

Digital Electronics

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CHAPTER-2

Minimization Techniques





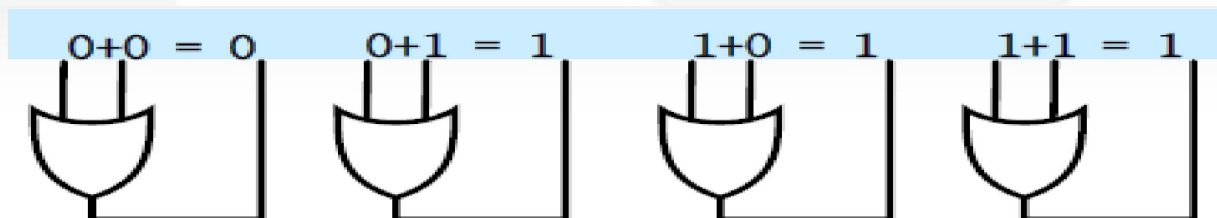
2.1 Boolean Algebra

- **Boolean Algebra:** In 1854 Logical algebra was published by **George Boole** known today as “Boolean Algebra”. It’s a convenient way and systematic way of expressing and analyzing the operation of logic circuits. 1938: **Claude Shannon** was the first to apply Boole’s work to the analysis and design of logic circuits.
- **Boolean Operations & Expressions:**
 - Variable: A symbol used to represent a logical quantity.
 - Complement: The inverse of a variable and is indicated by a bar over the variable.
 - Literal: A variable or the complement of a variable.
- **Boolean Addition:**

Boolean addition is equivalent to the OR operation.

A sum term is equal to 1 when one or more of the literals in the term are 1.

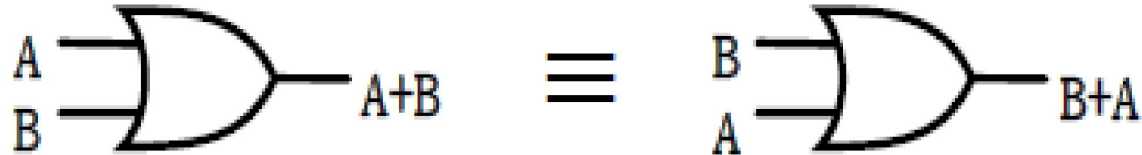
A sum term is equal to 0 only if each of the literals is 0.



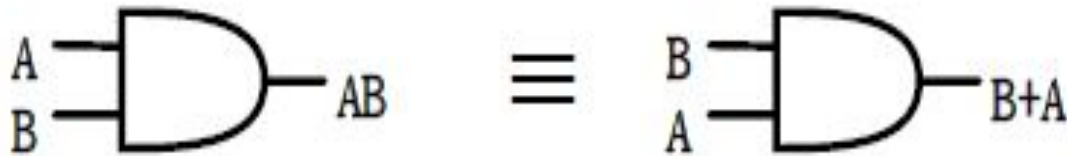
2.2 Laws & Rules of Boolean Algebra

Commutative Laws: The commutative law of addition for two variables is written as:

$$A+B = B+A.$$

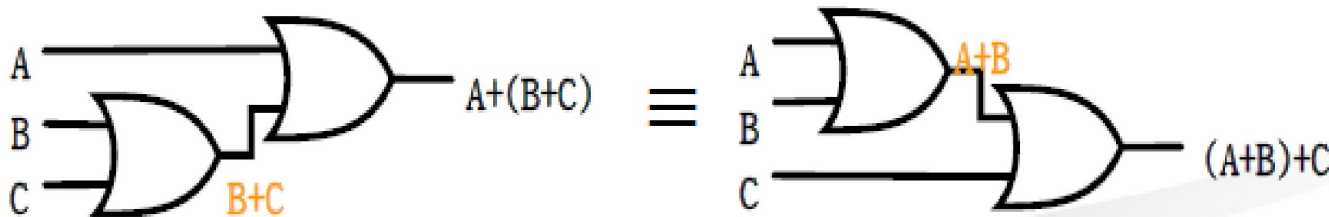


▪The commutative law of multiplication for two variables is written as: $AB = BA$.

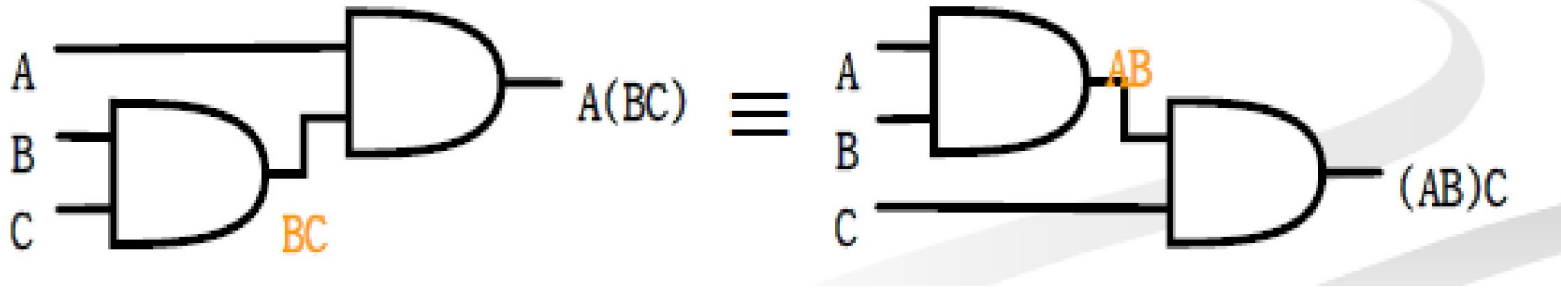


Associative Laws: The associative law of addition for 3 variables is written as:

