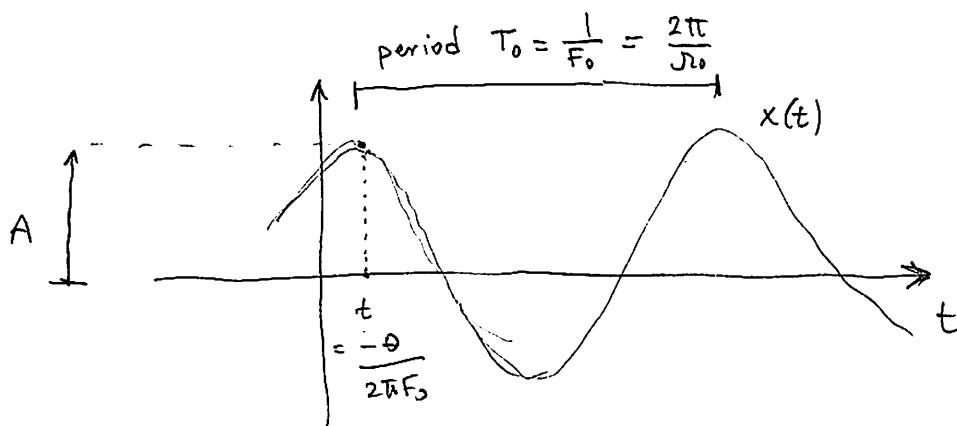


Lecture 9 (Fourier Transforms (FT), CT, DT)

Sinusoidal signals

① continuous time



$$\begin{aligned}
 x(t) &= [A] \cos(2\pi[F_0]t + [\theta]) \\
 &= A \cos(\underline{\omega_0} t + \theta) \\
 \omega_0 &= 2\pi F_0 \text{ (angular frequency)}
 \end{aligned}$$

$$2\pi F_0 t + \theta = 0 \Rightarrow t = \frac{-\theta}{2\pi F_0}$$

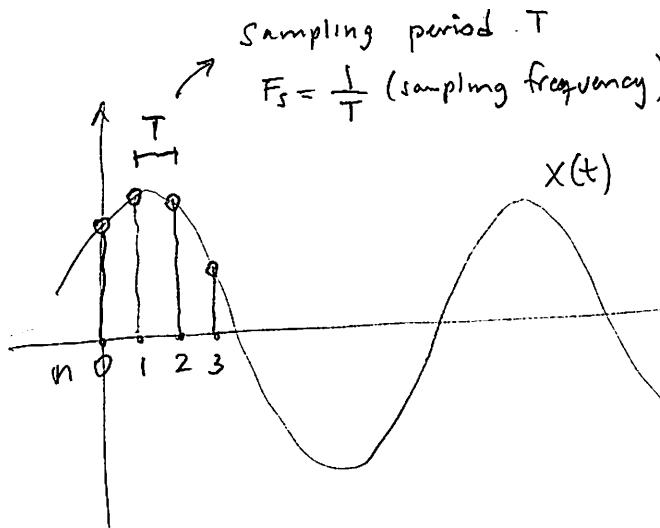
- periodic: $x(t + T_0) = A \cos(2\pi F_0(t + T_0) + \theta) = A \cos(2\pi F_0 t + 2\pi + \theta) = x(t)$
- Recall: Euler's identity: $e^{j\theta} = \cos\theta + j\sin\theta \Rightarrow \cos\theta = \frac{e^{j\theta} + e^{-j\theta}}{2}$

$$x(t) = A \cdot \frac{e^{j(\omega_0 t + \theta)} + e^{-j(\omega_0 t + \theta)}}{2}$$

$$= \frac{A}{2} e^{j\theta} \cdot e^{j\omega_0 t} + \frac{A}{2} e^{-j\theta} \cdot e^{j(-\omega_0)t}$$

(linear combination of complex sinusoids)

② Discrete-time: Sampling continuous-time sinusoid $x(t) = A \cos(\omega_0 t + \theta)$



$$x[n] = x(nT)$$

$$= A \cos(\omega_0 nT + \theta)$$

$$= A \cos(\omega n + \theta)$$

↳ normalized angular frequency

$$\omega = \omega_0 T$$

• periodic?
 $x[n] = A \cos(\omega n + \theta)$ is periodic iff $x[n] = x[n+N]$ for some $N \in \mathbb{Z}^+$

if and only if

$$\Leftrightarrow A \cos(\omega n + \theta) = A \cos(\omega n + \omega N + \theta) \Leftrightarrow \omega N = 2\pi \cdot k \text{ for some } k$$

