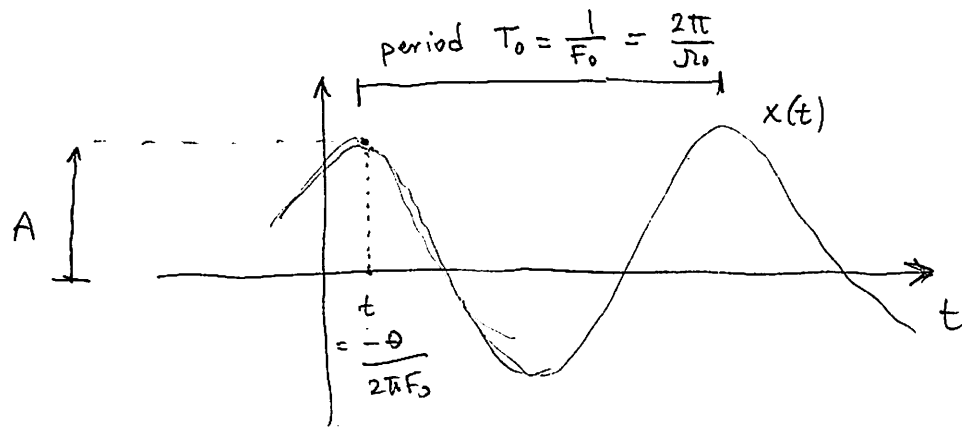


Lecture 9 (Fourier Transforms (FT), CT, DT)

Sinusoidal signals

① continuous time



$$x(t) = \boxed{A} \cos \left(2\pi \boxed{F_0} t + \boxed{\theta} \right)$$

amplitude frequency phase

$$= A \cos(\underbrace{\Omega_0 t}_{\text{angular frequency}} + \theta)$$

$$\Omega_0 = 2\pi F_0 \text{ (angular frequency)}$$

$$2\pi F_0 t + \theta = 0 \Rightarrow t = \frac{-\theta}{2\pi F_0}$$

• periodic: $x(t + T_0) = A \cos(2\pi F_0(t + T_0) + \theta) = A \cos(2\pi F_0 t + 2\pi + \theta) = x(t)$

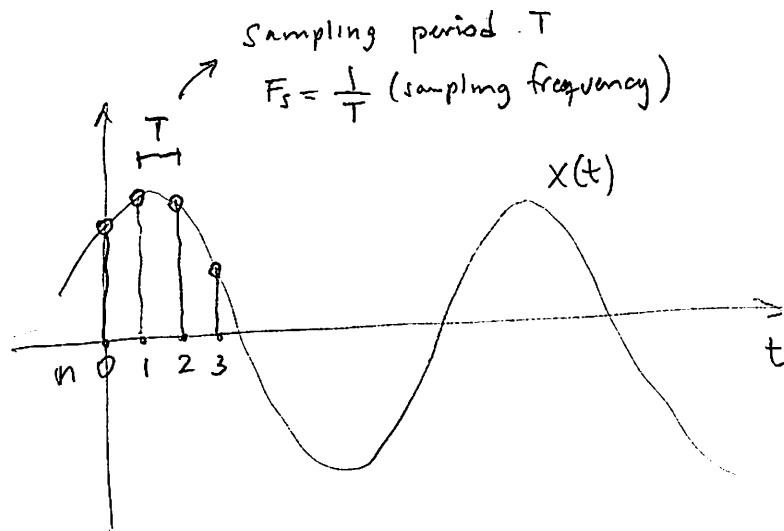
• Recall: Euler's identity: $e^{j\theta} = \cos \theta + j \sin \theta \Rightarrow \cos \theta = \frac{e^{j\theta} + e^{-j\theta}}{2}$

$$x(t) = A \cdot \frac{e^{j(\Omega_0 t + \theta)} + e^{-j(\Omega_0 t + \theta)}}{2}$$

$$= \frac{A}{2} e^{j\theta} \cdot e^{j\Omega_0 t} + \frac{A}{2} e^{-j\theta} \cdot e^{j(-\Omega_0)t}$$

(linear combination of complex sinusoids)

② Discrete-time: sampling continuous-time sinusoid $x(t) = A \cos(\Omega_0 t + \theta)$



$$x[n] = x(nT)$$

$$= A \cos(\Omega_0 nT + \theta)$$

$$= A \cos(\omega n + \theta)$$

↳ normalized angular frequency
 $\omega = \Omega_0 T$

• periodic?