$$A+(B+C) = (A+B)+C$$

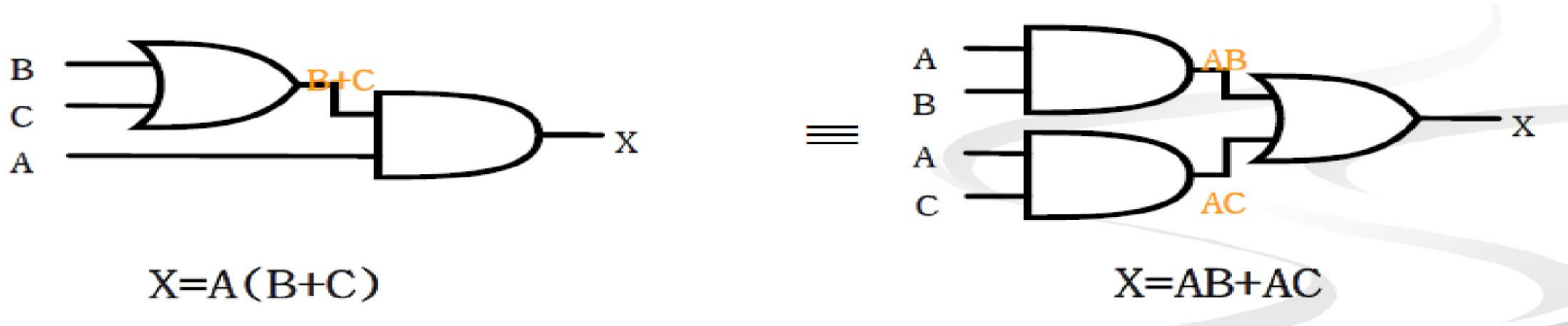


The associative law of multiplication for 3 variables is written as: $A(BC) = (AB)C$



Distributive Laws:

The distributive law is written for 3 variables as follows: $A(B+C) = AB + AC$



2.3 Basic Boolean Identities

- As with algebra, there will be Boolean operations that we will want to simplify. We apply the following Boolean identities to help.

Identity Name	AND Form	OR Form
Identity Law	$1x = x$	$0+x = x$
Null (or Dominance) Law	$0x = 0$	$1+x = 1$
Idempotent Law	$xx = x$	$x+x = x$
Inverse Law	$x\bar{x} = 0$	$x+\bar{x} = 1$
Commutative Law	$xy = yx$	$x+y = y+x$
Associative Law	$(xy)z = x(yz)$	$(x+y)+z = x+(y+z)$
Distributive Law	$x+yz = (x+y)(x+z)$	$x(y+z) = xy+xz$
Absorption Law	$x(x+y) = x$	$x+xy = x$
DeMorgan's Law	$(\overline{xy}) = \bar{x}+\bar{y}$	$(\overline{x+y}) = \bar{x}\bar{y}$
Double Complement Law	$\overline{\bar{x}} = x$	

2.4 Rules of Boolean Algebra

- As with algebra, there will be Boolean operations that we will want to simplify We apply the following Boolean identities to help

$$1. A + 0 = A$$

$$7. A \bullet A = A$$

$$2. A + 1 = 1$$

$$8. A \bullet \bar{A} = 0$$

$$3. A \bullet 0 = 0$$

$$9. \bar{\bar{A}} = A$$

$$4. A \bullet 1 = A$$

$$10. A + AB = A$$

$$5. A + A = A$$

$$11. A + \bar{A}B = A + B$$

$$6. A + \bar{A} = 1$$

$$12. (A + B)(A + C) = A + BC$$

2.5 DeMorgan's Theorems

- DeMorgan's theorems provide mathematical verification of:

1. The equivalency of the NAND and negative-OR gates.

The complement of two or more ANDed variables is equivalent to the OR of the complements of the individual variables.

The diagram shows the equivalence between a NAND operation and a Negative-OR operation. On the left, the expression $\overline{X \bullet Y}$ is enclosed in a red dashed oval with the word "NAND" in red above it. This is followed by an equals sign, and then the expression $\overline{X} + \overline{Y}$ is enclosed in a green dashed oval with the words "Negative-OR" in green above it.

$$\overline{X \bullet Y} = \overline{X} + \overline{Y}$$

2. The equivalency of the NOR and negative-AND gates.

The diagram shows the equivalence between a NOR operation and a Negative-AND operation. On the left, the expression $\overline{X + Y}$ is enclosed in an orange dashed oval with the word "NOR" in orange above it. This is followed by an equals sign, and then the expression $\overline{X} \bullet \overline{Y}$ is enclosed in a blue dashed oval with the words "Negative-AND" in blue above it.

$$\overline{X + Y} = \overline{X} \bullet \overline{Y}$$

The complement of two or more ORed variables is equivalent to the AND of the complements of the individual variables.

