

# Tiny Tapeout 03 Datasheet

Project Repository

<https://github.com/TinyTapeout/tinytapeout-03>

March 21, 2023

## Contents

<b>Render of whole chip</b>	<b>3</b>
<b>Projects</b>	<b>4</b>
0 : Test Inverter Project . . . . .	4
2 : 7 Segment Life . . . . .	5
3 : Another Piece of Pi . . . . .	7
4 : Wormy . . . . .	9
5 : Knight Rider Sensor Lights . . . . .	13
6 : Single digit latch . . . . .	15
7 : 4x4 Memory . . . . .	16
8 : KS-Signal . . . . .	17

9 : Hovalaag CPU . . . . .	19
10 : SKINNY SBOX . . . . .	20
11 : Stateful Lock . . . . .	21
12 : Ascon's 5-bit S-box . . . . .	22
13 : 8bit configurable galois lfsr . . . . .	23
14 : Sbox SKINNY 8 Bit . . . . .	24
15 : BinaryDoorLock . . . . .	25
16 : bad apple . . . . .	26
17 : TinyFPGA attempt for TinyTapeout3 . . . . .	27
18 : 4bit Adder . . . . .	28
19 : 12-bit PDP8 . . . . .	29
<b>Technical info</b>	<b>31</b>
Scan chain . . . . .	31
Clocking . . . . .	32
Clock divider . . . . .	33
Wait states . . . . .	33
Pinout . . . . .	33
Instructions to build GDS . . . . .	34
Changing macro block size . . . . .	35
<b>Verification</b>	<b>36</b>
Setup . . . . .	36
Simulations . . . . .	36
Top level tests setup . . . . .	37
Formal Verification . . . . .	38
Timing constraints . . . . .	38
Physical tests . . . . .	39
<b>Sponsored by</b>	<b>40</b>
<b>Team</b>	<b>40</b>

# Render of whole chip

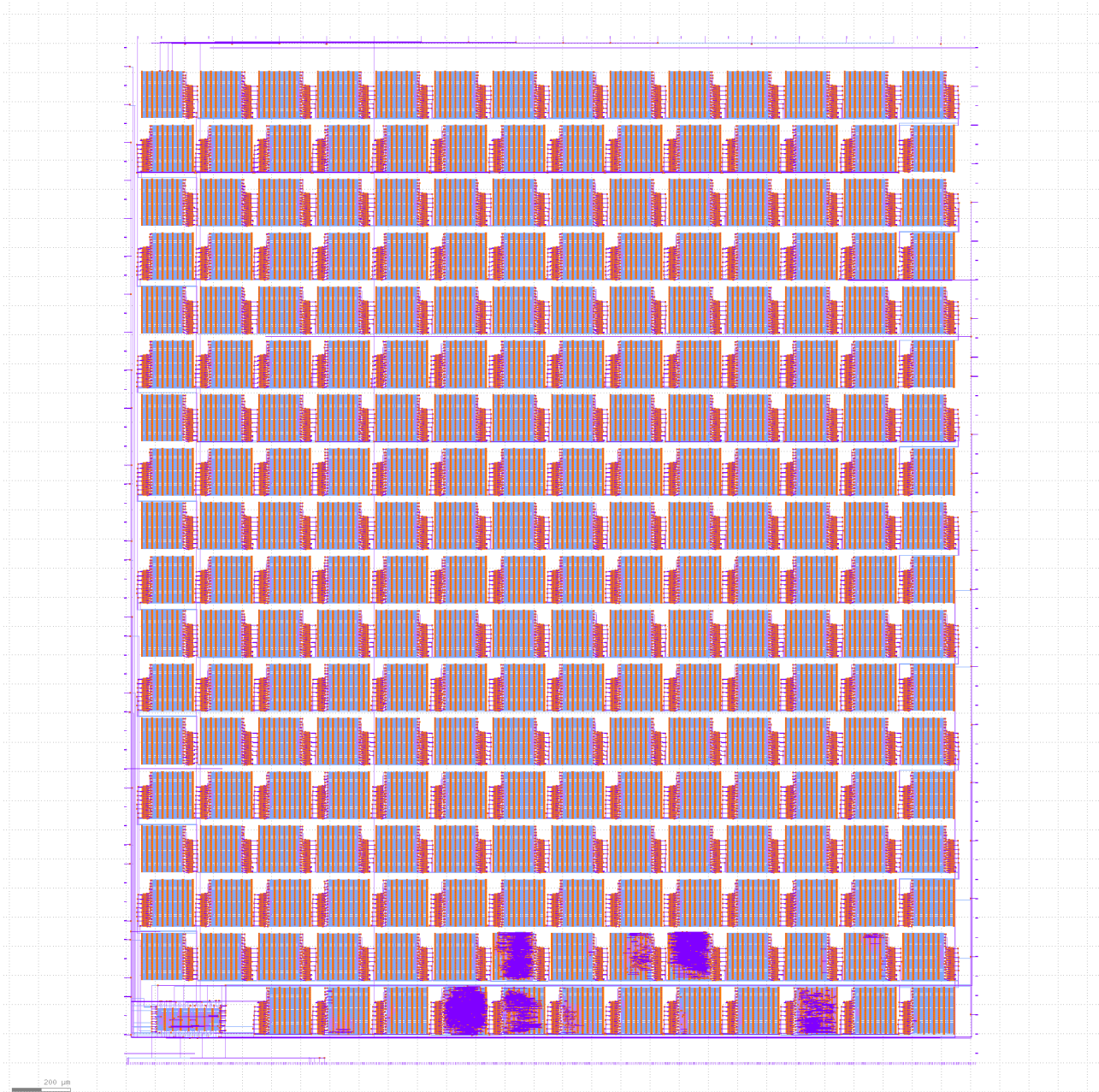


Figure 1: Full GDS

# Projects

## 0 : Test Inverter Project

- Author: Matt Venn
- Description: Inverts every line. This project is also used to fill any empty design spaces.
- GitHub repository
- Wokwi project
- Extra docs
- Clock: 0 Hz
- External hardware:

### How it works

Uses 8 inverters to invert every line.

### How to test

Setting the input switch to on should turn the corresponding LED off.

