonlinear constitutive laws for the tire C-elements can be used to model loss of contact.



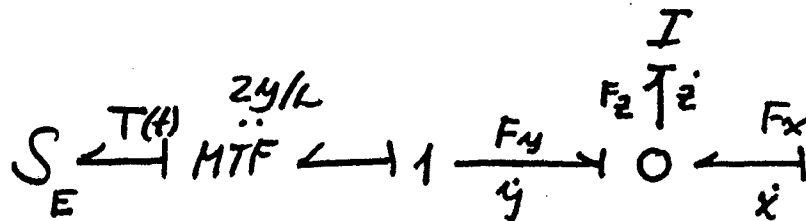
↑ zero force for negative deflection.

$$9-12 \quad \delta = 2(y^2 - L^2)^{1/2} - 2L$$

$$= 2L(1 + (\frac{y}{L})^2)^{1/2} - 2L$$

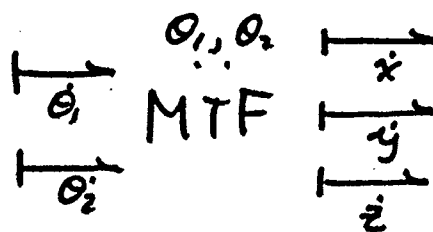
$$\approx 2L(1 + \frac{1}{2}(\frac{y}{L})^2) - 2L = y^2/L$$

$$\dot{\delta} = \frac{2y}{L} \dot{y} \quad F_y = -2 \frac{y}{L} T(t)$$



Note that the force source fixes causality on all bonds. This means that at any instant given $T(t)$ and y one can compute F_y . The graph of (a) does not show this necessary causality. Given F_y , it is not possible to compute y since $T(t)$ may even jump to zero at any time.

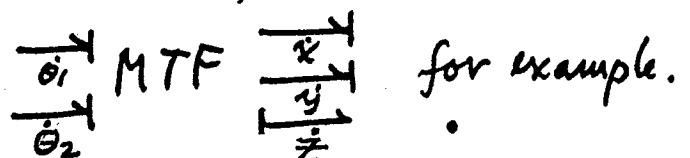
$$9-13 \quad \begin{bmatrix} \dot{x} \\ \dot{y} \\ \dot{z} \end{bmatrix} = \begin{bmatrix} L \cos \theta_1 \cos \theta_2 & -L \sin \theta_1 \sin \theta_2 \\ L \cos \theta_1 \sin \theta_2 & L \sin \theta_1 \cos \theta_2 \\ L \sin \theta_1 & 0 \end{bmatrix} \begin{bmatrix} \dot{\theta}_1 \\ \dot{\theta}_2 \end{bmatrix}$$



Corresponds to this causality

Integral causality on x, y, z bonds would mean inversion of a non-square matrix - clearly impossible

There are just two geometric degrees of freedom so this would work:



9-15

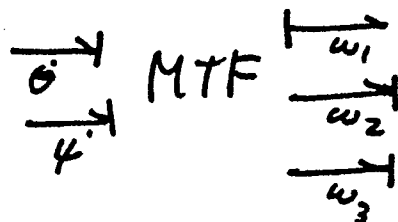
The MTF law is

9-12

$$\begin{bmatrix} \omega_1 \\ \omega_2 \\ \omega_3 \end{bmatrix} = \begin{bmatrix} 0 & \sin\theta \\ 0 & \cos\theta \\ 1 & 0 \end{bmatrix} \begin{bmatrix} \dot{\theta} \\ \dot{\psi} \end{bmatrix}$$

which shows immediately that $\omega_1, \omega_2, \omega_3$ cannot be chosen independently. It would, however, be possible to solve for $\dot{\theta}$ and $\dot{\psi}$ in terms of ω_2 and ω_3 and then to find ω_1 also in terms of ω_2 and ω_3 .

Thus, this causality is possible



The rest of the MTF laws are

$$\begin{bmatrix} \tau_\theta \\ \tau_\psi \end{bmatrix} = \begin{bmatrix} 0 & 0 & 1 \\ \sin\theta & \cos\theta & 0 \end{bmatrix} \begin{bmatrix} \tau_1 \\ \tau_2 \\ \tau_3 \end{bmatrix}$$

This MTF has two geometric degrees of freedom and 3 "force" degrees of freedom. Because of the 0's, not all sets of two flows will serve as inputs, however.

9-16

$$v_{x1} = v_x - \omega_y h + \omega_z a$$

$$v_{y1} = v_y + \omega_x h + \omega_z b$$

$$v_{z1} = v_z - \omega_x a - \omega_y b$$

$$v_{x2} = v_x - \omega_y h + \omega_z a$$

$$v_{y2} = v_y + \omega_x h - \omega_z c$$

$$v_{z2} = v_z - \omega_x a + \omega_y c$$

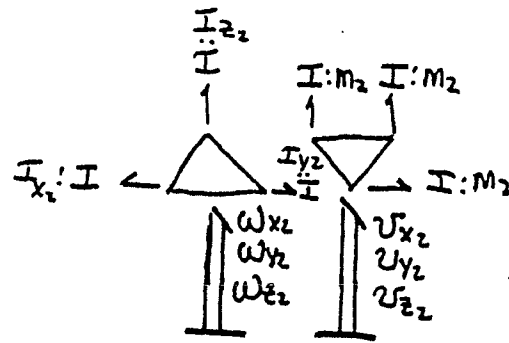
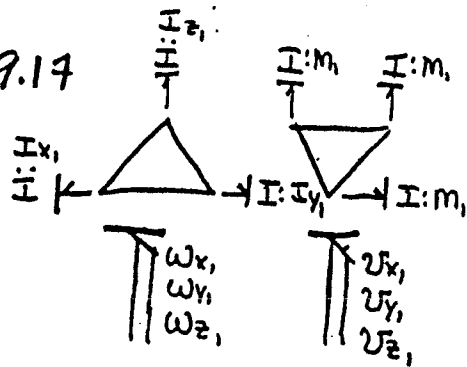
$$\begin{bmatrix} v_{x1} \\ v_{y1} \\ v_{z1} \\ v_{x2} \\ v_{y2} \\ v_{z2} \end{bmatrix} = \underbrace{\begin{bmatrix} 1 & 0 & 0 & 0 & -h & a \\ 0 & 1 & 0 & h & 0 & b \\ 0 & 0 & 1 & -a & -b & 0 \\ 1 & 0 & 0 & 0 & -h & a \\ 0 & 1 & 0 & h & 0 & -c \\ 0 & 0 & 1 & -a & c & 0 \end{bmatrix}}_{\underline{M}} \begin{bmatrix} v_x \\ v_y \\ v_z \\ \omega_x \\ \omega_y \\ \omega_z \end{bmatrix}$$

and

$$\begin{bmatrix} F_x \\ F_y \\ F_z \\ \tau_x \\ \tau_y \\ \tau_z \end{bmatrix} = \underline{M}^t \begin{bmatrix} F_{x1} \\ F_{y1} \\ F_{z1} \\ F_{x2} \\ F_{y2} \\ F_{z2} \end{bmatrix} \quad \leftarrow \text{amazing}$$

9-13

9.17



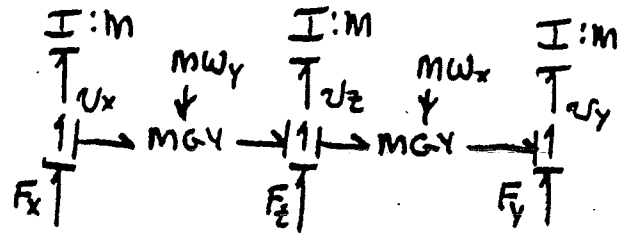
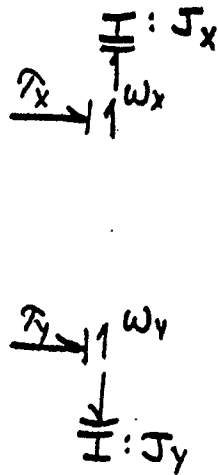
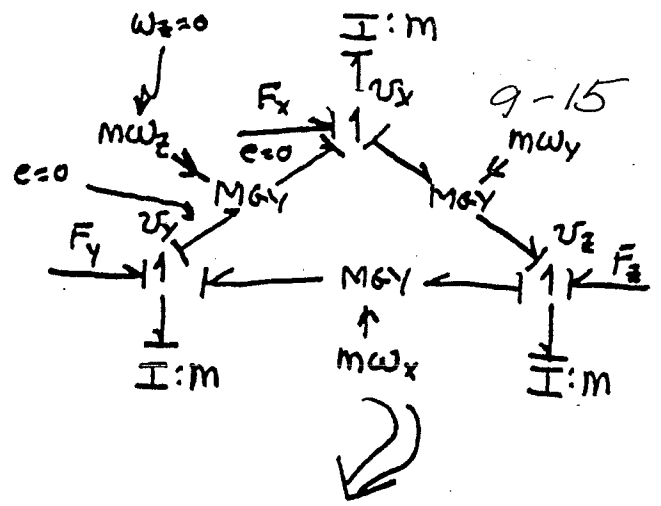
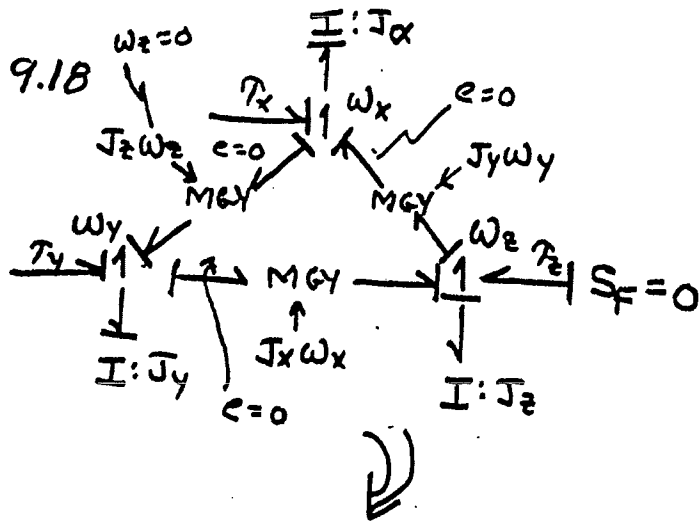
9-14

MTF \leftarrow These transformations \rightarrow MTF
 take body fixed, cg components into body fixed, attachment point components

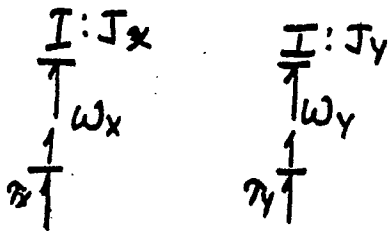
MTF \leftarrow The MTF's take \rightarrow MTF
 body fixed, attachment point components into inertial components

This 1-junction enforces constraint that inertial direction components at attachment point must be equal.

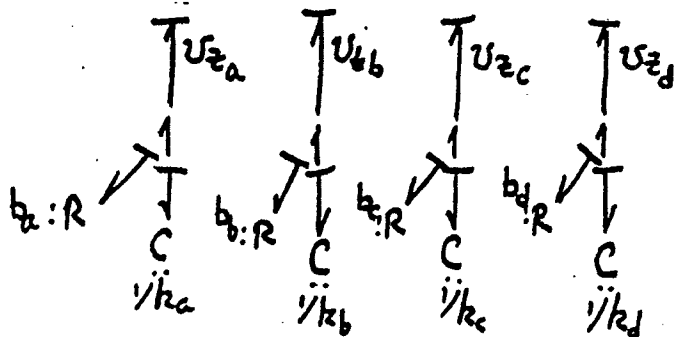
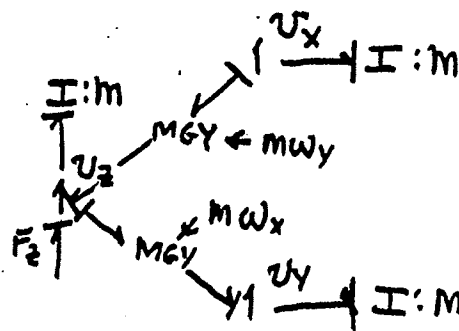
IF causality in body 1 is all integral, then some inertia element in body 2 must be in derivative causality, thus formulation is greatly complicated.



9.19



MTF



where

9-16

$$v_{za} = v_z + c_f \omega_x - W \omega_y$$

$$v_{zb} = v_z + c_f \omega_x + W \omega_y$$

$$v_{zc} = v_z - c_r \omega_x + W \omega_y$$

$$v_{zd} = v_z - c_r \omega_x - W \omega_y$$

$$\text{or } \begin{bmatrix} v_{za} \\ v_{zb} \\ v_{zc} \\ v_{zd} \end{bmatrix} = \begin{bmatrix} 1 & c_f & -W \\ 1 & c_f & W \\ 1 & -c_r & W \\ 1 & -c_r & -W \end{bmatrix} \begin{bmatrix} v_z \\ \omega_x \\ \omega_y \end{bmatrix}$$

9-20

From the bond graph

$$\dot{p}_x = F_x + m \omega_z v_y$$

$$\dot{p}_y = F_y - m \omega_z v_x$$

or

$$m(\dot{v}_x - \omega_z v_y) = F_x$$

$$m(\dot{v}_y + \omega_z v_x) = F_y$$

\therefore

$$a_x = \dot{v}_x - \omega_z v_y$$

$$a_y = \dot{v}_y + \omega_z v_x$$

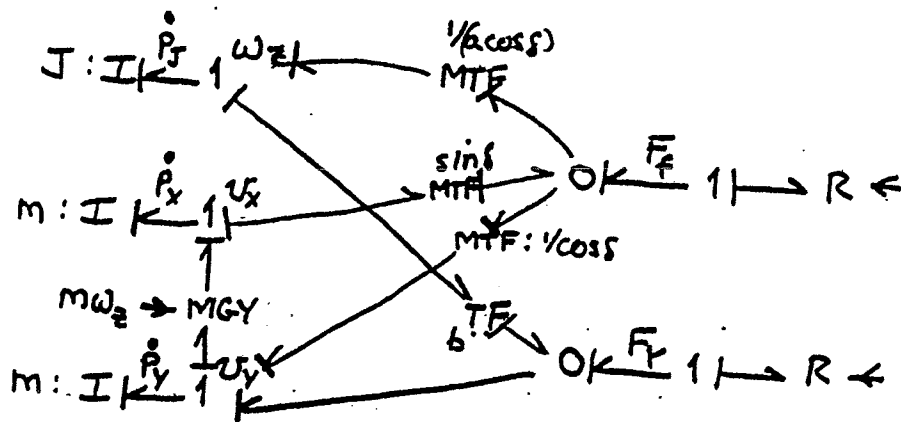
$$\text{but } v_x = \dot{r} \quad \omega_z = \dot{\theta}$$

$$v_y = r \dot{\theta}$$

$$\therefore a_x = \ddot{r} - \dot{\theta} r \dot{\theta} \leftarrow \text{same as } a_r$$

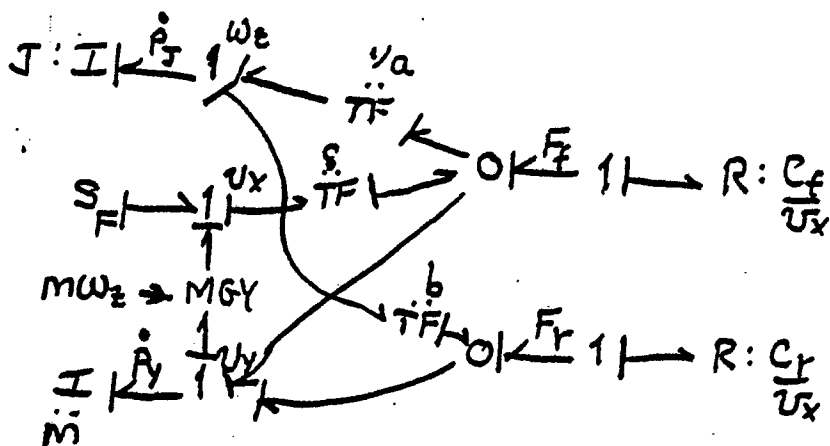
$$a_y = r \ddot{\theta} + \dot{r} \dot{\theta} + \dot{\theta} \dot{r} \leftarrow \text{same as } a_\theta$$

9-21



9-17

9-22



state vbls. P_J, P_Y

9-23

$$\dot{P}_J = -a \frac{C_f}{U_x} \left(a \frac{P_J}{J} + \frac{P_Y}{m} - v_x \delta \right) + b \frac{C_r}{U_x} \left(\frac{P_Y}{m} - b \frac{P_J}{J} \right)$$

$$\dot{P}_Y = -\frac{C_r}{U_x} \left(\frac{P_Y}{m} - b \frac{P_J}{J} \right) - \frac{C_f}{U_x} \left(a \frac{P_J}{J} + \frac{P_Y}{m} - \delta v_x \right) - m \frac{P_J}{J} v_x$$

$$\frac{d}{dt} \begin{bmatrix} P_Y \\ P_J \end{bmatrix} = \begin{bmatrix} \frac{-(C_r + C_f)}{m U_x} & \frac{(b C_r - a C_f)}{J U_x} - \frac{m v_x}{J} \\ \frac{b C_r - a C_f}{m U_x} & -\frac{(a^2 C_f + b^2 C_r)}{J U_x} \end{bmatrix} \begin{bmatrix} P_Y \\ P_J \end{bmatrix} + \begin{bmatrix} C_f \\ a C_f \end{bmatrix} \delta$$

$$9-24 \begin{bmatrix} s + \frac{C_r + C_f}{m v_x} & \frac{a C_f - b C_r}{J v_x} + \frac{m v_x}{J} \\ \frac{a C_f - b C_r}{m v_x} & s + \frac{a^2 C_f + b^2 C_r}{J v_x} \end{bmatrix} \begin{bmatrix} p_y \\ p_J \end{bmatrix} = 0 \quad 9-18$$

$$s^2 + \left[\frac{a^2 C_f + b^2 C_r}{J v_x} + \frac{C_r + C_f}{m v_x} \right] s + \frac{C_r + C_f}{m v_x} \left(\frac{a^2 C_f + b^2 C_r}{J v_x} \right) - \frac{(a C_f - b C_r)^2}{m J v_x^2} - \frac{(a C_f - b C_r) m v_x}{J} = 0$$

look at the constant term,

$$\frac{1}{m J v_x^2} \left\{ (a+b)^2 C_f C_r \right\} - \frac{(a C_f - b C_r)}{J} = \frac{1}{m J v_x^2} \left\{ (a+b)^2 C_f C_r + (b C_r - a C_f) m v_x^2 \right\}$$

thus

$$s^2 + \left[\frac{a^2 C_f + b^2 C_r}{J v_x} + \frac{C_r + C_f}{m v_x} \right] s + \frac{1}{m J v_x^2} \left[(a+b)^2 C_f C_r + (b C_r - a C_f) m v_x^2 \right] = 0$$

This equation will always have stable eigenvalues if all coefficients are positive. Thus if $b C_r > a C_f$, the system is stable. If

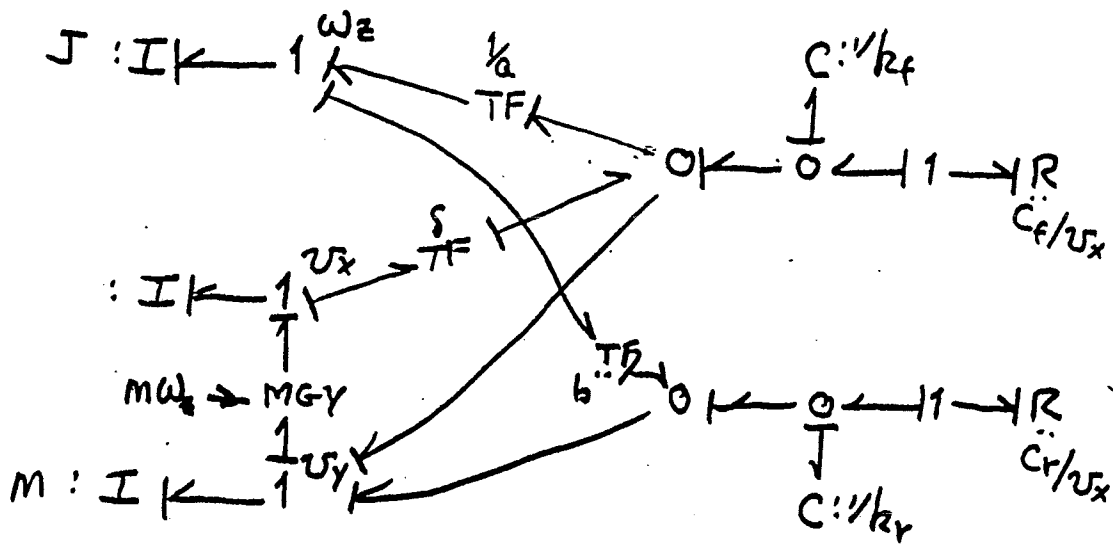
$$b C_r < a C_f$$

then last term is positive only if,

$$v_x^2 < \frac{(a+b)^2 C_f C_r}{m(a C_f - b C_r)} \quad \text{a critical speed.}$$

9-25

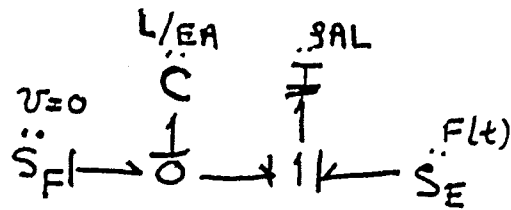
9-19



Note that the cornering forces are generated with the opposite causality from the previous problem.