2.6 Duality theorem

- It states that if you have a true **Boolean** equation then the dual of this statement is true.
- The dual of a boolean statement is found by replacing the statement's symbols with their counterparts. This means a “0” becomes a “1”, “1” becomes a “0”, “+” becomes a “.” and “.” becomes a “+”.

Given Expression	Dual
$0 = 1$	$1 = 0$
$0.1 = 0$	$1 + 0 = 1$
$A.0 = 0$	$A + 1 = 1$
$A.A = 0$	$A + A = 1$
$A.(B.C) = (A.B).C$	$A+(B+C) = (A+B) + C$

2.7 Minterm

- If a boolean function contains n variables than a product term which contains all the variables once in either complemented or uncomplemented form is called minterm.
- In the minterm main property of each minterm is that it will have value one only.
- For n variable expression the no. of minterms are 2^n like- if $n=2$ than minterms are 4, $n=3$ than minterms are 8.
- All the minterms are can be represented by m_0, m_1, m_2, \dots

Example:

X	Y	X.Y	Minterms	
0	0	0	$X'Y'$	m_0
0	1	0	$X'Y$	m_1
1	0	0	XY'	m_2
1	1	1	XY	m_3

2.8 Maxterm

- If a Boolean function contains n variables and the sum term contains all the possible combinations of n variables then the sum term is known as **maxterm**.
- This term is opposite of minterm because the value of each maxterm is 0. The maxterm can be represented as M_0, M_1, \dots
- The main property of each maxterm is that it has value 0 only for one combination of n input variables.

Example:

X	Y	$X + Y$	Maxterms	
0	0	0	$X + Y$	M_0
0	1	1	$X + Y'$	M_1
1	0	1	$X' + Y$	M_2
1	1	1	$X' + Y'$	M_3

2.9 Sum of Products (SOP)

- When we perform the sum of logically multiplied inputs than the resultant expression is called sum of product. The canonical or standard SOP is the sum of minterms.

Example: $Y(A,B,C) = m_1 + m_4 + m_5 + m_7$

A	B	C	Minterm	
0	0	0	$A'B'C'$	m_0
0	0	1	$A'B'C$	m_1
0	1	0	$A'BC'$	m_2
0	1	1	$A'BC$	m_3
1	0	0	$AB'C'$	m_4
1	0	1	$AB'C$	m_5
1	1	0	ABC'	m_6
1	1	1	ABC	m_7

$$= A'B'C + AB'C' + AB'C + ABC$$

$$= A'B'C + AB'C + AB'C' + ABC$$

$$= B'C(A' + A) + AB'C' + ABC$$

$$= B'C + AB'C' + ABC$$

2.10 Product of Sums (POS)

- The product of sum can be defined as the logical product of the maxterm for which it has the value 0.
- **Example:**

$$f(A,B,C) = (A' + B' + C) \cdot (A + C) \cdot (A' + B')$$

$$= (A' + B' + C) \cdot (A + C + BB') \cdot (A' + B' + CC')$$

$$= (A' + B' + C) \cdot ((A + C) + BB') \cdot ((A' + B') + CC')$$

By applying: $x + y \cdot z = (x + y) \cdot (x + z)$

$$= (A' + B' + C) \cdot (A + C + B) \cdot (A + C + B') \cdot (A' + B' + C) \cdot (A' + B' + C')$$

$$= M_6, M_0, M_2, M_6, M_7$$

$$= M_0, M_2, M_6, M_7$$

2.11 Product of Sums (POS)

- The product of sum can be defined as the logical product of the maxterm for which it has the value 0.
- **Example:**

$$f(A,B,C) = (A' + B' + C) \cdot (A + C) \cdot (A' + B')$$

$$= (A' + B' + C) \cdot (A + C + BB')$$

$$= (A' + B' + C) \cdot ((A + C) + BB') \cdot ((A' + B') + CC')$$

By applying: $x + y \cdot z = (x + y) \cdot (x + z)$

$$= (A' + B' + C) \cdot (A + C + B) \cdot (A + C + B') \cdot (A' + B' + C) \cdot (A' + B' + C')$$