## IO

#	Input	Output
0	a	segment a
1	b	segment b
2	c	segment c
3	d	segment d
4	e	segment e
5	f	segment f
6	g	segment g
7	dot	dot

## 2 : 7 Segment Life

- Author: icegoat9
- Description: Simple 7-segment cellular automaton
- GitHub repository
- Wokwi project
- Extra docs
- Clock: 0 Hz
- External hardware: None. Could add debounced momentary pushbuttons in parallel with dip switches 1,2,3 to make loading in new patterns and stepping through a run easier.

### How it works

See the Wokwi gate layout and simulation. At a high level...

- Seven flip-flops hold the cellular automaton's internal state, which is also displayed in the seven-segment display.
- Combinatorial logic generates the next state for each segment based on its neighbors, according to the ruleset...
  - Living segments with exactly one living neighbor (another segment that touches it end to end) survive, and all others die.
  - Dead segments with exactly two living neighbors come to life.
- When either the system clock or a user toggling the clock input go high, this new state is latched into the automaton's state.
- There's minor additional support logic to let the user manually shift in an initial condition and handle clock dividing.

### How to test

For full details and a few 'exercises for the reader', see the github README doc link. But at a high level, assuming the IC is mounted on the standard tynytapeout PCB which provides dip switches, clock, and a seven-segment display for output...

- Set all dip switches off and the clock slide switch to the 'manual' clock side.
- Power on the system. An arbitrary state may appear on the 7-segment display.
- Set dip switch 4 on ('run mode').
- Toggle dip switch 1 on and off to advance the automaton to the next state, you should see the 7-segment display update.

If you want to watch it run automatically (which may quickly settle on an empty state or a static pattern, depending on start state)...

- Set the PCB clock divider to the maximum clock division (255). With a system clock of 6.25kHz, the clock input should now be ~24.5Hz.
- Set dip switches 5 and 7 on to add a reasonable additional clock divider (see docs for more details on a higher or lower divider).
- Set dip switch 4 on.
- Switch the clock slide switch to the 'system clock' side. The display should advance at roughly 1.5Hz if I've done math correctly.

If you want to load an initial state...

- Set dip switch 4 off ('load mode').
- Toggle dip switches 2 and/or 3 on and off seven times total, to shift in 0 and 1 values to the automaton internal state.
- Set dip switch 4 on and run manually or automatically as above.

## IO

#	Input	Output
0	clock	7segmentA
1	load0	7segmentB
2	load1	7segmentC
3	runmode	7segmentD
4	clockdiv8	7segmentE
5	clockdiv4	7segmentF
6	clockdiv2	7segmentG
7	unused	7segmentDP

### 3 : Another Piece of Pi

- Author: Meinhard Kissich, EAS Group, Graz University of Technology
- Description: This design takes up the idea of James Ross [1], who submitted a circuit to Tiny Tapeout 02 that stores and outputs the first 1024 decimal digits of the number Pi (including the decimal point) to a 7-segment display. In contrast to his approach, a densely packed decimal encoding is used to store the data. With this approach, 1400 digits can be stored and output within the design area of 150um x 170um. However, at 1400 decimals and utilization of 38.99%, the limitation seems to be routing. Like James, I'm also interested to hear about better strategies to fit more information into the design with synthesizable Verilog code. [1] [https://github.com/jar/tt02\\_freespeech](https://github.com/jar/tt02_freespeech)
- GitHub repository
- HDL project
- Extra docs
- Clock: 0 Hz
- External hardware: 7-segment display

#### How it works

The circuit stores each triplet of decimals in a 10-bit vector encoded as densely packed decimals. An index vector selects the current digits to be output to the 7-segment display. It consists of an upper part `index[11:2]` that selects the triplet and a lower part `index[1:0]` that specifies the digit within the triplet. First, the upper part decides on the triplet, which is then decoded into three decimals. Afterwards, the lower part selects one of the three decimals to be decoded into 7-segment display logic and applied to the outputs. The index is incremented at each primary clock edge. However, when the lower part equals three, i.e., `index[1:0]==1'b10`, two is added, as the triplet consists of three (not four) digits.

- `index == 'b0000000000|00`: triplet[0], digit 0 within triplet
- `index == 'b0000000000|01`: triplet[0], digit 1 within triplet
- `index == 'b0000000000|10`: triplet[0], digit 2 within triplet
- `index == 'b0000000001|00`: triplet[1], digit 0 within triplet
- `index == 'b0000000001|01`: triplet[1], digit 1 within triplet
- `index == 'b0000000001|10`: triplet[1], digit 2 within triplet

There is one exception to the rule above: the decimal point. Another multiplexer

at the input of the 7-segment decoder can either forward a digit from the decoded triplet or a constant – the decimal point. Once the lower part of the index counter, i.e., `index[1:0]` reaches `2'b10` for the first time, the multiplexer selects the decimal point and pauses incrementing the index for one clock cycle.

- `index == 'b0000000000|00: triplet[0], digit 0 within triplet`
- `index == 'b0000000000|01: triplet[0], decimal point`
- `index == 'b0000000000|01: triplet[0], digit 1 within triplet`
- `index == 'b0000000000|10: triplet[0], digit 2 within triplet`
- `index == 'b0000000001|00: triplet[1], digit 0 within triplet`

## How to test

For simulation, please use the provided testbench and Makefile. It is important to run the `genmux.py` Python script first, as it generates the test vectors required by the Verilog testbench. For testing the physical chip, release the reset and compare the digits of Pi against a reference.

## IO

#	Input	Output
0	clk	segment a
1	reset	segment b
2	none	segment c
3	none	segment d
4	none	segment e
5	none	segment f
6	none	segment g
7	none	decimal LED



## 4 : Wormy

- Author: nqbit
- Description: MC Wormy Pants squirms like a worm and grows just as fast.
- GitHub repository
- HDL project
- Extra docs
- Clock: 300 Hz
- External hardware:

### How it works

Wormy is a very simple, addictive last person video game. This last person, open-world game takes you down the path of an earthworm. Wormy's world is made up of a 4x4 grid represented by 3x16-bit arrays: Direction[0], Direction[1], and Occupied. The Direction[x] maps keep track of which way a segment of worm moves, if it is on, and Occupied keeps track of if the grid location is occupied.

Example Occupied Grid:

```
-----  
| X X X X |  
|         X |  
|         X |  
|         |  
|-----|
```

In addition to direction and occupied there are also pointers to the head and the tail. The same grid would look something like the following:

Example Occupied Grid with head(H) and tail(T) highlighted:

```
-----  
| T X X X |  
|         X |  
|         H |  
|         |  
|-----|
```

BOOM! Wormy shouldn't run into itself. If its head hits any part of its body, that causes a collision. There is a collision if the location of the Wormy's head is occupied by another segment of the Wormy. To determine this, we keep track of the worm location, and specifically the current and future locations of the worm head (H) and tail (T). If the future location of H will occupy a location that will already be occupied, this causes a collision.

