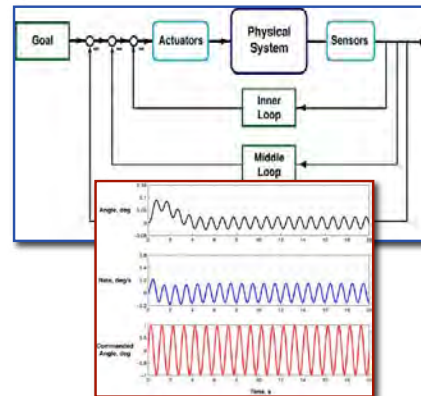


Dynamic Effects of Feedback Control

Robert Stengel

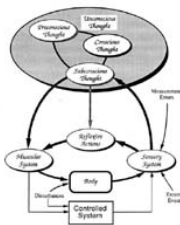
Robotics and Intelligent Systems MAE 345,
Princeton University, 2015

- Inner, Middle, and Outer Feedback Control Loops
- Step Response of Linear, Time-Invariant (LTI) Systems
- Position and Rate Control
- Transient and Steady-State Response to Sinusoidal Inputs



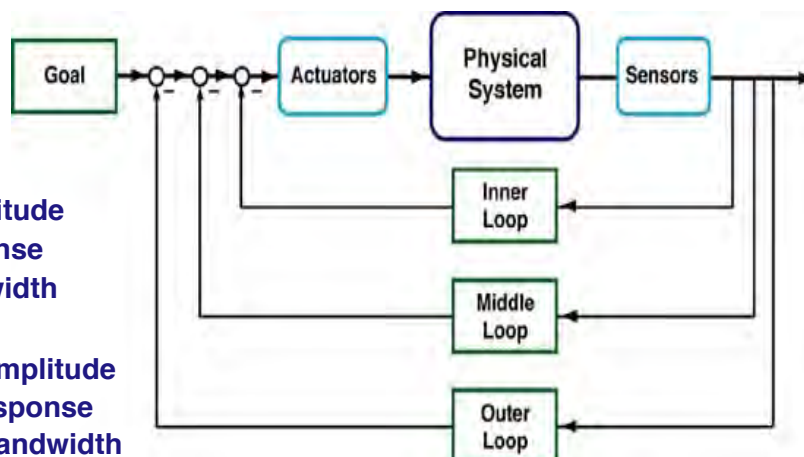
Copyright 2015 by Robert Stengel. All rights reserved. For educational use only.
<http://www.princeton.edu/~stengel/MAE345.html>

1



Outer-to-Inner-Loop Control Hierarchy

- **Inner Loop**
 - Small Amplitude
 - Fast Response
 - High Bandwidth
- **Middle Loop**
 - Moderate Amplitude
 - Medium Response
 - Moderate Bandwidth
- **Outer Loop**
 - Large Amplitude
 - Slow Response
 - Low Bandwidth



- **Feedback**
 - Error between command and feedback signal drives next inner-most loop

2

Natural Feedback Control

Inner Loop

Chicken Head Control - 1

http://www.youtube.com/watch?v=_dPlkFPowCc



Middle Loop

Hovering Red-Tail Hawks

<http://www.youtube.com/watch?v=-VPVZMSEvwU>



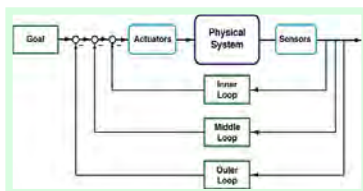
Outer Loop

Osprey Diving for Fish

<http://www.youtube.com/watch?v=qrgpl9-N6jY>



3



Outer-to-Inner-Loop Control Hierarchy of an Industrial Robot

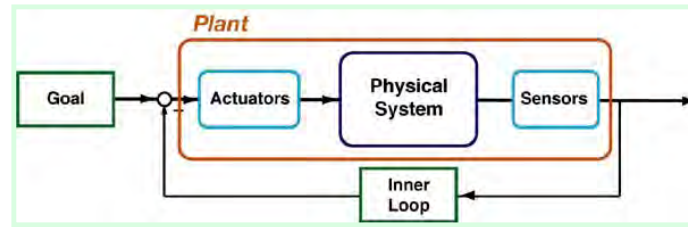
- **Inner Loop**
 - Focus on control of individual joints
- **Middle Loop**
 - Focus on operation of the robot
- **Outer Loop**
 - Focus on goals for robot operation



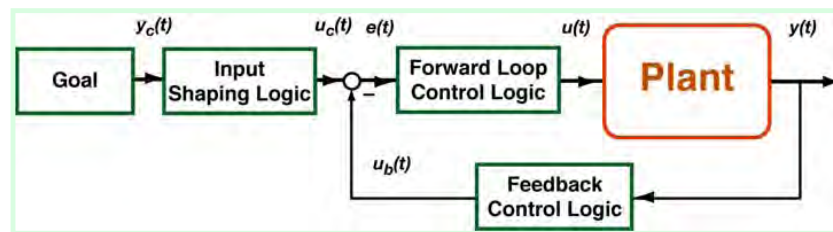
4

Inner-Loop Feedback Control

Feedback control design must account for actuator-system-sensor dynamics

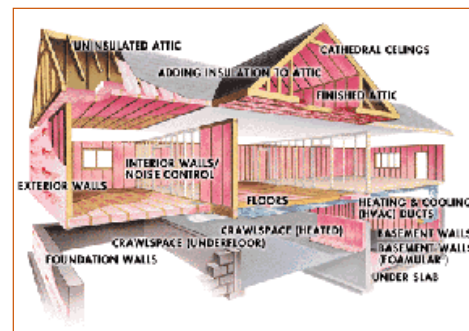
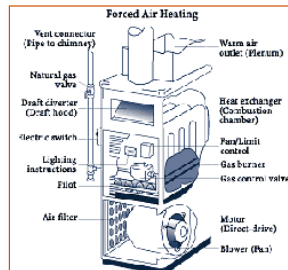
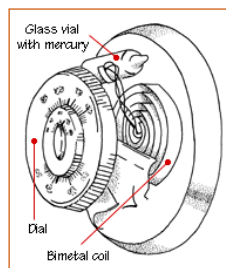