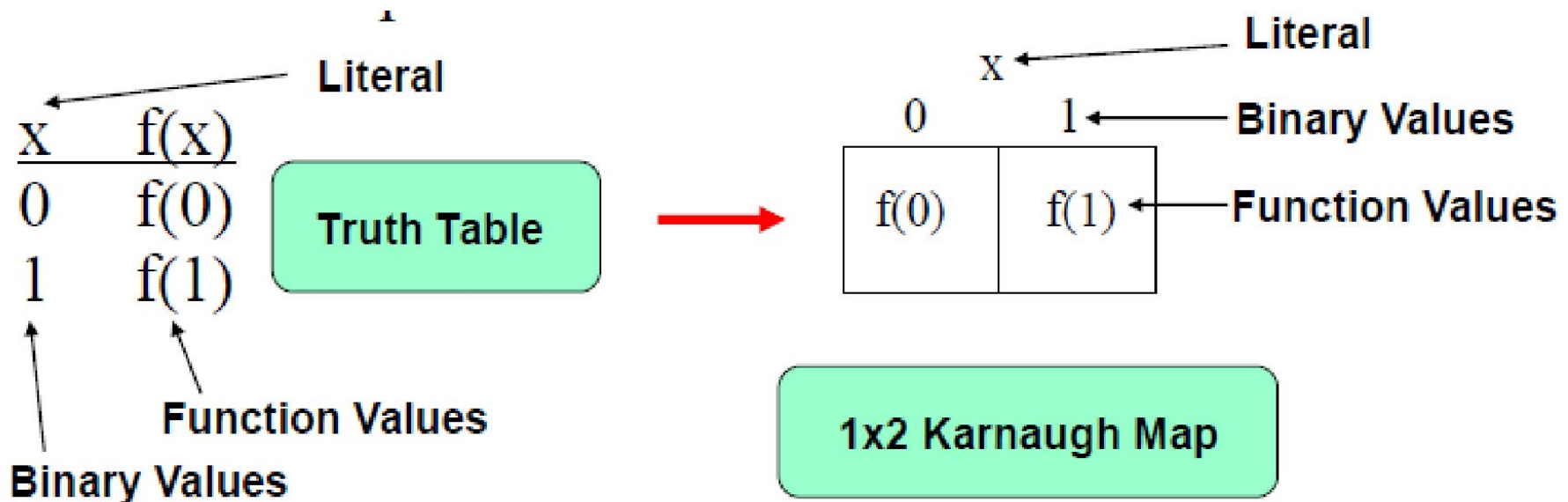
$$= M_6, M_0, M_2, M_6, M_7$$

$$= M_0, M_2, M_6, M_7$$

2.12 Karnough Map (K- map)

- Some times the Boolean theorems and laws makes the simplification logics of Boolean functions more complex. Than we have to use the k-map techniques.
- It is a graphical method which is used to simplify a Boolean function or to convert a truth table into its equivalent logic circuit.
- The k-map is designed by squares where each square represents a minterm or maxterm.
- We will determine these techniques by studying examples in order to establish the rules for map manipulation.

1 variable map:



2 variable map:

x	y	f(x,y)
0	0	f(0,0)
0	1	f(0,1)
1	0	f(1,0)
1	1	f(1,1)

Truth Table

		y	
		0	1
x	0	f(0,0)	f(0,1)
	1	f(1,0)	f(1,1)

2x2 Karnaugh Map

3 variable map:

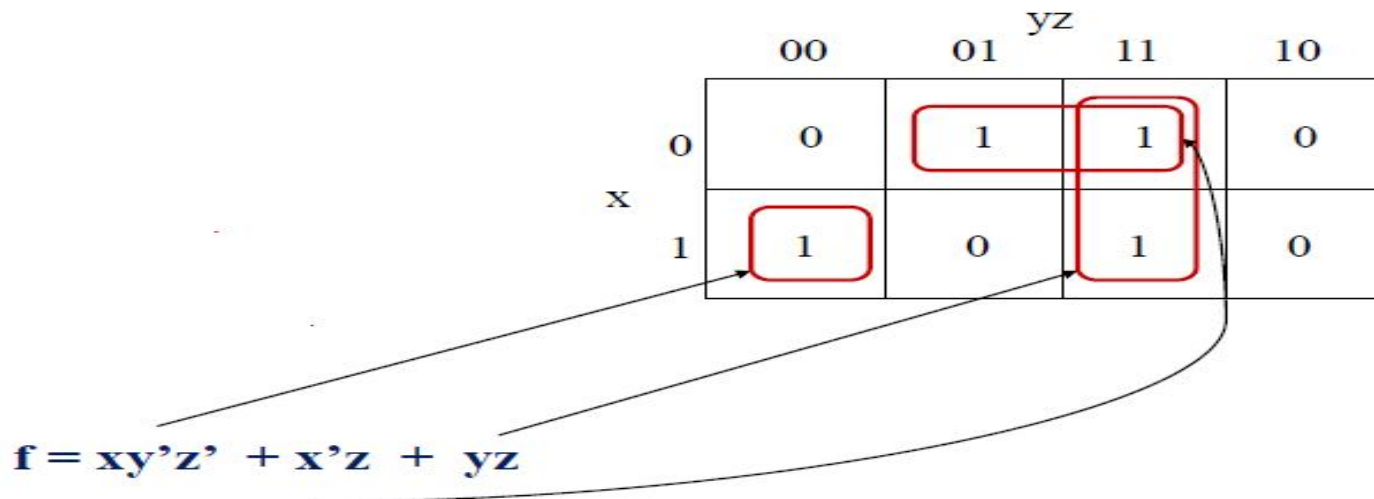
x	y	z	f(x,y,z)
0	0	0	f(000)
0	0	1	f(001)
0	1	0	f(010)
0	1	1	f(011)
1	0	0	f(100)
1	0	1	f(101)
1	1	0	f(110)
1	1	1	f(111)

		yz			
		00	01	11	10
x	0	f(000)	f(001)	f(011)	f(010)
	1	f(100)	f(101)	f(111)	f(110)

