

# Problem 4 Ex 10.6

## Problem

Ex. 7

Three periodic control tasks  $A$ ,  $B$  and  $C$  are executed on a CPU. The tasks have the following characteristics:

Task	$T_i$	$D_i$	$C_i$
A	4	4	1
B	5	5	2
C	$x$	$x$	3

- (a) Suppose that all of the CPU capability is available for  $A$ ,  $B$  and  $C$ . What is the minimum task period  $x$  for the task  $C$  to guarantee that all tasks  $A$ ,  $B$  and  $C$  are schedulable under earliest deadline first (EDF) scheduling? [2p]
- (b) Suppose  $x = 10$ . Try to schedule the above tasks using earliest deadline first (EDF) scheduling algorithm, and verify your result from (a). What is the worst response time for control task  $A$ ,  $B$  and  $C$ ? [3 p]
- (c) Suppose  $x = 10$ , again. Are the tasks  $A$ ,  $B$  and  $C$  schedulable under rate monotonic (RM) scheduling? Try to answer by computing the worst response time for control task  $A$ ,  $B$  and  $C$ . [3 p]
- (d) Comment on your results from (b) and (c). [2 p]

Ex 7.3

Ex 7.4

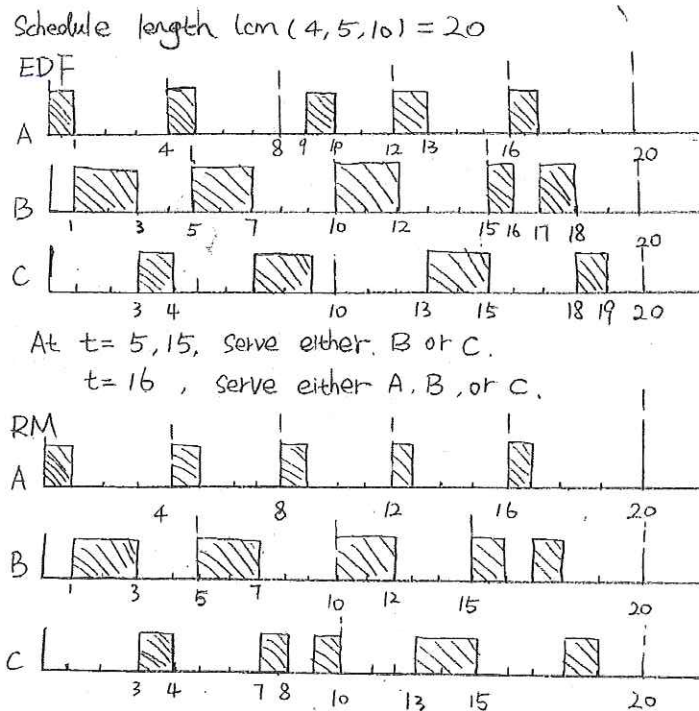
## Solution

- (a) In the case where EDF is used, the schedulability bound is one (sufficient and necessary). So

$$U = \sum_{i=1}^3 \frac{C_i}{T_i} = \frac{1}{4} + \frac{2}{5} + \frac{3}{x} \leq 1 \Rightarrow x \geq \frac{60}{7},$$

which gives the minimal task period for for task  $C$  is  $x = \frac{60}{7} \approx 8.57$ .

- (b) Tasks  $A$ ,  $B$  and  $C$  should be schedulable by EDF as  $10 > \frac{60}{7}$  from (a). The schedule length is given by  $\text{lcm}(T_A, T_B, T_C) = \text{lcm}(4, 5, 10) = 20$ . Below is the corresponding schedule:



# Problem 1

V22

Consider the model of normalized DC motor

$$G(s) = \frac{1}{s(s+1)}$$

- (a) find the state-space representation of  $G(s)$
- (b) Sample  $G(s)$  with sampling period  $h = \ln 2 \approx 0.7$
- (c) Determine the linear state-feedback controller

$$u(k) = [-l_1 \quad -l_2] x(k)$$

such that the characteristic polynomial of the closed-loop system is  $z^2 - 0.6z + 0.3 = 0$

- (d) Design a full-state observer for the state based on the output.