$\Leftrightarrow \omega$ must be a rational multiple of π

• A shift of 2π in the angular frequency, sinusoid doesn't change

$$x[n] = A \cos((\omega + 2\pi)n + \theta) = A \cos(\omega n + 2\pi n + \theta) = A \cos(\omega n + \theta)$$

Continuous-time Fourier Transform

$$x(t) = \int_{-\infty}^{\infty} \left| \frac{X(\omega)}{2\pi} \right| \cdot e^{j\omega t} d\omega$$

$\int_{-\infty}^{\infty}$

$\left| \frac{X(\omega)}{2\pi} \right|$ limit of a sum

$e^{j\omega t}$ complex sinusoid with frequency ω

$d\omega$ scaling/weight applied to frequency ω

$$X(\omega) = \int_{-\infty}^{\infty} x(t) e^{-j\omega t} dt$$

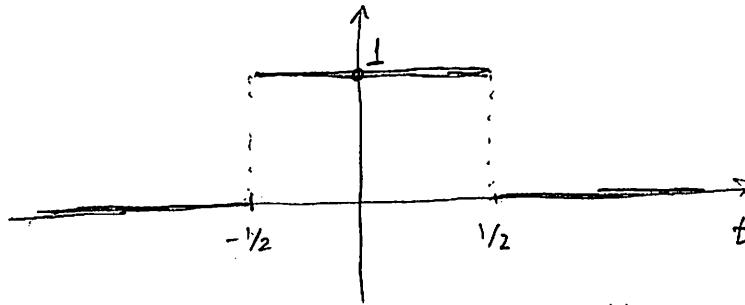
$\int_{-\infty}^{\infty}$

CTFT of $x(t)$

In words: Any signal $x(t)$ can be expressed as a linear combination of complex sinusoids $e^{j\omega t}$, each with $\frac{X(\omega)}{2\pi}$ as a weight

Ex: Rectangular signal

$$x(t) = \begin{cases} 1, & |t| < \frac{1}{2} \\ 0, & \text{otherwise} \end{cases}$$

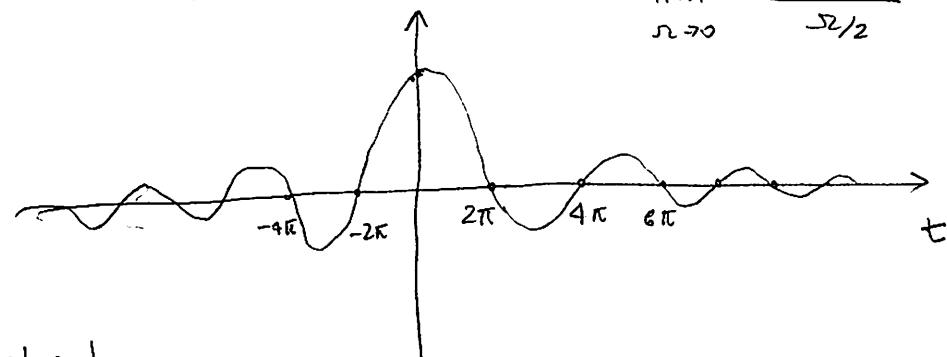


$$X(\omega) = \int_{-\infty}^{\infty} x(t) e^{-j\omega t} dt = \int_{-1/2}^{1/2} 1 \cdot e^{-j\omega t} dt = \left. \frac{e^{-j\omega t}}{-j\omega} \right|_{t=-1/2}^{1/2}$$

$$\begin{aligned} \sin \theta &= \frac{e^{j\theta} - e^{-j\theta}}{2j} \\ &= \frac{e^{-j\omega/2} - e^{-j\omega(-1/2)}}{-j\omega} = 2 \left(\frac{e^{j\omega/2} - e^{-j\omega/2}}{j\omega \cdot 2} \right) \\ &= 2 \frac{\sin(\omega/2)}{\omega} = \frac{\sin(\omega/2)}{\omega/2} \end{aligned}$$

In this case, $X(\omega)$ is real-valued.