This is made a bit trickier with growth (see below), because if the future state of H is set to occupy the current state of T - there is only a collision if growth is set to occur



grid of LEDs controlled by a multiplexer. Why multiplexing? With a multiplexed LED setup, we can control more display units (LEDs), with a limited number of outputs (8 on this TinyTapeout project).

To get multiplexing working the network of outputs is mapped to each display unit. This allows us to manipulate assigned outputs to control the state of each display unit, one at a time. We then cycle through each display unit quickly enough to display a persistent image to the last player.

Wires (A1-4, B1-4) map to each location on the game arena (4x4 grid):

```

A1|   |   |   |
-----
A2|   |   |   |
-----
A3|   |   |   |
-----
A4|   |   |   |
-----
  |B1 |B2 |B3 |B4

```

When B's voltage is OFF the LED's state changes to ON if A is also ON. - ON LED state: A(ON) — >| — B(OFF) - OFF LED state: A(ON) — >| — B(ON)

Example: The 3 filled squares below each represent a Wormy segment in the ON state as controlled by the multiplexer. Notice how each is lighter than the last. This is because the multiplexer cycles through each LED to update the state, creating one persistent image even though the LED's are not on over the entire period of time.

```

A1|   | 0 |   |
-----
A2|   | o |   |
-----
A3|   | . |   |
-----
A4|   |   |   |
-----
  |B1 |B2 |B3 |B4

```

Another Example: If you enter a dark cave and point a flashlight straight ahead at one point on the wall you have a very small visual field that is contained within the beam of light. However, you can expand your visual field in the cave by waving the flashlight back and forth across the wall. Despite the fact that the beam is moving over individual points on the wall, the entire wall can be seen at once. This is similar to the concept used in the Wormy display, since the multiplexer changes the state of

the worm occupied locations to ON one at a time, but in a cycle. The result is a solid image, made up of LEDs cycling through ON states to produce a persistent image of Wormy (that beautiful Lumbricina).

## How to test

After reset, you should see a single pixel moving along the display and it should grow every now and then.

## IO

#	Input	Output
0	clock	A0 - Multiplexer channel A to be tied to a an array of 16 multiplexed LEDs
1	reset	A1
2	button0	A2
3	button1	A3
4	button2	B0 - Multiplexer channel B to be tied to a an array of 16 multiplexed LEDs
5	button3	B1
6	none	B2
7	none	B3

## 5 : Knight Rider Sensor Lights

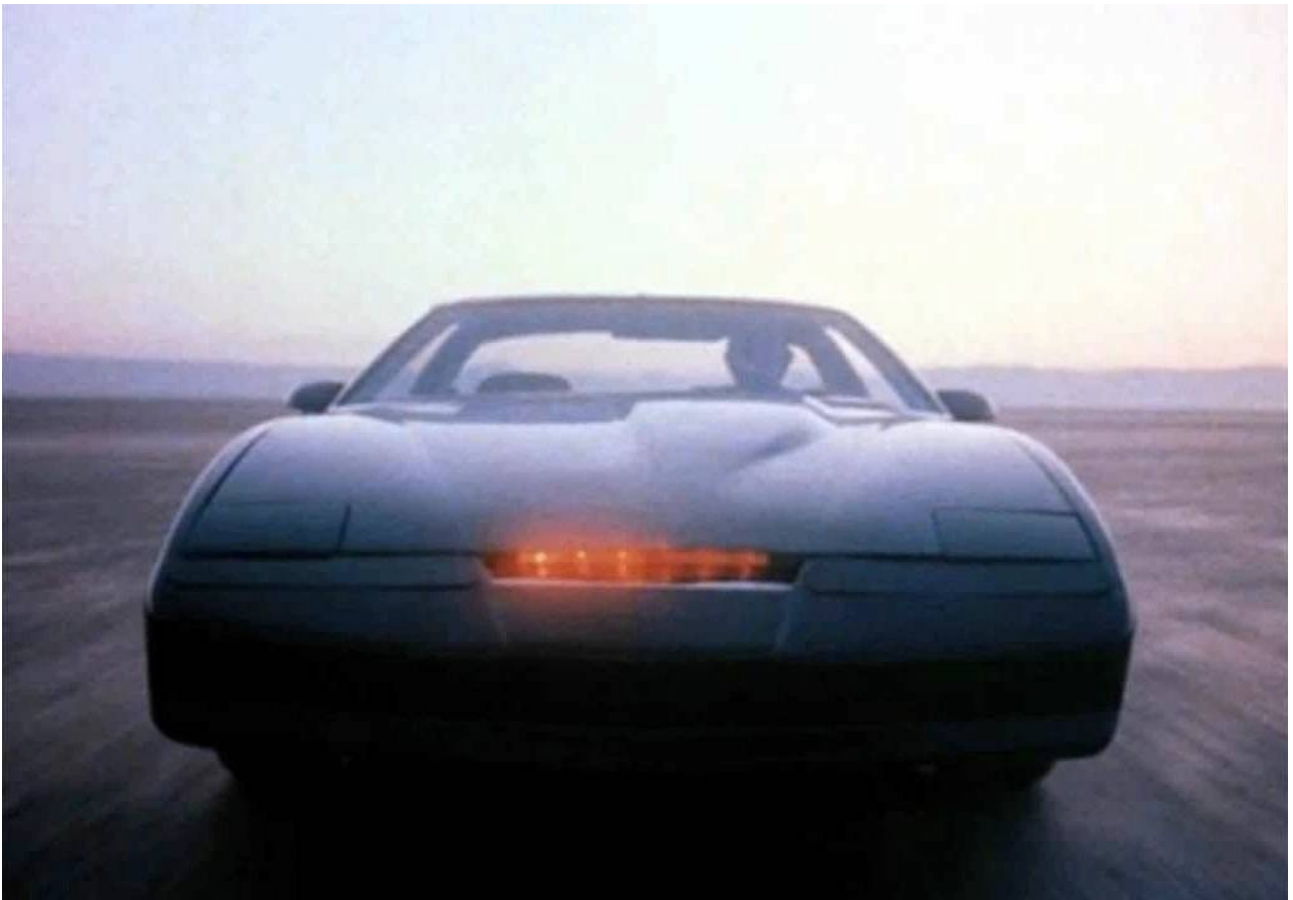


Figure 2: picture

- Author: Kolos Koblasz
- Description: The logic asserts output bits one by one, like KITT's sensor lights in Knight Rider.
- GitHub repository
- HDL project
- Extra docs
- Clock: 6000 Hz
- External hardware: Connect LEDs with ~1K-10K Ohm serial resistors to output pins and connect push button switches to Input[2] and Input[3] which drive the inputs with logic zeros when idle and with logic 1 when pressed. Rising edge on these inputs selects the next settings.