**Solution**

- (a)  $G(s) = \frac{1}{s^2+s}$ , the ~~observable~~ canonical form E4.1

$$\dot{x}_1 = \begin{bmatrix} 0 & 1 \\ 0 & -1 \end{bmatrix} x + \begin{bmatrix} 0 \\ 1 \end{bmatrix} u$$

$\quad \quad \quad \stackrel{\text{red}}{=} A \quad \quad \quad \stackrel{\text{red}}{=} B$

$$y = \begin{bmatrix} 1 & 0 \end{bmatrix} x$$

$\quad \quad \quad \stackrel{\text{red}}{=} C$

- (b) Sample the system by  $h = \ln 2$

$$x(k) = e^{Ah} x(k-1) + \int_0^h e^{As} ds \cdot B u(k-1)$$

$\quad \quad \quad \stackrel{\text{red}}{=} \Phi \quad \quad \quad \stackrel{\text{red}}{=} T$

$$y(k) = C x(k)$$

where

$$e^{Ah} = \mathcal{L}^{-1} \left\{ \begin{bmatrix} s & -1 \\ 0 & s+1 \end{bmatrix} \right\} \bigg|_{t=h} = \frac{1}{s(s+1)} \begin{bmatrix} s+1 & 1 \\ 0 & s \end{bmatrix}$$

$\quad \quad \quad \stackrel{\text{red}}{=} E1.2$

$$= \begin{bmatrix} \frac{1}{s} & \frac{1}{s(s+1)} \\ 0 & \frac{1}{s+1} \end{bmatrix} \underbrace{\mathcal{L}^{-1}}_{\substack{\text{red} \\ (1 \sim \frac{1}{s}, e^{at} \sim \frac{1}{s-a})}} \begin{bmatrix} 1 & 1 - e^{-t} \\ 0 & e^{-t} \end{bmatrix}$$

$$\text{Let } t = \ln 2 \Rightarrow e^{Ah} = \begin{bmatrix} 1 & 0.5 \\ 0 & 0.5 \end{bmatrix}$$

$$\int_0^h e^{As} ds = \int_0^h \begin{bmatrix} 1 & 1-e^{-\tau} \\ 0 & e^{-\tau} \end{bmatrix} d\tau = \begin{bmatrix} h & h-1+e^{-h} \\ 0 & 1-e^{-h} \end{bmatrix}$$

$$\int_0^h 1 d\tau = h, \quad \int_0^h (1-e^{-\tau}) d\tau = h - \int_0^h e^{-\tau} d\tau \\ = h - (-e^{-\tau}) \Big|_0^h = h - (1-e^{-h})$$

$$\text{Let } h = \ln 2 \approx 0.7 \Rightarrow \begin{bmatrix} 0.7 & 0.2 \\ 0 & 0.5 \end{bmatrix}$$

$\Rightarrow$  state-space model

$$x(k) = \begin{bmatrix} 1 & 0.5 \\ 0 & 0.5 \end{bmatrix} x(k-1) + \begin{bmatrix} 0.7 & 0.2 \\ 0 & 0.5 \end{bmatrix} \begin{bmatrix} 0 \\ 1 \end{bmatrix} u(k-1)$$

$$= \begin{bmatrix} 1 & 0.5 \\ 0 & 0.5 \end{bmatrix} x(k-1) + \begin{bmatrix} 0.2 \\ 0.5 \end{bmatrix} u(k-1)$$

$$y(k) = [1 \quad 0] x(k)$$

(c) the closed-loop system is obtained by

$$u(k-1) = [-l_1 \quad -l_2] x(k-1)$$

$$\Rightarrow x(k) = \begin{bmatrix} 1-0.2l_1 & 0.5-0.2l_2 \\ -0.5l_1 & 0.5-0.5l_2 \end{bmatrix} x(k-1) \quad \text{--- } G$$

Thus the characteristic polynomial is given by

$$z^2 + \left( \frac{l_1}{2} + \frac{l_2}{5} - \frac{3}{2} \right) z + \left( \frac{3l_2}{20} - \frac{l_1}{2} + \frac{1}{2} \right) = 0 \quad \text{--- } \det(zI - G)$$

$\text{--- } 0.6 \quad \quad \quad \text{--- } 0.3$

implies  $l_1 = 1, l_2 = 2$

(d) Full-order observer: E 3.11 (b)

The estimation error

$$\hat{\tilde{x}}(k+1|k) = (\Phi - KC) \hat{x}(k|k-1)$$

Let  $K = \begin{bmatrix} k_1 \\ k_2 \end{bmatrix}$

$$\begin{aligned} \Phi - KC &= \begin{bmatrix} 1 & 0.5 \\ 0 & 0.5 \end{bmatrix} - \begin{bmatrix} k_1 \\ k_2 \end{bmatrix} \begin{bmatrix} 1 & 0 \end{bmatrix} \\ &= \begin{bmatrix} 1-k_1 & 0.5 \\ -k_2 & 0.5 \end{bmatrix} \end{aligned}$$

the characteristic equation is given by

$$z^2 + (k_1 - 1.5)z + 0.5 - 0.5k_1 + 0.5k_2 = 0$$

If we set  $\beta_1 = \beta_2 = 0$  (deadbeat observer)

$$k_1 - 1.5 = 0$$

$$0.5 - 0.5k_1 + 0.5k_2 = 0$$

$\Rightarrow$

$$k_1 = 1.5$$

$$k_2 = 0.5$$

Thus the observer is given by

$$\hat{x}(k+1|k) = \underline{\Phi} x(k|k-1) + \underline{\Gamma} u(k) + \underline{K} [y(k) - \underline{C} \hat{x}(k|k-1)]$$



As a result,  $V$  can serve as a common Lyapunov function by choosing  $d > 0$ . Thus the system is globally asymptotically stable

$$(b) \quad A_3 = \begin{bmatrix} -3 & 1 \\ k_1 - 1 & 3 + k_2 \end{bmatrix}$$

$$A_4 = \begin{bmatrix} -1 & 3 \\ k_1 - 3 & k_2 - 5 \end{bmatrix}$$

$$V = x^T \begin{bmatrix} d_1 & 0 \\ 0 & d_2 \end{bmatrix} x$$

$$\Rightarrow A_3^T P + P A_3 = \begin{bmatrix} -6d_1 & d_1 + (k_1 - 1)d_2 \\ d_1 + (k_1 - 1)d_2 & 2(3 + k_2)d_2 \end{bmatrix}$$

$$A_4^T P + P A_4 = \begin{bmatrix} -2d_1 & 3d_1 + (k_1 - 3)d_2 \\ 3d_1 + (k_1 - 3)d_2 & 2(k_2 - 5)d_2 \end{bmatrix}$$

If  $d_1 = d_2 = d > 0$ ,  $k_1 = 0$ ,  $k_2 < -3$ , then it serves as a global Lyapunov function.

## Problem 2

EX 11

Consider the following switched system with  $i \in \{1, 2\}$

$$\dot{x}(t) = \begin{bmatrix} -2 & 2 \\ -2 & -1 \end{bmatrix} x(t) + \lambda_i \begin{bmatrix} 1 & 1 \\ -1 & -4 \end{bmatrix} x(t) + b u(t)$$

a) show that the system is asymptotically stable  
when  $\lambda_1 = 0, \lambda_2 = 1$  and  $b = 0$

~~b)~~ design feedback controller (no time)

$$u(t) = [k_1, k_2] x(t)$$

which stabilizes the system when  $\lambda_1 = -1, \lambda_2 = 1, b = [0 \ 1]^T$

Solution

(a) when  $b = 0$  and  $\lambda_1 = 0$

$$\dot{x}(t) = \begin{bmatrix} -2 & 2 \\ -2 & -1 \end{bmatrix} x = A_0 x$$

when  $\lambda_1 = \lambda_2 = 1$

$$\begin{aligned} \dot{x}(t) &= \begin{bmatrix} -2 & 2 \\ -2 & -1 \end{bmatrix} x + \begin{bmatrix} 1 & 1 \\ -1 & -4 \end{bmatrix} x \\ &= \begin{bmatrix} -1 & 3 \\ -3 & -5 \end{bmatrix} x = A_1 x \end{aligned}$$

In order to identify a global Lyapunov function, we assume a quadratic form

$$V = x^T \begin{bmatrix} d & 0 \\ 0 & d \end{bmatrix} x = x^T P x$$

By requiring  $A_0^T P + P A_0$  be negative definite,

$$A_0^T P + P A_0 = \begin{bmatrix} -4d & 0 \\ 0 & -2d \end{bmatrix} < 0 \Rightarrow d > 0$$

By requiring  $A_1^T P + P A_1$  be negative definite,

$$A_1^T P + P A_1 = \begin{bmatrix} -2d & 0 \\ 0 & -10d \end{bmatrix} < 0 \Rightarrow d > 0$$