Single-Input/Single-Output Example, with forward and feedback control logic ("compensation")



5

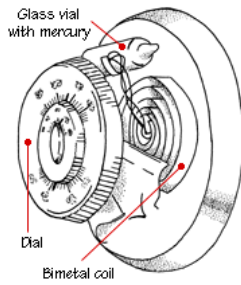
Thermostatic Temperature Control



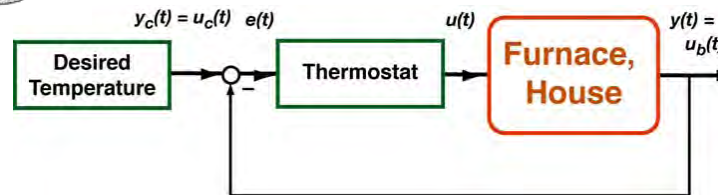
- **Dynamics**
 - Delays
 - Dead Zones
 - Saturation
 - Coupling
- **Structure**
 - Layout
 - Insulation
 - Circulation
 - Multiple Spaces
- **External Effects**
 - Solar Radiation
 - Air Temperature
 - Wind
 - Rain, Humidity

... all controlled by a simple (but nonlinear) on/off switch

6



Thermostat Control Logic



- **Control Law** [i.e., logic that drives the control variable, $u(t)$]

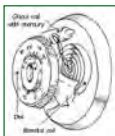
$$e(t) = y_c(t) - y(t) = u_c(t) - u_b(t)$$

< Thermostat >

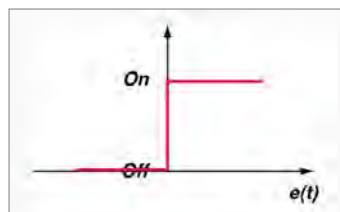
$$u(t) = \begin{cases} 1 \text{ (on)}, & e(t) > 0 \\ 0 \text{ (off)}, & e(t) \leq 0 \end{cases}$$

- y_c : Desired output variable (command)
- y : Actual output
- u : Control variable (forcing function)
- e : Control error

7



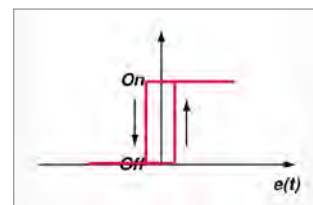
Thermostat Control Logic



$$u(t) = \begin{cases} 1 \text{ (on)}, & e(t) > 0 \\ 0 \text{ (off)}, & e(t) \leq 0 \end{cases}$$

- ...but control signal would “chatter” with slightest change of temperature
- **Solution:** Introduce **lag** to slow the switching cycle, e.g., **hysteresis**

$$u(t) = \begin{cases} 1 \text{ (on)}, & e(t) - T > 0 \\ 0 \text{ (off)}, & e(t) + T \leq 0 \end{cases}$$

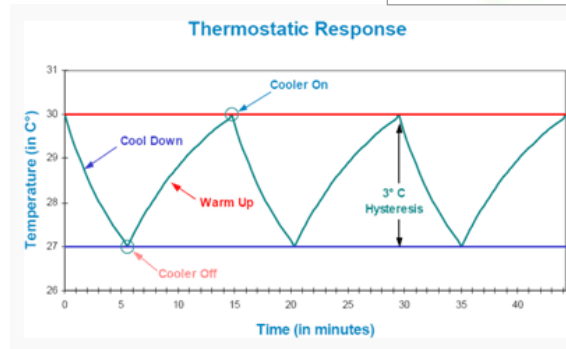
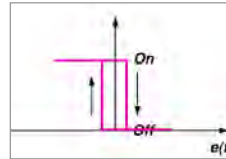


8

Thermostat Control Logic with Hysteresis

Hysteresis delays the response
System responds with a *limit cycle*

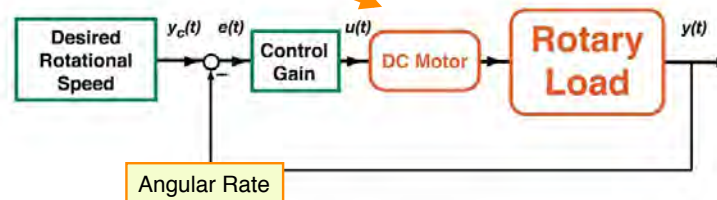
- Cooling control is similar with sign reversal



9



Speed Control of Direct-Current Motor



Linear Feedback Control Law (c = Control Gain)

$$u(t) = c e(t)$$

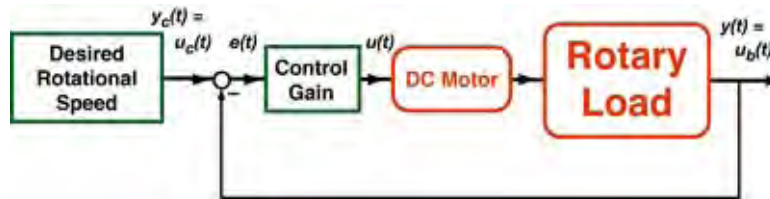
where

$$e(t) = y_c(t) - y(t)$$

How would $y(t)$ be measured?