### How it works

Uses several counters, shiftregisters to create a moving light. Input[2] and Input[3] can control speed and brightness respectively. Brightness control is achieved by PWM of the output bits at 50Hz. Simulated with 6KHz clock signal.

## How to test

After reset it starts moving the switched on LED. Input[0] is clk and Input[1] is reset (1=reset on, 0=reset off). By creating rising edges on Input[2] and Input[3] the two config spaces can be discovered. Conect LEDs with ~1K-10K Ohm serial resistors to output pins and connect push button switches to Input[2] and Input[3] which drive the inputs with logic zeros when idle and with logic 1 when pressed. Rising edge on these inputs selects the next settings.

## IO

#	Input	Output
0	clock	LED 0
1	reset	LED 1
2	speed control	LED 2
3	brightness control	LED 3
4	none	LED 4
5	none	LED 5
6	none	LED 6
7	none	LED 7

## 6 : Single digit latch

- Author: Dylan Garrett
- Description: Store a single digit 0-9 and display it on a 7-segment display
- GitHub repository
- Wokwi project
- Extra docs
- Clock: 0 Hz
- External hardware:

### How it works

Explain how your project works

### How to test

Explain how to test your project

## IO

#	Input	Output
0	clock	segment a
1	reset	segment b
2	none	segment c
3	none	segment d
4	none	segment e
5	none	segment f
6	none	segment g
7	none	dot

## 7 : 4x4 Memory

- Author: Yannick Reiß
- Description: Store 4x4 bits of memory.
- GitHub repository
- Wokwi project
- Extra docs
- Clock: 1000 Hz
- External hardware:

### How it works

On write enable, the four data-inputs are saved to the d-flipflops selected by the address inputs. The output is always read from the current selected flipflops.

### How to test

Connect a clock, buttons for reset and write\_enable and switches for address and data like shown below. The output can be read by using 4 LEDs or any other kind of binary output device.

## IO

#	Input	Output
0	clock	data1
1	reset	data2
2	write_enable	data3
3	addr1	data4
4	addr2	none
5	data1	none
6	data2	none
7	data3	none



## 8 : KS-Signal

- Author: Yannick Reiß
- Description: Set KS-Signal based on track information.
- GitHub repository
- Wokwi project
- Extra docs
- Clock: 10000 Hz
- External hardware: Input: Clock, 9 Buttons, Output: 3 yellow LEDs, 3 white LEDs, 1 green LED, 1 red LED, 1 orange LED

### How it works

Simply switches lamps on and off on toggle.

### How to test

Connect the clock, and the buttons as inputs. Connect the LEDs as outputs in the order shown below.

## IO

#	Input	Output
0	track1_free	Fahrt! (Green)
1	track2_free	Halt Erwarten! (Orange)
2	pre_signal	Halt! (Red)
3	none	Vorsicht! (All yellow)
4	none	pre signal indicator (white)
5	allow shunting	lower shunting indicator (white)
6	Vorsicht!	shorter breaking distance (white)
7	shorter breaking distance	none