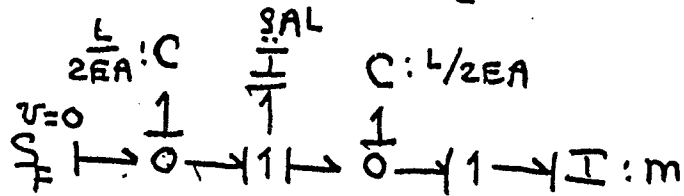
10.1

(a)

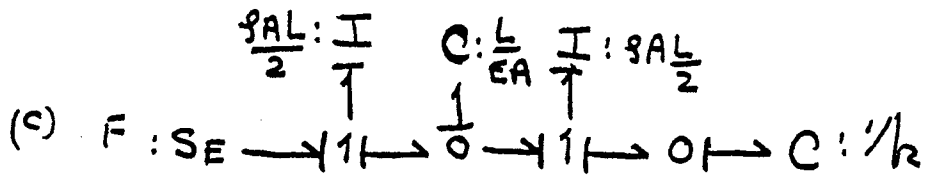


10-1

(b)

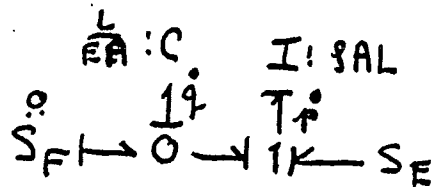


(c)



10.2

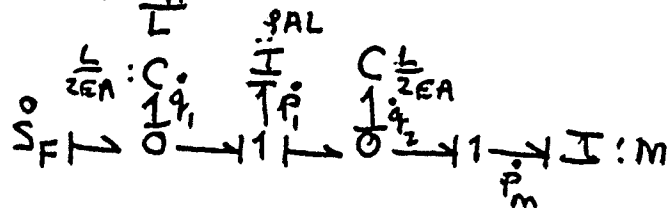
a)



$$\dot{q} = -\frac{P}{3AL}$$

$$P = F + q \frac{EA}{L}$$

b)

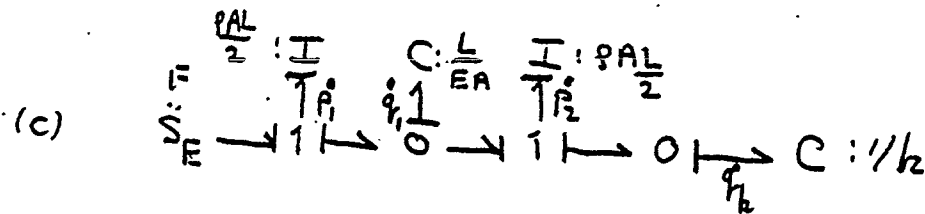


$$\dot{q}_1 = -\frac{P_1}{3AL}$$

$$P_1 = \frac{2EA}{L} q_1 - \frac{2EA}{L} q_2$$

$$\dot{q}_2 = \frac{P_1}{3AL} - \frac{P_m}{m}$$

$$P_m = \frac{2EA}{L} q_2$$



10-2

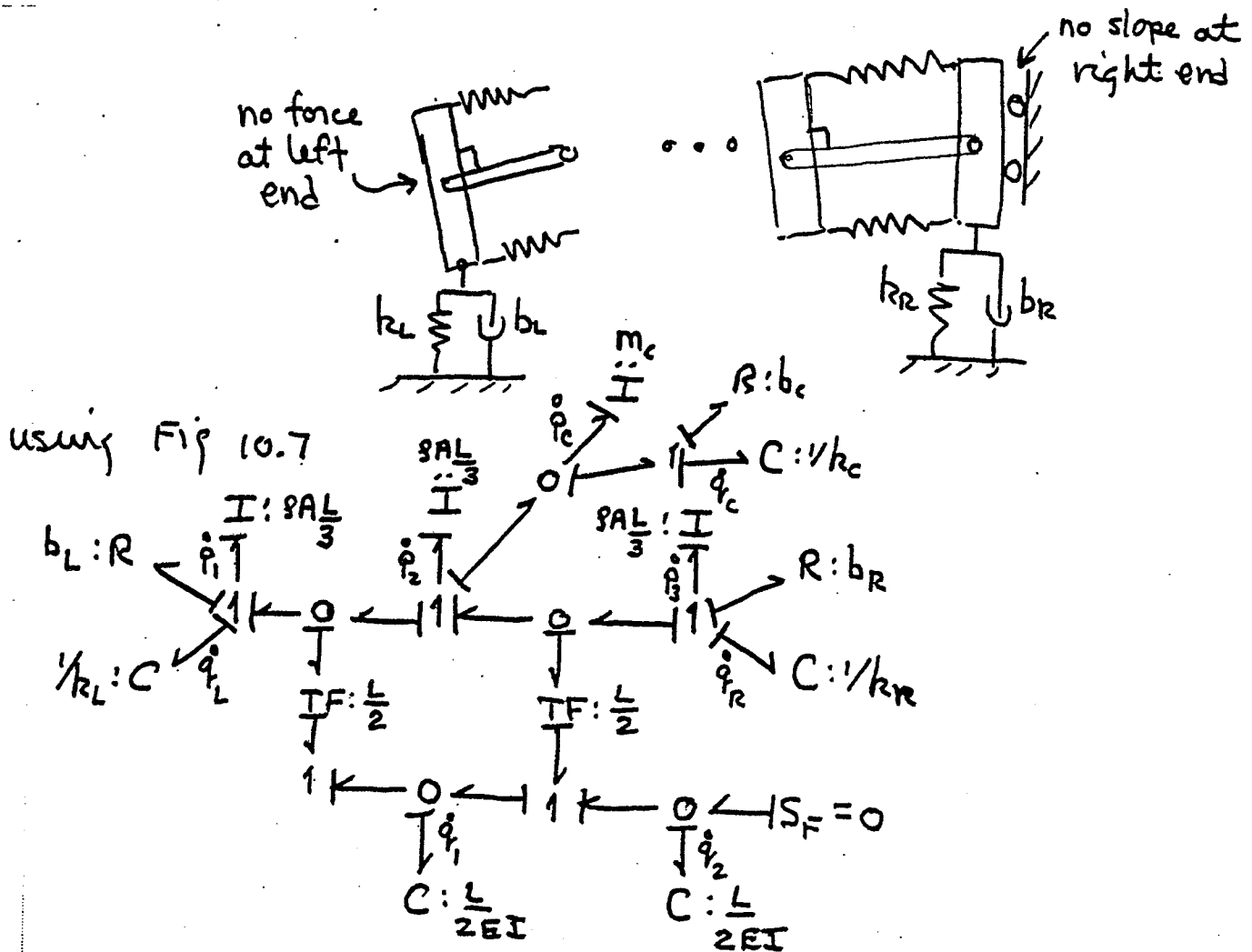
$$\dot{P}_1 = F - \frac{EA}{L} q_1$$

$$\dot{q}_1 = \frac{2P_1}{PAL} - \frac{2}{PAL} P_2$$

$$\dot{P}_2 = \frac{EA}{L} q_1 - k q_2$$

$$\dot{q}_2 = \frac{2P_2}{PAL}$$

10.3 It helps to draw the boundary lumps, so that modelling decisions are clear.



10.4 From soln. to P10.3,

10-3

$$\dot{P}_1 = -b_L \frac{P_1}{I} - k_L q_L + \frac{2(-1)}{L} \frac{q_1}{C}$$

$$\dot{P}_2 = +\frac{2}{L} \frac{q_1}{C} + \frac{2}{L} \left(\frac{q_1}{C} - \frac{q_2}{C} \right) - \left(k_C q_C + b_C \left(\frac{P_2}{I} - \frac{P_C}{m_C} \right) \right)$$

$$\dot{P}_3 = -b_R \frac{P_3}{I} - k_R q_R - \frac{2}{L} \left(\frac{q_1}{C} - \frac{q_2}{C} \right)$$

$$\dot{q}_1 = \frac{2}{L} \left(\frac{P_3}{I} - \frac{P_2}{I} \right) - \frac{2}{L} \left(\frac{P_2}{I} - \frac{P_1}{I} \right)$$

$$I = \rho A \frac{L}{3}$$

$$\dot{q}_2 = -\frac{2}{L} \left(\frac{P_3}{I} - \frac{P_2}{I} \right)$$

$$C = \frac{L}{2EI}$$

$$\dot{q}_L = \frac{P_1}{I} \quad \dot{q}_R = \frac{P_3}{I} \quad \dot{q}_C = \frac{P_2}{I} - \frac{P_C}{m_C}$$

10.5 $w(x,t) = Y(x) \cdot f(t)$ and $F = 0$ for now

$$\therefore \frac{\cancel{\rho} \frac{d^2 Y}{dx^2}}{\cancel{\rho Y f}} = \frac{\cancel{\rho Y} \frac{d^2 f}{dt^2}}{\cancel{\rho Y f}} = -\omega^2$$

$$\rho \frac{d^2 Y}{dx^2} + \rho \omega^2 Y = 0$$

or

$$\frac{d^2 Y}{dx^2} + \frac{\rho}{\rho} \omega^2 Y = 0$$

$$\therefore Y = A \cos kx + B \sin kx \quad k^2 = \frac{\rho}{\rho} \omega^2$$

$$\frac{dY}{dx}(0) = 0 = Bk \quad \therefore B = 0$$

$$\frac{dY}{dx}(L) = 0 = -Ak \sin kL = 0$$

$\therefore k = 0$ which is ok for this system
or $\sin kL = 0 \quad k_n L = n\pi \quad n = 0, 1, \dots$

and

$$\omega_n^2 = \frac{\pi}{g} \left(\frac{n\pi}{L} \right)^2 \quad n = 1, 2, 3, \dots$$

$$Y_n(x) = \cos n\pi \frac{x}{L}$$

10-4

for $\omega_n = 0$

$$\frac{d^2 Y_0}{dx^2} = 0$$

$$Y_0 = ax + b$$

but $\frac{dY_0}{dx}(0) = 0 = a$ ^{since} Y_0 must satisfy boundary conditions

$\therefore Y_0 = b$ where b is arbitrary, typically set to unity

$$\therefore \boxed{Y_0 = 1 \quad \omega_0 = 0}$$

10.6 $w(x, t) = \sum_n Y_n \eta_n(t)$

substitute

$$-\sum_{n=1}^{\infty} \pi \frac{d^2 Y_n}{dx^2} \eta_n + \sum_{n=1}^{\infty} g Y_n \frac{d^2 \eta_n}{dt^2} = F \delta(x-x_1)$$

but $\frac{d^2 Y_n}{dx^2} = -\frac{g}{\pi} \omega_n^2 Y_n$

$$\therefore \sum_{n=1}^{\infty} g \omega_n^2 Y_n \eta_n + \sum_{n=1}^{\infty} g Y_n \frac{d^2 \eta_n}{dt^2} = F \delta(x-x_1)$$

multiply by $Y_m(x)$ and $\int_0^L dx$

$$m_m \ddot{\eta}_m + m_m \omega_m^2 \eta_m = F Y_m(x_1)$$

$$m_m = \int_0^L g Y_m^2 dx = \int_0^L g \cos^2 m\pi \frac{x}{L} dx = \frac{gL}{2} \text{ for each mode.}$$

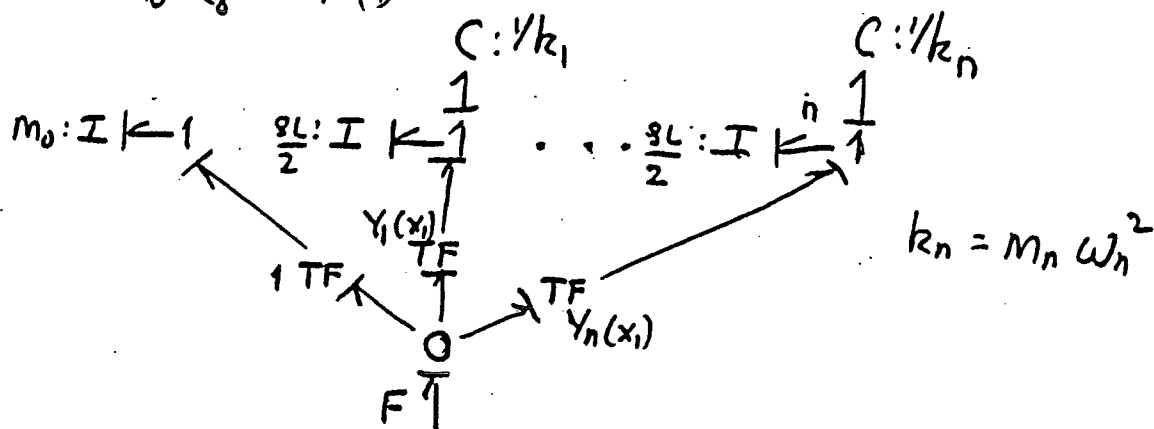
for zero freq. mode

$$m_0 = \int_0^L \frac{1}{2} \rho_0^2 dx = \rho L = \text{mass of string} \quad 10-5$$

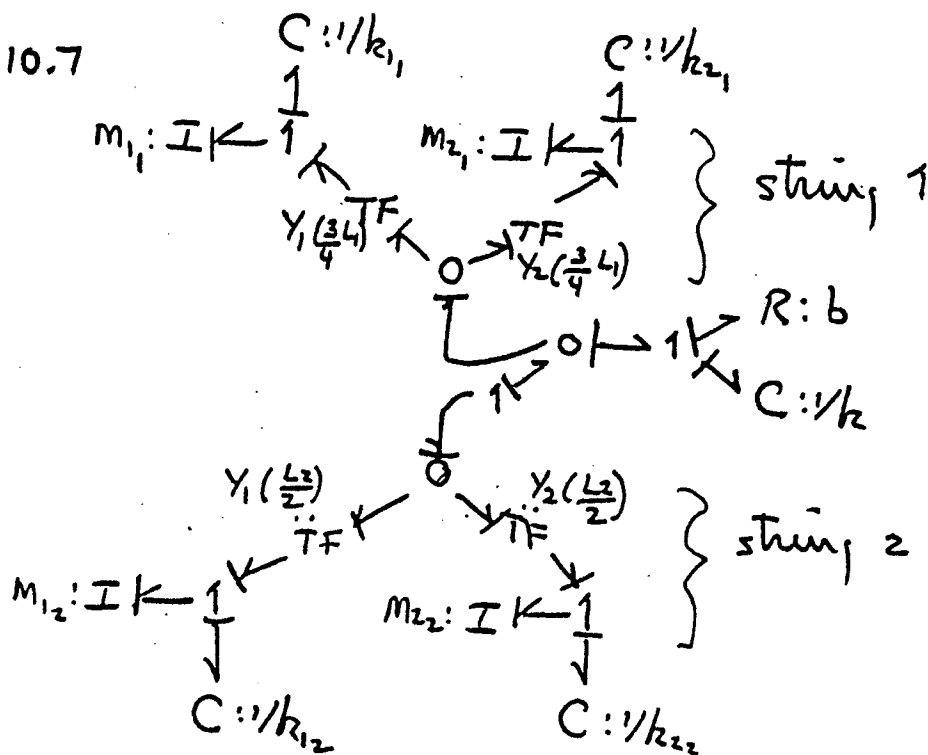
and

$$m_0 \ddot{\eta}_0 = F(1)$$

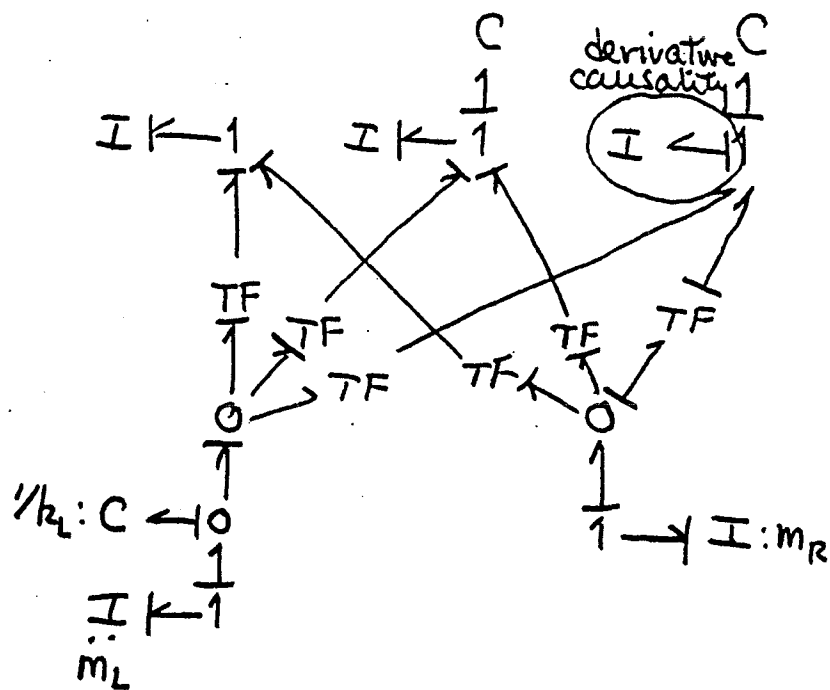
Thus



10.7

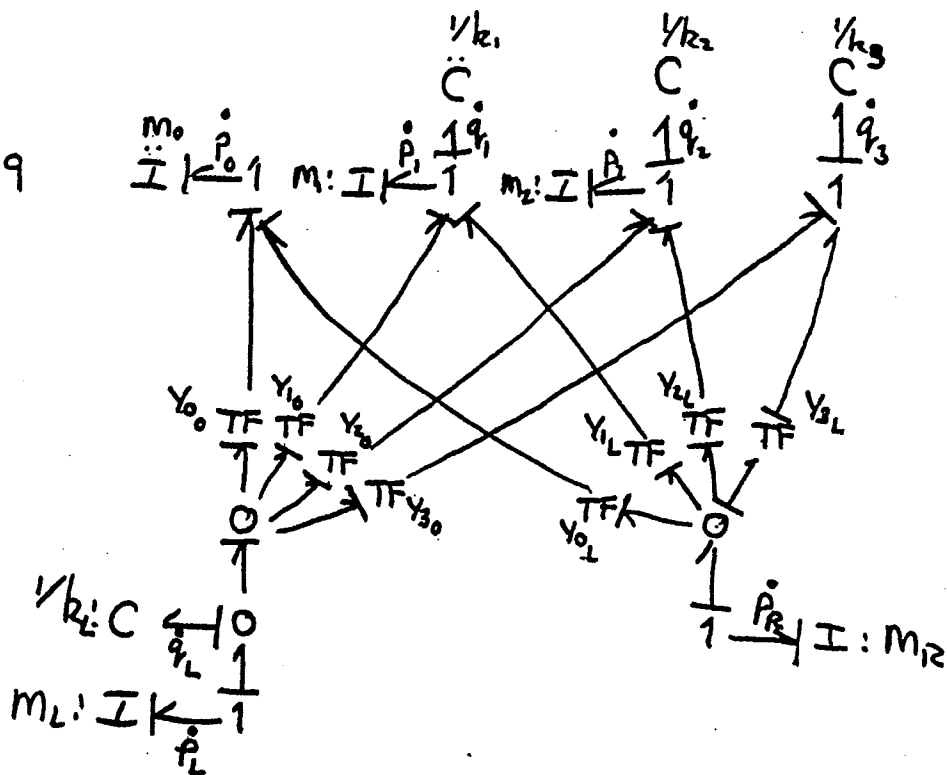


10.8



10-6

10.9



$$\begin{aligned}\ddot{p}_0 &= y_{00} k_L q_L + y_{0L} \frac{1}{Y_{3L}} (k_3 q_3 - y_{30} k_L q_L) \\ \ddot{p}_1 &= -k_1 q_1 + y_{10} k_L q_3 + \frac{y_{1L}}{Y_{3L}} (k_3 q_3 - y_{30} k_L q_L) \\ \ddot{p}_2 &= -k_2 q_2 + y_{20} k_L q_L + \frac{y_{2L}}{Y_{3L}} (k_3 q_3 - y_{30} k_L q_L)\end{aligned}$$