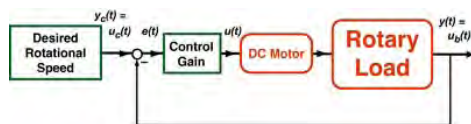
10

Characteristics of the Model



- **Simplified Dynamic Model**
 - Rotary inertia, J , is the sum of motor and load inertias
 - Internal damping neglected
 - **Output speed, $y(t)$, rad/s, is an integral of the control input, $u(t)$**
 - Motor control torque is proportional to $u(t)$
 - **Desired speed, $y_c(t)$, rad/s, is constant**

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Model of Dynamics and Speed Control

First-order LTI ordinary differential equation

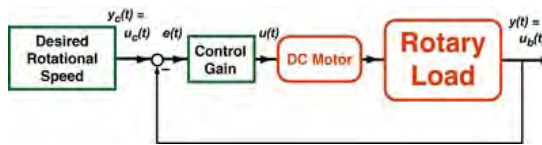
$$\frac{dy(t)}{dt} = \frac{1}{J} u(t) = \frac{c}{J} e(t) = \frac{c}{J} [y_c(t) - y(t)], \quad y(0) \text{ given}$$

Integral of the equation, with $y(0) = 0$

$$y(t) = \frac{1}{J} \int_0^t u(t) dt = \frac{c}{J} \int_0^t e(t) dt = \frac{c}{J} \int_0^t [y_c(t) - y(t)] dt$$

- Direct integration of $y_c(t)$
- Negative feedback of $y(t)$

12



Step Response of Speed Controller

Step input :

$$y_c(t) = \begin{cases} 0, & t < 0 \\ 1, & t \geq 0 \end{cases}$$

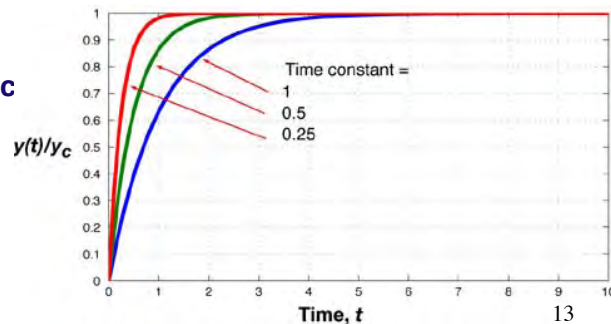
- Solution of the integral

$$y(t) = y_c \left[1 - \exp^{-\left(\frac{c}{J}\right)t} \right] = y_c \left[1 - \exp^{\lambda t} \right] = y_c \left[1 - \exp^{-t/\tau} \right]$$

- where

- $\lambda = -c/J$ = eigenvalue or root of the system (rad/sec)
- $\tau = J/c$ = time constant of the response (sec)

What does the shaft angle response look like?



Angle Control of Direct-Current Motor



- Simplified Dynamic Model

- Rotary inertia, J , is the sum of motor and load inertias
- Output angle, $y(t)$, is a double integral of the control, $u(t)$
- Desired angle, $y_c(t)$, is constant

Feedback Control Law

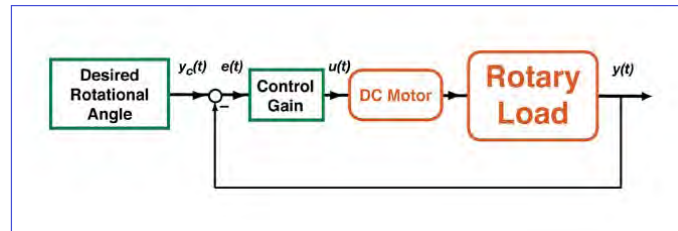
$$u(t) = c e(t)$$

where

$$e(t) = y_c(t) - y(t)$$

How would $y(t)$ be measured?

Model of Dynamics and Angle Control



Output angle, $y(t)$, as a function of time

$$y(t) = \frac{1}{J} \int_0^t \int_0^t u(t) dt dt = \frac{c}{J} \int_0^t \int_0^t e(t) dt dt = \frac{c}{J} \int_0^t \int_0^t [y_c - y(t)] dt dt$$

Associated 2nd-order, linear, time-invariant ordinary differential equation

$$\frac{d^2 y(t)}{dt^2} = \ddot{y}(t) = \frac{c}{J} [y_c - y(t)]$$

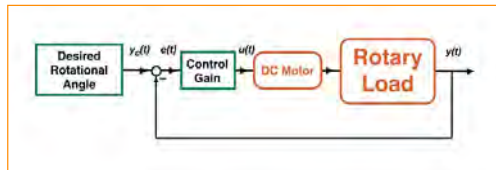
15

Model of Dynamics and Angle Control

- Corresponding set of 1st-order equations, with
 - Angle: $x_1(t) = y(t)$
 - Angular rate: $x_2(t) = dy(t)/dt$

$$\begin{aligned} \dot{x}_1(t) &= x_2(t) \\ \dot{x}_2(t) &= \frac{u(t)}{J} = \frac{c}{J} [y_c - y(t)] = \frac{c}{J} [y_c - x_1(t)] \end{aligned}$$

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State-Space Model of the DC Motor