$$\lim_{\omega \rightarrow 0} \frac{\sin(\omega/2)}{\omega/2} = 1$$

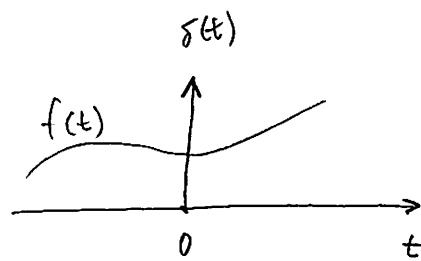


In general, $X(\omega)$ is complex-valued

and we need to plot magnitude and phase

The Dirac delta function

$$\delta(t) = \begin{cases} \infty & \text{at } t=0 \\ 0 & \text{otherwise} \end{cases} \quad \text{and} \quad \int_{-\infty}^{\infty} \delta(t) dt = 1$$



key property:

$$\int_{-\infty}^{\infty} f(t) \delta(t) dt = f(t) \Big|_{t=0} = f(0)$$

$$\int_{-\infty}^{\infty} f(t) \delta(t-t_0) dt = f(t_0)$$

$$\text{Ex: } \int_{-\infty}^{\infty} \delta(5t) f(t) dt = \int_{-\infty}^{\infty} \delta(u) f\left(\frac{u}{5}\right) \cdot \frac{1}{5} du = \frac{1}{5} \int_{-\infty}^{\infty} \delta(u) f\left(\frac{u}{5}\right) du$$

\uparrow

$u = 5t$
 $du = 5dt$

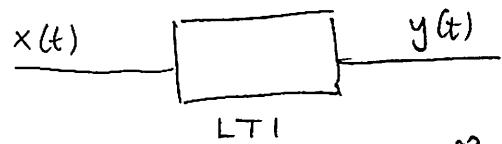
$$= \frac{1}{5} \cdot f\left(\frac{0}{5}\right) = \frac{1}{5} f(0)$$

$$\text{Ex 2: } \int_{-\infty}^{\infty} \delta(5t-15) f(t) dt = \int_{-\infty}^{\infty} \delta(5(t-3)) f(t) dt = \frac{1}{5} f(3)$$

CTFT of $\delta(t)$: $x(t) = \delta(t)$, $X(\omega) = \int_{-\infty}^{\infty} x(t) e^{-j\omega t} dt = e^{-j\omega \cdot 0} = 1$

CTFT of $\delta(t-t_0)$: $X(\omega) = e^{-j\omega t_0}$

Continuous-time convolution



$$y(t) = (x * h)(t) = \int_{-\infty}^{\infty} x(\tau) h(t-\tau) d\tau$$

If $x(t) = \delta(t)$, then

$$y(t) = \int_{-\infty}^{\infty} \delta(\tau) h(t-\tau) d\tau = h(t-\tau) \Big|_{\tau=0} = \underline{h(t)}$$

impulse response

We notice that convolution with $\delta(t)$ doesn't change signal:

$$x(t) * \delta(t) = x(t)$$

Ex: $e^{-t} u(t) * \delta(5t - 15)$, $u(t) = \begin{cases} 1 & \text{for } t \geq 0 \\ 0 & \text{otherwise} \end{cases}$

$$\delta(5t - 15) = \frac{1}{5} \delta(t - 3)$$

$$\Rightarrow e^{-t} u(t) * \frac{1}{5} \cdot \delta(t-3) = \frac{1}{5} e^{-(t-3)} u(t-3)$$

CTFT

$$x(t)$$

CTFT

$$X(\omega)$$

$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(\omega) e^{j\omega t} d\omega$$

$$X(\omega) = \int_{-\infty}^{\infty} x(t) e^{-j\omega t} dt$$

DTFT

$$x[n]$$

DTFT

$$X(e^{j\omega}) \quad (\begin{matrix} \text{continuous} \\ \text{function of } \omega \end{matrix})$$

$$x[n] = \frac{1}{2\pi} \int_{-\pi}^{\pi} X(e^{j\omega}) e^{j\omega n} d\omega$$

$$X(e^{j\omega}) = \sum_{n=-\infty}^{\infty} x[n] e^{-j\omega n} \quad X_d(\omega)$$

$$\begin{aligned} X(e^{j(\omega + 2\pi k)}) &= \sum_{n=-\infty}^{\infty} x[n] e^{-j(\omega + 2\pi k)n} = \sum_{n=-\infty}^{\infty} x[n] e^{-j\omega n} \cdot \underbrace{e^{-j2\pi kn}}_1 \\ &= X(e^{j\omega}) \end{aligned}$$

Hence $X(e^{j\omega})$ is a periodic function
of ω with period 2π

