

PK

8th LECTURE, SUGGESTED SOLUTION.

1.

- Design a TYPE-I FIR filter with $\omega_c = \pi/4$ using the window method. The filter should comply with the specifications for the LP-filter designed in lecture 7.

Desired frequency response;

$$H_d(e^{j\omega}) = \begin{cases} e^{-j\omega M/2} & |\omega| \leq \pi/4 \\ 0 & \pi/4 < |\omega| \leq \pi \end{cases}$$

from which we can derive the ideal impulse response

$$h_d[n] = \frac{1}{2\pi} \int_{-\pi/4}^{\pi/4} e^{-j\omega M/2} \cdot e^{j\omega n} d\omega$$

$$= \frac{1}{2\pi} \int_{-\pi/4}^{\pi/4} \left(\cos\left\{(n - \frac{M}{2})\omega\right\} + j \sin\left\{(n - \frac{M}{2})\omega\right\} \right) d\omega$$

$$= \frac{1}{2\pi} \left[\sin\left\{(n - \frac{M}{2})\omega\right\} \right]_{-\pi/4}^{\pi/4} \cdot \frac{1}{n - \frac{M}{2}}$$

$$= \frac{1}{2\pi(n - \frac{M}{2})} \cdot \left\{ \sin\left(\frac{\pi}{4}(n - \frac{M}{2})\right) - \sin\left(-\frac{\pi}{4}(n - \frac{M}{2})\right) \right\}$$

\Downarrow

$$h_d[n] = \frac{\sin\left(\frac{\pi}{4}(n - \frac{M}{2})\right)}{\pi(n - \frac{M}{2})} \quad -\infty < n < \infty$$

Now, $h_d[n]$ is $\begin{cases} \text{Symmetric around } M/2 \\ \text{infinite} \\ \text{non-causal} \end{cases}$

Thus we truncate $h_d[n]$ with a symmetric window function;

$$h[n] = h_d[n] \cdot w[n], \quad w[n] \neq 0 \quad 0 \leq n \leq M$$

Using eqn. 140a (p.343), we can now write the frequency response for the filter (TYPE-I)

$$H(e^{j\omega}) = e^{-j\omega M/2} \left(\sum_{k=0}^{M/2} a[k] \cdot \cos(\omega k) \right)$$

where

$$\begin{aligned} a[0] &= h[M/2] \\ a[k] &= 2 \cdot h[M/2 - k] \quad \text{for } k=1, 2, \dots, M/2 \end{aligned}$$

Due to the symmetric impulse response $h[n]$, the filter has linear phase response.

Thus we need to concern only about the Amplitude response

$$|H(e^{j\omega})| = \left| \sum_{k=0}^{M/2} a[k] \cdot \cos(\omega k) \right|$$

In order to fulfill the OdB DC Gain requirement we need to find G given as;

$$\Downarrow \quad G \cdot |H(e^{j\omega})| = 1 \quad \Big|_{\omega=0}$$

$$\Downarrow \quad G \cdot \left| \sum_{k=0}^{M/2-1} a[k] \right| = 1$$

$$\Downarrow \quad G = \frac{1}{\left| \sum_{k=0}^{M/2-1} a[k] \right|}$$

and thus the overall Amplitude response is

$$|H(e^{j\omega})| = \frac{\left| \sum_{k=0}^{M/2-1} a[k] \cdot \cos(\omega k) \right|}{\left| \sum_{k=0}^{M/2-1} a[k] \right|}$$

Now, for given Window function $w[n]$ our task is to find the smallest value of M which satisfies the design specifications