Open-loop dynamic equation

$$\begin{bmatrix} \dot{x}_1(t) \\ \dot{x}_2(t) \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ 0 & 0 \end{bmatrix} \begin{bmatrix} x_1(t) \\ x_2(t) \end{bmatrix} + \begin{bmatrix} 0 \\ 1/J \end{bmatrix} u(t)$$

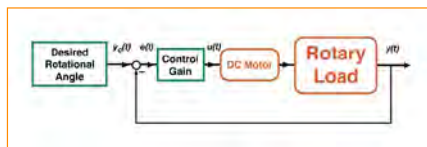
Feedback control law

$$u(t) = c[y_c(t) - y_1(t)] = c[y_c(t) - x_1(t)]$$

Closed-loop dynamic equation

$$\begin{bmatrix} \dot{x}_1(t) \\ \dot{x}_2(t) \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ -c/J & 0 \end{bmatrix} \begin{bmatrix} x_1(t) \\ x_2(t) \end{bmatrix} + \begin{bmatrix} 0 \\ c/J \end{bmatrix} y_c$$

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Step Response with Angle Feedback

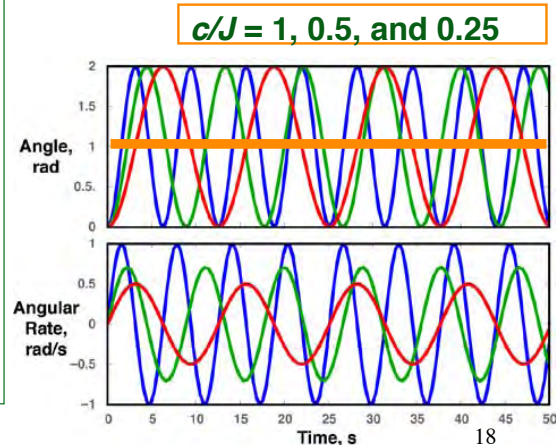
$$\begin{bmatrix} \dot{x}_1(t) \\ \dot{x}_2(t) \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ -c/J & 0 \end{bmatrix} \begin{bmatrix} x_1(t) \\ x_2(t) \end{bmatrix} + \begin{bmatrix} 0 \\ c/J \end{bmatrix} y_c$$

% Step Response of Undamped Angle Control

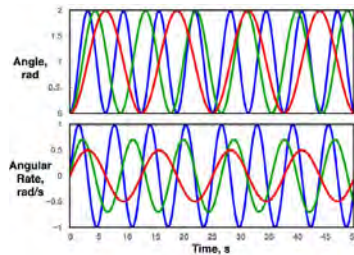
```
F1 = [0 1;-1 0];
G1 = [0;1];
F2 = [0 1;-0.5 0];
G2 = [0;0.5];
F3 = [0 1;-0.25 0];
G3 = [0;0.25];
Hx = [1 0;0 1];
```

```
Sys1 = ss(F1,G1,Hx,0);
Sys2 = ss(F2,G2,Hx,0);
Sys3 = ss(F3,G3,Hx,0);
```

```
step(Sys1,Sys2,Sys3)
```



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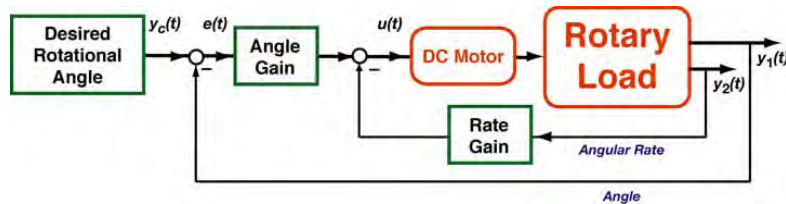


What Went Wrong?

- **No damping!**
- **Solution: Add rate feedback in the control law**

- **Control law with rate feedback**

$$u(t) = c_1 [y_c(t) - y_1(t)] - c_2 y_2(t)$$



Closed-loop dynamic equation

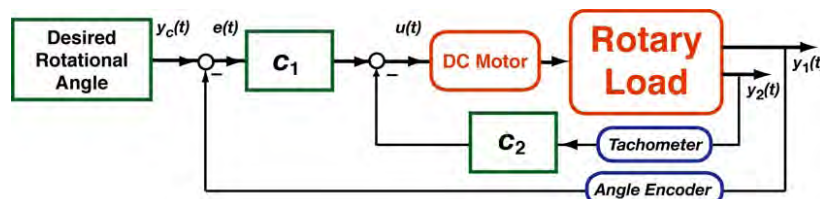
$$\begin{bmatrix} \dot{x}_1(t) \\ \dot{x}_2(t) \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ -c_1/J & -c_2/J \end{bmatrix} \begin{bmatrix} x_1(t) \\ x_2(t) \end{bmatrix} + \begin{bmatrix} 0 \\ c_1/J \end{bmatrix} y_c$$

19

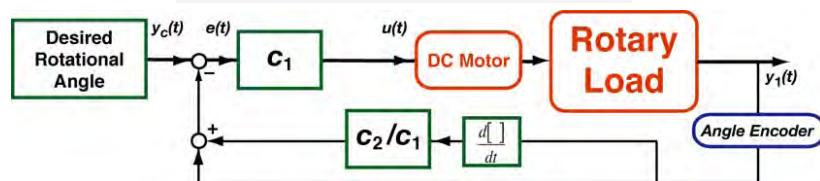
Alternative Implementations of Rate Feedback

$$u(t) = c_1 [y_c(t) - y_1(t)] - c_2 y_2(t) = c_1 [y_c(t) - y_1(t)] - c_2 \frac{dy_1(t)}{dt}$$