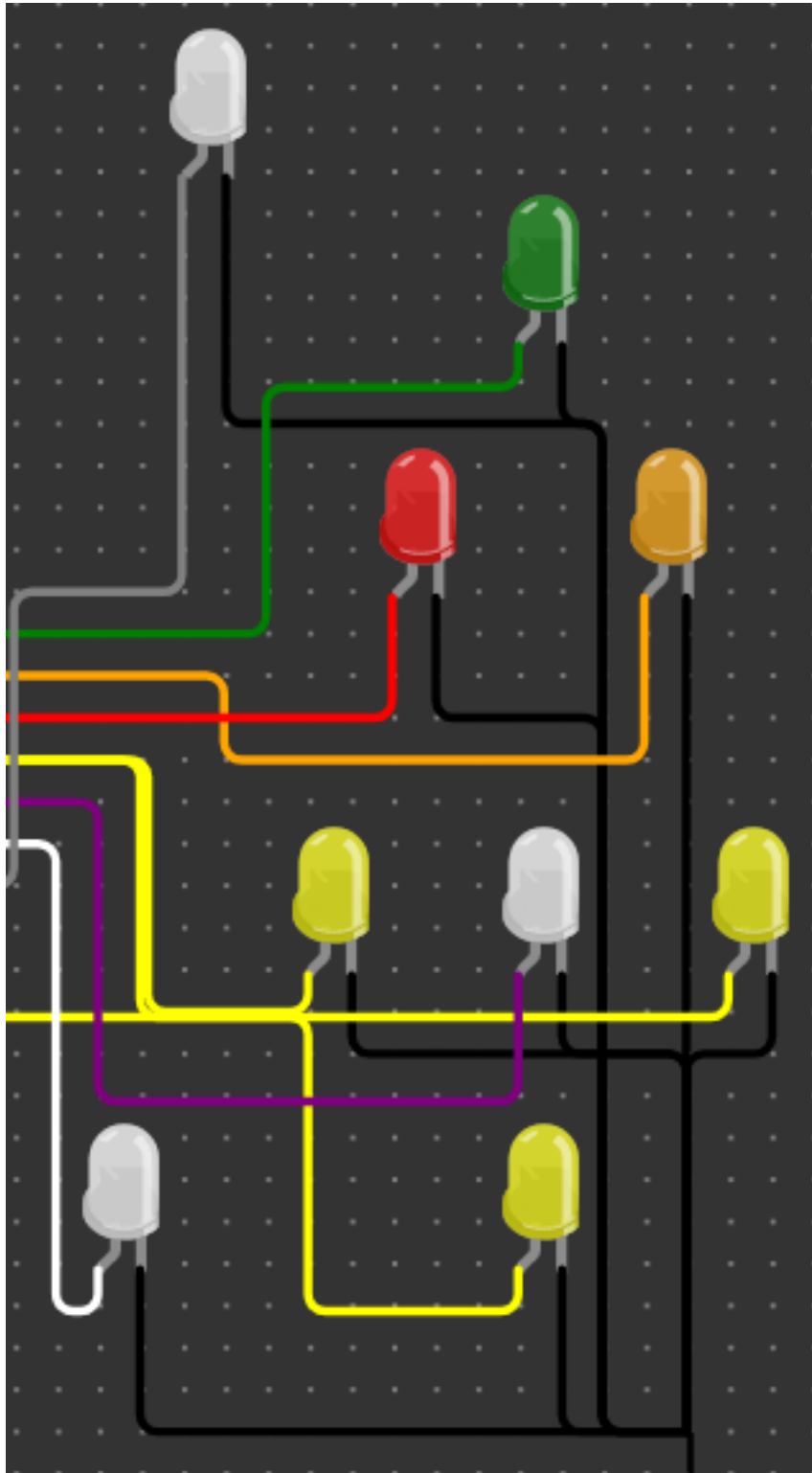


Figure 3: picture

## 9 : Hovalaag CPU

- Author: Mike Bell
- Description: Implementation of the CPU from HOVALAAG
- GitHub repository
- HDL project
- Extra docs
- Clock: 12500 Hz
- External hardware:

### How it works

HOVALAAG (Hand-Optimizing VLIW Assembly Language as a Game) is a free Zachlike game.

This is an implementation of the VLIW processor from the game. Thank you to @nothings for the fun game, making the assembler public domain, and for permission to create this hardware implementation.

The processor uses 32-bit instructions and has 12-bit I/O. The instruction and data are therefore passed in and out over 10 clocks per processor clock.

More details in the github repo.

### How to test

The assembler can be downloaded to generate programs.

The subcycle counter can be reset independently of the rest of the processor, to ensure you can get to a known state without clearing all registers.

## IO

#	Input	Output
0	Clock	Output 0
1	Reset enable	Output 1
2	Input 0 or Reset	Output 2
3	Input 1 or Reset subcycle count	Output 3
4	Input 2	Output 4
5	Input 3	Output 5
6	Input 4	Output 6
7	Input 5	Output 7

## 10 : SKINNY SBOX

- Author: Niklas Fassbender
- Description: Implementation of a 4-Bit Sbox for SKINNY
- GitHub repository
- Wokwi project
- Extra docs
- Clock: 0 Hz
- External hardware:

### How it works

Explain how your project works

### How to test

Explain how to test your project

## IO

#	Input	Output
0	bit_0_lsb	bit_0_lsb
1	bit_1	bit_1
2	bit_2	bit_2
3	bit_3_msb	bit_3_msb
4	none	none
5	none	none
6	none	none
7	none	none

## 11 : Stateful Lock

- Author: Tim Henkes
- Description: A little combination lock which requires three codes in the correct order to unlock
- GitHub repository
- Wokwi project
- Extra docs
- Clock: 0 Hz
- External hardware:

### How it works

A little two-bit state machine decides which code is required to enter the next state. The fourth state equals to the lock being open. A wrong code in any state resets to the first state.

### How to test

To test the project, refer to its Wokwi, which is public.

## IO

#	Input	Output
0	unused	lock status
1	reset	unused
2	enter	unused
3	code digit 0	unused
4	code digit 1	unused
5	code digit 2	unused
6	code digit 3	unused
7	code digit 4	unused

## 12 : Ascon's 5-bit S-box

- Author: Fabio Campos
- Description: Ascon's 5-bit S-box
- GitHub repository
- Wokwi project
- Extra docs
- Clock: 0 Hz
- External hardware:

### How it works

See Figure 4a in <https://eprint.iacr.org/2021/1574.pdf>

### How to test

See Section 3.2 and Table 4 in <https://eprint.iacr.org/2021/1574.pdf>

## IO

#	Input	Output
0	x0	x0
1	x1	x1
2	x2	x2
3	x3	x3
4	x4	x4
5	none	none
6	none	none
7	none	none

## 13 : 8bit configurable galois lfsr

- Author: Alexander Schönborn
- Description: A 8bit configurable galois lfsr.
- GitHub repository
- HDL project
- Extra docs
- Clock: 1000 Hz
- External hardware:

### How it works

Uses 2 cycles each to set the shift and xor mask registers. It has 3 operation modes: normal shift,

### How to test

After reset, set the shift register and xor state, after that normal shift mode

## IO

#	Input	Output
0	clock	output[0] lsb normal lfsr output
1	reset	output[1] other 7 bits to see the full state
2	mode[0] 00 normal shift mode, 01 set register mode,	output[2]
3	mode[1] 10 set mode registers, 11 unused	output[3]
4	data_in[0] is used for both filling register and xor mask state	output[4]
5	data_in[1] needs 2 cycles to fill all 8 bits	output[5]
6	data_in[2] first cyle is lower 4 bits, 2nd upper 4 bits	output[6]
7	data_in[3] see above	output[7]

## 14 : Sbox SKINNY 8 Bit

- Author: Thorsten Knoll
- Description: A circuit for the substitution of 8 bits. Made for the SKINNY cipher algorithm.
- GitHub repository
- Wokwi project
- Extra docs
- Clock: 0 Hz
- External hardware:

### How it works

Explain how your project works

### How to test

Explain how to test your project

## IO

#	Input	Output
0	in_0	out_0
1	in_1	out_1
2	in_2	out_2
3	in_3	out_3
4	in_4	out_4
5	in_5	out_5
6	in_6	out_6
7	in_7	out_7



## 15 : BinaryDoorLock

- Author: Marcus Michaely
- Description: Input is 8-Bit and only one combination opens the door
- GitHub repository
- Wokwi project
- Extra docs
- Clock: 0 Hz
- External hardware:

### How it works

Explain how your project works

### How to test

Explain how to test your project

## IO

#	Input	Output
0	Bit_0	Output
1	Bit_1	none
2	Bit_2	none
3	Bit_3	none
4	Bit_4	none
5	Bit_5	none
6	Bit_6	none
7	Bit_7	none

## 16 : bad apple

- Author: shadow1229
- Description: Plays bad apple over a Piezo Speaker connected across io\_out[1:0].  
Based on <https://github.com/meriac/tt02-play-tune>
- GitHub repository
- HDL project
- Extra docs
- Clock: 12000 Hz
- External hardware: Piezo speaker connected across io\_out[1:0]

### How it works

Converts an RTTL ringtone into verilog and plays it back using differential PWM modulation.