In order to do that, write a program and conduct experiments with the rectangular as well as the Hamming window.


```
% The program calculates the amplitude response of an M'th order TYPE-I FIR
% filter, given the desired impulse response h_d[n]
%
% The Rectangular Window is used for truncation of h_d[n]
clear

% Filter order (has to be even for TYPE-I FIR filters)
M = 12;

% Frequency sweep from 0 til PI with 1000 points
for i=0:999,
    omega(i+1) = pi*i/999;
end;

% The amplitude response is now calculated for the 1000 discrete frequency
% values.
%
% The amplitude response has the form  $|H| = N(\omega,k)/D(k)$ 

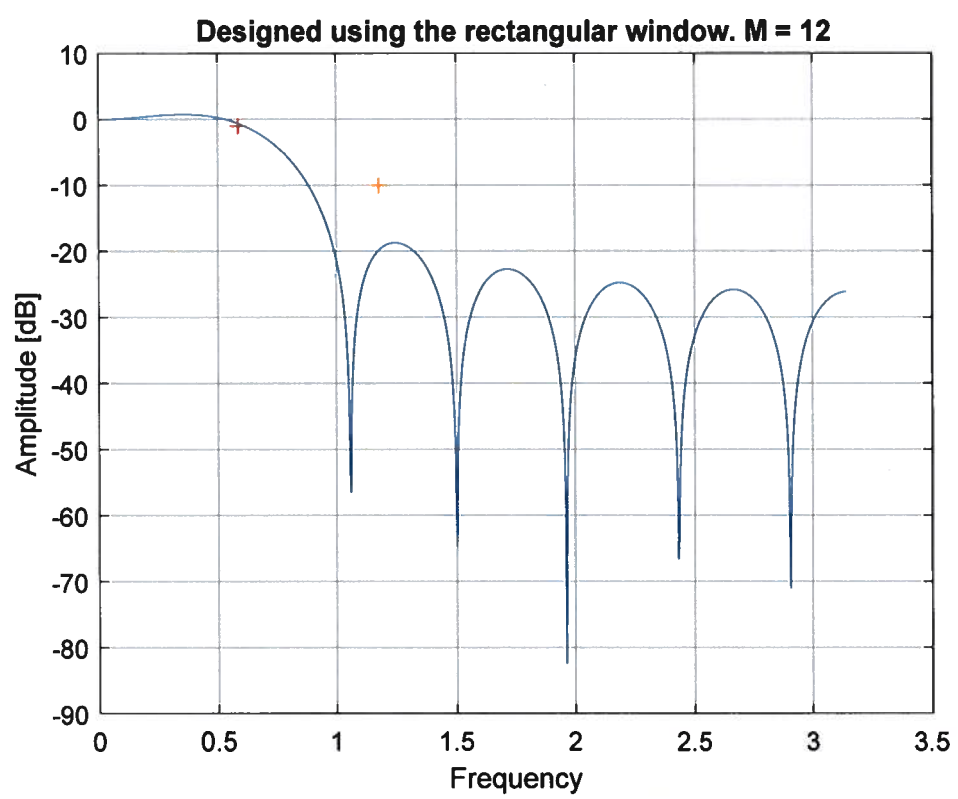
for i=0:999
    % The numerator N is calculated using the summation variable sum_n
    sum_n = 0;
    % The denominator D is calculated using the summation variable sum_d
    sum_d = 0;
    % Each sum has a total of M/2+1 terms
    for k=1:(M/2)
        % First calculate the ideal impulse response sample. Note that the
        % symmetri point (k=0) is not included. This final term will be added
        % later.
        h(k) = (sin((pi/4)*k)/(pi*k));
        a(k) = 2*h(k);

        % Now both N and D are updated
        sum_n = sum_n + (a(k) * cos(k * omega(i+1)));
        sum_d = sum_d + a(k);
    end

    % Finally, add the contribution from the symmetri point, i.e., k=0
    % For the given impulse response, this value equals 1/4
    sum_n = sum_n + 0.25;
    sum_d = sum_d + 0.25;

    % Finally, the amplitude value at the actual frequency point is calculated
    amp(i+1) = abs(sum_n)/abs(sum_d);
end,

% Plot the amplitude response (in dB) together with the specifications.
plot(omega,20*log10(amp),omega(187),-1,'+',omega(375),-10,'+')
grid;
xlabel('Frequency')
ylabel('Amplitude [dB]')
title(sprintf('Designed using the rectangular window. M = %.0f', M))
```



```
% The program calculates the amplitude response of an M'th order TYPE-I FIR
% filter, given the desired impulse response h_d[n]
%
% The Hamming Window is used for truncation of h_d[n]
clear

% Filter order (has to be even for TYPE-I FIR filters)
M = 28;

% Frequency sweep from 0 til PI with 1000 points
for i=0:999,
    omega(i+1) = pi*i/999;
end;

% The amplitude response is now calculated for the 1000 discrete frequency
% values.
%
% The amplitude response has the form  $|H| = N(\omega, k)/D(k)$ 

for i=0:999
    % The numerator N is calculated using the summation variable sum_n
    sum_n = 0;
    % The denominator D is calculated using the summation variable sum_d
    sum_d = 0;
    % Each sum has a total of M/2+1 terms
    for k=1:(M/2)
        % First calculate the ideal impulse response sample. Note that the
        % symmetri point (k=0) is not included. This final term will be added
        % later.
        h_d(k) = (sin((pi/4)*k)/(pi*k));
        % The impulse response is now multiplied with the Hamming window
        h(k) = h_d(k) * (0.54 - 0.46*cos((2*pi)*(k + M/2)/M));
        a(k) = 2*h(k);

        % Now both N and D are updated
        sum_n = sum_n + (a(k) * cos(k * omega(i+1)));
        sum_d = sum_d + a(k);
    end

    % Finally, add the contribution from the symmetri point, i.e., k=0
    % For the given impulse response, this value equals 1/4
    sum_n = sum_n + 0.25;
    sum_d = sum_d + 0.25;

    % Finally, the amplitude value at the actual frequency point is calculated
    amp(i+1) = abs(sum_n)/abs(sum_d);
end,

% Plot the amplitude response (in dB) together with the specifications.
plot(omega, 20*log10(amp), omega(187), -1, '+', omega(375), -10, '+')
grid;
xlabel('Frequency')
ylabel('Amplitude [dB]')
title(sprintf('Designed using the Hamming window. M = %.0f', M))
```

