

Lecture 18 - FFT Algorithm (fast way to compute DFT)

Recall

DFT_N

$$\{x[n]\}_{n=0}^{N-1}$$

DFT

$$\{X[k]\}_{k=0}^{N-1}$$

$$X[k] = \sum_{n=0}^{N-1} x[n] e^{-j \frac{2\pi}{N} kn}$$

$\underbrace{\phantom{\sum_{n=0}^{N-1} x[n] e^{-j \frac{2\pi}{N} kn}}_{W_N}}$

$$\begin{aligned} \text{Note. } X[k+N] &= \sum_{n=0}^{N-1} x[n] e^{-j \frac{2\pi}{N} (k+N)n} \\ &= \sum_{n=0}^{N-1} x[n] e^{j \frac{2\pi}{N} kn} e^{-j 2\pi kn} = X[k] \end{aligned}$$

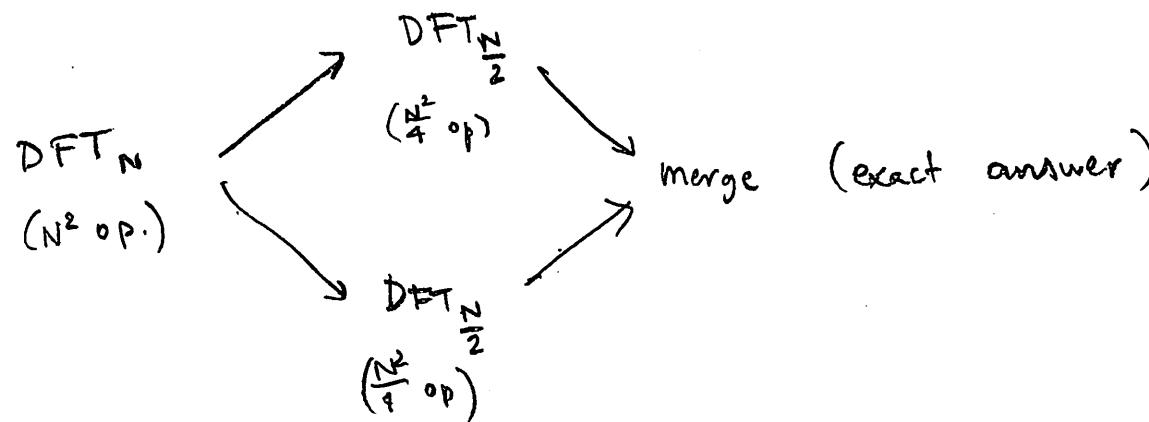
$\Rightarrow X[k]$ as N -periodic



How many operations do we need?

We need N^2 multiply-accumulate operations
 \rightarrow bad scaling! Too slow.

Main idea of the FFT: divide and conquer!



FFT (radix-2, decimation in time)

$$\text{DFT}_N : X[k] = \sum_{n=0}^{N-1} x[n] W_N^{kn} = \sum_{n \text{ even}} + \sum_{n \text{ odd}}$$

assume N
is even

$$= \sum_{m=0}^{\frac{N}{2}-1} \underbrace{x[2m]}_{a[m]} W_N^{k \cdot 2m} + \sum_{m=0}^{\frac{N}{2}-1} \underbrace{x[2m+1]}_{b[m]} W_N^{k(2m+1)}$$

$$\text{Observe: } \textcircled{1} \quad W_N^{2m} = e^{-j \frac{2\pi}{N} \cdot 2m} = e^{-j \frac{2\pi}{N} \cdot m} = W_{\frac{N}{2}}^m$$

$$\textcircled{2} \quad W_N^{2m+1} = W_N \cdot W_N^{2m} = W_N \cdot W_{\frac{N}{2}}^m$$

$$\Rightarrow X[k] = \underbrace{\sum_{m=0}^{\frac{N}{2}-1} a[m] W_{\frac{N}{2}}^{km}}_{A[k]} + W_N^k \cdot \underbrace{\sum_{m=0}^{\frac{N}{2}-1} b[m] W_{\frac{N}{2}}^{km}}_{B[k]}$$

For $k=0, \dots, \frac{N}{2}-1$, this formula is fine: $X[k] = A[k] + W_N^k B[k]$

For other values, we use periodicity of DFT.

