States to a Circuit

Kunal Adhia and Abhishek Patel

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Contents

L	Introduction	
2	Background	2
3	The States Diagram	2
1	Simulation	3
5	From States to Circuit 5.1 The Quine McCluskey Algorithm	3 3 8
3	The Final Circuit	9
7	Additional Features 7.1 WinCupl Programming	10 10
3	What's Next	11
)	Bibliography	12

1 Introduction

Memory devices and sequential logic are underlying concepts of almost every electronic device today, from digital clocks and timers to computers and cell phones. There are many pieces of software which can make the process of designing these sometimes sophisticated circuits simpler, but not one can convert a pseudo-design into the final circuit. However, Kunal Adhia and Abhishek Patel, sophomores at the Middlesex County Academy for Science, Mathematics, and Engineering Technologies in Edison, New Jersey, under electrical engineering instructor and supervisor Enzo Paterno, have completed the task and have designed a product which electrical engineers from all levels and categories can use to eliminate the tedious process of conversion. A simple states diagram can

now be quickly and simply made, simulated, and converted into a circuit which implements a D flip-flop, one of the most basic memory devices. The program is called States to a Circuit, or STaC, and was created with Python, a high-level programming language.

2 Background

Just as there are organizers and planners for writing an essay, a states diagram is the first step in designing a fully functional circuit using sequential logic. The diagram consists of "states", each with an assigned binary value. The states are the outputs of the circuit, and are dependent on the input. However, a sequential logic design isn't as simple as a combinational logic design, where there is one output for each input. A sequential logic design is one in which the next state, or output, is determined by both the input and the current state. Since the current state is necessary for the feedback design, a form of electronic memory is required. These "flip-flops" hold the current state until a change is inputted. However, the changes are synchronized by a clock, or a square wave alternating between a logical high state and a logical low state with a constant frequency. The current state is linked to the next state with a transition, which represents the input. If the transition and current state conditionals are both satisfied, the next state will occur. These circuits are very tedious to create, but since the process is repetitious, it can be programmed.

3 The States Diagram

The states diagram is a simple visual or textual pseudo-design of how the output reflects the input. STaC provides a graphical user interface in which the user can visually create, save, and view their states diagrams. The states are positioned in a regular polygon, and the transitions are represented by arrows, black or red depending on the input condition. These states can be moved around to an orientation of the user's choice with STaC's drag-and-drop capability. One can add and delete states, add and delete transitions, set the initial condition, or the first output state when the circuit is active, and relabel the states for a more complete understanding of their states diagram.

A common example of a states diagram which STaC can easily depict is a counter. For example, the last digit of a clock continuously progresses through the digits 0-9, and a timer counts backwards, from 9-0. This is actually much more complicated to create a circuit for than it seems, but it can now be done with ease using the STaC states diagram:

The states are represented by the yellow circles, the labels, which can be changed, are the outputs, and the arrows represent the transitions. The above design can be used both for the last digit of a clock or a timer. To create a clock, the input condition is set to high, and to create a timer, the input condition is set to low. The arrow points from the current state to the next state on

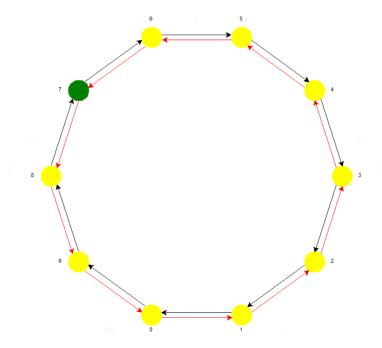


Figure 1: A Modulo 10 Up/Down Counter Created with STaC

the condition of the arrow. The green state signifies the initial state. When the circuit is activated, the green state will be visited first, and the circuit will operate from there.

4 Simulation

If a clock has a frequency of one Hertz, the flip-flop will receive the clock signal every second. The simulation option in STaC allows one to enter a data input and view the resulting output sequence to determine if the output matches the user's desires. The user can track the sequence and easily understand any errors he or she may have made before converting the states diagram to the circuit.

5 From States to Circuit

5.1 The Quine McCluskey Algorithm

Karnaugh Maps are a very effective way to minimize Boolean expressions, but only under minimal variables. These maps can also be prone to mistakes, as they rely on visual patterns to find prime implicants. Overall, they do not provide an algorithm to be implemented.

The Quine-McCluskey Algorithm, developed by W.V. Quine, and later expanded on by Edward J. McCluskey, works for any number of variables, and can be proven and explained by the following example.

$$f(d, c, b, a) = \sum (0, 1, 2, 5, 6, 7, 8, 9, 10, 14)$$

We can take this 4-bit summation of minterms and write each one out with its 4-bit binary counterpart. In this example:

0	0000
1	0001
2	0010
5	0101
6	0110
7	0111
8	1000
9	1001
10	1010
14	1110

Table 1: Binary Representation of Minterms

Number of 1s	Minterms	Decimal
Group 1	0000	0
Group 2	0001	1
	0010	2
	1000	8
Group 3	0101	5
	0110	6
	1001	9
	1010	10
Group 4	0111	7
	1110	14

Table 2: Groupings Pass 1

In this algorithm, the first way of solving an unsimplified canonical expression is to group it by the number of '1's in each minterm. Upon further inspection, we see that there are 4 distinct groups in this table.