2x4 Karnaugh Map

		yz			
		00	01	11	10
x	0	0	1	1	0
	1	1	0	1	0

- Circle all 1 entries that, taken together, form a rectangle.
- Start with the largest rectangle, then proceed to smaller rectangle.
- Generally speaking, there will be more than one independent rectangle. each reflecting a different prime implicant.
- Each rectangle corresponds to a prime implicant term. Gathering all terms in SOP form,



4 variable map:

<u>w</u>	<u>x</u>	<u>y</u>	<u>z</u>	<u>f(w,x,y,z)</u>
0	0	0	0	f(0000)
0	0	0	1	f(0001)
0	0	1	0	f(0010)
0	0	1	1	f(0011)
0	1	0	0	f(0100)
0	1	0	1	f(0101)
0	1	1	0	f(0110)
0	1	1	1	f(0111)
1	0	0	0	f(1000)
1	0	0	1	f(1001)
1	0	1	0	f(1010)
1	0	1	1	f(1011)
1	1	0	0	f(1100)
1	1	0	1	f(1101)
1	1	1	0	f(1110)
1	1	1	1	f(1111)

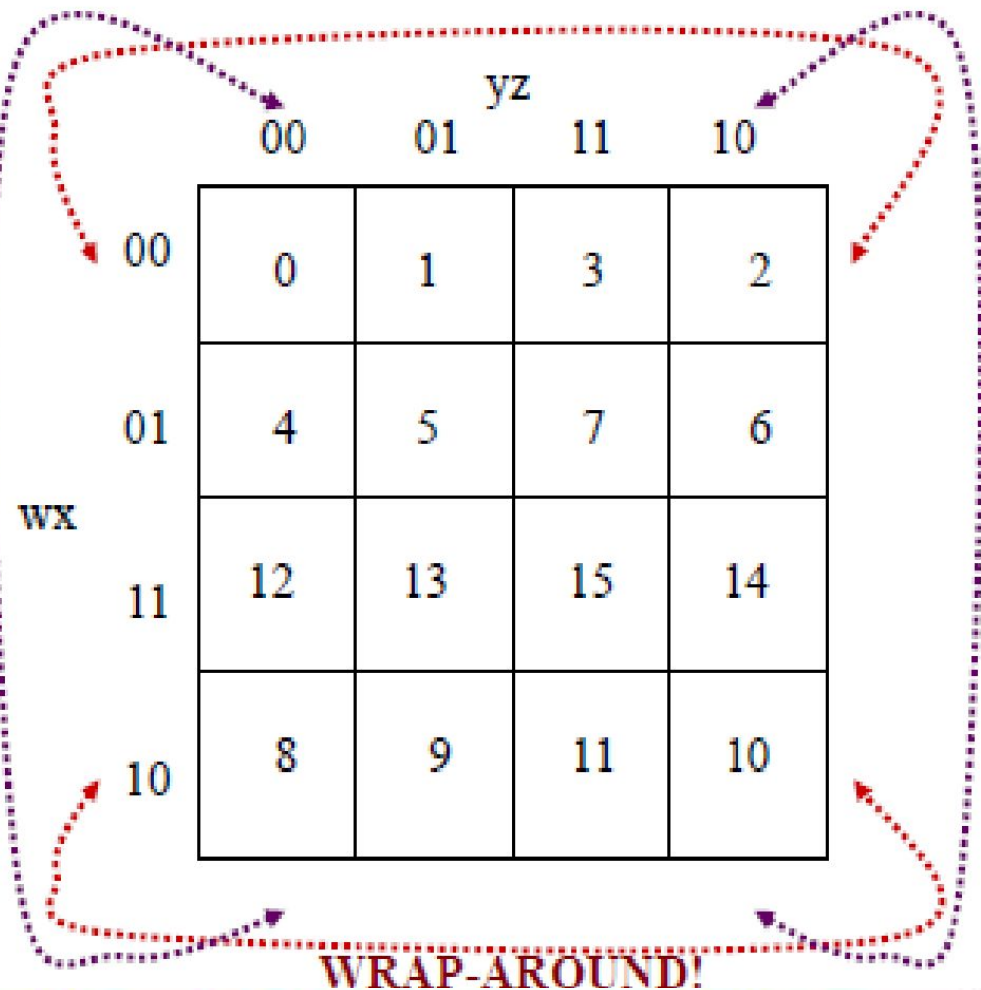
		yz			
		00	01	11	10
wx	00	f(0000)	f(0001)	f(0011)	f(0010)
	01	f(0100)	f(0101)	f(0111)	f(0110)
	11	f(1100)	f(1101)	f(1111)	f(1110)
	10	f(1000)	f(1001)	f(1011)	f(1010)

4 variable map:

w	x	y	z	f(w,x,y,z)
0	0	0	0	f(0000)
0	0	0	1	f(0001)
0	0	1	0	f(0010)
0	0	1	1	f(0011)
0	1	0	0	f(0100)

Note the way that both the row and column indices change by only 1 bit at a time.