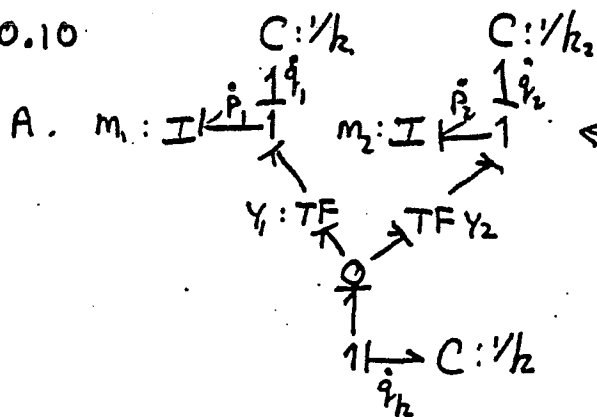
$$\ddot{q}_1 = \frac{P_1}{m_1} \quad \ddot{q}_2 = \frac{P_2}{m_2} \quad \ddot{q}_3 = \frac{1}{Y_{3L}} \left[\frac{P_R}{m_R} - Y_{0L} \frac{P_0}{m_0} - Y_{1L} \frac{P_1}{m_1} - Y_{2L} \frac{P_2}{m_2} \right]$$

$$\ddot{q}_L = \frac{P_L}{m_L} - Y_{00} \frac{P_0}{m_0} - Y_{10} \frac{P_1}{m_1} - Y_{20} \frac{P_2}{m_2} - \frac{Y_{30}}{Y_{3L}} \left[\frac{P_R}{m_R} - Y_{0L} \frac{P_0}{m_0} - Y_{1L} \frac{P_1}{m_1} - Y_{2L} \frac{P_2}{m_2} \right]$$

$$\ddot{P}_L = -k_L q_L$$

$$\ddot{P}_R = (-1) \frac{1}{Y_{3L}} \left[k_3 q_3 - Y_{30} k_L q_L \right]$$

10.10



$$\text{where } m_1 = m_2 = \frac{\rho A L}{2}$$

Fixed-free modes

$$k_i = m_i \omega_i^2 \quad \omega_i^2 = \frac{E}{\rho} (2i-1)^2 \left(\frac{\pi}{2L} \right)^2$$

$$Y_i = \sin(2i-1) \frac{\pi}{2} \quad (\text{at } x=L)$$

$$\ddot{P}_1 = -k_1 q_1 + Y_1 (-k q_h)$$

$$\ddot{P}_2 = -k_2 q_2 + Y_2 (-k q_h)$$

$$\ddot{q}_1 = P_1/m_1 \quad \ddot{q}_2 = P_2/m_2$$

$$\ddot{q}_h = Y_1 \frac{P_1}{m_1} + Y_2 \frac{P_2}{m_2}$$

$$\begin{bmatrix} s & 0 & k_1 & 0 & Y_1 k \\ 0 & s & 0 & k_2 & Y_2 k \\ -1/m_1 & 0 & s & 0 & 0 \\ 0 & -1/m_2 & 0 & s & 0 \\ -Y_1/m_1 & -Y_2/m_2 & 0 & 0 & s \end{bmatrix} \begin{bmatrix} P_1 \\ P_2 \\ q_1 \\ q_2 \\ q_h \end{bmatrix} = 0$$

expand determinant in terms of cofactors (looks bad, but requires only a little patience)

$$s \begin{bmatrix} s & 0 & k_1 & 0 & Y_1 k \\ 0 & s & 0 & k_2 & Y_2 k \\ -1/m_1 & 0 & s & 0 & 0 \\ 0 & -1/m_2 & 0 & s & 0 \\ -Y_1/m_1 & -Y_2/m_2 & 0 & 0 & s \end{bmatrix} + k_1 \begin{bmatrix} 0 & s & k_1 & 0 & Y_1 k \\ 0 & s & 0 & k_2 & Y_2 k \\ -1/m_1 & 0 & s & 0 & 0 \\ 0 & -1/m_2 & 0 & s & 0 \\ -Y_1/m_1 & -Y_2/m_2 & 0 & 0 & s \end{bmatrix} + Y_1 k \begin{bmatrix} 0 & s & 0 & k_1 & 0 \\ -1/m_1 & 0 & s & 0 & 0 \\ 0 & -1/m_2 & 0 & s & 0 \\ -Y_1/m_1 & -Y_2/m_2 & 0 & 0 & s \end{bmatrix}$$

again

$$S^2 \begin{bmatrix} s & k_2 & y_2 k \\ -\frac{1}{m_2} s & 0 & 0 \\ -\frac{y_2}{m_2} & 0 & s \end{bmatrix} + \frac{k_1}{m_1} \begin{bmatrix} s & k_2 & y_1 k \\ -\frac{1}{m_2} s & 0 & 0 \\ -\frac{y_1}{m_2} & 0 & s \end{bmatrix} - y_1 k s \begin{bmatrix} 0 & s & k_2 \\ 0 & -\frac{1}{m_2} s & 0 \\ -\frac{y_1}{m_1} & -\frac{y_2}{m_2} & 0 \end{bmatrix}$$

10-8

again

$$s^3 s^2 + \frac{k_2}{m_2} s^2 s + \frac{y_2^2 k}{m_2} s^2 s + \frac{k_1}{m_1} s^2 s + \frac{k_1}{m_1} \frac{k_2}{m_2} s + \frac{k_1}{m_1} \frac{y_2^2 k}{m_2} s + \frac{y_1^2 k}{m_1} s (s^2 + \frac{k_2}{m_2}) = 0$$

now let $k \rightarrow \infty$ and retain

$$k \left[\frac{y_2^2}{m_2} s^3 + \frac{k_1}{m_1} \frac{y_2^2}{m_2} s + \frac{y_1^2}{m_1} s (s^2 + \frac{k_2}{m_2}) \right] = 0$$

$$k s \left[\left(\frac{y_2^2}{m_2} + \frac{y_1^2}{m_1} \right) s^2 + \frac{k_1}{m_1} \frac{y_2^2}{m_2} + \frac{k_2}{m_2} \frac{y_1^2}{m_1} \right] = 0$$

thus, the remaining frequency is

$$\omega_n^2 = \frac{\frac{k_1}{m_1} \frac{y_2^2}{m_2} + \frac{k_2}{m_2} \frac{y_1^2}{m_1}}{\frac{y_2^2}{m_2} + \frac{y_1^2}{m_1}} = \frac{\omega_z^2 + \omega_1^2 \left(\frac{y_2}{y_1} \right)^2 \frac{m_1}{m_2}}{1 + \left(\frac{y_2}{y_1} \right)^2 \frac{m_1}{m_2}}$$

using the fixed-free modal parameters

$$\omega_1^2 = \frac{E}{\rho} \left(\frac{\pi}{2L} \right)^2 \quad \omega_2^2 = 9 \frac{E}{\rho} \left(\frac{\pi}{2L} \right)^2 \quad y_1 = \sin \frac{\pi}{2} = 1 \quad y_2 = \sin \frac{3\pi}{2} = -1$$

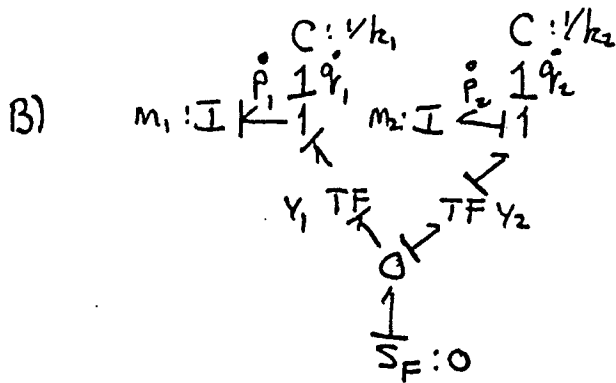
$$\omega_n^2 = \frac{E}{\rho} \left(\frac{\pi}{2L} \right)^2 \left[\frac{9 + 1}{1 + 1} \right] = 5 \frac{E}{\rho} \left(\frac{\pi}{2L} \right)^2 = \frac{5}{4} \frac{E}{\rho} \left(\frac{\pi}{L} \right)^2 \quad m_1 = m_2$$

The actual fixed-fixed mode frequency is

$$\omega_{act}^2 = \frac{E}{\rho} \left(\frac{\pi}{L} \right)^2$$

thus our 2 mode model, based upon fixed-free modes yields

$$\omega_n = 1.12 \sqrt{\frac{E}{\rho}} \frac{\pi}{L} \quad \text{while} \quad \omega_{act} = \sqrt{\frac{E}{\rho}} \frac{\pi}{L} \quad (12\% \text{ error})$$



$$P_2 = m_2 \frac{1}{y_2} (-v) y_1 \frac{P_1}{m_1} = -\frac{y_1}{y_2} P_1 \quad 10-9$$

$$\begin{aligned} \ddot{P}_1 &= -k_1 q_1 + \frac{y_1}{y_2} (k_2 q_2 + \ddot{P}_2) \\ &= -k_1 q_1 + \frac{y_1}{y_2} k_2 q_2 - \left(\frac{y_1}{y_2}\right)^2 \ddot{P}_1 \end{aligned}$$

$$(1) \ddot{P}_1 = \frac{-k_1}{1 + \left(\frac{y_1}{y_2}\right)^2} q_1 + \frac{y_1/y_2}{1 + \left(\frac{y_1}{y_2}\right)^2} k_2 q_2$$

$$(2) \ddot{q}_1 = +\frac{P_1}{m_1} \quad (3) \ddot{q}_2 = \frac{1}{y_2} (-v) y_1 \frac{P_1}{m_1}$$

in Laplace domain

$$\begin{bmatrix} s & \frac{1}{1 + \left(\frac{y_1}{y_2}\right)^2} & \frac{-k_2 y_1/y_2}{1 + \left(\frac{y_1}{y_2}\right)^2} \\ -\frac{1}{m_1} & s & 0 \\ \frac{y_1}{y_2} \frac{1}{m_1} & 0 & s \end{bmatrix} \begin{bmatrix} P_1 \\ q_1 \\ q_2 \end{bmatrix} = 0$$

expand

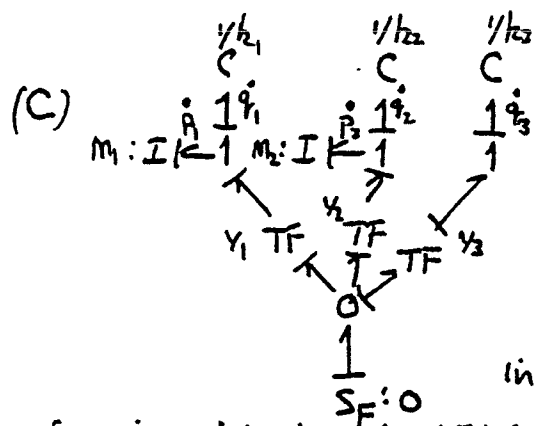
$$s^3 + \frac{k_1}{1 + \left(\frac{y_1}{y_2}\right)^2} \frac{s}{m_1} + \frac{\left(\frac{y_1}{y_2}\right)^2 k_2}{1 + \left(\frac{y_1}{y_2}\right)^2} \frac{s}{m_1} = 0$$

$$s \left[s^2 + \frac{k_1/m_1}{1 + \left(\frac{y_1}{y_2}\right)^2} + \frac{k_2}{m_1} \frac{\left(\frac{y_1}{y_2}\right)^2}{1 + \left(\frac{y_1}{y_2}\right)^2} \right] = 0$$

thus

$$\omega_n^2 = \frac{\omega_1^2 + \left(\frac{y_1}{y_2}\right)^2 \frac{m_2}{m_1} \omega_2^2}{1 + \left(\frac{y_1}{y_2}\right)^2}$$

← this result is identical to part A for $m_1/m_2 = 1$



$$\ddot{p}_1 = -k_1 q_1 + \frac{y_1}{y_3} k_3 q_3$$

$$\ddot{p}_2 = -k_2 q_2 + \frac{y_2}{y_3} k_3 q_3$$

$$\ddot{q}_1 = \frac{p_1}{m_1} \quad \ddot{q}_2 = \frac{p_2}{m_2} \quad \ddot{q}_3 = \frac{1}{y_3} \left[-y_1 \frac{p_1}{m_1} - y_2 \frac{p_2}{m_2} \right]$$

10-10

in Laplace domain

$$\begin{bmatrix} s & 0 & k_1 & 0 & -\frac{y_1}{y_3} k_3 \\ 0 & s & 0 & k_2 & -\frac{y_2}{y_3} k_3 \\ -\frac{1}{m_1} & 0 & s & 0 & 0 \\ 0 & -\frac{1}{m_2} & 0 & s & 0 \\ \frac{y_1}{y_3} \frac{1}{m_1} & \frac{y_2}{y_3} \frac{1}{m_2} & 0 & 0 & s \end{bmatrix} \begin{bmatrix} p_1 \\ p_2 \\ q_1 \\ q_2 \\ q_3 \end{bmatrix}$$

here we go again

$$s \begin{bmatrix} s & 0 & k_2 & -\frac{y_2}{y_3} k_3 \\ 0 & s & 0 & 0 \\ -\frac{1}{m_2} & 0 & s & 0 \\ \frac{y_2}{y_3} \frac{1}{m_2} & 0 & 0 & s \end{bmatrix} + k_1 \begin{bmatrix} 0 & s & k_2 & -\frac{y_2}{y_3} k_3 \\ -\frac{1}{m_1} & 0 & 0 & 0 \\ 0 & -\frac{1}{m_2} & s & 0 \\ \frac{y_1}{y_3} \frac{1}{m_1} & \frac{y_2}{y_3} \frac{1}{m_2} & 0 & s \end{bmatrix} - \frac{y_1}{y_3} k_3 \begin{bmatrix} 0 & s & 0 & k_2 \\ -\frac{1}{m_1} & 0 & s & 0 \\ 0 & -\frac{1}{m_2} & 0 & s \\ \frac{y_1}{y_3} \frac{1}{m_1} & \frac{y_2}{y_3} \frac{1}{m_2} & 0 & 0 \end{bmatrix}$$

again

$$s^2 \begin{bmatrix} s & k_2 & -\frac{y_2}{y_3} k_3 \\ -\frac{1}{m_2} & s & 0 \\ \frac{y_2}{y_3} \frac{1}{m_2} & 0 & s \end{bmatrix} + \frac{k_1}{m_1} \begin{bmatrix} s & k_2 & -\frac{y_2}{y_3} k_3 \\ -\frac{1}{m_2} & s & 0 \\ \frac{y_2}{y_3} \frac{1}{m_2} & 0 & s \end{bmatrix} + \frac{y_1}{y_3} k_3 s \begin{bmatrix} 0 & s & k_2 \\ 0 & -\frac{1}{m_2} & s \\ \frac{y_1}{y_3} \frac{1}{m_1} & \frac{y_2}{y_3} \frac{1}{m_2} & 0 \end{bmatrix}$$

again

$$\left[s^5 + s^3 \frac{k_2}{m_2} + \left(\frac{y_2}{y_3} \right)^2 \frac{k_2}{m_2} s^3 \right] + \left[\frac{k_1}{m_1} s^3 + \frac{k_1}{m_1} \frac{k_2}{m_2} s + \frac{k_1}{m_1} \left(\frac{y_2}{y_3} \right)^2 \frac{k_2}{m_2} s \right] + \left(\frac{y_1}{y_3} \right)^2 \frac{k_3}{m_1} s \left(s^2 + \frac{k_2}{m_2} \right) = 1$$

$$\text{or } s \left[s^4 + \left(\frac{k_2}{m_2} + \left(\frac{y_2}{y_3} \right)^2 \frac{k_2}{m_2} + \frac{k_1}{m_1} + \left(\frac{y_1}{y_3} \right)^2 \frac{k_2}{m_1} \right) s^2 + \frac{k_1}{m_1} \frac{k_2}{m_2} + \left(\frac{y_2}{y_3} \right)^2 \frac{k_1}{m_1} \frac{k_2}{m_2} + \left(\frac{y_1}{y_3} \right)^2 \frac{k_3}{m_1} \frac{k_2}{m_2} \right] = 0$$

to obtain the predicted natural frequencies, let $s \rightarrow j\omega$ 10-1.1
and noting that $m_1 = m_2 = m_3$

$$\omega^4 - \left[\omega_2^2 + \omega_1^2 + \omega_3^2 \left(\left(\frac{Y_2}{Y_3} \right)^2 + \left(\frac{Y_1}{Y_3} \right)^2 \right) \right] \omega^2 + \omega_1^2 \omega_2^2 + \omega_3^2 \left(\left(\frac{Y_2}{Y_3} \right)^2 \omega_1^2 + \left(\frac{Y_1}{Y_3} \right)^2 \omega_2^2 \right) = 0$$

For fixed-free modes, $Y_1 = \sin \frac{\pi}{2} = 1$ $Y_2 = \sin \frac{3\pi}{2} = -1$ $Y_3 = \sin \frac{5\pi}{2} = 1$

$$\omega_1^2 = \frac{E}{\rho} \left(\frac{\pi}{2L} \right)^2 \quad \omega_2^2 = 9 \frac{E}{\rho} \left(\frac{\pi}{2L} \right)^2 \quad \omega_3^2 = 25 \frac{E}{\rho} \left(\frac{\pi}{2L} \right)^2$$

use in above,

$$\omega^4 - \left[1 + 9 + 25(1+1) \right] \frac{E}{\rho} \left(\frac{\pi}{2L} \right)^2 \omega^2 + \left[9 + 25(1+9) \right] \left(\frac{E}{\rho} \right)^2 \left(\frac{\pi}{2L} \right)^4 = 0$$

$$\omega^4 - 60 \frac{E}{\rho} \left(\frac{\pi}{2L} \right)^2 \omega^2 + 259 \left(\frac{E}{\rho} \right)^2 \left(\frac{\pi}{2L} \right)^4 = 0$$

$$\omega_{1,2}^2 = \left[30 \pm \sqrt{900 - 259} \right] \frac{E}{\rho} \left(\frac{\pi}{2L} \right)^2$$

$$\omega_1^2 = 4.68 \frac{E}{\rho} \left(\frac{\pi}{2L} \right)^2$$

$$\omega_1 = 1.08 \sqrt{\frac{E}{\rho}} \frac{\pi}{L}$$

$$\omega_2^2 = 55.32 \frac{E}{\rho} \left(\frac{\pi}{2L} \right)^2$$

$$\omega_2 = 3.72 \sqrt{\frac{E}{\rho}} \frac{\pi}{L}$$

} predicted
from
fixed-free
modes.