### How to test

Provide 12kHz clock on io\_in[0], briefly hit reset io\_in[1] (L->H->L) and io\_out[1:0] will play a differential sound wave over piezo speaker (Bad Apple)

## IO

#	Input	Output
0	clock	piezo_speaker_p
1	reset	piezo_speaker_n
2	none	none
3	none	none
4	none	none
5	none	none
6	none	none
7	none	none

## 17 : TinyFPGA attempt for TinyTapeout3

- Author: Emilian Miron
- Description: FPGA attempt
- GitHub repository
- HDL project
- Extra docs
- Clock: 1000 Hz
- External hardware:

### How it works

TODO

### How to test

After reset, the counter should increase by one every second.

### IO

#	Input	Output
0	clock	segment a
1	reset	segment b
2	none	segment c
3	none	segment d
4	none	segment e
5	none	segment f
6	none	segment g
7	none	slow clock output

## 18 : 4bit Adder

- Author: Carin Schreiner
- Description: This tiny tape out project takes two four bit numberbs and adds them.
- GitHub repository
- Wokwi project
- Extra docs
- Clock: 0 Hz
- External hardware:

### How it works

Explain how your project works

### How to test

Explain how to test your project

## IO

#	Input	Output
0	a0	r0
1	a1	r1
2	a2	r2
3	a3	r3
4	b0	carry
5	b1	none
6	b2	none
7	b3	none

## 19 : 12-bit PDP8

- Author: Paul Campnell
- Description: PDP8 core
- GitHub repository
- HDL project
- Extra docs
- Clock: 1000 Hz
- External hardware:

### How it works

This is a 12-bit basic PDP8 cpu - it doesn't have the extended arithmetic unit (so no multiply or divide). Included is an assembler (mostly for test). Bus interface is a 5-clock to get 12 bits of address and 12 bits of data though 8-bit interfaces. Address is 2 beats of 6 bits each, data is 3 beats of 4 bits each, I/O cycles have an extra beat

output bits

7 6 5 4 3 2 1 0

1 0 A A A A A A address hi

1 1 A A A A A A address lo

0 1 1 I I 4 2 1 IO cycle intro

either

0 0 0 0 - - - - read data high nibble

0 0 1 0 - - - - read data med nibble

0 1 0 0 - - - - read data low nibble

or

0 0 0 1 D D D D write data high nibble

0 0 1 1 D D D D write data med nibble

0 1 0 1 D D D D write data low nibble

Input bits are ignored except during read beats, interrupts are sampled during the first address beat

### How to test

code in test-bench, assembler in asm dir

## IO

#	Input	Output
0	clock	address_0_data_mux_0
1	reset	address_1_data_mux_1
2	none	address_2_data_mux_2
3	none	address_3_data_mux_3
4	data_in_0	address_4_rw_mux
5	data_in_1	address_5_phase_lo_select_mux
6	data_in_2	phase_hi_select
7	data_in_3	address_data_select

# Technical info

## Scan chain

All 250 designs are joined together in a long chain similar to JTAG. We provide the inputs and outputs of that chain (see pinout below) externally, to the Caravel logic analyser, and to an internal scan chain driver.

The default is to use an external driver, this is in case anything goes wrong with the Caravel logic analyser or the internal driver.

The scan chain is identical for each little project, and you can read it here.