$x[n] = A \cos(\omega n + \theta)$ is periodic iff $x[n] = x[n+N]$ for some $N \in \mathbb{Z}_+$

$$\Leftrightarrow A \cos(\omega n + \theta) = A \cos(\omega n + \omega N + \theta) \Leftrightarrow \omega N = 2\pi \cdot k \text{ for some } k$$

$$\Leftrightarrow \omega \text{ must be a rational multiple of } \pi$$

• A shift of 2π in the angular frequency, sinusoid doesn't change

$$x[n] = A \cos((\omega + 2\pi)n + \theta) = A \cos(\omega n + 2\pi n + \theta) = A \cos(\omega n + \theta)$$

Continuous-time Fourier Transform

\mathcal{F}

$$x(t) = \int_{-\infty}^{\infty} \left[\frac{X(\Omega)}{2\pi} \right] \cdot \boxed{e^{j\Omega t}} d\Omega$$

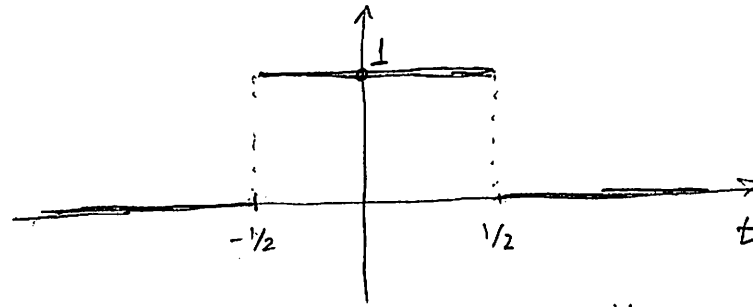
limit of a sum \rightarrow scaling/weight applied to frequency Ω \rightarrow complex sinusoid with frequency Ω

$$X(\Omega) = \underbrace{\int_{-\infty}^{\infty} x(t) e^{-j\Omega t} dt}_{\text{CTFT of } x(t)}$$

In words: Any signal $x(t)$ can be expressed as a linear combination of complex sinusoids $e^{j\Omega t}$, each with $\frac{X(\Omega)}{2\pi}$ as a weight

Ex: Rectangular signal

$$x(t) = \begin{cases} 1, & |t| < 1/2 \\ 0, & \text{otherwise} \end{cases}$$



$$X(\Omega) = \int_{-\infty}^{\infty} x(t) e^{-j\Omega t} dt = \int_{-1/2}^{1/2} 1 \cdot e^{-j\Omega t} dt = \left. \frac{e^{-j\Omega t}}{-j\Omega} \right|_{t=-1/2}^{1/2}$$

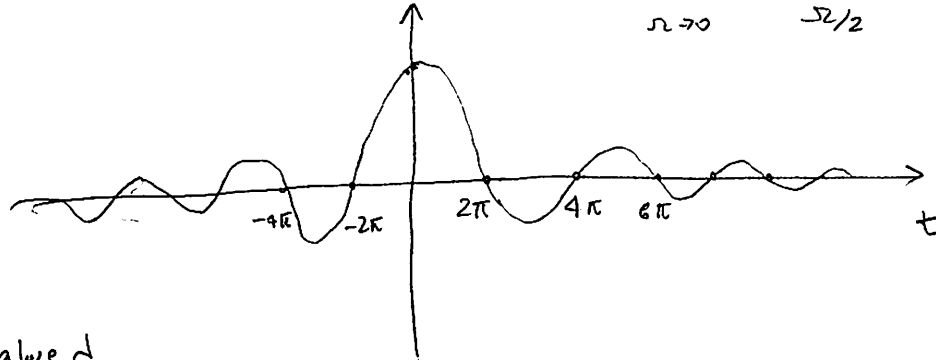
$$\sin \theta = \frac{e^{j\theta} - e^{-j\theta}}{2j}$$

$$= \frac{e^{-j\Omega/2} - e^{-j\Omega(-1/2)}}{-j\Omega} = 2 \left(\frac{e^{j\Omega/2} - e^{-j\Omega/2}}{j\Omega \cdot 2} \right)$$

$$= 2 \frac{\sin(\Omega/2)}{\Omega} = \frac{\sin(\Omega/2)}{\Omega/2}$$

In this case, $X(\Omega)$ is real-valued.

$$\lim_{\Omega \rightarrow 0} \frac{\sin(\Omega/2)}{\Omega/2} = 1$$



In general, $X(\Omega)$ is complex-valued

and we need to plot magnitude and phase

The Dirac delta function

$$\delta(t) = \begin{cases} \infty & \text{at } t=0 \\ 0 & \text{otherwise} \end{cases}$$

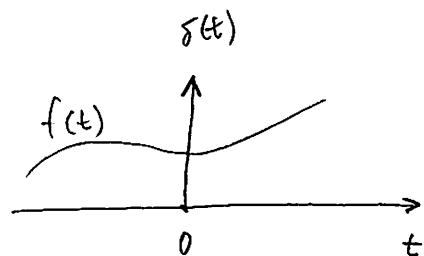
and

$$\int_{-\infty}^{\infty} \delta(t) dt = 1$$

key property:

$$\int_{-\infty}^{\infty} f(t) \delta(t) dt = f(t) \Big|_{t=0} = f(0)$$

$$\int_{-\infty}^{\infty} f(t) \delta(t-t_0) dt = f(t_0)$$



$$\begin{aligned} \text{Ex: } \int_{-\infty}^{\infty} \delta(5t) f(t) dt &= \int_{-\infty}^{\infty} \delta(u) f\left(\frac{u}{5}\right) \cdot \frac{1}{5} \cdot du = \frac{1}{5} \int_{-\infty}^{\infty} \delta(u) f\left(\frac{u}{5}\right) du \\ &= \frac{1}{5} \cdot f\left(\frac{0}{5}\right) = \frac{1}{5} f(0) \end{aligned}$$

$\begin{matrix} u = 5t \\ du = 5 dt \end{matrix}$

$$\text{Ex 2: } \int_{-\infty}^{\infty} \delta(5t-15) f(t) dt = \int_{-\infty}^{\infty} \delta(5(t-3)) f(t) dt = \frac{1}{5} f(3)$$

$$\text{CTFT of } \delta(t): \quad x(t) = \delta(t), \quad X(\omega) = \int_{-\infty}^{\infty} \overset{\delta(t)}{x(t)} e^{-j\omega t} dt = e^{-j\omega \cdot 0} = 1$$

$$\text{CTFT of } \delta(t-t_0): \quad X(\omega) = e^{-j\omega t_0}$$

Continuous-time convolution



$$y(t) = (x * h)(t) = \int_{-\infty}^{\infty} x(\tau) h(t-\tau) d\tau$$

If $x(t) = \delta(t)$, then

$$y(t) = \int_{-\infty}^{\infty} \delta(\tau) h(t-\tau) d\tau = h(t-\tau) \Big|_{\tau=0} = \underbrace{h(t)}_{\text{impulse response}}$$

We notice that convolution with $\delta(t)$ doesn't change signal:

$$x(t) * \delta(t) = x(t)$$

Ex: $e^{-t} u(t) * \delta(5t-15)$, $u(t) = \begin{cases} 1 & \text{for } t \geq 0 \\ 0 & \text{otherwise} \end{cases}$

$$\delta(5t-15) = \frac{1}{5} \delta(t-3)$$

$$\Rightarrow e^{-t} u(t) * \frac{1}{5} \delta(t-3) = \frac{1}{5} e^{-(t-3)} u(t-3)$$

CTFT

$x(t)$

CTFT

$X(\Omega)$

$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(\Omega) e^{j\Omega t} d\Omega$$

$$X(\Omega) = \int_{-\infty}^{\infty} x(t) e^{-j\Omega t} dt$$

DTFT

$x[n]$

DTFT

$X(e^{j\omega})$ (continuous function of ω)

equivalent notation: $X_d(\omega)$

$$x[n] = \frac{1}{2\pi} \int_{-\pi}^{\pi} X(e^{j\omega}) e^{j\omega n} d\omega$$

$$X(e^{j\omega}) = \sum_{n=-\infty}^{\infty} x[n] e^{-j\omega n}$$

$$\begin{aligned} X(e^{j(\omega+2\pi k)}) &= \sum_{n=-\infty}^{\infty} x[n] e^{-j(\omega+2\pi k)n} = \sum_{n=-\infty}^{\infty} x[n] e^{-j\omega n} \cdot \underbrace{e^{-j2\pi kn}}_1 \\ &= X(e^{j\omega}) \end{aligned}$$

Hence $X(e^{j\omega})$ is a periodic function of ω with period 2π