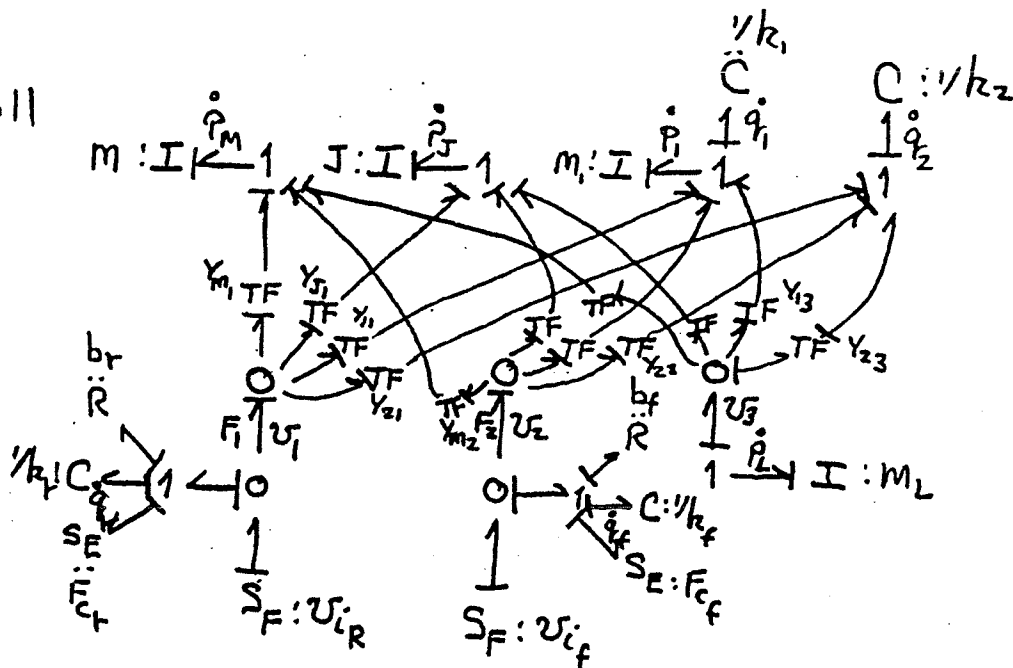
actual fixed-fixed frequencies

$$\omega_{1a} = \sqrt{\frac{E}{\rho}} \frac{\pi}{L} \quad 8\% \text{ error in first freq.}$$

$$\omega_{2a} = 2 \sqrt{\frac{E}{\rho}} \frac{\pi}{L} \quad 86\% \text{ error in second freq.}$$

10.11

10-12



$$\ddot{P}_M = Y_{M1} (F_{c_r} + k_r q_r + b_r (v_{i_r} - v_1)) + Y_{M2} (F_{c_f} + k_f q_f + b_f (v_{i_f} - v_2)) + \frac{Y_{13}}{Y_{23}} (k_2 q_2 - Y_{22} F_2 - Y_{21} F_1)$$

$$\ddot{P}_J = Y_{J1} F_1 + Y_{J2} F_2 + \frac{Y_{J3}}{Y_{23}} (k_2 q_2 - Y_{22} F_2 - Y_{21} F_1)$$

$$\ddot{P}_1 = k_1 q_1 + Y_{11} F_1 + Y_{12} F_2 + \frac{Y_{13}}{Y_{23}} (k_2 q_2 - Y_{22} F_2 - Y_{21} F_1)$$

$$\ddot{P}_L = (-1) \frac{1}{Y_{23}} (k_2 q_2 - Y_{22} F_2 - Y_{21} F_1)$$

$$\ddot{q}_1 = P_1 / M_1 \quad \ddot{q}_2 = \frac{1}{Y_{23}} \left(\frac{P_L}{M_L} - Y_{13} \frac{P_1}{M_1} - Y_{J3} \frac{P_J}{J} - Y_{M3} \frac{P_M}{M} \right)$$

$$\ddot{q}_r = v_{i_r} - v_1$$

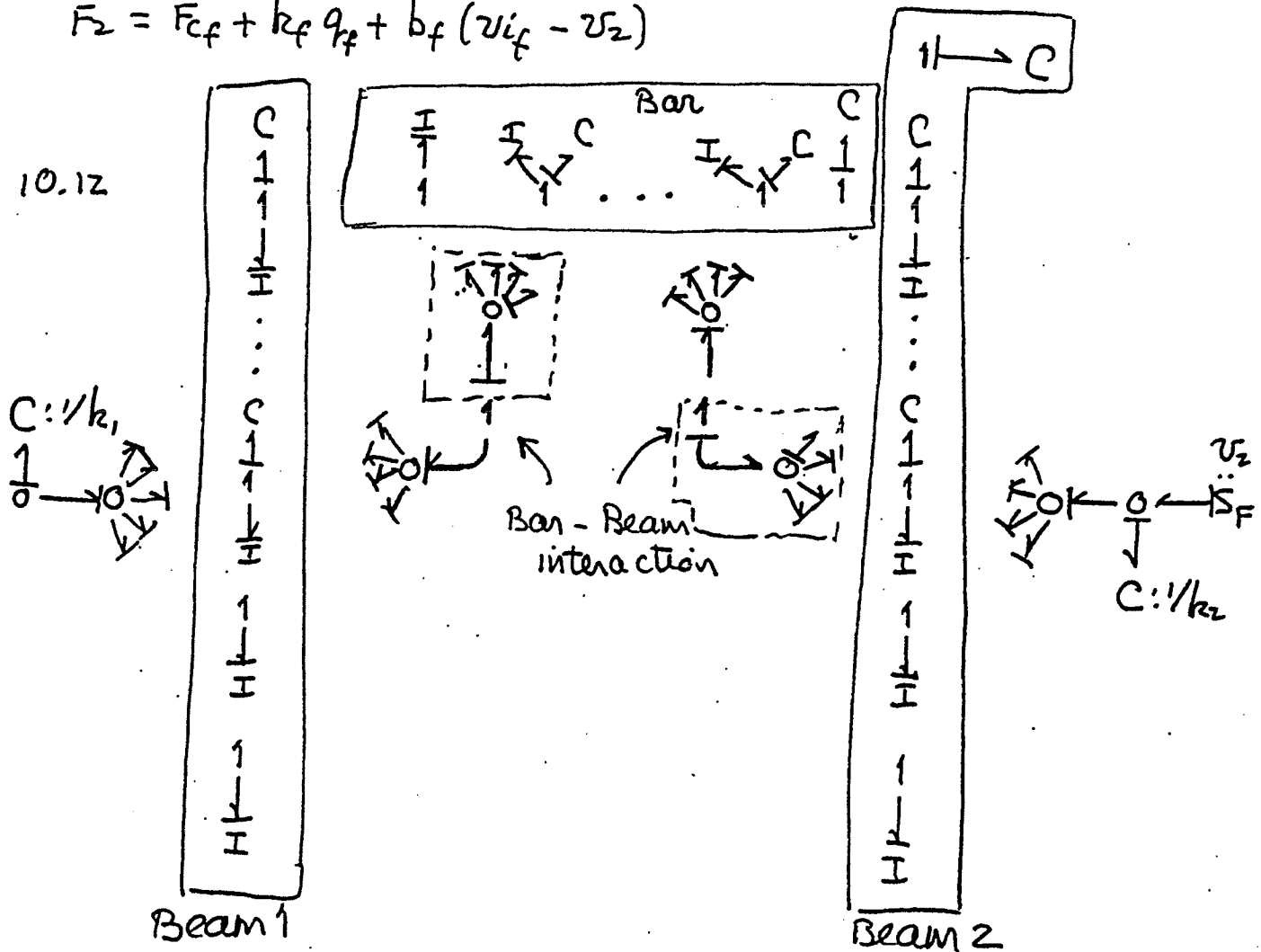
$$\ddot{q}_f = v_{i_f} - v_2$$

$$v_1 = Y_{M1} \frac{P_M}{M} + Y_{J1} \frac{P_J}{J} + Y_{11} \frac{P_1}{M_1} + \frac{Y_{21}}{Y_{23}} \left(\frac{P_L}{M_L} - Y_{13} \frac{P_1}{M_1} - Y_{J3} \frac{P_J}{J} - Y_{M3} \frac{P_M}{M} \right)$$

$$v_2 = Y_{M2} \frac{P_M}{M} + Y_{J2} \frac{P_J}{J} + Y_{12} \frac{P_1}{M_1} + \frac{Y_{22}}{Y_{23}} \left(\frac{P_L}{M_L} - Y_{13} \frac{P_1}{M_1} - Y_{J3} \frac{P_J}{J} - Y_{M3} \frac{P_M}{M} \right)$$

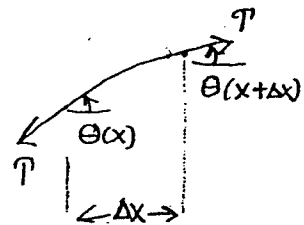
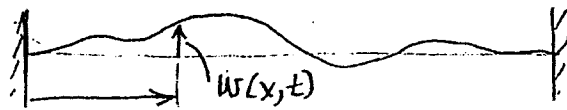
$$F_1 = F_{c1} + k_r q_r + b_r (v_{i1} - v_1)$$

$$F_2 = F_{c2} + k_f q_f + b_f (v_{i2} - v_2)$$



causal flow input at left end of bar dictates additional c-element in bar modes. Causal flow input into beam 2 dictates additional c-element in beam modes.

10.13



$$T \theta(x+\Delta x) - T \theta(x) = \rho \Delta x \frac{\partial^2 w}{\partial t^2}$$

$$\theta(x+\Delta x) = \theta(x) + \frac{\partial \theta}{\partial x} \Delta x \quad \text{and} \quad \theta \sim \frac{\partial w}{\partial x}$$

∴

$$T \frac{\partial^2 w}{\partial x^2} = \rho \frac{\partial^2 w}{\partial t^2} \quad \leftarrow \text{eqn. of motion}$$

let

$$w(x, t) = Y(x) \cdot f(t)$$

$$T \frac{d^2 Y}{dx^2} f = \rho \frac{d^2 f}{dt^2} Y \quad \text{or} \quad \frac{T}{\rho} \frac{d^2 Y}{dx^2} \frac{1}{Y} = \frac{d^2 f}{dt^2} \frac{1}{f} = -\omega^2$$

$$\text{or} \quad \frac{d^2 Y}{dx^2} + \frac{\rho \omega^2}{T} Y = 0$$

$$\text{or} \quad \frac{d^2 Y}{dx^2} + \beta^2 Y = 0 \quad \beta^2 = \frac{\rho}{T} \omega^2$$

$$Y(x) = A \cos \beta x + B \sin \beta x$$

apply boundary conditions

$$\left. \begin{array}{l} 0 = A \\ 0 = B \sin \beta L \end{array} \right\} \sin \beta_n L = 0 \quad \beta_n L = n\pi \quad n=1, 2, 3, \dots$$

$$\therefore \omega_n^2 = \frac{T}{\rho} \beta_n^2 = \frac{T}{\rho} \left(\frac{n\pi}{L} \right)^2$$

$$Y_n(x) = B_n \sin n\pi \frac{x}{L}$$

10.13 (continued)

The modal mass, m_n , is

10-15

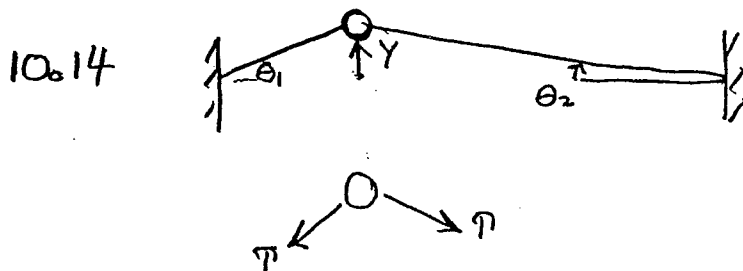
$$m_n = \int_0^L \rho Y_n^2(x) dx = \rho \int_0^L B_n^2 \sin^2 n\pi x \frac{dx}{L} = \frac{\rho L B_n^2}{2}$$

$$k_n = m_n \omega_n^2 = \frac{\rho L B_n^2}{2} \frac{\rho (n\pi)^2}{L}$$

B_n is arbitrary. We usually choose $B_n = 1$, or perform some normalizing operation such as,

$$m_n = 1 = \frac{\rho L B_n^2}{2} \quad \leftarrow \text{let all modal masses} = 1$$

$$B_n = \sqrt{\frac{2}{\rho L}}$$

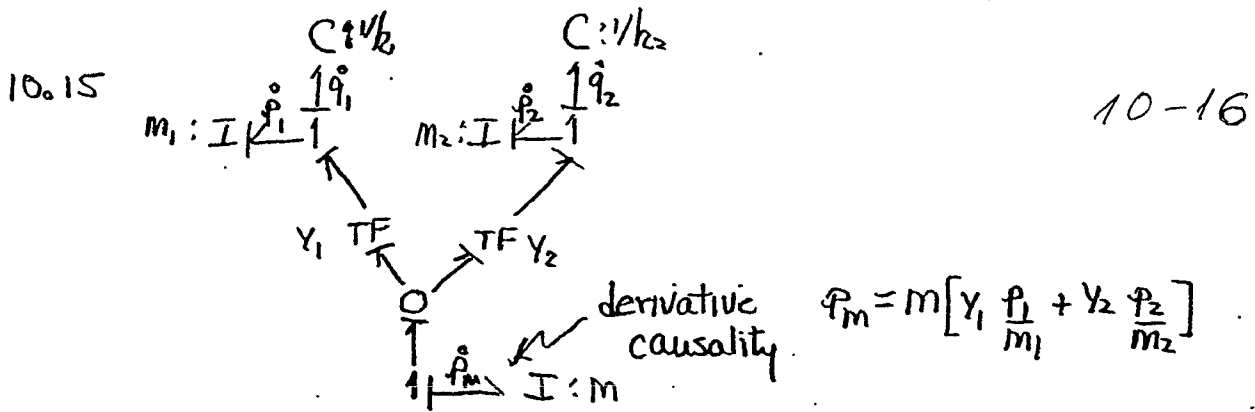


$$-\tau \theta_1 - \tau \theta_2 = m \ddot{y}$$

$$\theta_1 \sim \frac{y}{\frac{1}{4}L} \quad \theta_2 \sim \frac{y}{\frac{3}{4}L}$$

$$\therefore -\tau \left[4 \frac{y}{L} + \frac{4y}{3L} \right] = m \ddot{y}$$

$$m \ddot{y} + \frac{16}{3} \frac{\tau}{L} y = 0 \quad \therefore \omega_n^2 = \frac{16}{3} \frac{\tau}{L} \frac{1}{m}$$



$$\ddot{p}_1 = -k_1 q_1 - y_1 \ddot{p}_m = -k_1 q_1 - y_1 \left[\frac{m}{m_1} y_1 \ddot{p}_1 + \frac{m}{m_2} y_2 \ddot{p}_2 \right]$$

$$\ddot{p}_2 = -k_2 q_2 - y_2 \ddot{p}_m = -k_2 q_2 - y_2 \left[\frac{m}{m_1} y_1 \ddot{p}_1 + \frac{m}{m_2} y_2 \ddot{p}_2 \right]$$

$$q_1 = p_1 / m_1$$

$$q_2 = p_2 / m_2$$

in the s-domain

$$\begin{bmatrix} \left[1 + \frac{m}{m_1} y_1^2 \right] s & \frac{m}{m_2} y_1 y_2 s \\ \frac{m}{m_1} y_1 y_2 s & \left[1 + \frac{m}{m_2} y_2^2 \right] s \\ -1/m_1 & 0 \\ 0 & -1/m_2 \end{bmatrix} \begin{bmatrix} k_1 \\ 0 \\ s \\ 0 \end{bmatrix} \begin{bmatrix} 0 \\ k_2 \\ 0 \\ s \end{bmatrix} \begin{bmatrix} p_1 \\ p_2 \\ q_1 \\ q_2 \end{bmatrix} = 0$$

derive characteristic eqn:

$$\left(1 + \frac{m}{m_1} y_1^2 \right) s^2 \left[\left(1 + \frac{m}{m_2} y_2^2 \right) s^2 + \frac{k_2}{m_2} \right] - \frac{m}{m_1} (y_1 y_2)^2 s^4 + \frac{k_1}{m_1} \left[\left(1 + \frac{m}{m_2} y_2^2 \right) s^2 + \frac{k_2}{m_2} \right] = 0$$

$$\left[\left(1 + \frac{m}{m_1} y_1^2 \right) \left(1 + \frac{m}{m_2} y_2^2 \right) - \frac{m}{m_1} y_1^2 y_2^2 \frac{m}{m_2} \right] s^4 + \left[\left(1 + \frac{m}{m_1} y_1^2 \right) \frac{k_2}{m_2} + \left(1 + \frac{m}{m_2} y_2^2 \right) \frac{k_1}{m_1} \right] s^2 + \frac{k_1}{m_1} \frac{k_2}{m_2} = 0$$

$$\left[1 + \frac{m}{m_1} y_1^2 + \frac{m}{m_2} y_2^2 \right] s^4 + \left[\left(1 + \frac{m}{m_1} y_1^2 \right) \frac{k_2}{m_2} + \left(1 + \frac{m}{m_2} y_2^2 \right) \frac{k_1}{m_1} \right] s^2 + \frac{k_1}{m_1} \frac{k_2}{m_2} = 0$$

10.15 (continued)

10-1.7

let $\frac{k_1}{m_1} = \omega_1^2$, $\frac{k_2}{m_2} = \omega_2^2$, $m_1 = m_2 = \frac{m_s}{2} = m_m$, $m_s = \text{mass of string} = \mu L$

let $s = j\omega$ to introduce frequency eqn:

$$\omega^4 = \underbrace{\left(1 + \frac{m}{m_m} \gamma_1^2\right) \omega_2^2 + \left(1 + \frac{m}{m_m} \gamma_2^2\right) \omega_1^2}_{b} \omega^2 + \underbrace{\frac{\omega_1^2 \omega_2^2}{1 + \frac{m}{m_m} \gamma_1^2 + \frac{m}{m_m} \gamma_2^2}}_c = 0$$

$$\omega^4 - b\omega^2 + c = 0$$

$$\omega^2 = \frac{b}{2} \pm \sqrt{\left(\frac{b}{2}\right)^2 - c} = \frac{b}{2} \pm \frac{b}{2} \sqrt{1 - 4 \frac{c}{b^2}} = \frac{b}{2} \left\{ 1 \pm \sqrt{1 - 4 \frac{c}{b^2}} \right\}$$

Let's do a little trick here. It is likely that $\frac{m}{m_m} \gg 1$, i.e. the mass element is more massive than the string. Then, as long as γ_1 and γ_2 are not zero (which they are not at $x = \frac{L}{4}$), it is reasonable to have,

$$b = \frac{\gamma_1^2 \omega_2^2 + \gamma_2^2 \omega_1^2}{\gamma_1^2 + \gamma_2^2} \quad c = \frac{\omega_1^2 \omega_2^2}{\frac{m}{m_m} (\gamma_1^2 + \gamma_2^2)}$$

Then,

$$\omega^2 = \frac{b}{2} \left\{ 1 \pm \sqrt{1 - 4 \underbrace{\frac{\omega_1^2 \omega_2^2}{\frac{m}{m_m} (\gamma_1^2 + \gamma_2^2)} (\gamma_1^2 \omega_2^2 + \gamma_2^2 \omega_1^2)^2}_{\epsilon}} \right\}$$

The next trick is that if $\frac{m}{m_m} \gg 1$, the radical looks like

$$\sqrt{1 - \epsilon} \approx 1 - \frac{\epsilon}{2} \quad \text{for } \epsilon \ll 1$$

10.15 (continued)

10-18

Then

$$\omega^2 = \frac{b}{2} \left\{ 1 \pm \left(1 - \frac{\epsilon}{2} \right) \right\}$$

and the lowest frequency would be,

$$\omega^2 = \frac{b}{2} \frac{\epsilon}{2} = \frac{1}{4} \frac{Y_1^2 \omega_2^2 + Y_2^2 \omega_1^2}{Y_1^2 + Y_2^2} \cdot \frac{4 \omega_1^2 \omega_2^2}{\frac{M}{m_m}} \frac{(Y_1^2 + Y_2^2)}{(Y_1^2 \omega_2^2 + Y_2^2 \omega_1^2)^2}$$

$$\omega^2 = \frac{1}{\frac{M}{m_m}} \frac{\omega_1^2 \omega_2^2}{(Y_1^2 \omega_2^2 + Y_2^2 \omega_1^2)}$$

$$\omega_1^2 = \frac{\mathcal{P}}{S} \frac{\pi^2}{L^2}$$

$$\omega_2^2 = \frac{\mathcal{P}}{S} \frac{4\pi^2}{L^2}$$

$$Y_1\left(\frac{L}{4}\right) = \sin \frac{\pi}{4} = 0.71$$

$$m_m = \frac{8L}{2}$$

$$Y_2\left(\frac{L}{4}\right) = \sin 2\frac{\pi}{4} = 1.0$$

(see soln to Prob 10.13)