1	0	1	1	f(1011)
1	1	0	0	f(1100)
1	1	0	1	f(1101)
1	1	1	0	f(1110)
1	1	1	1	f(1111)



This implies that two rows, or columns, whose indices differ by only 1 bit value, are adjacent.

2.13 Minimization of logic functions using K-map

Example: Minimize the below boolean equation using K-map

$$F = \overline{A}.\overline{B}.\overline{C}.\overline{D} + \overline{A}.B.\overline{C}.D + \overline{A}.B.\overline{C}.\overline{D} + A.\overline{B}.\overline{C}.\overline{D} + \overline{A}.\overline{B}.C.\overline{D} + A.\overline{B}.C.\overline{D} + \overline{A}.\overline{B}.\overline{C}.D$$

AB \ CD	00	01	11	10
00	1	1		1
01	1	1		
11				
10	1			1

$$F = \overline{B}.\overline{D} + \overline{A}.\overline{C}$$

2.14 Don't care conditions using K-map

- When constructing the terms in the simplification procedure, we can choose to either cover or not cover the don't care conditions.
- To distinguish the don't care conditions from 1's and 0's, an **X** will be used.
- Don't care conditions are part of function specification. They can be used for both sum-of-product and product-of-sum forms of functions.

$$f = \sum m(\dots) + \sum d(\dots) \quad f = \prod M(\dots) \prod D(\dots)$$

Example: Minimize the below boolean equation using K-map:

$$F = \sum m(1, 3, 7) + \sum d(0, 5)$$

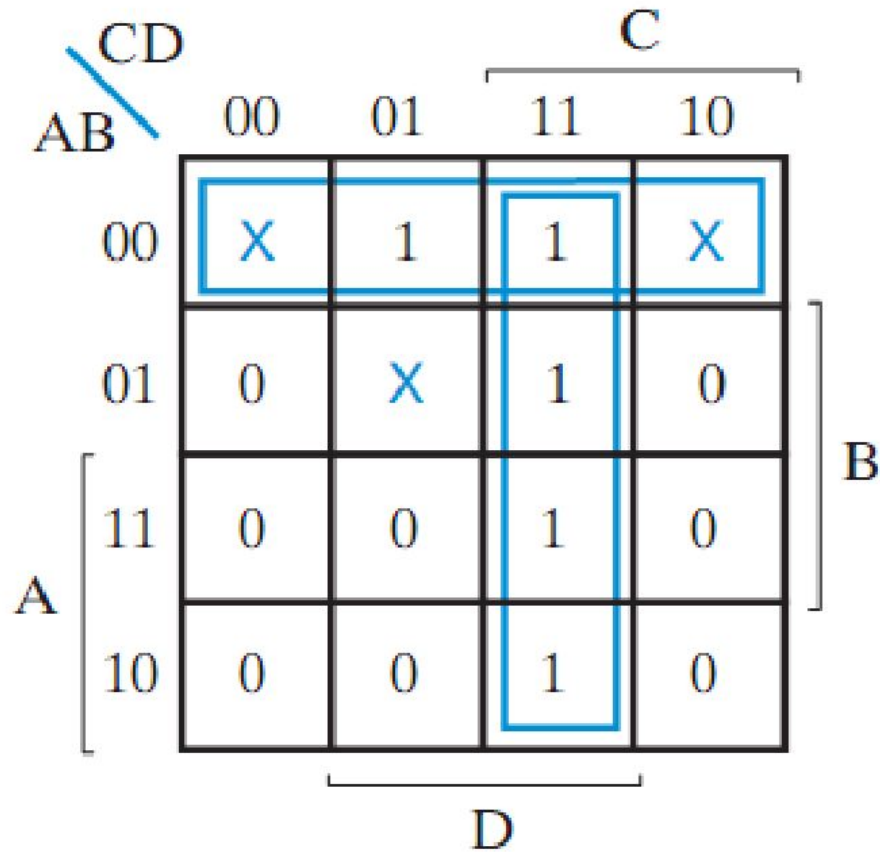
- Circle the x's that help get bigger groups of 1's (or 0's if POS).

