

# SIGNAL PROCESSING

## DEPARTMENT OF ELECTRICAL ENGINEERING

E 102      **Digital Signalbehandling**      EQ2300/ 2E1340

Final Examination 2007–12–15, 14.00–19.00

- Literature:** Hayes: *Statistical Digital Signal Processing and Modeling*  
or  
Proakis, Manolakis: *Digital Signal Processing*  
Bengtsson: *Complementary Reading in Digital Signal Processing*  
*Copies of the slides and two lecture hand outs*  
*Beta – Mathematics Handbook*  
*Collection of Formulas in Signal Processing, KTH*  
Unprogrammed pocket calculator.
- Notice:** Answer in Swedish or English.  
At most one problem should be treated per page.  
Motivate each step in the solutions (also for the multi-choice questions).  
Answers without motivation / justification carry no rewards  
Write your name and *personnummer* on each page.  
Write the number of solution pages on the cover page.  
  
The exam consists of five problems with a maximum of 10 points each.  
For a passing grade, 24 points are normally required.
- Contact:** Bhavani Shankar, Signal Processing, 08-790 84 35
- Results:** Will be reported within three working weeks at “My pages”.
- Solutions:** Will be available on the course homepage after the exam.
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***Good luck!***

- Don't forget to fill in the course evaluation form! Follow the link at the course WWW page.
- Don't forget to pick up your graded project report for Project 2 from STEX on 21 December 2007.

1. (a) In the class, we had considered zero padding as a tool to convert circular convolution into linear convolution. We shall now consider the converse. Let  $x(n)$  be a length 4 sequence (i.e,  $x(n) = 0, n > 3 \text{ or } n < 0$ ) and  $h(n)$  is a length 3 sequence. Define sequences

$$\begin{aligned} z_1 &= \{ \underset{\uparrow}{x(3)}, x(0), x(1), \dots, x(3) \} \\ z_2 &= \{ x(2), \underset{\uparrow}{x(3)}, x(0), x(1), \dots, x(3) \} \\ z_3 &= \{ x(1), x(2), x(3), x(0), \underset{\uparrow}{x(1)}, \dots, x(3) \} \end{aligned}$$

- i. Define  $y(n) = z_3 * h(n)$  (\* denotes Linear Convolution). Show that  $\{y(3), y(4), y(5), y(6)\} = x(n) \textcircled{4} h(n)$  ( $\textcircled{4}$  denotes 4 point circular convolution). (2p)

- ii. Can you extract  $x(n) \textcircled{4} h(n)$  from

- $z_1 * h(n)$
- $z_2 * h(n)$  (1p)

- (b) If the  $z$  transform of  $x(n) = \rho^n u(n), |\rho| < 1$  is  $X(z) = \frac{1}{1-\rho z^{-1}}$ , find the sequence whose  $z$  transform is,  $Y(z) = \frac{1}{(1-\rho z^{-1})^2}$ . (3p)

- (c) A student is asked to verify the DFT and IDFT operations. He correctly obtains the DFT of  $\{ \underset{\uparrow}{a}, \underset{\uparrow}{b}, \underset{\uparrow}{c}, \underset{\uparrow}{d}, \underset{\uparrow}{e} \}$  as  $\{ \underset{\uparrow}{A}, \underset{\uparrow}{B}, \underset{\uparrow}{C}, \underset{\uparrow}{D}, \underset{\uparrow}{E} \}$ . Instead of computing the IDFT, he computes the DFT of  $\{ \underset{\uparrow}{A}, \underset{\uparrow}{B}, \underset{\uparrow}{C}, \underset{\uparrow}{D}, \underset{\uparrow}{E} \}$  again. What would his answer be? (4p)

2. The received signal at a radar is given by,

$$y(n) = A e^{j2\pi f_0 n} + w(n)$$

where  $w(n)$  is a zero mean white noise with variance  $\sigma_w^2$ . A genie provides the following values for the auto-correlation function,  $r_{yy}(k) = E \{y(n+k)[y(n)]^*\}$ ,

$$\begin{aligned} r_{yy}(0) &= 3.0300 \\ r_{yy}(1) &= 1.5 (\sqrt{3} + i) \end{aligned}$$

From these data, determine the following using Pisarenko method,

- (a) frequency of operation,  $f_0$
- (b) Amplitude,  $A$
- (c) Noise variance,  $\sigma_w^2$

(10p)

3. For the following signal model,

$$y(n) = A \sin(\omega_0 n + \phi) + w(n), \quad n = 0, 1, 2, \dots, N - 1$$

where  $w(n)$  is a white noise, consider the problem of estimating the amplitude  $A$  and phase  $\phi$  of the signal ( $\omega_0$  is assumed to be known). This is a nonlinear estimation problem that can be solved using a nonlinear least squares approach. However, the signal model can be reparameterized using the following trigonometric rule

$$\sin(\alpha + \beta) = \sin(\alpha) \cos(\beta) + \cos(\alpha) \sin(\beta)$$

such that the estimation problem becomes linear in two new parameters.

- (a) Do the appropriate reparameterization and show how to calculate  $A$  and  $\phi$  from the new parameters.
- (b) Write the estimation problem as a linear least squares problem in the new parameters (use vectors and matrices). Find the least squares solution.
- (c) Simplify the least squares solution for  $\omega_0 = \pi/2$ .