The goal of the first step of this method, called "Finding Prime Implicants", is achieved by grouping together minterms which differ by one bit.

A few examples:

- 0000 0001
- 0010 0110
- 0111 1111

The pattern soon emerges that using the groups formed earlier, each minterm can only be combined with a term from an adjacent group. if we expand each possible combination of the first group, we end up with

0000	0001
0000	0010
0000	1000

Table 3: Combination Pattern of Group 1

As we do this for each group we can simplify each equation to whatever the binary representation is. Group 1 also does not compare with Group 3, as they are 2 bits apart, and would no have only 1 different bit.

After we expand through every group, the minterm list would be:

X = "Don't Care" (can be either 1 or 0) Values

Minterms	Binary
0,1	000X
0,2	00X0
0,8	X000
1,5	0X01
1,9	X001
2,6	0X10
2,10	X010
8,9	100X
8,10	10X0
5,7	01X1
6,7	011X
6,14	X110
10,14	1X10

Table 4: Combinations After Pass 1

We then take a second pass, completing the same task on this group of numbers

Group	Decimal	Minterms
Group 1	000X	0,1
	00X0	0,2
	X000	0,8
Group 2	0X01	1,5
	X001	1,9
	0X10	2,6
	X010	2,10
	100X	8,9
	10X0	8,10
Group 3	01X1	5,7
	011X	6,7
	X110	6,14
	1X10	10,14

Table 5: Groupings Pass 2

Thus, taking the combinations of all the minterms with the adjacent groups, we minimize the equations to:

If any combinations yield A,B,C,D and A,B,D,C, these can be simplified into 1 as either of the two, by the Cumulative Property.

Minterms	Binary
0,1,8,9	X00X
0,2,8,10	X0X0
2,6,10,14	XX10

Table 6: Combinations After Pass 2

Now there can be no more combinations, as none of the grouping vary by only 1 bit. Therefore, we can produce an SOP to implement the final solution. The MSB, or Most Significant Bit, is usually set to be the highest letter value, so in this case, that would be D. We substitute the final minterm groupings with letters instead of bits, omitting the Xs. For example, X00X would represents C'B', as 1111 would be DCBA. We also have to remember to include the groups from before that did not make a match when combined at all, which in this example would be the rows with the Minterms (1,5), (5,7), and (6,7).

Turning each of these binary values into letters would result in:

$$f(d, c, b, a) = BA' + C'A' + C'B' + D'CB + D'CA + D'B'A$$

We then need to simplify this further, as there are many redundant terms. We know that the minterms are the following:

Row	Minterms
1	0,1,8,9
2	0,2,8,10
3	2,6,10,14
4	1,5
5	5,7
6	6,7

Table 7: Minterms

As we parse through the list of minterms, we see that there is only one occurrence of 9 and 14, which proves that the product of minterms in that row will be needed. Since Row 1 and Row 3 cover minterms 0,1,2,6,8,9,10, and 14, these are no longer needed in the final SOP.

The new table is as follows:

Row	Minterms
1	0,1,8,9
2	0,2,8,10
3	2,6,10,14
4	1,(5)
5	(5),(7)
6	6,(7)

Table 8: Minterms Remaining - ()

As we can see, Row 2 has been removed, and is therefore not in the final Sum. Now, only 2 minterms are left to be implemented, which are 5 and 7. We see that Row 5 covers both of them, and that finishes the SOP.

The Final Simplified Equation is:

$$f(d, c, b, a) = C'B' + BA' + D'CA$$

There is one more problem, however. If there are no essential prime implicants (when there are no single occurrence minterms), we have to use Petrick's Method.

5.2 Petrick's Method

If there are no essential implicants, this method solves the problem of redundant terms. First, we use the number of the rows to create a POS (Product of Sums). For example, lets say we have:

$$f(c,b,a) = \sum (0,1,2,5,6,7)$$

We make the chart and simplify using the 1 bit difference rule, and obtain:

Row	Minterms
1	0,1
2	0,2
3	1,5
4	2,6
5	5,7
6	6,7

Table 9: Minterms

We then create Sums for each of the Minterms, as 0 occurs in Row 1 and

Row 2, the Sum would be (R1 + R2). This continues until we iterate through all the minterms, and finally reach a final equation.

$$f(c,b,a) = (R1+R2)(R1+R3)(R2+R4)(R3+R5)(R4+R6)(R5+R6)$$

Ordered by the Commutative Property, for aesthetics:

$$f(c, b, a) = (R1 + R2)(R1 + R3)(R4 + R2)(R4 + R6)(R5 + R3)(R5 + R6)$$

Then, by the law of $(X + Y)(X + Z) = (X + YZ)$

We deduce this equation to be:

$$f(c, b, a) = (R1 + R2R3)(R4 + R2R6)(R5 + R3R6)$$

Then, by the Distributive Property:

$$f(c,b,a) = (R1R4R5) + (R1R2R5R6) + (R2R3R4R5) + (R2R3R5R6) + (R1R3R4R6) + (R1R2R3R6) + (R1R2R3R6) + (R2R3R4R6) + (R2R3R4R6) + (R2R3R4R6) + (R2R3R4R6)$$

Now that it is in SOP form, we want to find the ones with the least number of variables. This leaves us with:

$$f(c, b, a) = (R1R4R5) + (R2R3R6)$$

This tells us that to get the SOP, we must include the equations from Row 1, 4 and 5, or the equations from Row 2, 3 and 6.

