

## Blackboard Notes for lecture - 2

slide 6:

To figure out the component over  $V_{be}$ , we need to find out the relationship between  $i_b$  and  $V_{be}$  in this circuit branch.

slide 7:

We know that:

$$i_b = \frac{i_c}{\beta} = \frac{I_s}{\beta} \exp\left(\frac{V_{BE}}{V_T}\right)$$

therefore, the component over  $V_{be}$  is a PN diode.

slide 13:

When a small perturbation is added in the input side:

$$V_{in} = V_m \sin \omega t, \text{ then}$$

$$I_c = I_s \exp\left(\frac{V_0 + V_{in}}{V_T}\right) = \underline{I_s \exp\left(\frac{V_0}{V_T}\right)} \cdot \exp\left(\frac{V_m \sin \omega t}{V_T}\right)$$

$$= I_{c0} \cdot \exp\left(\frac{V_m \sin \omega t}{V_T}\right)$$

$$\approx I_{c0} \cdot \left(1 + \frac{V_m \sin \omega t}{V_T}\right)$$

$$= I_{c0} + \frac{I_{c0}}{V_T} V_m \sin \omega t$$

$$= I_{c0} + g_m V_m \sin \omega t$$

$$= \underline{I_{c0} + g_m V_{in}}$$

~~we~~ we use an approximation here:

$$e^x \approx 1 + x, \text{ when } x \ll 1$$

slide 14:

according to  $I_c = I_{co} + g_m V_{in}$ ,

the circuit is therefore divided into two parts.

slide 15:

In small-signal model:

$$i_B = \frac{g_m V_{in}}{\beta}$$

$$R = \frac{V_{in}}{i_B} = \frac{\beta}{g_m}$$

therefore: a resistor with a value of  $\beta/g_m$  is determined, which is different from the large-signal model that is modeled with a PN junction.