One input, two outputs



One input, one output



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Step Response with Angle and Rate Feedback

$$c_1 / J = 1$$

$$c_2 / J = 0, 1.414, 2.828$$

% Step Response of Damped Angle Control

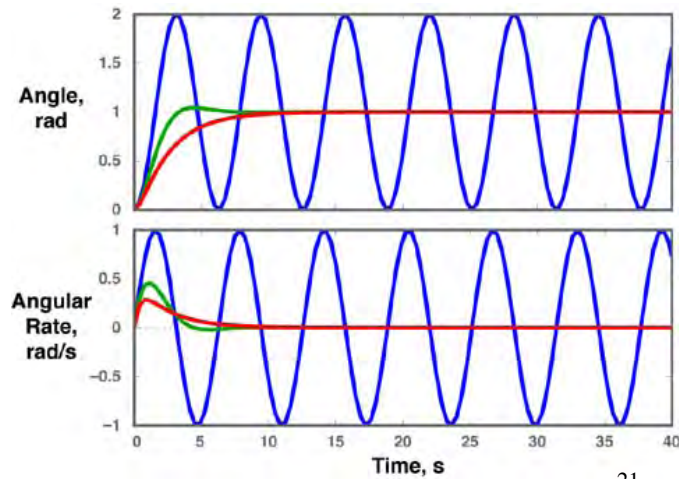
```
F1 = [0 1;-1 0];
G1 = [0;1];

F1a = [0 1;-1 -1.414];
F1b = [0 1;-1 -2.828];

Hx = [1 0;0 1];

Sys1 = ss(F1,G1,Hx,0);
Sys2 = ss(F1a,G1,Hx,0);
Sys3 = ss(F1b,G1,Hx,0);

step(Sys1,Sys2,Sys3)
```



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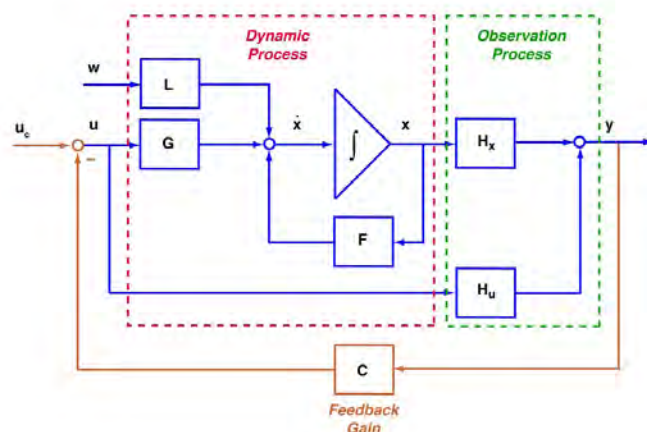
LTI Model with Feedback Control

- Command input, u_c , has dimensions of u

$$u(t) = u_c(t) - Cy(t)$$

$$\dot{x}(t) = Fx(t) + Gu(t) + Lw(t)$$

$$y(t) = H_x x(t) + H_u u(t)$$



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Effect of Feedback Control on the LTI Model

$$\begin{aligned}\dot{\mathbf{x}}(t) &= \mathbf{F}\mathbf{x}(t) + \mathbf{G}\mathbf{u}(t) = \mathbf{F}\mathbf{x}(t) + \mathbf{G}[\mathbf{u}_c(t) - \mathbf{C}\mathbf{y}(t)] \\ &= \mathbf{F}_{open\ loop}\mathbf{x}(t) + \mathbf{G}\{\mathbf{u}_c(t) - \mathbf{C}[\mathbf{H}_x\mathbf{x}(t)]\}\end{aligned}$$

$$\begin{aligned}&= [\mathbf{F} - \mathbf{GCH}_x]\mathbf{x}(t) + \mathbf{G}\mathbf{u}_c(t) \\ &\triangleq \mathbf{F}_{closed\ loop}\mathbf{x}(t) + \mathbf{G}\mathbf{u}_c(t)\end{aligned}$$

Feedback modifies the stability matrix of the closed-loop system

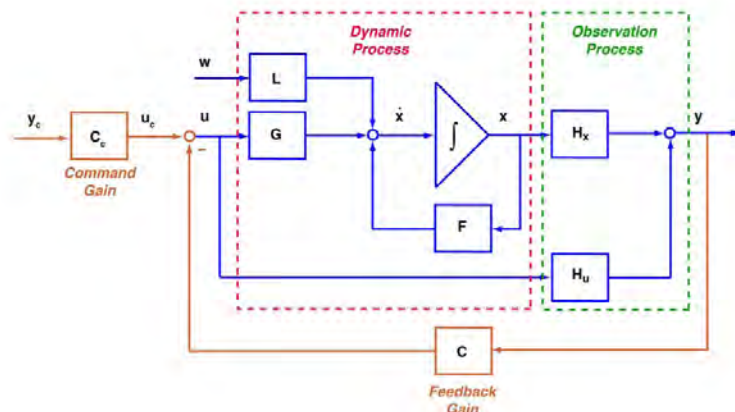
Convergence or divergence

Envelope of transient response

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LTI Model with Feedback Control and Command Gain

Command input, \mathbf{y}_c , is “shaped” by \mathbf{C}_c



$$\begin{aligned}\mathbf{u}(t) &= \mathbf{u}_c(t) - \mathbf{C}\mathbf{y}(t) \\ &= \mathbf{C}_c\mathbf{y}_c(t) - \mathbf{C}\mathbf{y}(t)\end{aligned}$$