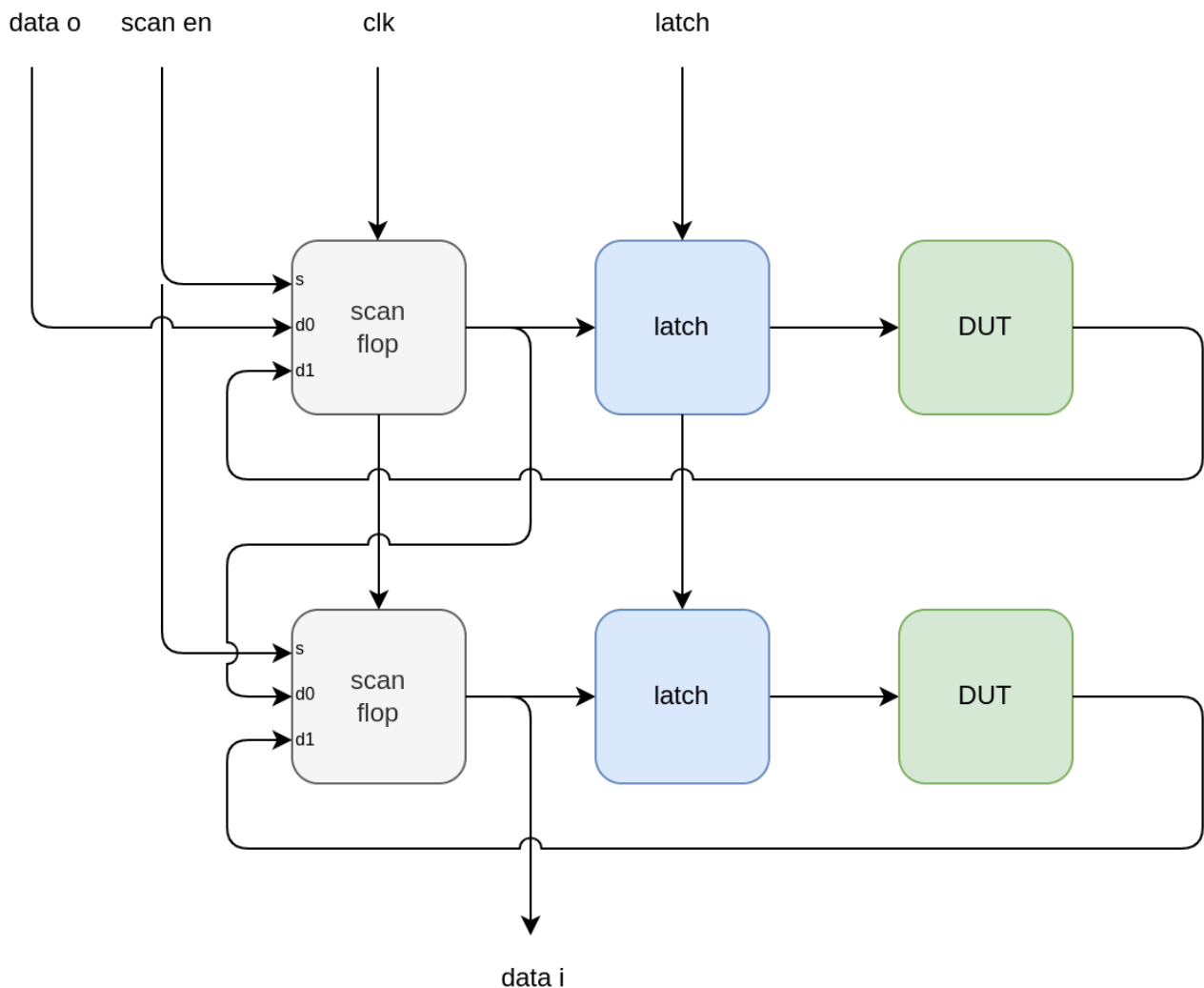


Figure 4: block diagram

## Updating inputs and outputs of a specified design

A good way to see how this works is to read the FSM in the scan controller. You can also run one of the simple tests and check the waveforms. See how in the scan chain

verification doc.

- Signal names are from the perspective of the scan chain driver.
- The desired project shall be called DUT (design under test)

Assuming you want to update DUT at position 2 (0 indexed) with inputs = 0x02 and then fetch the output. This design connects an inverter between each input and output.

- Set scan\_select low so that the data is clocked into the scan flops (rather than from the design)
- For the next 8 clocks, set scan\_data\_out to 0, 0, 0, 0, 0, 0, 1, 0
- Toggle scan\_clk\_out 16 times to deliver the data to the DUT
- Toggle scan\_latch\_en to deliver the data from the scan chain to the DUT
- Set scan\_select high to set the scan flop's input to be from the DUT
- Toggle the scan\_clk\_out to capture the DUT's data into the scan chain
- Toggle the scan\_clk\_out another 8 x number of remaining designs to receive the data at scan\_data\_in

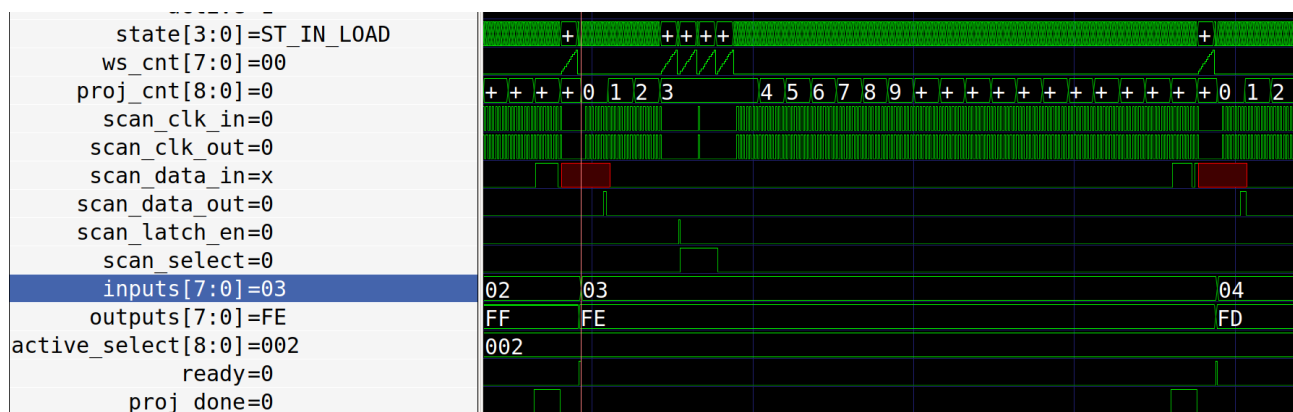


Figure 5: update cycle

### Notes on understanding the trace

- There are large wait times between the latch and scan signals to ensure no hold violations across the whole chain. For the internal scan controller, these can be configured (see section on wait states below).
- The input looks wrong (0x03) because the input is incremented by the test bench as soon as the scan controller captures the data. The input is actually 0x02.
- The output in the trace looks wrong (0xFE) because it's updated after a full refresh, the output is 0xFD.

## Clocking

Assuming:



- 100MHz input clock
- 8 ins & 8 outs
- 2 clock cycles to push one bit through the scan chain (scan clock is half input clock rate)
- 250 designs
- scan controller can do a read/write cycle in one refresh

So the max refresh rate is  $100\text{MHz} / (8 * 2 * 250) = 25000\text{Hz}$ .

## Clock divider

A rising edge on the `set_clk_div` input will capture what is set on the input pins and use this as a divider for an internal slow clock that can be provided to the first input bit.

The slow clock is only enabled if the `set_clk_div` is set, and the resulting clock is connected to `input0` and also output on the `slow_clk` pin.

The slow clock is synced with the scan rate. A divider of 0 mean it toggles the `input0` every scan. Divider of 1 toggles it every 2 cycles. So the resultant slow clock frequency is  $\text{scan\_rate} / (2 * (N+1))$ .

See the `test_clock_div` test in the scan chain verification.

## Wait states

This dictates how many wait cycle we insert in various state of the load process. We have a sane default, but also allow override externally.

To override, set the wait amount on the inputs, set the `driver_sel` inputs both high, and then reset the chip.

See the `test_wait_state` test in the scan chain verification.

## Pinout

PIN	NAME	DESCRIPTION
20:12	<code>active_select</code>	9 bit input to set which design is active
28:21	<code>inputs</code>	8 inputs
36:29	<code>outputs</code>	8 outputs
37	<code>ready</code>	goes high for one cycle everytime the scanchain is loaded
10	<code>slow_clk</code>	slow clock from internal clock divider
11	<code>set_clk_div</code>	enable clock divider

9:8	driver_sel	which scan chain driver: 00 = external, 01 =
21	ext_scan_clk_out	for external driver, clk input
22	ext_scan_data_out	data input
23	ext_scan_select	scan select
24	ext_scan_latch_en	latch
29	ext_scan_clk_in	clk output from end of chain
30	ext_scan_data_in	data output from end of chain

## Instructions to build GDS

To run the tool locally or have a fork's GitHub action work, you need the GH\_USERNAME and GH\_TOKEN set in your environment.

GH\_USERNAME should be set to your GitHub username.

To generate your GH\_TOKEN go to <https://github.com/settings/tokens/new> . Set the checkboxes for repo and workflow.