Reduced form : $F = C$

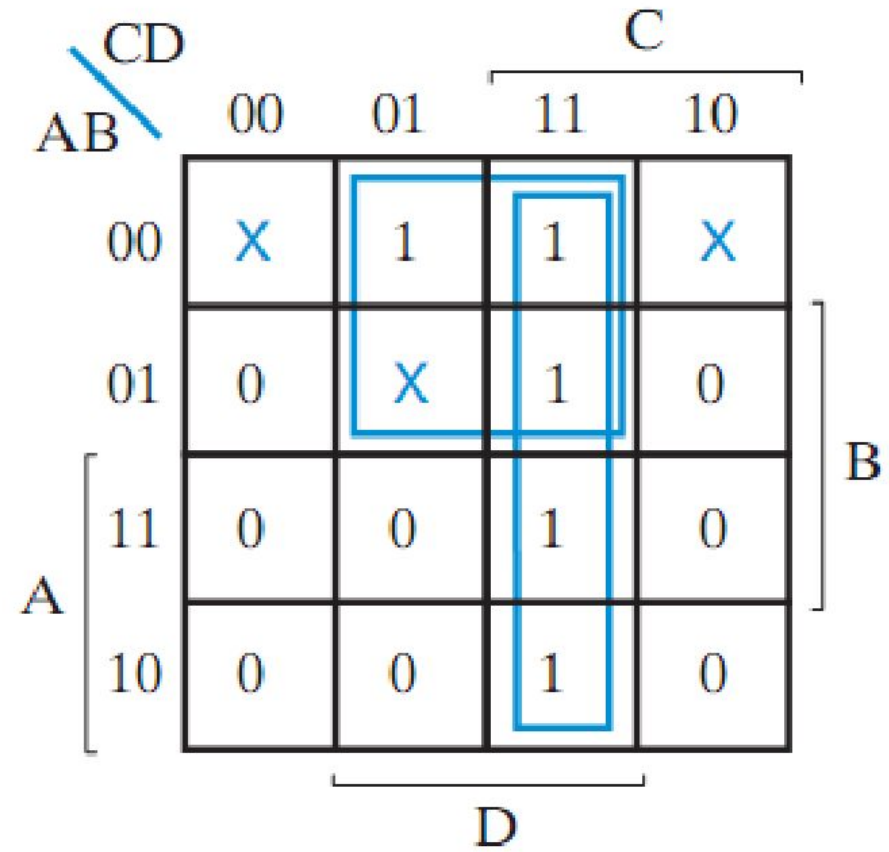
BC					
A		00	01	11	10
0	0	X	1	1	2
1	4		X	1	6

Example: Minimize the below boolean equation using K-map:

$$F(A, B, C, D) = \sum m(1, 3, 7, 11, 15) + \sum d(0, 2, 5)$$



(a) $F = CD + \bar{A} \bar{B}$



(b) $F = CD + \bar{A} D$

Here two solutions, both are right but second solution is right for circuit design.

2.15 Quine-McCluskey method of minimization (Tabular Method)

- Compute all prime implicants.
- Find a minimum expression for Boolean functions.
- No visualization of prime implicants.
- Can be programmed and implemented in a computer.

Example: Minimize the below boolean equation:

$$F(W, X, Y, Z) = \sum m(0, 3, 5, 6, 7, 10, 12, 13) + \sum d(2, 9, 15)$$

Step 1 : Divide all the minterms(and don't cares) of a function into groups

**For
Minterms:**












Minterm ID	W	X	Y	Z
0	0	0	0	0
3	0	0	1	1
5	0	1	0	1
6	0	1	1	0
7	0	1	1	1
10	1	0	1	0
12	1	1	0	0
13	1	1	0	1

**For don't
cares:**

Minterm ID	W	X	Y	Z
2	0	0	1	0
9	1	0	0	1
15	1	1	1	1









Groups	Minterm ID	W	X	Y	Z	Merge Mark
G0	0	0	0	0	0	
G1	2	0	0	1	0	
G2	3	0	0	1	1	
	5	0	1	0	1	
	6	0	1	1	0	
	9	1	0	0	1	
	10	1	0	1	0	
	12	1	1	0	0	
G3	7	0	1	1	1	
	13	1	1	0	1	
G4	15	1	1	1	1	

Step 2: Merge minterms from adjacent groups to form a new implicant table

Groups	Minterm ID	W	X	Y	Z	Merge Mark
G0	0	0	0	0	0	
G1	2	0	0	1	0	
G2	3	0	0	1	1	
	5	0	1	0	1	
	6	0	1	1	0	
	9	1	0	0	1	
	10	1	0	1	0	
	12	1	1	0	0	
G3	7	0	1	1	1	
	13	1	1	0	1	
G4	15	1	1	1	1	

Groups	Minterm ID	W	X	Y	Z
G0'	0, 2	0	0	d	0
G1'	2, 3	0	0	1	d
	2, 6	0	d	1	0
	2, 10	d	0	1	0
G2'	3, 7	0	d	1	1
	5, 7	0	1	d	1
	6, 7	0	1	1	d
	5, 13	d	1	0	1
	9, 13	1	d	0	1
	12, 13	1	1	0	d
G3'	7, 15	d	1	1	1
	13, 15	1	1	d	1

Step 3: Repeat step 2 until no more merging is possible

Groups	Minterm ID	W	X	Y	Z	Merge Mark
G0'	0, 2	0	0	d	0	
G1'	2, 3	0	0	1	d	
	2, 6	0	d	1	0	
	2, 10	d	0	1	0	
G2'	3, 7	0	d	1	1	
	5, 7	0	1	d	1	
	6, 7	0	1	1	d	
	5, 13	d	1	0	1	
	9, 13	1	d	0	1	
	12, 13	1	1	0	d	
G3'	7, 15	d	1	1	1	
	13, 15	1	1	d	1	