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Effect of Command Gain on LTI Model

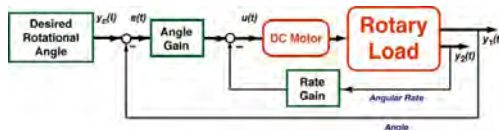
$$\begin{aligned}\dot{\mathbf{x}}(t) &= \mathbf{F}\mathbf{x}(t) + \mathbf{G}\mathbf{u}(t) = \mathbf{F}\mathbf{x}(t) + \mathbf{G}\{\mathbf{C}_c\mathbf{y}_c(t) - \mathbf{C}\mathbf{y}(t)\} \\ &= \mathbf{F}\mathbf{x}(t) + \mathbf{G}\{\mathbf{C}_c\mathbf{y}_c(t) - \mathbf{C}[\mathbf{H}_x\mathbf{x}(t)]\} \\ &= [\mathbf{F} - \mathbf{GCH}_x]\mathbf{x}(t) + \mathbf{G}\mathbf{C}_c\mathbf{y}_c(t)\end{aligned}$$

- Steady-state response of the system $\dot{\mathbf{x}}(t) = \mathbf{0}$

$$\mathbf{x}^*(t) = -[\mathbf{F} - \mathbf{GCH}_x]^{-1} \mathbf{G}\mathbf{C}_c\mathbf{y}_c^*(t)$$

- Command gain adjusts the steady-state response
- Has no effect on the stability of the system

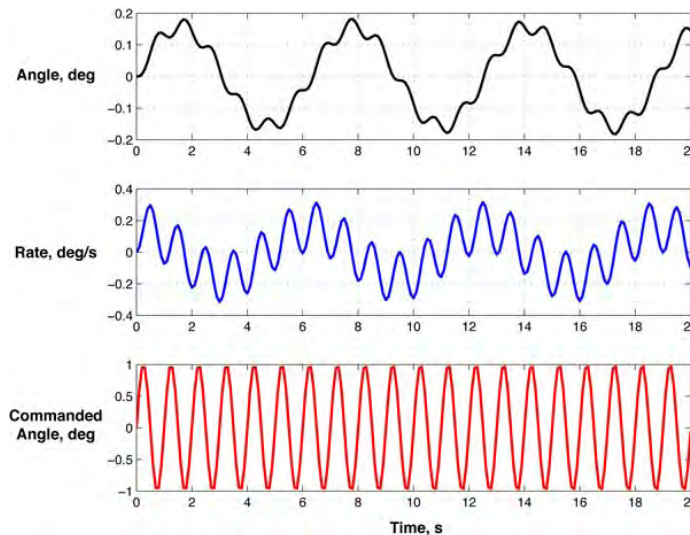
25



**Response to Sine Wave
Input with Angle Feedback:
No Damping**

$$y_c(t) = \sin(\omega t) = \sin(6.28 t), \text{ deg}$$

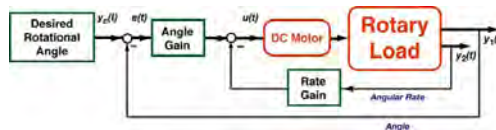
$$c_1/J = 1; c_2/J = 0$$



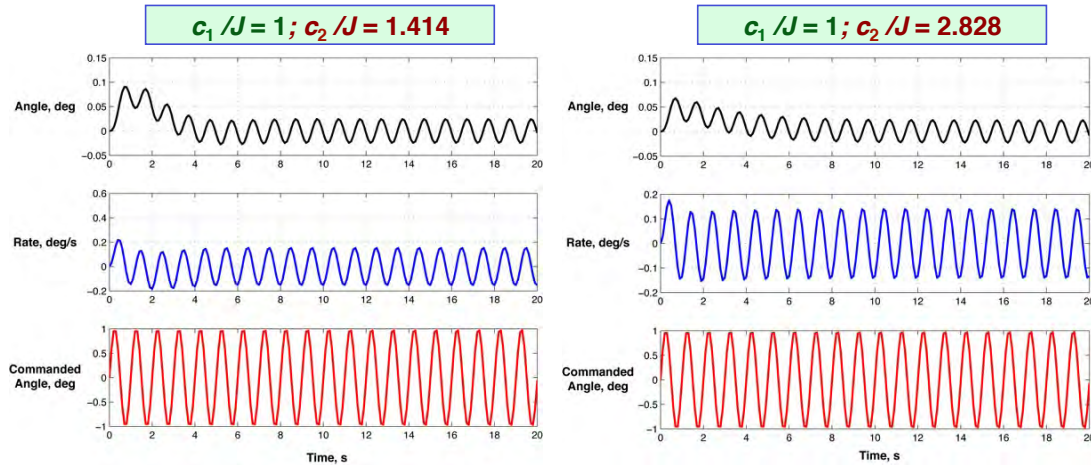
- Why are there 2 oscillations in the output?
 - Undamped transient response to the input
 - Long-term dynamic response to the input
- System has a natural frequency of oscillation, ω_n
- Long-term response to a sine wave is a sine wave

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Response to Sine Wave Input with Rate Damping



$$y_c(t) = \sin(\omega t) = \sin(6.28 t), \text{ deg}$$



- With damping, transient response decays
- In this case, damping has negligible effect on long-term response