To run locally, make a file like this:

```
export GH_USERNAME=<username>
export GH_TOKEN=<token>
```

And then source it before running the tool.

## Fetch all the projects

This goes through all the projects in project\_urls.py, and fetches the latest artifact zip from GitHub. It takes the verilog, the GL verilog, and the GDS and copies them to the correct place.

```
./configure.py --clone-all --fetch-gds
```

## Configure Caravel

Caravel needs the list of macros, how power is connected, instantiation of all the projects etc. This command builds these configs and also makes the README.md index.

```
./configure.py --update-caravel
```

## Build the GDS

To build the GDS and run the simulations, you will need to install the Sky130 PDK and OpenLane tool. It takes about 5 minutes and needs about 3GB of disk space.

```
export PDK_ROOT=<some dir>/pdk
export OPENLANE_ROOT=<some dir>/openlane
cd <the root of this repo>
make setup
```

Then to create the GDS:

```
make user_project_wrapper
```

## Changing macro block size

After working out what size you want:

- adjust `configure.py` in `CaravelConfig.create_macro_config()`.
- adjust the PDN spacing to match in `openlane/user_project_wrapper/config.tcl`:
  - `set ::env(FP_PDN_HPITCH)`
  - `set ::env(FP_PDN_HOFFSET)`

# Verification

We are not trying to verify every single design. That is up to the person who makes it. What we want is to ensure that every design is accessible, even if some designs are broken.

We can split the verification effort into functional testing (simulation), static tests (formal verification), timing tests (STA) and physical tests (LVS & DRC).

See the sections below for details on each type of verification.

## Setup

You will need the GitHub tokens setup as described in INFO.

The default of 250 projects takes a very long time to simulate, so I advise overriding the configuration:

```
# fetch the test projects
./configure.py --test --clone-all
# rebuild config with only 20 projects
./configure.py --test --update-caravel --limit 20
```

You will also need iVerilog & cocotb. The easiest way to install these are to download and install the oss-cad-suite.

## Simulations

- Simulation of some test projects at RTL and GL level.
- Simulation of the whole chip with scan controller, external controller, logic analyser.
- Check wait state setting.
- Check clock divider setting.

### Scan controller

This test only instantiates user\_project\_wrapper (which contains all the small projects). It doesn't simulate the rest of the ASIC.

```
cd verilog/dv/scan_controller
make test_scan_controller
```

The Gate Level simulation requires scan\_controller and user\_project\_wrapper to be re-hardened to get the correct gate level netlists:

- Edit `openlane/scan_controller/config.tcl` and change `NUM_DESIGNS=250` to `NUM_DESIGNS=20`.
- Then from the top level directory:  
`make scan_controller make user_project_wrapper`
- Then run the GL test  
`cd verilog/dv/scan_controller make test_scan_controller_gl`

## **single**

Just check one inverter module. Mainly for easy understanding of the traces.

```
make test_single
```

## **custom wait state**

Just check one inverter module. Set a custom wait state value.

```
make test_wait_state
```

## **clock divider**

Test one inverter module with an automatically generated clock on input 0. Sets the clock rate to 1/2 of the scan refresh rate.

```
make test_clock_div
```

## **Top level tests setup**

For all the top level tests, you will also need a RISC-V compiler to build the firmware.

You will also need to install the 'management core' for the Caravel ASIC submission wrapper. This is done automatically by following the PDK install instructions.

## **Top level test: internal control**

Uses the scan controller, instantiated inside the whole chip.

```
cd verilog/dv/scan_controller_int
make coco_test
```

## Top level test: external control

Uses external signals to control the scan chain. Simulates the whole chip.

```
cd verilog/dv/scan_controller_ext  
make coco_test
```

## Top level test: logic analyser control

Uses the RISC-V co-processor to drive the scanchain with firmware. Simulates the whole chip.

```
cd verilog/dv/scan_controller_la  
make coco_test
```

## Formal Verification

- Formal verification that each small project's scan chain is correct.
- Formal verification that the correct signals are passed through for the 3 different scan chain control modes.

## Scan chain

Each GL netlist for each small project is proven to be equivalent to the reference scan chain implementation. The verification is done on the GL netlist, so an RTL version of the cells used needed to be created. See [here](#) for more info.

## Scan controller MUX

In case the internal scan controller doesn't work, we also have ability to control the chain from external pins or the Caravel Logic Analyser. We implement a simple MUX to achieve this and formally prove it is correct.

## Timing constraints

Due to limitations in OpenLane - a top level timing analysis is not possible. This would allow us to detect setup and hold violations in the scan chain.

Instead, we design the chain and the timing constraints for each project and the scan controller with this in mind.

- Each small project has a negedge flop at the end of the shift register to reclock the data. This gives more hold margin.

- Each small project has SDC timing constraints
- Scan controller uses a shift register clocked with the end of the chain to ensure correct data is captured.
- Scan controller has its own SDC timing constraints
- Scan controller can be configured to wait for a programmable time at latching data into the design and capturing it from the design.
- External pins (by default) control the scan chain.

## Physical tests

- LVS
- DRC
- CVC

### LVS

Each project is built with OpenLane, which will check LVS for each small project. Then when we combine all the projects together we run a top level LVS & DRC for routing, power supply and macro placement.

The extracted netlist from the GDS is what is used in the formal scan chain proof.

### DRC

DRC is checked by OpenLane for each small project, and then again at the top level when we combine all the projects.

### CVC

Mitch Bailey' CVC checker is a device level static verification system for quickly and easily detecting common circuit errors in CDL (Circuit Definition Language) netlists. We ran the test on the final design and found no errors.

- See the paper here.
- Github repo for the tool: <https://github.com/d-m-bailey/cvc>

**Sponsored by**



## **Team**

Tiny Tapeout would not be possible without a lot of people helping. We would especially like to thank:

- Uri Shaked for wokwi development and lots more
- Sylvain Munaut for help with scan chain improvements
- Mike Thompson for verification expertise
- Jix for formal verification support
- Propy for help with GitHub actions
- Maximo Balestrini for all the amazing renders and the interactive GDS viewer
- James Rosenthal for coming up with digital design examples
- All the people who took part in TinyTapeout 01 and volunteered time to improve docs and test the flow
- The team at YosysHQ and all the other open source EDA tool makers
- Efabless for running the shuttles and providing OpenLane and sponsorship
- Tim Ansell and Google for supporting the open source silicon movement
- Zero to ASIC course community for all your support