(10p)

4. A filter has been implemented digitally as shown in Figure 1. The discrete-time input signal  $x(n)$  [Volts] is A/D converted (quantized) using quantization step size  $\Delta$  [Volts]. The signal is filtered using finite precision arithmetics (the precision is  $\Delta$ ). Finally, the signal is D/A converted back to analog discrete-time domain,  $y(n)$ , again using step size  $\Delta$  [Volts]. We wish to determine the unknown gain,  $k$ , and the

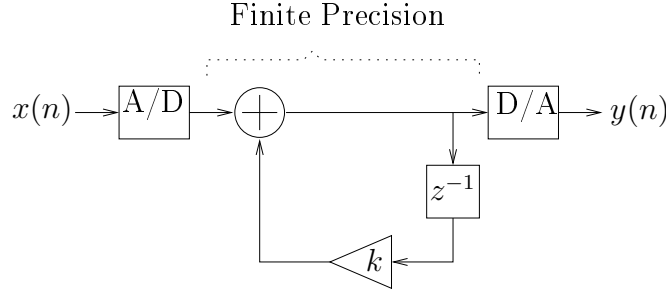


Figure 1: The digital filter.

quantization step-size,  $\Delta$ , by feeding the filter with a sinusoid

$$x(n) = \cos(\omega_0 n + \phi) \quad [\text{Volts}], \quad n = 0, \dots, N-1,$$

that has a known frequency  $\omega_0 = \pi/3$  [rad/sample], and an unknown phase,  $\phi$ , uniformly distributed in  $[0, 2\pi)$ . Figure 2 shows the estimated spectrum of  $y(n)$  using periodogram averaging:  $K = 100$  periodograms were averaged using  $L = 1000$  samples for each periodogram. The total number of output samples is  $N = KL$ .

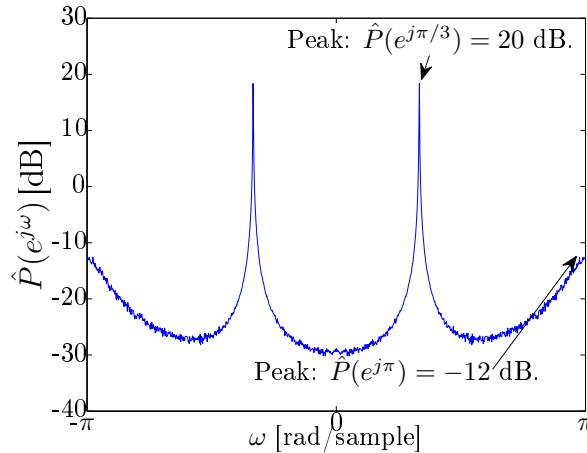


Figure 2: The averaged periodogram.

- (a) Determine the filter transfer function,  $H(z)$ , as a function of  $k$ , disregarding the quantization effects. (1p)
- (b) Show that the periodogram peak at frequency  $\omega_0$  can be approximated as

$$\hat{P}(e^{j\omega_0}) \approx \frac{L|H(e^{j\omega_0})|^2}{4}$$

*Hint:* Note that the quantization noise power is negligible at frequency  $\omega_0$ . Further more, you may assume that  $\frac{1}{L} \sum_{n=0}^{L-1} \exp(j\omega n) \approx 0$ , for all  $\omega \neq 0$ . (3p)

- (c) Using the result of part (b), obtain the value of  $k$ . (3p)
- (d) Using suitable data provided in Figure 2, estimate the quantization step size,  $\Delta$ . Assume that no overflow occurs in the summation. (3p)

5. Consider the system in Figure 3, where the input to the system is a sinusoidal signal with two odd overtones (harmonics). That is

$$x(n) = \sum_{k=1}^3 A_k \sin(2\pi(2k-1) f_{x,0} n)$$

where  $f_{x,0} = 1/20$ . Figure 4 shows the periodogram of the output  $y(m)$  calculated using  $N = 10^3$  data samples.

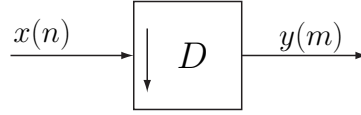


Figure 3: Downsampling

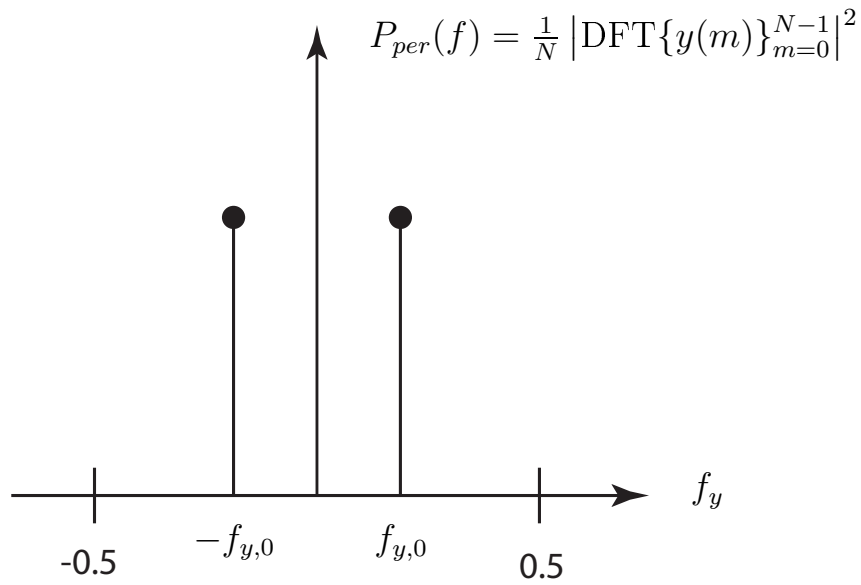


Figure 4: Periodogram of the down sampled signal.

Determine the smallest down sampling factor  $D$  that has been used in figure 3. What are the other possible values of  $D$ ? (10p)

*Hint:* Due to aliasing caused by down sampling, all signals have the same frequency after down sampling.