Substituting the equations in, we can have either of two equations:

$$f(c, b, a) = (C'B') + (BA') + (AC)$$
 or
$$f(c, b, a) = (A'C') + (AB') + (BC)$$

Either of these two equations work as a final SOP, and both are simplified with the same number of terms in each.

6 The Final Circuit

A sequential logic consists of two major components: the flip-flops and the gates. The flip-flops hold the present state until the next state is activated on the clock signal. The gates take the current state and determine the flip-flops' input, depending on the next state. We chose the D flip-flop, the most basic and widely used flip-flop. A high input sets the memory to high, and vice versa. The logic used was sum-of-product, or SOP, simplification. Taking the final simplified expression, worked out by the Quine-McCluskey algorithm,

each input is a sum of a series of products of the outputs. If at least one of the products is true, the sum will also be true. To draw the circuit, we made each of the flip-flops outputs, along with their complements, the input signal and complement, and the clock signal a bus. To access these signals, a wire simply had to be connected to the bus. This allowed us to avoid a messy circuit and easily access one signal multiple times.

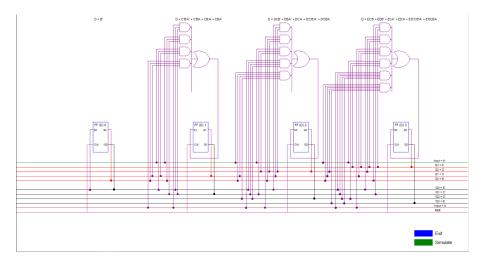


Figure 2: A Modulo 16 Up/Down Circuit Created with STaC

7 Additional Features

7.1 WinCupl Programming

In addition to transforming a states diagram into a circuit, STaC can convert a states diagram to equations which can be used to code a programmable logic device (PLD), which can be used in real-life circuits. For example, WinCupl, an application to program a PLD, can be programmed to implement a sequential logic design. This requires the use of Boolean equations for each output. Our program can create the .pld using the derived equations in WinCupl format, so all the user needs to do is compile the code and burn the PLD. This empowers the user to take an electronic plan and quickly insert it into a real-life circuit.

```
Name STaC;
PartNo 00;
Date 01/27/2016
Revision 01;
Designer STaC;
Company STaC;
Assembly None;
```

```
Location ;
Device g22v10 ;

/*Input Pins*/
PIN 2 = a;

/*Output Pins*/
PIN 14 = b;
PIN 15 = c;
PIN 16 = d;
PIN 17 = e;

/*Equations*/

b.D = !b ;
c.D = !e&!c&b&!a # c&!b&!a # c&b&a # d&!c&!b&a # e&!b&a ;
d.D = d&c&a # d&!c&b # d&!b&!a # e&!b&!a # e&!b&a ;
e.D = !e&!d&!c&!b&a # d&c&b&!a # e&!b&!a # e&b&a ;
```

The user is also given the option to simulate the circuit to visualize the binary output of each of the D flip-flops, given a one second clock period.

8 What's Next

The STaC application is a work-in-progress, as there are so many uses of this product. What we have right now is just the skeleton of what we plan to add to the application in the future.

The graphical user interface can only support up to sixteen states due to the fact that the circuit can be so sophisticated, it may not fit onto the window. However, we plan to eliminate the current limit on the number of states and make the circuit more than just a simple window. We plan create the circuit so that the user can scroll to capture the full view of the circuit.

Additionally, we plan to allow the user the freedom to change the binary output of any state of their choice. This would eliminate the number of independent states in the diagram, allowing for a more organized diagram.

One of the major changes we plan to implement is the option to have multiple inputs. Having only one input for transitions is not very practical for sophisticated designs, as a sequential design can be affected by more than one input. Having the option to use multiple inputs would empower the user to create just about any sequential design possible, making our application incredibly powerful and useful.

Another important addition we could implement is a non-graphic textual states creator. The visual diagram can be condensed into one numerical array which the user can manually type out to create the same circuit. This would be beneficial for programmers who intend to create a series of circuits automatically, or professionals who understand what the states diagram means numerically. We intend to provide an option in which a user can create a text file, upload it to the STaC application, and have the circuits and simplified equations created without graphically creating the states diagram. All of these future proposals demonstrate the power of STaC.

9 Bibliography

- [1]329, CMPUT. "Quine-McCluskey." Quine-McCluskey: An Algorithm. N.p., n.d. Web.
- [2] Adams, Douglas, and Eoin Colfer. "The Hitchhiker's Guide to the Galaxy." Google Books. Pan Books, n.d. Web. 26 Jan. 2016.