Then

$$\omega^2 = \frac{1}{\frac{M}{m}} \frac{8L}{2} \left(\frac{\mathcal{P}}{S}\right)^2 4 \frac{\pi^4}{L^4} \frac{1}{(0.71)^2 \frac{\mathcal{P}}{S} \frac{4\pi^2}{L^2} + \frac{\mathcal{P}}{S} \frac{\pi^2}{L^2}}$$

$$= \frac{1}{m} \cancel{2.84} \frac{\mathcal{P}}{S} \frac{\pi^2}{L^2} \frac{1}{1 + 4(0.71)^2}$$

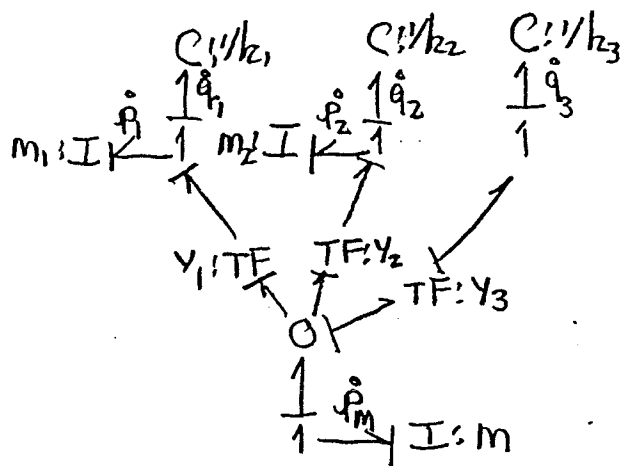
$$= 2 \frac{\mathcal{P}}{mL} \frac{\pi^2}{1 + 4(0.71)^2} = 6.54 \frac{\mathcal{P}}{mL}$$

$$\therefore \omega = 2.56 \sqrt{\frac{\mathcal{P}}{mL}}$$

From Prob 10.14

$$\omega_n = 2.31 \sqrt{\frac{\mathcal{P}}{mL}}$$

10.16



10-19

no derivative causality

$$\dot{p}_1 = -k_1 q_1 + \frac{y_1}{y_3} k_3 q_3$$

$$\dot{p}_2 = -k_2 q_2 + \frac{y_2}{y_3} k_3 q_3$$

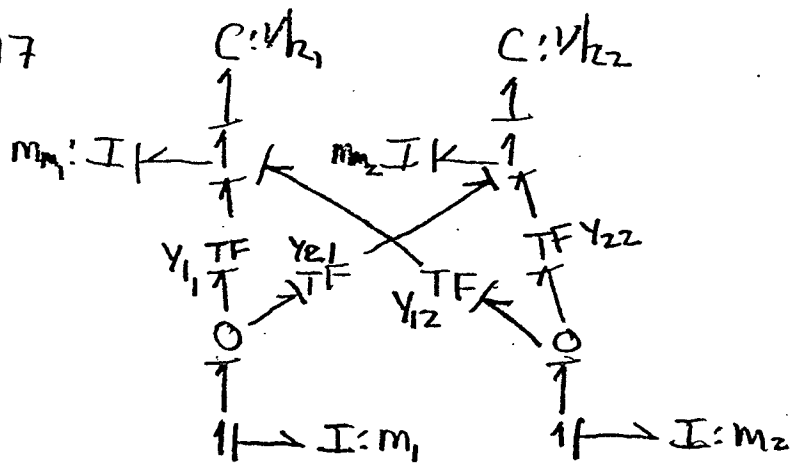
$$\dot{p}_m = -\frac{1}{y_3} k_3 q_3$$

$$\dot{q}_1 = p_1 / m_1$$

$$\dot{q}_2 = p_2 / m_2$$

$$\dot{q}_3 = \frac{1}{y_3} \left[\frac{p_m}{m} - y_1 \frac{p_1}{m_1} - y_2 \frac{p_2}{m_2} \right]$$

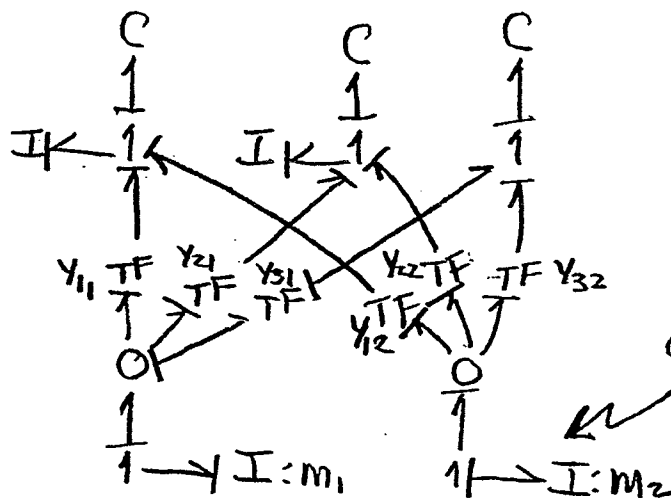
10.17



10-20

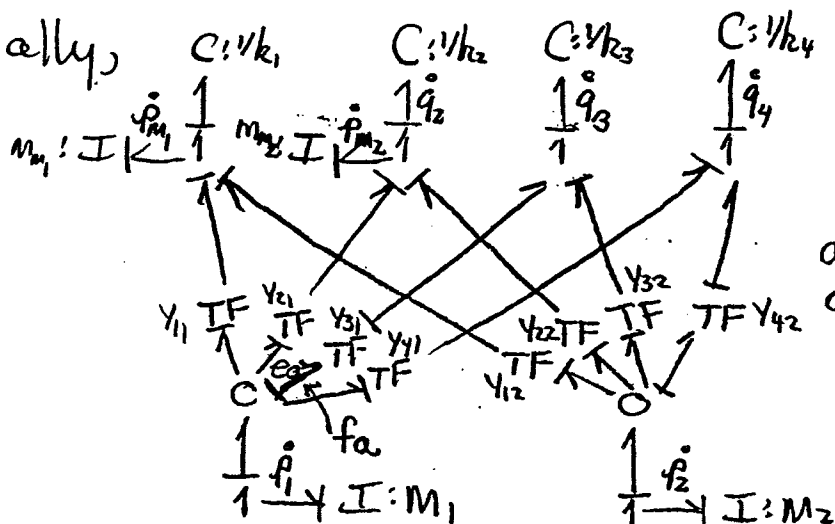
Both masses in derivative causality

Append additional modal compliance



only 1 mass in derivative causality.

Finally,



bond with e_a was arbitrarily assigned, thus, algebraic loop

10.17 (continued)

10-21

can relate e_a to itself, and f_a to itself

$$e_a = \frac{1}{Y_{31}} \left[k_3 q_3 - \frac{Y_{32}}{Y_{42}} (k_4 q_4 - Y_{41} e_a) \right] \quad f_a = \frac{f_1}{m_1} - \frac{Y_{11} f_{m1}}{m_{m1}} - \frac{Y_{21} f_{m2}}{m_{m2}} - \frac{Y_{41}}{Y_{42}} \left(\frac{f_2}{m_2} - \frac{Y_{12} f_{m1}}{m_{m1}} - \frac{Y_{22} f_{m2}}{m_{m2}} - \frac{Y_{32} f_a}{Y_{31}} \right)$$

$$e_a \left[1 - \frac{Y_{32} Y_{41}}{Y_{31} Y_{42}} \right] = \frac{1}{Y_{31}} k_3 q_3 - \frac{Y_{32}}{Y_{31} Y_{42}} k_4 q_4 \quad \text{solve for } f_a$$

Derive eqns.

$$\ddot{p}_{m1} = -k_1 q_1 + Y_{11} e_a + \frac{Y_{12}}{Y_{42}} (k_4 q_4 - Y_{41} e_a)$$

$$\ddot{p}_{m2} = -k_2 q_2 + Y_{21} e_a + \frac{Y_{22}}{Y_{42}} (k_4 q_4 - Y_{41} e_a)$$

$$\ddot{p}_1 = -e_a$$

$$\ddot{p}_2 = -\frac{1}{Y_{42}} (k_4 q_4 - Y_{41} e_a)$$

$$\ddot{q}_1 = \ddot{p}_{m1} / m_{m1}$$

$$\ddot{q}_2 = \ddot{p}_{m2} / m_{m2}$$

$$\ddot{q}_3 = \frac{1}{Y_{31}} \ddot{f}_a$$

$$\ddot{q}_4 = \frac{1}{Y_{42}} \left[\frac{f_2}{m_2} - \frac{Y_{12} f_{m1}}{m_{m1}} - \frac{Y_{22} f_{m2}}{m_{m2}} - \frac{Y_{32} f_a}{Y_{31}} \right]$$

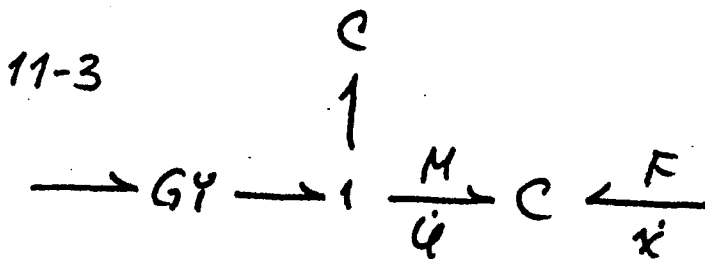
Substitute for e_a + f_a and a complete state representation exists.

$$11-1 \quad L = \frac{N^2}{R}; \quad R = \left(\frac{l}{\mu A} \right)_{\text{core}} + \left(\frac{l}{\mu_0 A} \right)_{\text{gap}} \quad 11-1$$

$$11-2 \quad L = \frac{N^2}{R_{\text{total}}} \quad \begin{array}{l} \text{at 1-junctions } R's \text{ add} \\ \text{at 0-junctions } \frac{1}{R's} = \rho's \text{ add} \end{array}$$

$$R_{\text{TOTAL}} = R_1 + R_5 + \frac{1}{\frac{1}{R_3} + \frac{1}{R_2 + R_4 + R_6}}$$

where all R 's are of the form $\left(\frac{l}{\mu A} \right)$.



In this case, $R = R(x)$ because the area for the flux changes with x rather than because the length of an air gap as in Fig. 11.9. This is a more gentle change in R so from Eq. (11.20), F will be smaller, but will act through a longer stroke.

$$R(x) \cong \frac{l_0}{\mu A(x)} = \frac{l_0}{\mu (A_0 - \frac{x}{w})}$$

for $x < w = \text{width of core.}$

11.4

$$l_g = 2 \times 10^{-3} \text{ m}$$

11-2

$$A_g = \pi \cdot 30 \times 10^{-3} \cdot 10 \times 10^{-3} \text{ m}^2$$

$$A_m = \frac{\pi (30 \times 10^{-3})^2}{4} \text{ m}^2$$

$$l_m = \frac{A_m}{A_g} l_g \cdot \frac{B_m}{\mu_0 H_m}$$

$$= \frac{\pi (30 \times 10^{-3})^2 (2 \times 10^{-3})}{4 \pi (30 \times 10^{-3}) \cdot 10 \times 10^{-3}}$$

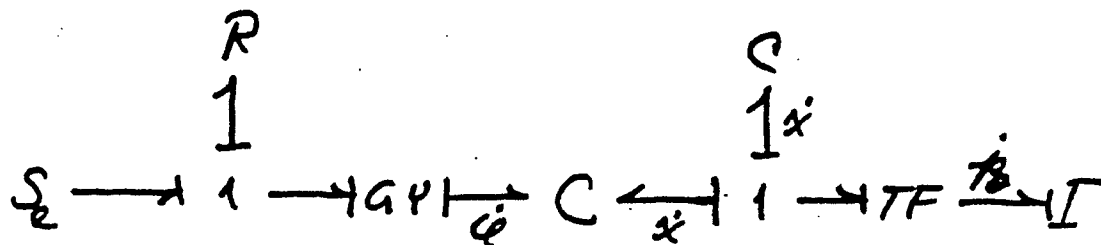
$$= \frac{6}{4} \times 10^{-3} \text{ m} = \underline{\underline{1.5 \text{ mm}}}$$

$$B_g \approx \frac{A_m}{A_g} \cdot B_m = \frac{\pi (30 \times 10^{-3})^2 (0.5)}{4 \pi (30 \times 10^{-3}) (10 \times 10^{-3})} \text{ T}$$

$$\approx \frac{3}{4} \cdot (0.5) = \underline{\underline{0.375 \text{ T}}}$$

11.5

11-3



$$\dot{\varphi} = \frac{1}{N} \left[V_0(t) - R \cdot \frac{1}{N} \left(R_0 + \frac{\kappa}{\mu_0 A} \right) \varphi \right]$$

$$\dot{x} = l p_0 / J_0$$

$$\dot{p}_0 = l \left[-k(x - x_0) - \frac{\varphi^2}{2\mu_0 A} \right]$$

11.6

$$F = \frac{\varphi^2}{2\mu_0 A}, \Delta M = NI = IR\varphi = \frac{\kappa}{\mu_0 A} \varphi$$

$$\varphi = \frac{NI\mu_0 A}{\kappa}, F(x, I) = \frac{\mu_0 N^2 A}{2} \frac{I^2}{x^2}$$

Fig. 11.10 $\frac{e}{i} \rightarrow \text{AY} \rightarrow \text{C} \leftarrow \frac{\tau}{\omega}$

No permanent magnet - only variable reluctance effect.

Fig. 11.12 $\frac{e}{i} \rightarrow \text{MAY} \rightarrow \overset{\text{C}}{\uparrow} 1 \frac{e}{\theta}$

For round rotor, variable reluctance effect would be small.

Fig. 11.13a $\frac{e}{i} \rightarrow \text{AY} \rightarrow \text{C} \leftarrow \frac{F}{V}$

No permanent magnet

Fig. 11.13b $\frac{e}{i} \rightarrow \text{AY} \xrightarrow{\frac{F}{\kappa}}$

Permanent magnet doesn't move, reluctance constant, MAY necessary only for large motions.

Fig. 11.14 $\frac{e}{i} \rightarrow \text{MAY} \rightarrow \overset{\text{C}}{\uparrow} 1 \xrightarrow{\frac{F}{\kappa}}$

Reluctance does not change much with position.
For moderate excursions MAY \rightarrow AY

$$11.8 \quad F = \frac{\phi^2}{2\mu_0 A}, \quad \phi = BA$$

11-5

$$\frac{F}{A} = \frac{B^2}{2\mu_0}$$

$$\text{if } B = 0.5, A = 10^{-4} \text{ m}^2, \mu_0 = 4\pi \times 10^{-7} \text{ Tm/A}$$

$$F = \frac{(0.5)^2 \times 10^{-4}}{2 \cdot 4\pi \times 10^{-7}} = 9.95 \text{ N}$$

$m = 1.02 \text{ kg}$ could be lifted

11.9 Maximum current

$$i_{\max} = V_{\max} / R = 10/8 = 1.25 \text{ A}$$

$$A_w = i_{\max} / J = 1.25 / 20 \times 10^6$$

$$= 6.25 \times 10^{-8} \text{ m}^2$$

$$l = \frac{R A_w}{\rho_{cu}} = \frac{8 \cdot 6.25 \times 10^{-8}}{1.72 \times 10^{-8}} = 29.07 \text{ m}$$

$$F_{\max} = B l i_{\max} = 0.5 \cdot 29.07 \cdot 1.25 = 18.17 \text{ N}$$

or $\boxed{F_m = \frac{B V^2}{R \rho_{cu} J} = 18.17 \text{ N}}$

$$11.10 \quad E = \frac{1}{2} R(\theta) \varphi^2$$

11-7

$$M_2 = \frac{\partial E}{\partial \varphi} = R(\theta) \varphi$$

$$\tau = \frac{\partial E}{\partial \theta} = \frac{1}{2} \varphi^2 \frac{dR(\theta)}{d\theta}$$

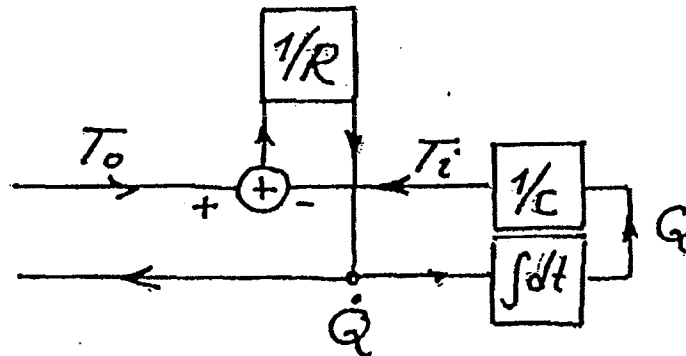
$$M_1 = -M_2 = \left(\frac{\varphi}{\varphi_0} - 1 \right) M_0 = -R(\theta) \varphi$$

$$\text{so } \varphi = \frac{M_0 \varphi_0}{\varphi_0 R(\theta) + M_0}$$

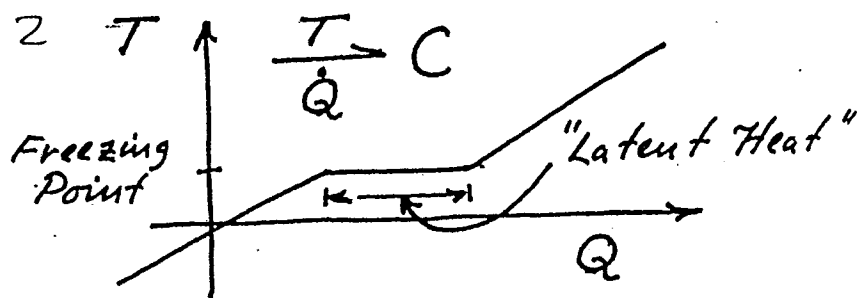
$$\tau = \frac{1}{2} \frac{M_0^2 \varphi_0^2}{(\varphi_0 R(\theta) + M_0)^2} \frac{dR(\theta)}{d\theta}$$

12-1

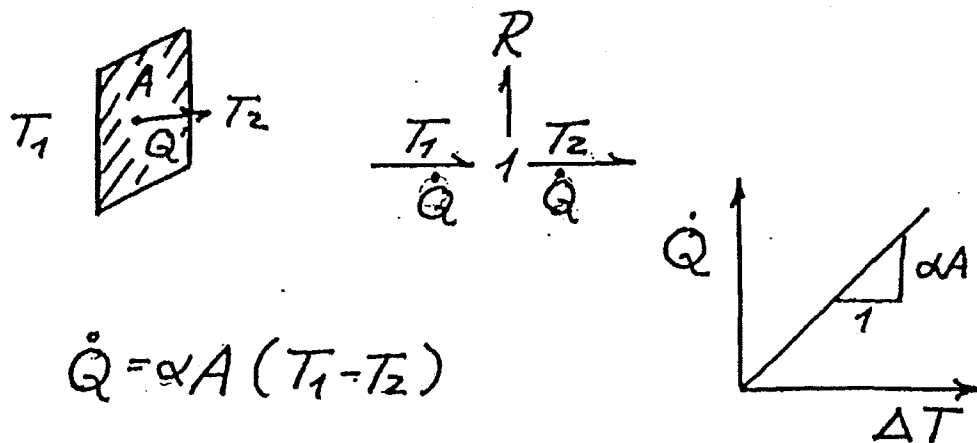
12-1



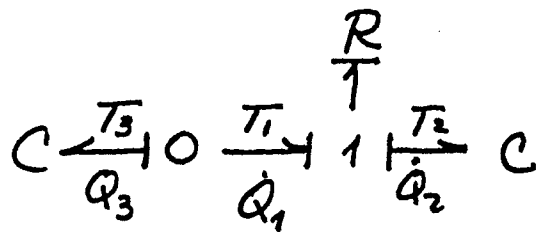
12-2



12-3



12-4



12-2

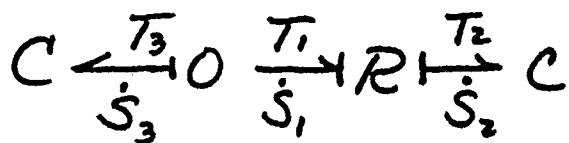
Assume $\dot{Q}_1 = \dot{Q}_2 = -\dot{Q}_3 = H(T_1 - T_2)$ as in Eq. (9.92)

$T_1 = T_3 = T_{30} + \frac{Q_3}{C_3}$; $T_2 = T_{20} + \frac{Q_2}{C_2}$ as in Eq. (9.98)

State Eqs. $\dot{Q}_2 = H(T_3 - T_2)$

$$= H\left(T_{30} + \frac{Q_3}{C_3} - T_{20} - \frac{Q_2}{C_2}\right)$$

$$\dot{Q}_3 = -H\left(T_{30} + \frac{Q_3}{C_3} - T_{20} - \frac{Q_2}{C_2}\right)$$



Corresponding assumptions: $\dot{S}_2 = \frac{(T_3 - T_2) \cdot H}{T_2}$ } Eqn. (12.35)

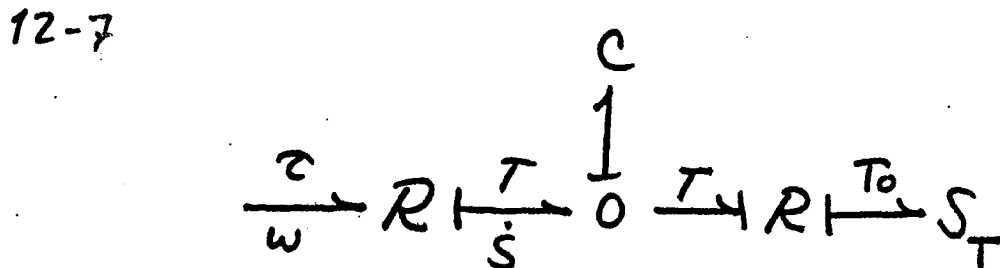
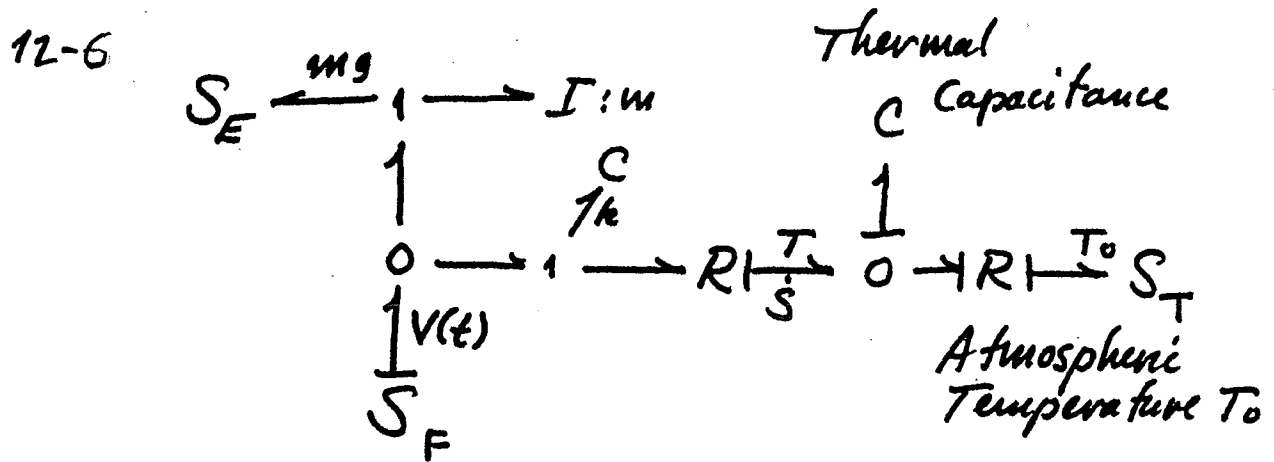
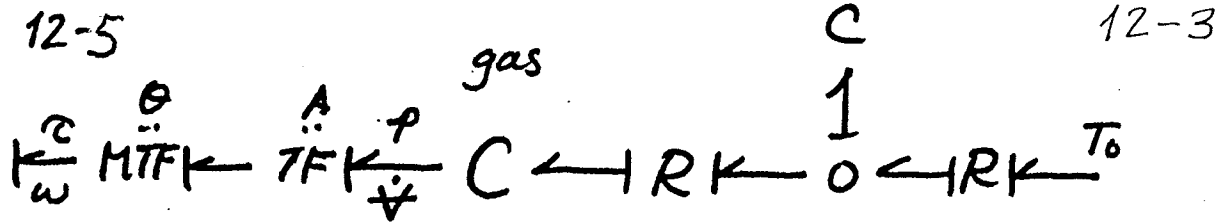
$$\left. \begin{array}{l} T_1 = T_3 = T_{30} e^{S_3/C_3} \\ T_2 = T_{20} e^{S_2/C_2} \end{array} \right\} \text{Eqn. (9.100)} \quad \dot{S}_1 = \frac{(T_3 - T_2) H}{T_1}$$

State Eqs. $\dot{S}_2 = H \frac{T_{30} e^{S_3/C_3} - T_{20} e^{S_2/C_2}}{T_{20} e^{S_2/C_3}}$

$$\dot{S}_3 = -H \frac{(T_{30} e^{S_3/C_3} - T_{20} e^{S_2/C_2})}{T_{30} e^{S_3/C_3}}$$

$$\dot{S}_2 + \dot{S}_3 = \dot{S}_2 - \dot{S}_1 = \frac{(T_2 - T_3)^2 H}{T_2 T_3} \geq 0 \text{ for any } T_2, T_3$$

Cylinder Capacitance



12-8 (a) $\rho V = \text{const.}$ $\rho \Delta V + V \Delta \rho = 0$

$$\frac{\Delta \rho}{\rho} = - \frac{\Delta V}{V} \quad \text{so} \quad \Delta p = -B \frac{\Delta V}{V}$$

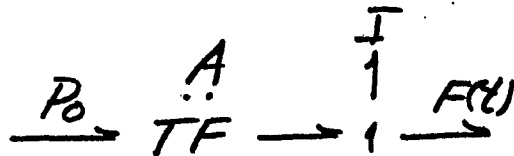
(b) $I = \rho A l$, $C = A l / B$

(c) λ should be $> l$

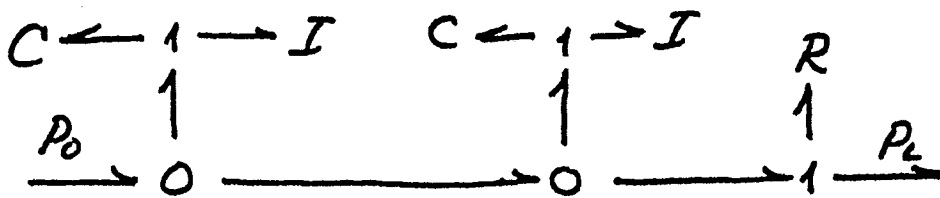
or $f = \frac{c}{\lambda}$ should be $< \frac{c}{l}$

$$\omega < \frac{2\pi c}{l}$$

12-9 As Fig. 12.10 shows the inclusion of a Bernoulli resistor with $A_2 > A_1$ for inflow will give static power recovery i.e. the pressure acting on the mass would be greater than P_0 (+). To model complete dynamic pressure loss, simply leave the Bernoulli resistor out.

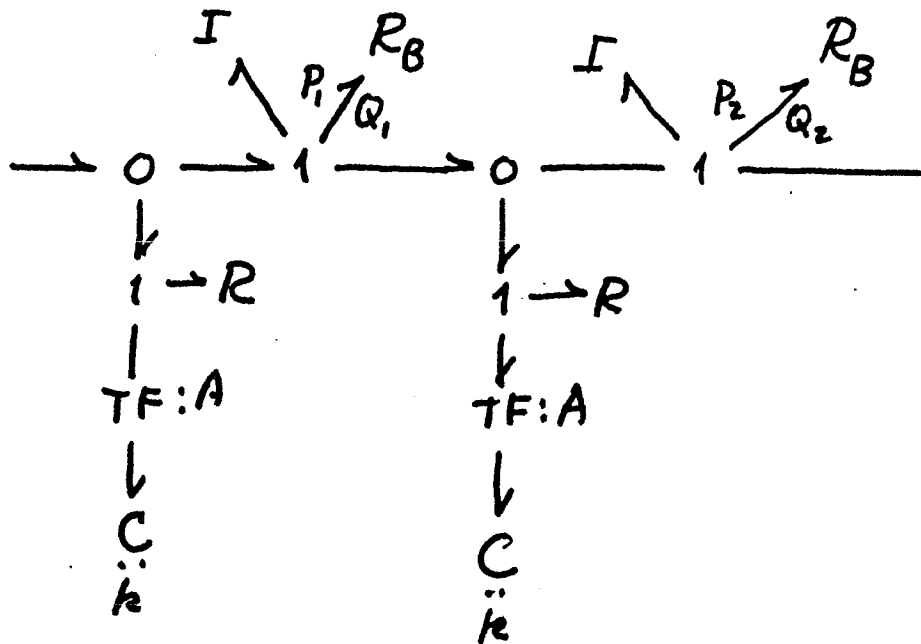


12-10



$$I_i = \frac{\rho_0 l_i}{A_i} ; \quad C_i = \frac{V_i}{B} = \frac{V_i}{\rho_0 c^2}$$

12-11

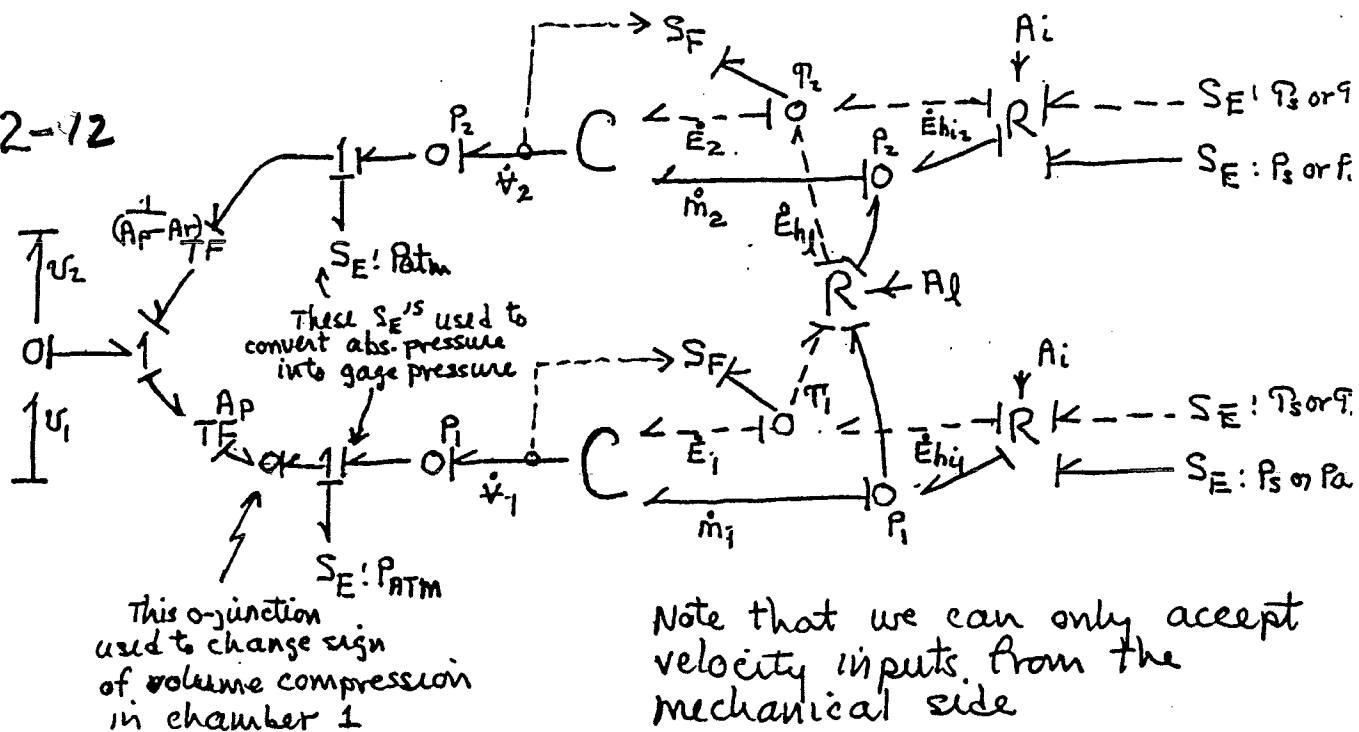


12-5

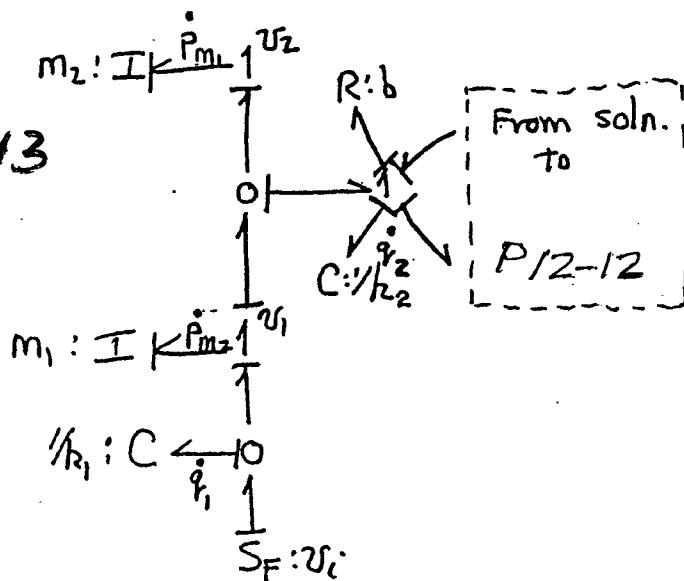
$$P_1 = \rho \frac{Q_1^2}{2} \left(\frac{1}{A_v^2} - \frac{1}{A_m^2} \right) ; P_2 = \rho \frac{Q_2^2}{2} \left(\frac{1}{A_m^2} - \frac{1}{A_v^2} \right)$$

when $Q_1 = Q_2$ $P_1 = -P_2$ and all dynamic pressure is recovered. For steady flow, the static pressure is less in the venturi than in the main pipe so the differing piston displacements are a measure of the flow.

12-12



12-13



state eqns.

$$\dot{P}_{m1} = k_2 q_2 + b \left(\frac{P_{m1}}{m_1} - \frac{P_{m2}}{m_2} \right) - (P_2 - P_{atm})(A_P - A_R) + (P_1 - P_{atm}) A_P$$

$$\dot{P}_{m2} = k_1 q_1 - k_2 q_2 - b \left(\frac{P_{m1}}{m_1} - \frac{P_{m2}}{m_2} \right) + (P_2 - P_{atm})(A_P - A_R) - (P_1 - P_{atm}) A_P$$

$$\dot{q}_1 = u_1 - P_{m2}/m_2$$

$$\dot{q}_2 = \frac{P_{m2}}{m_2} - \frac{P_{m1}}{m_1}$$

where

$$P_2 = \frac{R}{C_v} \left(\frac{E_2}{V_2} \right)$$

12-7

$$P_1 = \frac{R}{C_v} \left(\frac{E_1}{V_1} \right)$$

and

$$\dot{V}_1 = -A_P \left(\frac{P_{M1}}{M_1} - \frac{P_{M2}}{M_2} \right)$$

$$\dot{V}_2 = (A_P - A_L) \left(\frac{P_{M1}}{M_1} - \frac{P_{M2}}{M_2} \right)$$

$$\dot{E}_1 = -P_1 \dot{V}_1 + \dot{E}_{hi1} - \dot{E}_{hl}$$

$$\dot{E}_2 = -P_2 \dot{V}_2 + \dot{E}_{hl} + \dot{E}_{hi2}$$

where \dot{E}_{hi1} and \dot{E}_{hi2} come from isentropic nozzle relations for intake
and \dot{E}_{hl} comes from isentropic nozzle, 4 port R-element for leakage.

Isentropic nozzle relations need P_1 and P_2 where

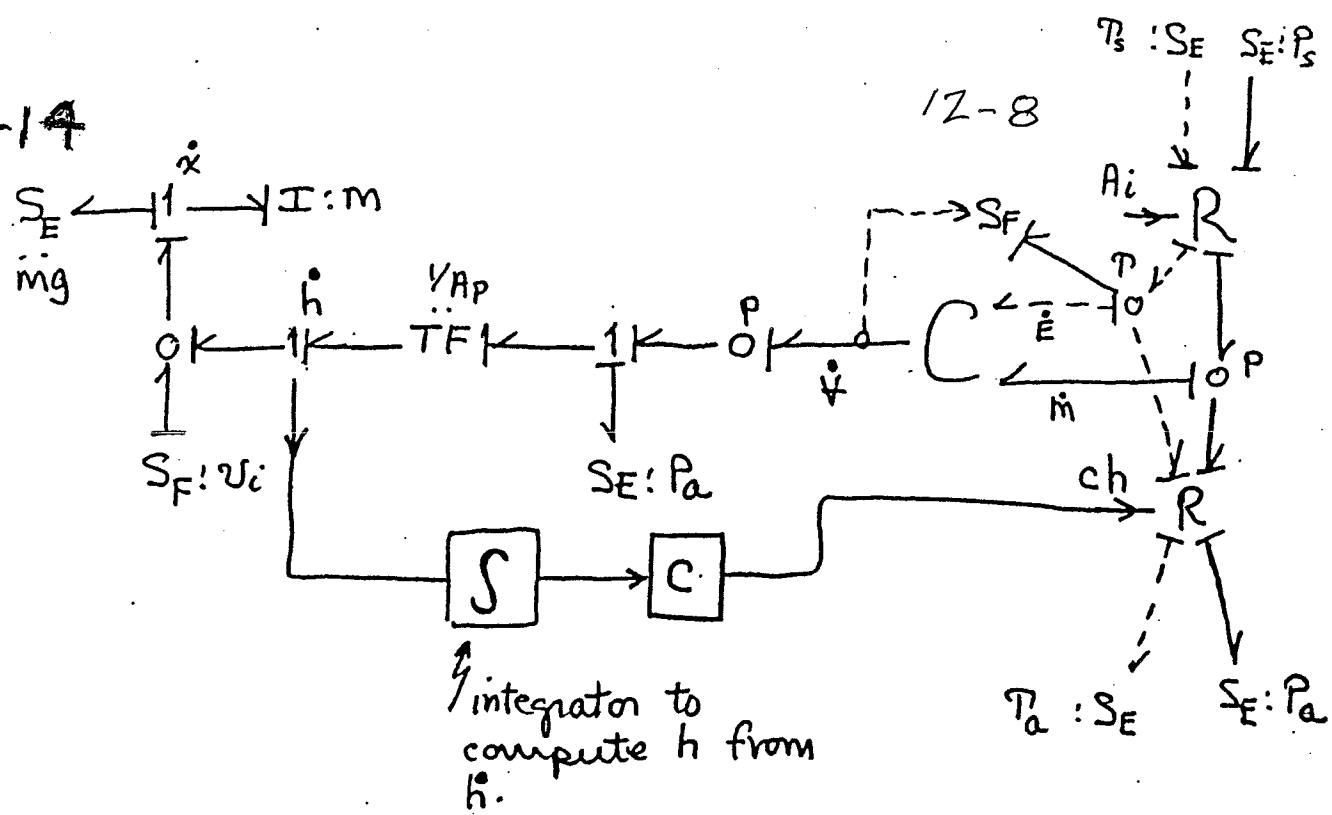
$$P_1 = \frac{1}{C_v} \frac{E_1}{M_1}$$

$$P_2 = \frac{1}{C_v} \frac{E_2}{M_2}$$

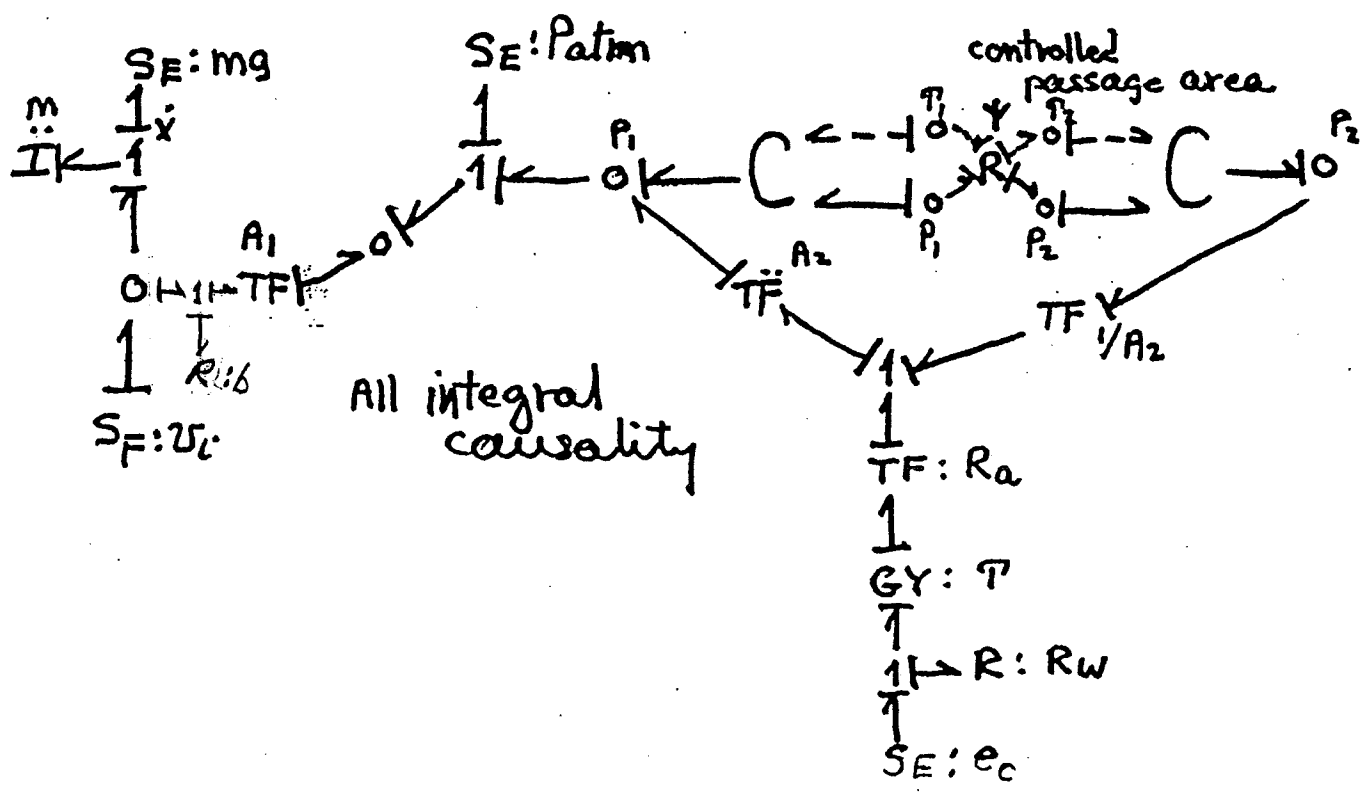
and $\dot{M}_1 = [\text{from 4 port R for intake 1}]$

$\dot{M}_2 = [\text{from 4 port R for intake 2}]$

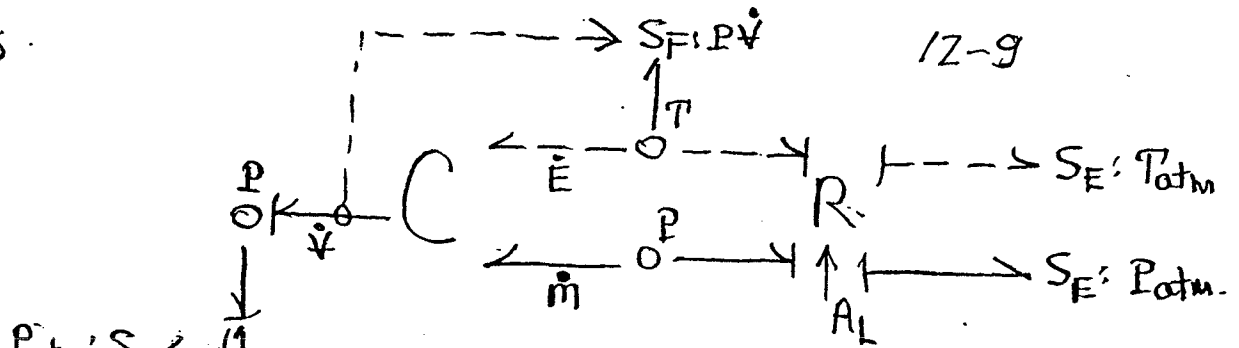
12-14



12-15



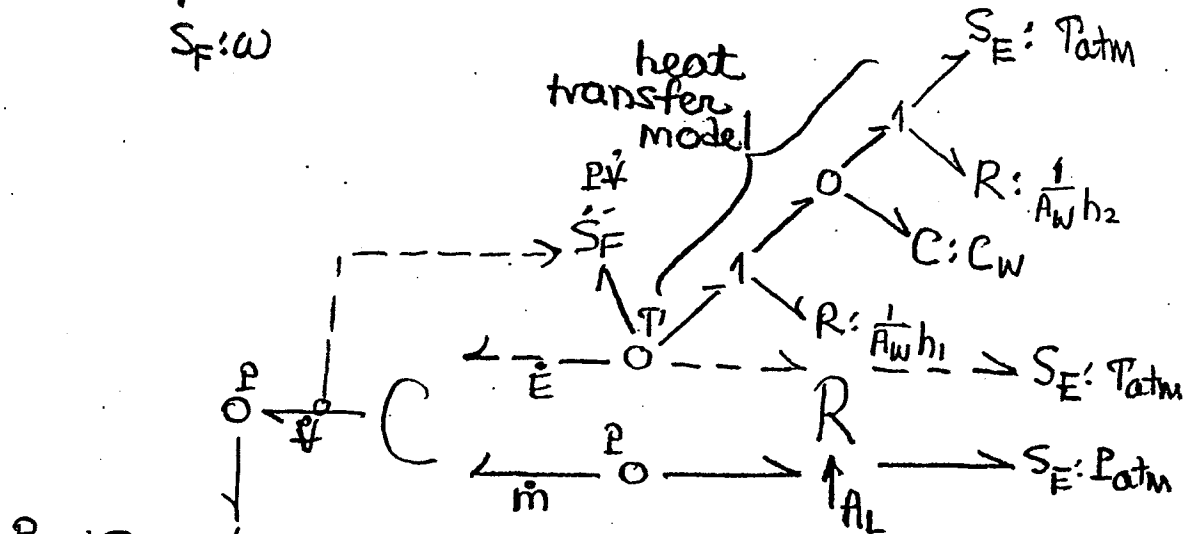
12-16



T and P are set by the constitutive laws, Eqs. (12.81)(12.82)

m(θ) is the modulating function from sect. 8.1

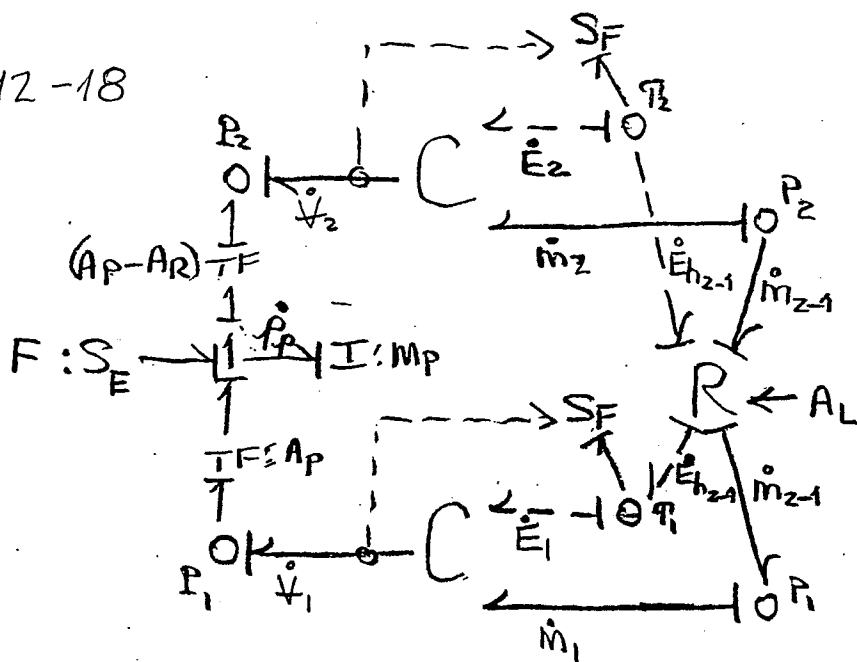
12-17



A_w ≡ cylinder wall area

12-18

12-10



state variables

 $E_1, m_1, v_1, E_2, m_2, v_2, p_p$

$$\dot{E}_1 = -P_1 \dot{v}_1 + \dot{E}_{h2-1}$$

$$\dot{m}_1 = \dot{m}_{2-1}$$

$$\dot{v}_1 = A_p \frac{p_p}{m_p}$$

$$\dot{E}_2 = -P_2 \dot{v}_2 - \dot{E}_{h2-1}$$

$$\dot{m}_2 = -\dot{m}_{2-1}$$

$$\dot{v}_2 = -(A_p - A_r) \frac{p_p}{m_p}$$

$$\dot{E}_{h2-1} = \dot{E}_{h2-1}(T_1, P_1, T_2, P_2) \left\{ \text{see} \right.$$

$$\dot{m}_{h2-1} = \dot{m}_{h2-1}(T_1, P_1, T_2, P_2) \left\{ \text{sect. 12.5.2} \right.$$

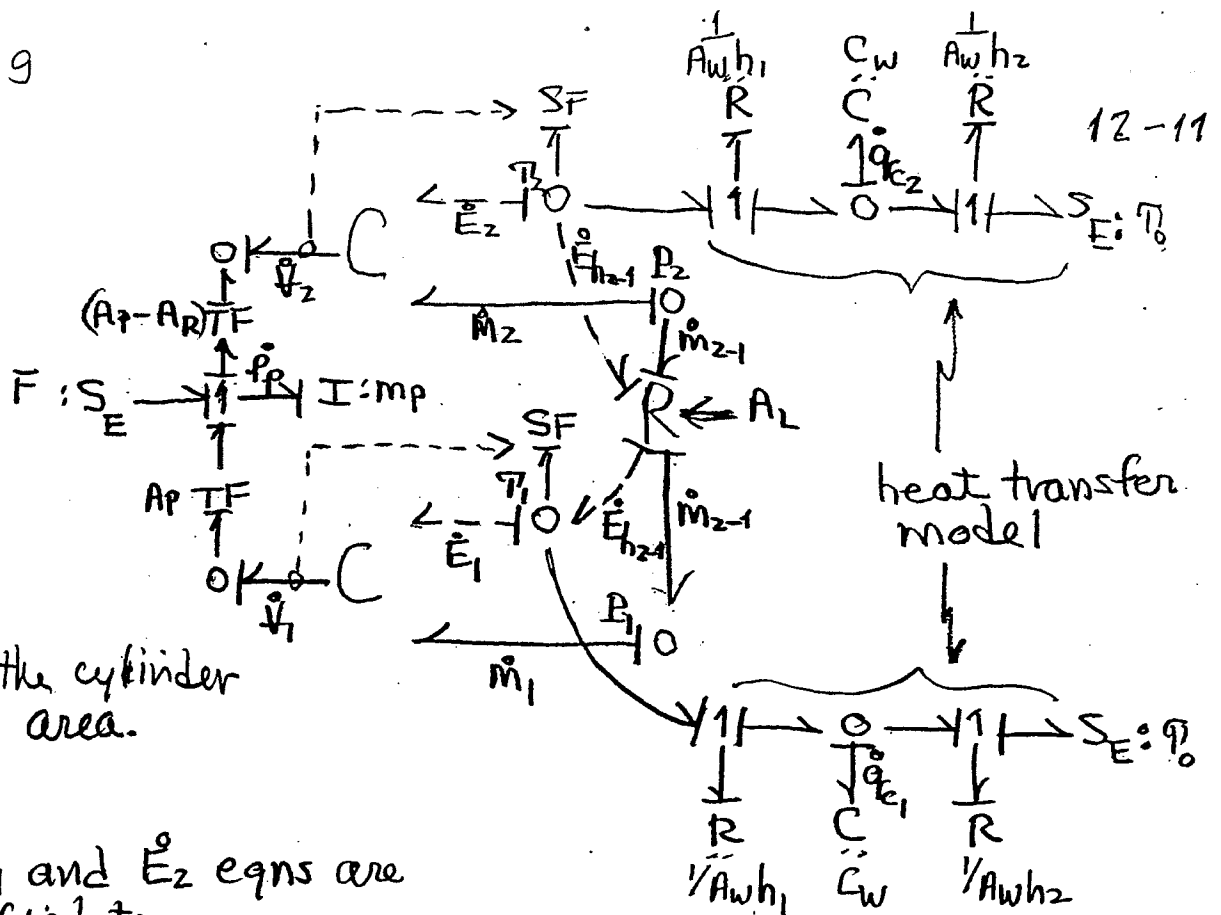
$$T_1 = \frac{1}{C_v} \frac{E_1}{m_1}$$

$$P_1 = \frac{R}{C_v} \frac{E_1}{v_1}$$

$$T_2 = \frac{1}{C_v} \frac{E_2}{m_2}$$

$$P_2 = \frac{R}{C_v} \frac{E_2}{v_2} \left\{ \begin{array}{l} \text{Eqs. (12.78)} \\ \text{and (12.79)} \end{array} \right.$$

12-19



The \dot{E}_1 and \dot{E}_2 eqns are modified to,

$$\dot{E}_1 = -P_1 \dot{V}_1 + \dot{E}_{h_{2-1}} - A_w h_1 (T_1 - \frac{q_{c1}}{C_w})$$

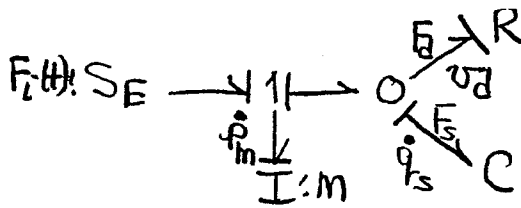
$$\dot{E}_2 = -P_2 \dot{V}_2 - \dot{E}_{h_{2-1}} - A_w h_1 (T_2 - \frac{q_{c2}}{C_w})$$

plus new state eqns.,

$$\dot{q}_{c1} = A_w h_1 (T_1 - \frac{q_{c1}}{C_w}) - A_w h_2 (\frac{q_{c1}}{C_w} - T_0)$$

$$\dot{q}_{c2} = A_w h_1 (T_2 - \frac{q_{c2}}{C_w}) - A_w h_2 (\frac{q_{c2}}{C_w} - T_0)$$

13.1



state variables
 p_m, q_s

13-1

$$\dot{p}_m = F_i - F_s$$

$$F_s = g_s q_s^3$$

$$\dot{q}_s = \frac{p_m}{m} - v_d$$

$$v_d = \left\{ \frac{|F_d|}{g_d} \right\}^{1/3} \cdot \text{sign}(F_d)$$

$$F_d = F_s$$

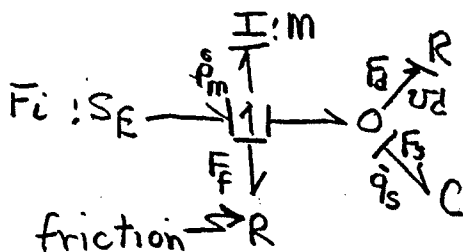
Note that causality on $\rightarrow R$ requires v_d as output and F_d as input. Also, care must be used in inverting the damper constitutive law.

13.2

For the spring, $F_s = g_s q_s |q_s|$ to handle sign of F_s

For the damper, $v_d = \left\{ \frac{|F_d|}{g_d} \right\}^{1/2} \cdot \text{sign}(F_d)$

13.3



$$\dot{p}_m = F_i - F_f - F_s$$

$$F_s = g_s q_s^3$$

$$\dot{q}_s = \frac{p_m}{m} - v_d$$

$$F_d = F_s, \quad v_d = \left\{ \frac{|F_d|}{g_d} \right\}^{1/3} \cdot \text{sign}(F_d)$$

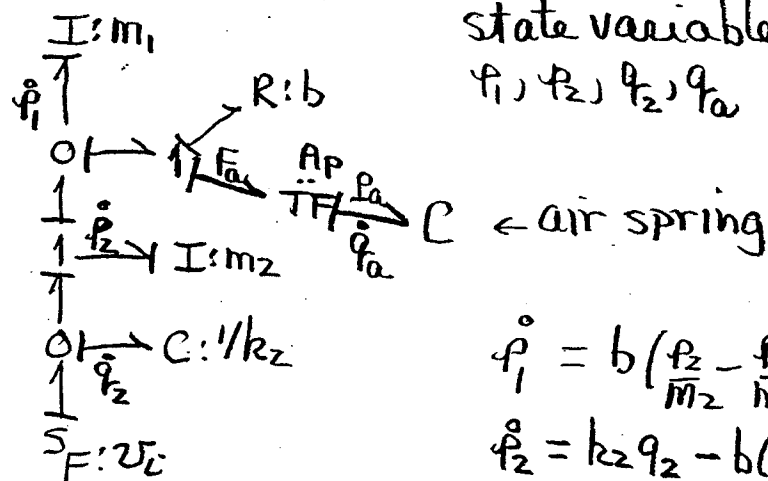
For friction characterized as $F_f = \mu N$ where μ = friction coefficient and N is the weight of m , it is usually sufficient to use,

$$F_f = \mu N \frac{v_m}{|v_m| + \epsilon}$$

$$v_m = p_m/m$$

ϵ is a "small" number ($10^{-4}, 10^{-5}$) which allows $v_m = 0$

13.4



For the air spring

$$\delta = \delta_{in} + q_a$$

$$\text{if } \delta \leq 0 \quad \delta = 0$$

$$F_a = P_0 A_p \left[\frac{1 - \left(\frac{A_p \delta}{V_0} \right)^\gamma}{\left(\frac{A_p \delta}{V_0} + \epsilon \right)^\gamma} \right]$$

ϵ is small number ($10^{-4}, 10^{-5}$) that allows for $\delta = 0$

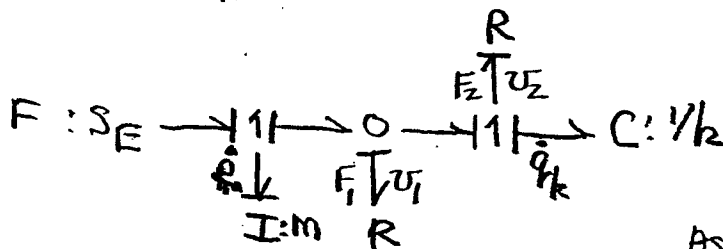
$$\dot{p}_1 = b \left(\frac{p_2}{m_2} - \frac{p_1}{m_1} \right) + F_a$$

$$\dot{p}_2 = k_2 q_2 - b \left(\frac{p_2}{m_2} - \frac{p_1}{m_1} \right) - F_a$$

$$\dot{q}_2 = v_2 - p_2/m_2$$

$$\dot{q}_a = A_p \left[\frac{p_2}{m_2} - \frac{p_1}{m_1} \right]$$

13.5



Incomplete causality indicates algebraic loop. Causality was completed with the result shown.

As indicated by causality,

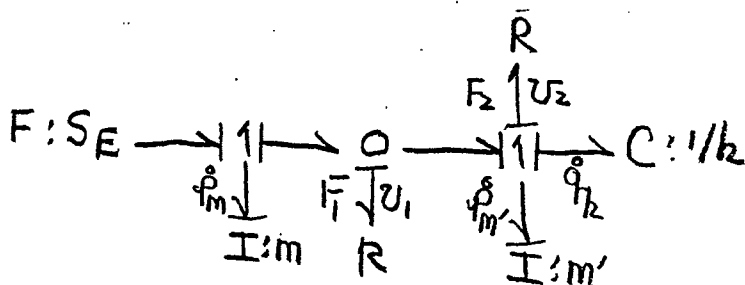
$$F_1 = g_1 v_1^3, \quad v_2 = \left\{ \frac{|F_2|}{g_2} \right\}^{1/2} \text{sign}(F_2)$$

Algebraic loop involves,

$$F_1 = g_1 \left[\frac{q_m}{m} - v_2 \right]^3$$

$$v_2 = \left\{ \frac{|F_1 - k q_k|}{g_2} \right\}^{1/2} \text{sign}(F_1 - k q_k)$$

← complicated to solve for F_1 or v_2

Introduce some mass, m' , at the junction of the dampers,

no algebraic loop

13.5 (continued)

$$\dot{p}_m = F - F_1$$

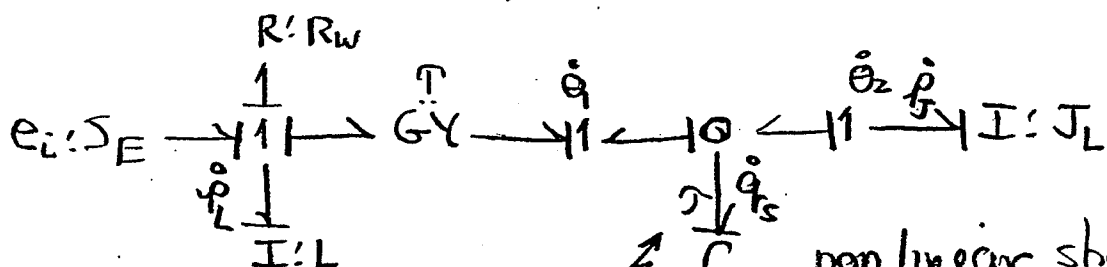
$$\dot{q}_k = \frac{p_{m'}}{m'}$$

$$\dot{p}_{m'} = -k q_k + F_1 - F_2$$

$$F_1 = g_1 \left(\frac{p_m}{m} - \frac{p_{m'}}{m'} \right)^3 \quad 13-3$$

$$F_2 = g_2 \underbrace{\frac{p_{m'}}{m'}}_{v_2} \left| \frac{p_{m'}}{m'} \right|$$

13.6



state vbls, p_L, p_J

derivative causality
non linear shaft

procedure from chap 5 suggests

$$q_s = \left\{ \frac{1}{g} \right\}^{1/3} \text{sign}(\tau)$$

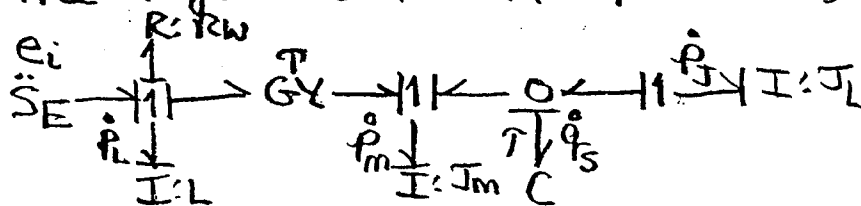
$$\tau = (-1)^n \frac{p_L}{L}$$

$$\dot{p}_L = e_i - \frac{R_w}{L} p_L - \tau \left(\frac{p_J}{J_L} - \dot{q}_s \right)$$

$$\dot{p}_J = \frac{\tau}{L} p_L$$

we need \dot{q}_s which comes from differentiating q_s

This could be done for this problem, but instead, append the rotary inertia of the motor, J_m , to the 1-junction with $\dot{\theta}_1$. Yields,



$$\dot{p}_L = e_i - \frac{R_w}{L} p_L - \frac{\tau}{J_m} p_m$$

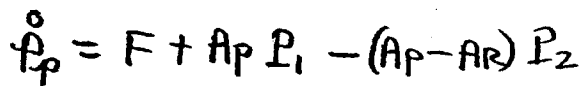
$$\dot{p}_J = (-1)^n \tau$$

$$\dot{p}_m = \frac{\tau}{L} p_L + \tau$$

$$\dot{q}_s = \frac{p_J}{J_L} - \frac{p_m}{J_m}$$

$$\tau = g q_s^3$$

13-4

$$p_p, v_1, v_2, E_2, m_2, E_1, m_1$$


$$\dot{E}_2 = -P_2 \dot{V}_2 - \dot{E}_{h_{2,1}}$$

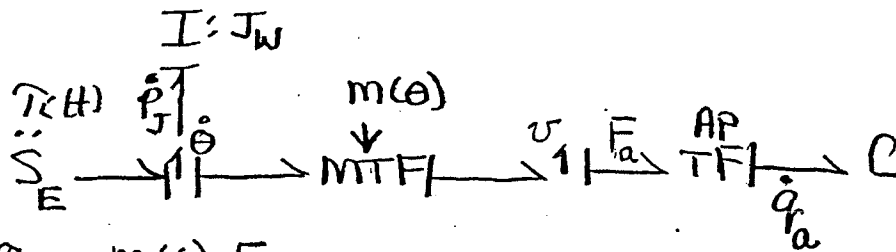
$$\dot{E}_1 = -P_1 \dot{V}_1 + \dot{E}_{h2,1}$$

$\dot{m}_{z,1}$, $\dot{E}_{h,z-1}$ are dependent upon $P_1, \dot{Q}_1, P_2, \dot{Q}_2$, and are calculated as shown in sect. 12.4.2

P_1, Q_1, P_2, Q_2 come from constitutive laws, Eqs. (12.81) and (12.82).

13.8

13-5



$$\dot{p}_J = \tau_c - m(\theta) F_a$$

$$\dot{q}_a = A_P m(\theta) \frac{\dot{p}_J}{J_W}$$

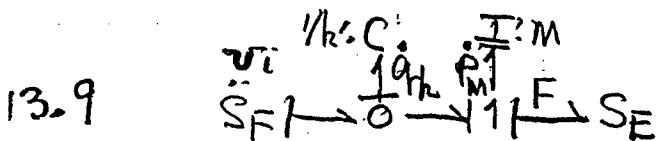
$m(\theta)$ comes from eq. 13.59, but we need θ . Expand the state space with,

$$\dot{\theta} = \frac{\dot{p}_J}{J_W}$$

F_a comes from prob. 13.4. Since q_a is positive if the air volume is decreasing, it is convenient to write,

$$\dot{s} = -\dot{q}_a.$$

This way we can assign an independent initial condition to s , and use the formula from prob 13.4.



$$\dot{p}_M = k q_h - F$$

$$\dot{q}_h = v_i - \frac{\dot{p}_M}{m}$$

Let v be the velocity just before the wall is reached. The momentum at that instant is $p_i = mv$. We desire the momentum just after impact to be $p_f = -mv$. Therefore the $\Delta p = p_f - p_i = -2mv$. And the force required is

$$F = -\frac{2mv}{\Delta t}$$

13-9 (continued)

13-6

For the sign of the force used in the bond graph and resulting equations,

$$F = \frac{2m\dot{x}}{\Delta t} = \frac{2\dot{p}_m}{\Delta t}$$

We need to bookkeep the location of the mass, so we expand the state space with,

$$\dot{x} = \frac{\dot{p}_m}{m},$$

then add,

$$\text{IF } x < d \quad KK=0, F=0$$

$$\text{IF } x > d \text{ and } KK=0 \text{ then}$$

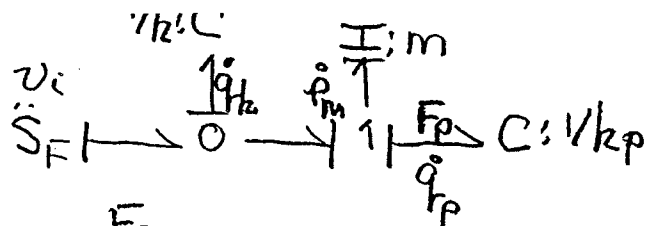
$$p_{m_0} = p_m, t_0 = t, KK=1, F = 2\dot{p}_{m_0}/\Delta t$$

(else

$$\text{IF } t < t_0 + \Delta t \quad F = 2\dot{p}_{m_0}/\Delta t$$

These steps will apply a force for almost exactly Δt seconds or until the mass is moving away from the wall.

13.10



13-7

$$\dot{p}_m = k q_k - F_p$$

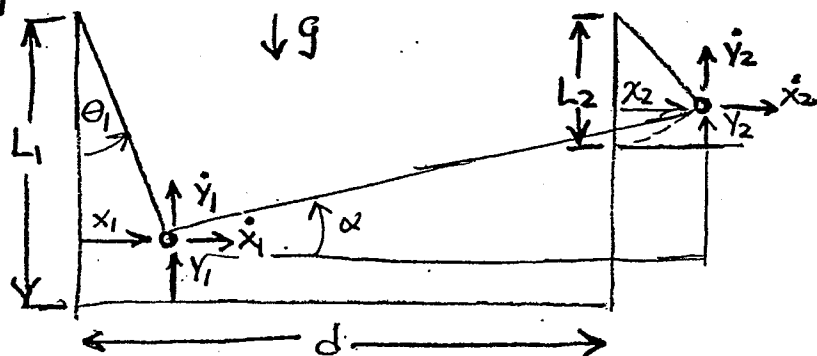
$$\dot{q}_k = v_c - \frac{p_m}{m}$$

$$\dot{q}_p = \frac{p_m}{m}$$

then add

$$\text{if } q_p > d \quad F_p = k_p(q_p - d)$$

13.11



some kinematics

$$\dot{x}_1 = L_1 \dot{\theta}_1 \cos \theta_1$$

$$\dot{y}_1 = L_1 \dot{\theta}_1 \sin \theta_1$$

$$\dot{x}_2 = L_2 \dot{\theta}_2 \cos \theta_2$$

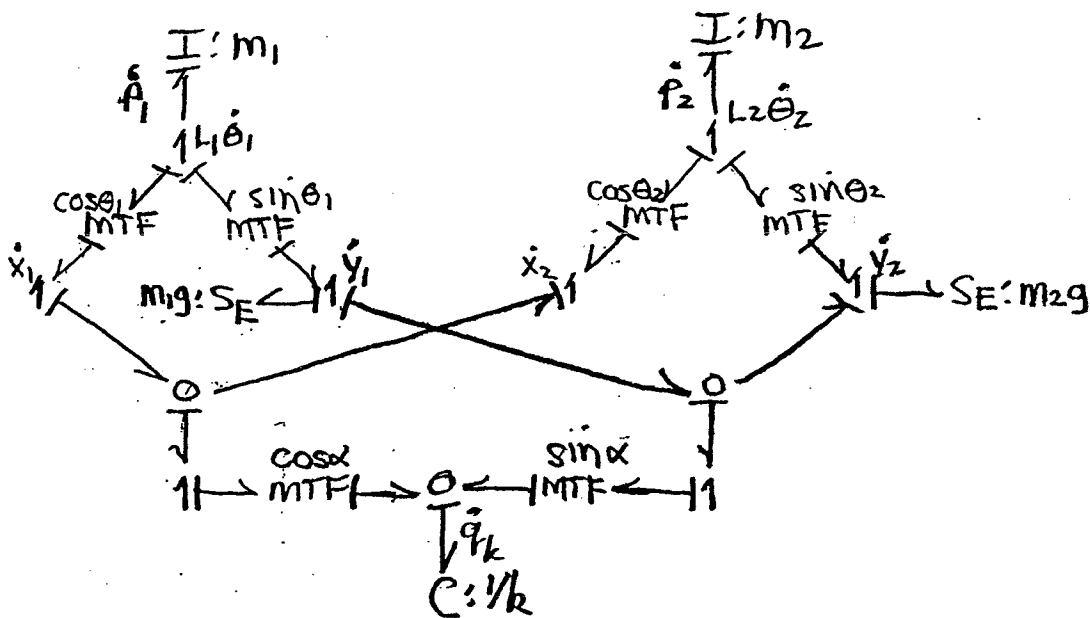
$$\dot{y}_2 = L_2 \dot{\theta}_2 \sin \theta_2$$

$$\tan \alpha = \frac{L_1 - L_2 + y_2 - y_1}{d + x_2 - x_1}$$

$$v_s \equiv \text{relative velocity across spring} = (\dot{x}_1 - \dot{x}_2) \cos \alpha + (\dot{y}_1 - \dot{y}_2) \sin \alpha$$

13.11 (continued)

13-8



$$\dot{p}_1 = -\cos\theta_1 \cos\alpha k q_k - \sin\theta_1 [m_1 g + \sin\alpha k q_k]$$

$$\dot{p}_2 = +\cos\theta_2 \cos\alpha k q_k - \sin\theta_2 [m_2 g - \sin\alpha k q_k]$$

$$\dot{q}_k = \cos\alpha \left[\cos\theta_1 \frac{p_1}{m_1} - \cos\theta_2 \frac{p_2}{m_2} \right] + \sin\alpha \left[\sin\theta_1 \frac{p_1}{m_1} - \sin\theta_2 \frac{p_2}{m_2} \right]$$

plus the additional eqns,

$$\dot{\theta}_1 = \frac{1}{L_1} \frac{p_1}{m_1} \quad \dot{\theta}_2 = \frac{1}{L_2} \frac{p_2}{m_2}$$

$$\dot{x}_1 = \cos\theta_1 \frac{p_1}{m_1}$$

$$\dot{x}_2 = \cos\theta_2 \frac{p_2}{m_2}$$

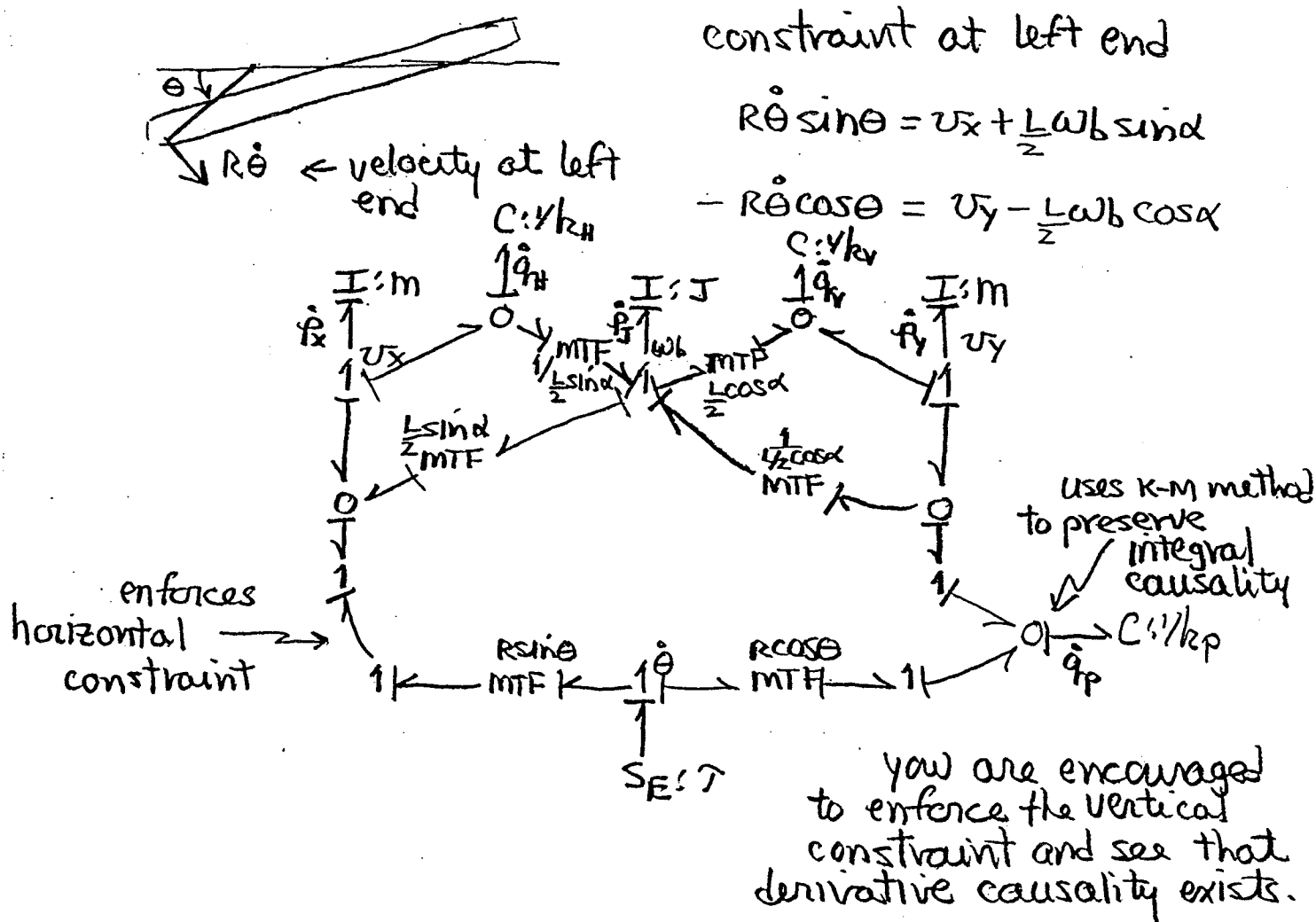
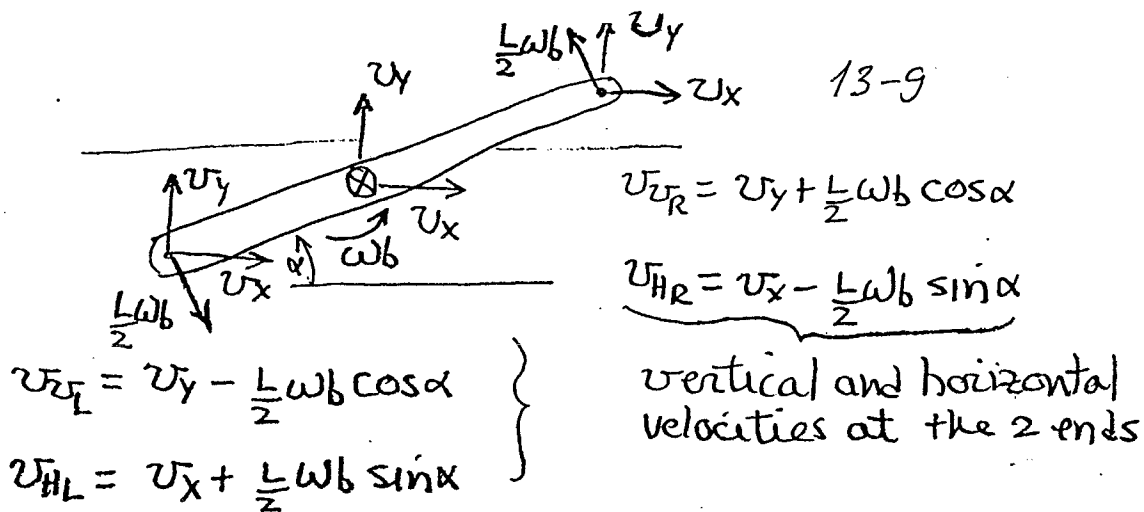
$$\dot{y}_1 = \sin\theta_1 \frac{p_1}{m_1}$$

$$\dot{y}_2 = \sin\theta_2 \frac{p_2}{m_2}$$

These provide availability of $\theta_1, \theta_2, x_1, y_1, x_2, y_2$

$$\alpha = \tan^{-1} \left(\frac{L_1 - L_2 + y_2 - y_1}{d + x_2 - x_1} \right)$$

13.12



13.12 (continued)

13-10

$$\dot{p}_x = -k_H q_H + \frac{1}{R \sin \theta} (\tau - R \cos \theta k_P q_P)$$

$$\dot{p}_y = -k_V q_V - k_P q_P$$

$$\dot{p}_J = \frac{L}{2} \sin \alpha k_H q_H - \frac{L}{2} \cos \alpha k_V q_V + \frac{L}{2} \sin \alpha \frac{1}{R \sin \theta} (\tau - R \cos \theta k_P q_P)$$

$$\dot{q}_H = \frac{p_x}{m} - \frac{L}{2} \sin \alpha \frac{p_J}{J}$$

$$\dot{q}_V = \frac{L}{2} \cos \alpha \frac{p_J}{J} + \frac{p_y}{m}$$

$$\dot{q}_P = \frac{p_y}{m} - \frac{L}{2} \cos \alpha \frac{p_J}{J} + \frac{R \cos \theta}{R \sin \theta} \left(\frac{p_x}{m} + \frac{L}{2} \sin \alpha \frac{p_J}{J} \right)$$

We also need

$$\dot{\theta} = \frac{1}{R \sin \theta} \left(\frac{p_x}{m} + \frac{L}{2} \sin \alpha \frac{p_J}{J} \right)$$

$$\dot{\alpha} = \frac{p_J}{J}$$

This is a complete state representation. The parameters k_P and k_V must be "stiff" to simulate the original slider-crank device.

Also, care must be taken during simulation to avoid $\theta = 0$ (exactly). Notice that this formulation requires division by $\sin \theta$. This can be avoided by using the K-M method to relieve the horizontal constraint at the left end.