

DSP

8/28/15

frequency domain discrete time:

①

observe: complex exponentials $x[n] = e^{j\omega n}$ are "eigen vectors" of LTI there exists a scalar $\lambda \in \mathbb{C}$ that does not depend on time "n" s.t. $T_x = \lambda x$

proof: $y[n] = \sum_{k=-\infty}^{\infty} h[k] x[n-k] = \sum_{k=-\infty}^{\infty} x[k] e^{j\omega(n-k)}$ only 1

$$y[n] = e^{j\omega n} \sum_{k=-\infty}^{\infty} h[k] e^{j\omega k} = H(e^{j\omega})$$

so,

$$\boxed{\text{@ } \omega \quad y[n] = H(e^{j\omega}) e^{j\omega n}}$$

example:

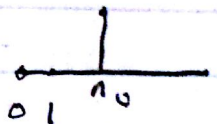
LTI

delay

$$y[n] = x[n-n_0]$$

 $n_0 \in \mathbb{N}$

$$h[n] = \delta[n-n_0]$$

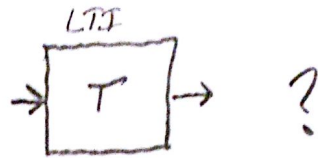
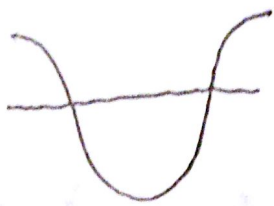


so,

$$H(e^{j\omega}) = \sum_{k=-\infty}^{\infty} h[k] e^{-j\omega k}$$

$$H(e^{j\omega}) = e^{-j\omega n_0}$$

and $|H(e^{j\omega})| = 1$ the phase encodes delay



- output will Δ amplitude
- shift = $\phi + \Delta H(e^{j\omega})$

sinusoidal response of LTI with a real impulse

$$x[n] = A \cos(\omega n + \phi) = \frac{A}{2} \left\{ e^{j(\omega n + \phi)} + e^{-j(\omega n + \phi)} \right\}$$

$$= \frac{A}{2} \left\{ e^{j\phi} e^{j\omega n} + e^{-j\phi} e^{-j\omega n} \right\}$$

$$T_x[n] = \frac{A}{2} e^{j\phi} T(e^{j\omega n}) + \frac{A}{2} e^{-j\phi} T(e^{-j\omega n})$$

$$= \frac{A}{2} e^{j\phi} H(e^{j\omega}) e^{j\omega n} + \frac{A}{2} e^{-j\phi} H(e^{-j\omega}) e^{-j\omega n}$$

$$H(e^{-j\omega}) = \sum_{k=-\infty}^{\infty} h[k] e^{j\omega k} = \left[\sum_{k=-\infty}^{\infty} h[k] e^{-j\omega k} \right]^*$$

$$T_x[n] = \frac{A}{2} e^{j\phi} H(e^{j\omega}) e^{j\omega n} + \frac{A^*}{2^*} (e^{j\phi})^* H^*(e^{j\omega}) (e^{j\omega n})^*$$

$$z + z^* = z \operatorname{Re}(z) \quad \text{so,}$$

$$T_x[n] = \operatorname{Re} \left(A e^{j\phi} H(e^{j\omega}) e^{j\omega n} \right)$$

$$H(e^{j\omega}) = |H(e^{j\omega})| e^{j\Delta H(e^{j\omega})} \quad \text{so,}$$

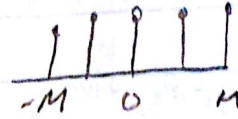
$$T_x[n] = A \operatorname{Re} \left(|H(e^{j\omega})| e^{j(\omega n + \phi + \Delta H(e^{j\omega}))} \right)$$

$$y[n] = A |H(e^{j\omega})| \cos(\omega n + \phi + \Delta H(e^{j\omega}))$$

frequency response "windowing"

1. moving average: $y[n] = \frac{1}{2M+1} \sum_{m=-M}^M x[n-m]$

$$h[n] = \frac{1}{2M+1} \sum_{m=-M}^M \delta[n-m]$$



"window"

"window" at zero is low pass filter

$$H(e^{j\omega}) = \sum_{k=-\infty}^{\infty} h[k] e^{-j\omega k} = \frac{1}{2M+1} \sum_{k=-M}^M e^{-j\omega k}$$

$$H(e^{j\omega}) = \left(\frac{e^{j\omega M}}{2M+1} \right) \left(\frac{e^{-j\omega(2M+1)} - 1}{e^{-j\omega} - 1} \right) = \left(\frac{e^{-j\omega(M+1)} - e^{j\omega M}}{e^{-j\omega} - 1} \right) \left(\frac{1}{2M+1} \right)$$

$$= \left(\frac{1}{2M+1} \right) \left(\frac{e^{-j\omega/2} (e^{-j\omega(M+1/2)} - e^{j\omega(M+1/2)}) / 2}{e^{-j\omega/2} (e^{-j\omega/2} - e^{j\omega/2}) / 2} \right)$$

$$= \frac{1}{2M+1} \frac{\sin[(M+1/2)\omega]}{\sin(\omega/2)} \quad \text{when } \omega \neq 0$$

$$\sin(\omega/2) \approx \omega/2$$

$$\boxed{|H(e^{j\omega})| \approx \frac{1}{2M+1} \frac{\sin((M+1/2)\omega)}{\omega/2}}$$

DIRICHLET KERNEL

2π periodic