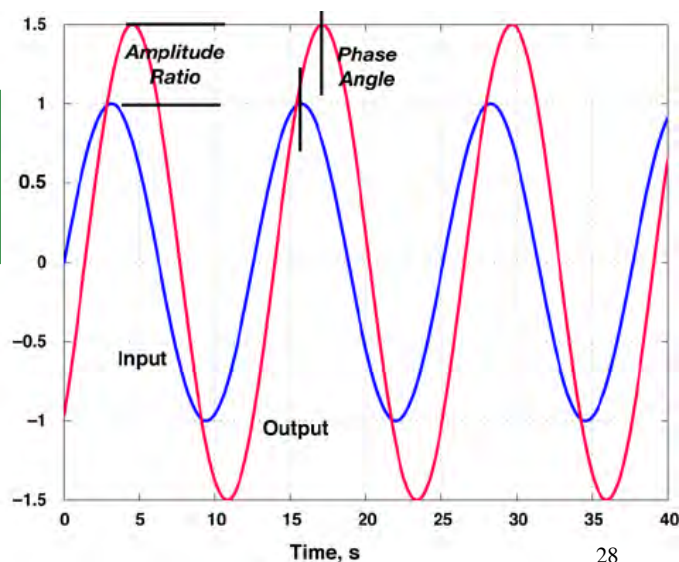
27

System Dynamics Produces Differences in Amplitude and Phase Angle of Input and Output

$$\text{Amplitude Ratio (AR)} = \frac{y_{peak}}{y_{c_{peak}}}$$

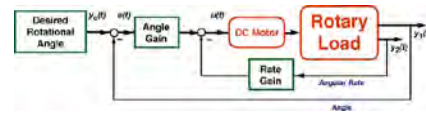
$$\text{Phase Angle} = -360 \frac{\Delta t_{peak}}{\text{Period}}, \text{ deg}$$

- Amplitude ratio and phase angle characterize the system model



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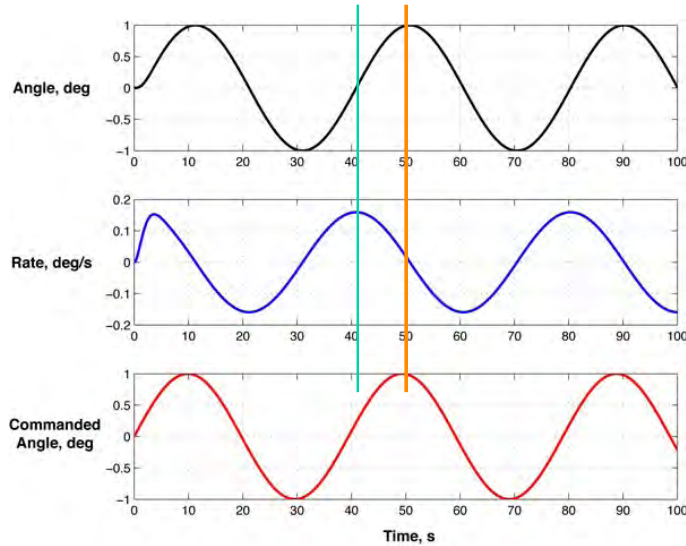
Effect of Input Frequency on Output Amplitude and Phase Angle



$$y_c(t) = \sin(t / 6.28), \text{ deg}$$

$$c_1 / J = 1; c_2 / J = 1.414$$

- With low input frequency, input and output amplitudes are about the same
- Lag of angle output oscillation, compared to input, is small
- Rate oscillation "leads" angle oscillation by ~90 deg

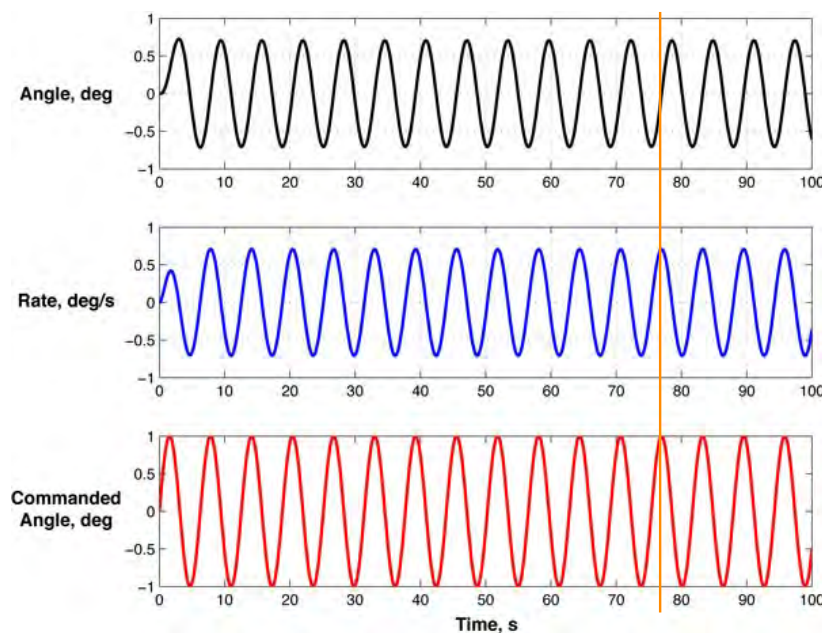


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At Higher Frequency, Phase Angle Lag Increases