Groups	Minterm ID	W	X	Y	Z
G1''	2, 3, 6, 7	0	d	1	d
	2, 6, 3, 7	0	d	1	d
G2''	5, 7, 13, 15	d	1	d	1
	5, 7, 13, 15	d	1	d	1

Step 3: Repeat step 2 until no more merging is possible

Groups	Minterm ID	W	X	Y	Z	Merge Mark
G0''	0, 2	0	0	d	0	
G1''	2, 3, 6, 7	0	d	1	d	
	2, 10	d	0	1	0	
G2''	5, 7, 13, 15	d	1	d	1	
	9, 13	1	d	0	1	
	12, 13	1	1	0	d	

No more merging possible!

Step 4: Put all prime implicants in a cover table (don't cares excluded)

Minterm ID	$\overline{W} \overline{X} \overline{Z}$	$\overline{W}Y$	$\overline{X}Y\overline{Z}$	XZ	$WX\overline{Y}$	$W\overline{Y}Z$
0	1					
3		1				
5				1		
6		1				
7		1		1		
10			1			
12					1	
13				1	1	1

Step 5: Identify essential minterms, and hence essential prime implicants

Minterm ID	$\bar{W} \bar{X} \bar{Z}$	$\bar{W} Y$	$\bar{X} Y \bar{Z}$	XZ	$W X \bar{Y}$	$W \bar{Y} Z$
0	1					
3		1				
5				1		
6		1				
7		1		1		
10			1			
12					1	
13				1	1	1

 E.M.T E.P.I

Step 6: Add prime implicants to the minimum expression of F until all minterms of F are covered (Already covered)

Minterm ID	$\overline{W} \overline{X} \overline{Z}$	$\overline{W} Y$	$\overline{X} Y \overline{Z}$	XZ	$W X \overline{Y}$	$W \overline{Y} Z$
0	1					
3		1				
5				1		
6		1				
7		1		1		
10			1			
12					1	
13				1	1	1

 E.M.T E.P.I

So after simplification through QM method, a minimum expression for $F(W, X, Y, Z)$ is:

$$F(W, X, Y, Z) = \overline{W} \overline{X} \overline{Z} + \overline{W} Y + \overline{X} Y \overline{Z} + XZ + W X \overline{Y}$$

2.16 Variable Entered Maps:

- Due to the difficulty in managing K map exceeding 4 variables like 5 or 6 variables we have the technique called map entered variables[VEM].
- As we all know for n variables a K map has 2^n variables. To allow a smaller map to handle a larger number of variables that is to reduce the number of squares the concept of variable entered mapping was introduced.
- This is done by choosing one variable as map entered variable and thus including it in the k map along with the zeros, ones and the don't care conditions.

Example: Minimize the below boolean equation using K-map:

$$f(A, B, C) = \sum m_i(0,1,4,5,7)$$

A	B	C	Y
0	0	0	1
0	0	1	1
0	1	0	0
0	1	1	0
1	0	0	1
1	0	1	1
1	1	0	0
1	1	1	1

A	B	Y
0	0	1
0	1	0
1	0	1
1	1	C

Truth table considering C as MEV is

MEV K-map is

		B	
		0	1
A	0	1	0
	1	1	C

So grouping in our MEV K-map is as shown and equation is

		B	
		0	1
A	0	1	0
	1	1	C

Grouping in vem

		B	
		0	1
A	0	C	C
	1	0	1

$Y = \bar{A}C + AB$

		B	
		0	1
A	0	\bar{C}	\bar{C}
	1	1	0

$Y = A\bar{B} + \bar{A}\bar{C}$

		B	
		0	1
A	0	0	\bar{C}
	1	0	1

$Y = AB + B\bar{C}$

		B	
		0	1
A	0	\bar{C}	\bar{C}
	1	C	1

$Y = \bar{A}\bar{C} + AC + B\bar{C}$

you can group C and C's, 1 and C's, 1 and C's but your cannot group C and C's.

$$Y = \bar{B} + \bar{A}\bar{C}$$

Verification using K-map

BC					
A		00	01	11	10
	0	1	1	0	1
1	1	1	0	0	0

$$Y = \bar{B} + \bar{A}\bar{C}, \text{ verified!}$$

2.17 Realizing Logic Function with Gates:

Implement the below Boolean equation using logic gates:

$$F(A,B,C,D)=A'B'C'D'+A'B'C'D+A'BC'D+AB'C'D'+AB'C'D+AB'CD'+AB'CD'$$

AB \ CD	00	01	11	10
00	1	1	0	1
01	0	1	0	0
11	0	0	0	0
10	1	1	0	1

$$F(A,B,C)=B'C'+B'D'+A'C'D$$

So, we can see that function reduces to simple form. Now, we implement this derived function using gates which gives o/p exactly same as real function.

K-map Simplification Circuit:

