

Final Exam

8:00-11:00AM, Wednesday, May 9, 2018

Name: _____

Section: 9:00 AM 12:00 PM 3:00 PM

NetID: _____

Score: _____

Problem	Pts.	Score
1	6	
2	4	
3	8	
4	4	
5	4	
6	6	
7	4	
8	6	
9	6	
10	6	
11	4	
12	6	
13	8	
14	6	
15	6	
16	6	
Total	90	

Instructions

- You may not use any books, calculators, or notes other than four handwritten two-sided sheets of 8.5" x 11" paper.
 - Show all your work to receive full credit for your answers.
 - When you are asked to "calculate", "determine", or "find", this means providing closed-form expressions (i.e., without summation or integration signs).
 - Neatness counts. If we are unable to read your work, we cannot grade it.
 - Turn in your entire booklet once you are finished. No extra booklet or papers will be considered.
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(6 Pts.)

1. Answer “True” or “False” for the following statements.

- (a) If the unit pulse response, $h[n]$, of an arbitrary LSI system is zero for $n < 0$, the system must be causal. **T/F**
- (b) A longer FIR filter can be designed to achieve the transition bandwidth at least as narrow as another shorter FIR filter. **T/F**
- (c) If a system is BIBO stable then it must be causal. **T/F**
- (d) The response of a BIBO *unstable* LSI system to any non-zero input is always unbounded. **T/F**
- (e) A serial connection of two BIBO stable systems is necessarily stable. **T/F**
- (f) A parallel connection of two BIBO stable systems is necessarily stable. **T/F**

(4 Pts.)

2. A sequence $x[n]$ has the z -transform

$$X(z) = \frac{z}{(z - \frac{1}{2})(z - 2)}, \quad \text{ROC : } |z| > 2.$$

Circle all correct $x[n]$ in the following list of answers.

- (a) $x[n] = -\frac{1}{3}(\frac{1}{2})^n u[n] + \frac{4}{3}2^n u[n]$
- (b) $x[n] = -\frac{1}{3}(\frac{1}{2})^{n-1} u[n-1] + \frac{4}{3}2^{n-1} u[n-1]$
- (c) $x[n] = \frac{1}{3}(\frac{1}{2})^{n-1} u[-n] - \frac{4}{3}2^{n-1} u[-n]$
- (d) $x[n] = \frac{1}{3}(\frac{1}{2})^{n-1} u[n-1] + \frac{4}{3}2^{n-1} u[n-1]$

(8 Pts.)

3. Given a **causal** LTI system with the transfer function:

$$H(z) = \frac{4}{1 - \frac{1}{2}z^{-1}} + \frac{2}{1 - \frac{1}{4}z^{-1}},$$

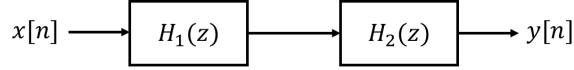
- (a) Determine each of the following:
- Locations of the all poles and zeros of the system
 - Region of convergence of $H(z)$
 - Is the system is BIBO stable?
- (b) What is the impulse response $\{h[n]\}$ of the given system?
- (c) Determine a *causal* linear constant coefficient difference equation (LCCDE) whose transfer function is as above.

(4 Pts.)

4. Consider the following cascaded system with two causal subsystems:

$$H_1(z) = \frac{z - 0.25}{z^2 - 1.5z + 0.5}, \quad \text{and}$$
$$H_2(z) = \frac{z + c}{(z - 0.25)^2}.$$

Determine c so that the following cascaded system is BIBO stable.



(4 Pts.)

5. Recall that the convolution of two discrete signals $\{x[n]\}$ and $\{h[n]\}$ is denoted as:

$$(x * h)[n] = \sum_{k=-\infty}^{\infty} x[k]h[n - k].$$

Prove that $((x * h_1) * h_2)[n] = (x * (h_1 * h_2))[n]$ for all n . Draw equivalent system block diagrams for the left side and right side of the equation.

(6 Pts.)

6. An FIR filter is described by the difference equation

$$y[n] = x[n] - x[n - 6]$$

- (a) Determine the frequency response of the system, $H_d(\omega)$.

- (b) The filter's response to the input

$$x[n] = \cos\left(\frac{\pi}{10}n\right) + 3 \sin\left(\frac{\pi}{3}n + \frac{\pi}{10}\right).$$

is given by

$$y[n] = A_1 \cos\left(\frac{\pi}{10}n + \phi_1\right) + A_2 \sin\left(\frac{\pi}{3}n + \phi_2\right).$$

Determine A_1 , A_2 , ϕ_1 , and ϕ_2 .

$$A_1 =$$

$$A_2 =$$

$$\phi_1 =$$

$$\phi_2 =$$

(4 Pts.)

7. Let $(X[k])_{k=0}^{99}$ be the 100-point **DFT** of a **real-valued** sequence $(x[n])_{n=0}^{99}$ and $X_d(\omega)$ be the **DTFT** of $x[n]$ zero-padded to infinite length. Circle all correct equations in the following list.

- (a) $X[70] = X_d\left(-\frac{6\pi}{10}\right)$
- (b) $X[70] = X_d\left(\frac{70\pi}{50}\right)$
- (c) $|X[70]| = |X_d\left(\frac{70\pi}{100}\right)|$
- (d) $\angle X[70] = -\angle X_d\left(\frac{3\pi}{5}\right)$

(6 Pts.)

8. Let $\{x[n]\}_{n=0}^{N-1}$ be a **real-valued** N -point sequence with N -point DFT $\{X[k]\}_{n=0}^{N-1}$.

- (a) Show that $X[N/2]$ is real-valued if N is even.