$$c_1 / J = 1; c_2 / J = 1.414$$

$$y_c(t) = \sin(t), \text{ deg}$$

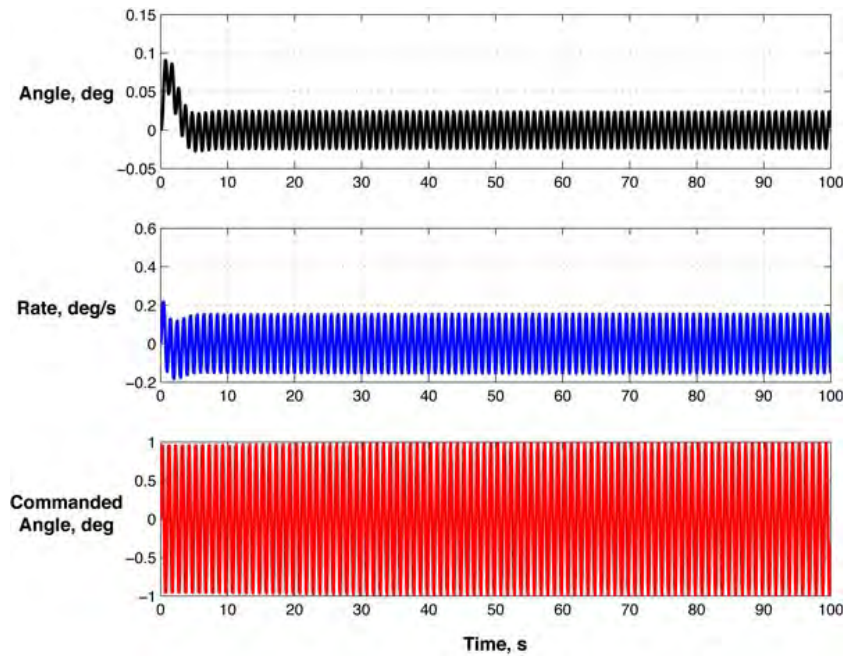


30

$$c_1/J = 1; c_2/J = 1.414$$

$$y_c(t) = \sin(6.28 t), \text{ deg}$$

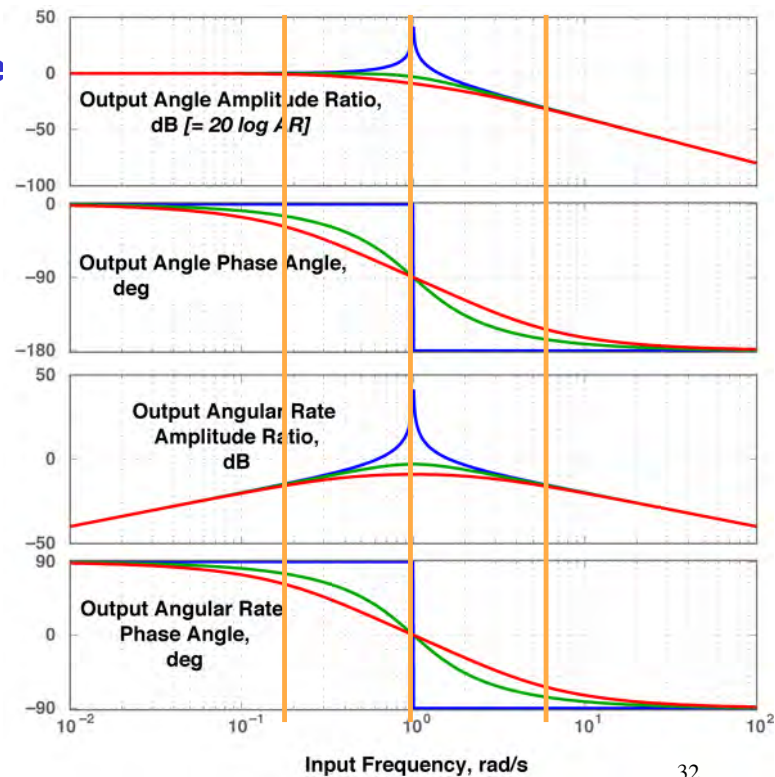
At Even Higher Frequency,
Amplitude Ratio Decreases



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Frequency Response of the DC Motor with Feedback Control

- Long-term response to sinusoidal inputs over range of frequencies
 - Determine experimentally or
 - from the transfer function
- Transfer function based on the Laplace transform of the system
- Frequency response depicted in the Bode Plot



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***Next Time:
Analog and Digital
Control Systems***

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Supplemental Material

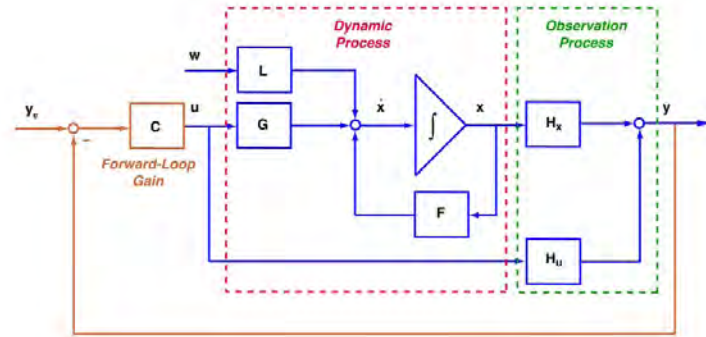
34

LTI Control with Forward-Loop Gain

$$\mathbf{u}(t) = \mathbf{C}[\mathbf{y}_c(t) - \mathbf{y}(t)]$$

$$\dot{\mathbf{x}}(t) = \mathbf{F} \mathbf{x}(t) + \mathbf{G} \mathbf{u}(t) + \mathbf{L} \mathbf{w}(t)$$

$$\mathbf{y}(t) = \mathbf{H}_x \mathbf{x}(t) + \mathbf{H}_u \mathbf{u}(t)$$



With $\mathbf{C}_c = \mathbf{C}$, command input, \mathbf{y}_c , has dimensions of \mathbf{y}