

# MEEC

## ELECTRONICS FOR MICRO-SYSTEMS

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Individual Assignment Problems Solving Analysis [EN]

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# 1 Problem 1

## 1.1 Voltage Gain

Considering an ideal OPAMP working on the linear region, the following assumptions were made.

- $Z_{in} = +\infty$
- $Z_{out} = 0$
- $A_d = +\infty$
- $V_+ = V_-$

In order to obtain the value of  $V_{out}(V_{in})$ , it is necessary to get the circuit equations.

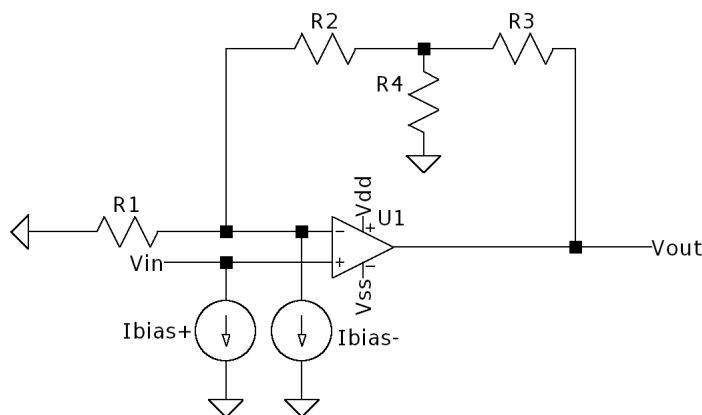
$$\begin{cases} i_t = \frac{V_-}{R_1} \\ V_x = V_- + R_2 \cdot i_t \\ V_{out} = V_x + R_3 \cdot i_3 \\ i_3 = i_t + i_4 = i_t + \frac{V_x}{R_4} \end{cases} \quad (1)$$

Where  $i_t$  is the current passing through  $R_1$  and  $R_2$ ,  $V_x$  is voltage between  $R_3$  terminals. This results in the following equation.

$$V_{out} = \frac{R_1 R_4 + R_2 R_4 + R_3 (R_1 + R_2 + R_4)}{R_1 R_4} \cdot V_i \quad (2)$$

## 1.2 Input Current Bias

In order to calculate the impact of the current bias, a current supply is placed in parallel on the input terminals of the OPAMP. As shown in the figure



**Figure 1:** Circuit with input bias current

Using superposition, to evaluate the current effect on the output, the following system of equations is obtained.

$$\begin{cases} V_- = V_+ = 0 \\ I_{R_1} = 0 \\ \frac{V_x}{R_4} + \frac{V_x - V_{out}}{R_3} + I_{bias-} \\ V_{out} = I_{bias-} - \frac{V_x - V_{out}}{R_3} \end{cases} \quad (3)$$

## 2 Problem 5

### 2.1 Considerations

- $T1 = 42\text{ }^{\circ}\text{C} = 315.15\text{ }K$
- $T2 = 42.5\text{ }^{\circ}\text{C} = 315.65\text{ }K$

### 2.2 NTC

Using the beta model:

$$R = R_0 e^{\beta \left( \frac{1}{T} - \frac{1}{T_0} \right)} \quad (4)$$

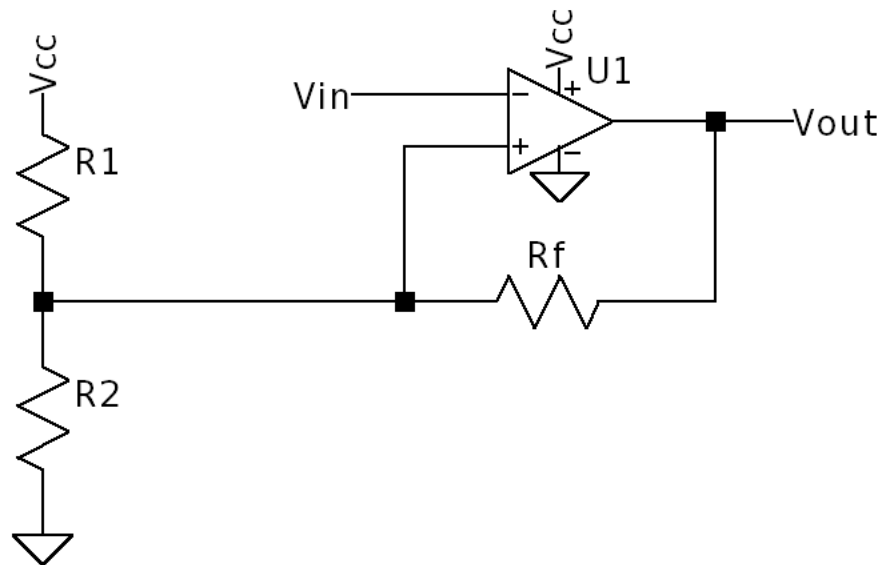
From the datasheet :

- $\beta = 3988$
- $R(25^{\circ}) = 5k\Omega$

Therefore:

$$R = 5K \cdot e^{3988 \cdot \left( \frac{1}{T} - \frac{1}{298.15} \right)} \quad (5)$$

Hence:  $R(T1) = R(T2) =$

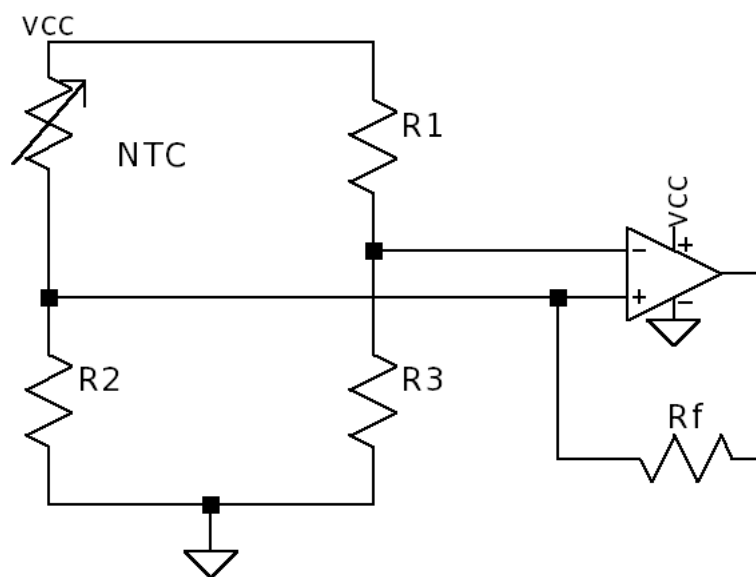


**Figure 2:** Comparator with hysteresis circuit [1]

For circuit dimensioning the following equations were used [2].

$$\begin{cases} \frac{R_f}{R_1} = \frac{V_L}{V_H - V_L} \\ \frac{R_2}{R_1} = \frac{V_L}{V_{CC} - V_H} \end{cases} \quad (6)$$

But since this circuit is inverting.



**Figure 3:** NTC with non-inverting comparator

In this circuit  $V_{in} = V_{CC} \cdot \frac{R_3}{R_3 + R_1}$ ,  $R_1 = R_{NTC}$ .

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## References

- [1] TDK, “Ntc thermistors for temperature measuremen,” [https://product.tdk.com/system/files/dam/doc/product/sensor/ntc/ntc\\_element/data\\_sheet/50/db/ntc/ntc\\_mini\\_sensors\\_s861.pdf](https://product.tdk.com/system/files/dam/doc/product/sensor/ntc/ntc_element/data_sheet/50/db/ntc/ntc_mini_sensors_s861.pdf), 2018.
- [2] A. Kay and T. Claycomb, “Comparator with hysteresis reference design,” Texas Instruments, Technical Note TIDU020A, June 2014, revised Edition. [Online]. Available: <https://www.ti.com/lit/ug/tidu020a/tidu020a.pdf>