- (b) Show that $X[\langle N - k \rangle_N] = X^*[k]$ where $\langle n \rangle_N$ denotes n modulo N .

(6 Pts.)

9. The z-transform of $x[n]$ is

$$X(z) = \frac{1}{1 - \frac{1}{2}z^{-1}}.$$

Compute the DTFT of

(a) $\{y[n]\}$ where $y[n] = x[n] e^{-j\pi n/4}$ for all n .

(b) $\{v[n]\}$ where $v[n] = x[2n + 1]$ for all n .

(6 Pts.)

10. The signal $x_a(t) = 3 \sin(40\pi t) + 2 \cos(60\pi t)$ is sampled at a sampling period T to obtain the discrete-time signal $x[n] = x_a(nT)$.
- (a) Compute and sketch the magnitude of the continuous-time Fourier transform of $x_a(t)$.
- (b) Compute and sketch the magnitude of the discrete-time Fourier transform of $x[n]$ for 1) $T = 10 \text{ ms}$, 2) $T = 20 \text{ ms}$.
- (c) For $T = 10 \text{ ms}$ and $T = 20 \text{ ms}$, determinate whether the original continuous-time signal $x_a(t)$ can be recovered from $x[n]$.

(4 Pts.)

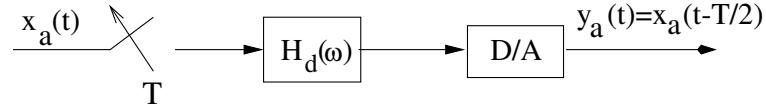
11. An analog signal $x_a(t) = \cos(200\pi t) + \sin(500\pi t)$ is to be processed by a digital signal processing (DSP) system with a digital filter sandwiched between an ideal A/D and an ideal D/A converters with the sampling frequency 1 kHz. Suppose that we want to pass the second component but stop the first component of $x_a(t)$. Sketch the specification of the digital filter in such a DSP system, and identify the transition band of that desired digital filter.

(6 Pts.)

12. Let $x[n]$ be the input to a D/A convertor with $T = 5\text{ms}$. Sketch the output signal $x_a(t)$ for the following cases. **Label your axis tick marks and units clearly.**
- The D/A convertor is a ZOH and $x[n] = 2\delta[n] + 3\delta[n - 7]$.
 - The D/A convertor is an “ideal” D/A and $x[n] = 3\delta[n - 7]$

(8 Pts.)

13. Consider the following system:



where the D/A convertor is an ideal D/A. Assume that $x_a(t)$ is bandlimited to Ω_{\max} (rad/sec), T is chosen to be $T < \frac{\pi}{\Omega_{\max}}$ and the impulse response of overall system is $h(t) = \delta(t - T/2)$ (or $H_a(\Omega) = e^{-j\Omega T/2}$).

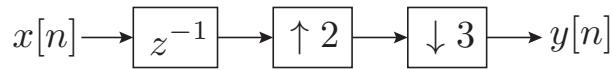
(a) Determine the frequency response $H_d(\omega)$ of the **desired** digital filter.

(b) Determine the unit pulse response $h[n]$ of the **desired** digital filter.

(c) Determine a length-2 FIR filter $g[n]$ that approximates the above desired filter $h[n]$ using a rectangular window design. Is this **designed** FIR filter $g[n]$ LP or GLP?

(4 Pts.)

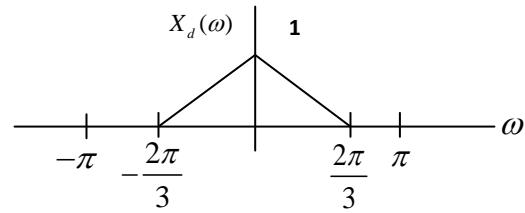
14. Consider the system in the figure below. For each of the following statements determine whether they are true or false. If false, give an example of $x[n]$ and the corresponding $y[n]$ that violate the property.



- (a) The system is linear
- (b) The system is time-invariant
- (c) The system is causal
- (d) The system is BIBO stable.

(6 Pts.)

15. Figure below shows the DTFT of a sequence $h[n]$.

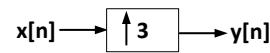


- (a) Consider the downsampling operation shown below,



Compute the maximum integer value of M you can use so that no aliasing occurs.

- (c) Consider the upsampling operation shown below. Sketch the DTFT of $y[n]$.



(6 Pts.)

16. Let \mathcal{S} be a **stable LSI** system that maps an input signal $\{x[n]\}$ to a output signal $\{y[n]\}$ as:

$$\{x[n]\} \longrightarrow \boxed{\mathcal{S}} \longrightarrow \{y[n]\}$$

Suppose that instead of $\{x[n]\}$, we input $\{\tilde{x}[n]\}$ into the system \mathcal{S} , where $\{\tilde{x}[n]\}$ differs from $\{x[n]\}$ only at one sample; that is, for some finite constants n_0 and E ,

$$\tilde{x}[n] = \begin{cases} x[n] & \text{for all } n \neq n_0, \\ x[n] + E & \text{for } n = n_0. \end{cases}$$

This produces the output signal $\{\tilde{y}[n]\}$. Show that for sufficiently large n , $\tilde{y}[n]$ approach $y[n]$, which means

$$\lim_{n \rightarrow +\infty} |\tilde{y}[n] - y[n]| = 0.$$

Hint: $\sum_{n \in \mathbb{Z}} |h[n]| < +\infty$ implies $\lim_{n \rightarrow +\infty} |h[n]| = 0$.