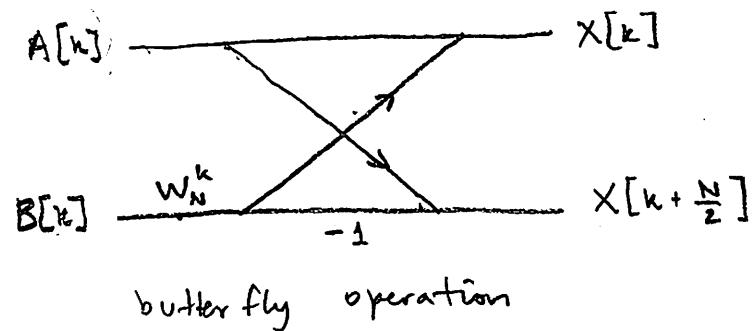
$$X\left[k + \frac{N}{2}\right] = A\left[k + \frac{N}{2}\right] + W_N^{k+\frac{N}{2}} B\left[k + \frac{N}{2}\right] = A[k] + W_N^k \cdot \underbrace{W_N^{\frac{N}{2}}}_{e^{-j \frac{2\pi}{N} \cdot \frac{N\pi}{2}}} B[k] = A[k] - B[k]$$

$$e^{-j \frac{2\pi}{N} \cdot \frac{N\pi}{2}} = -1$$

For $k=0, \dots, \frac{N}{2}-1$

$$X[k] = A[k] + W_N^k B[k]$$

$$X[k + \frac{N}{2}] = A[k] - W_N^k B[k]$$



This holds as long as N is even.

\Rightarrow we choose $N = 2^0$ so that we can apply this recursively

$$\{A[k]\}_{n=0}^{\frac{N}{2}-1} = \text{DFT}_{\frac{N}{2}} \{x[0], x[2], \dots, x[N-2]\}$$

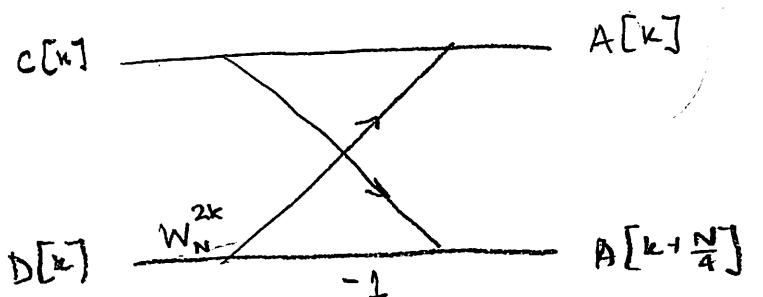
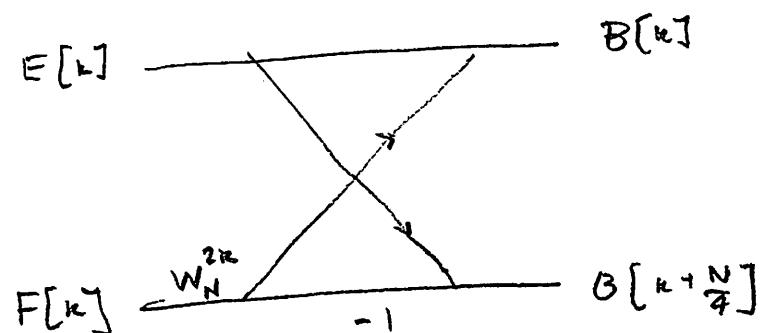
reverse \downarrow

$$A[k] = C[k] + W_{\frac{N}{2}}^k D[k]$$

$$A[k + \frac{N}{4}] = C[k] - W_{\frac{N}{2}}^k D[k] = W_N^{2k}$$

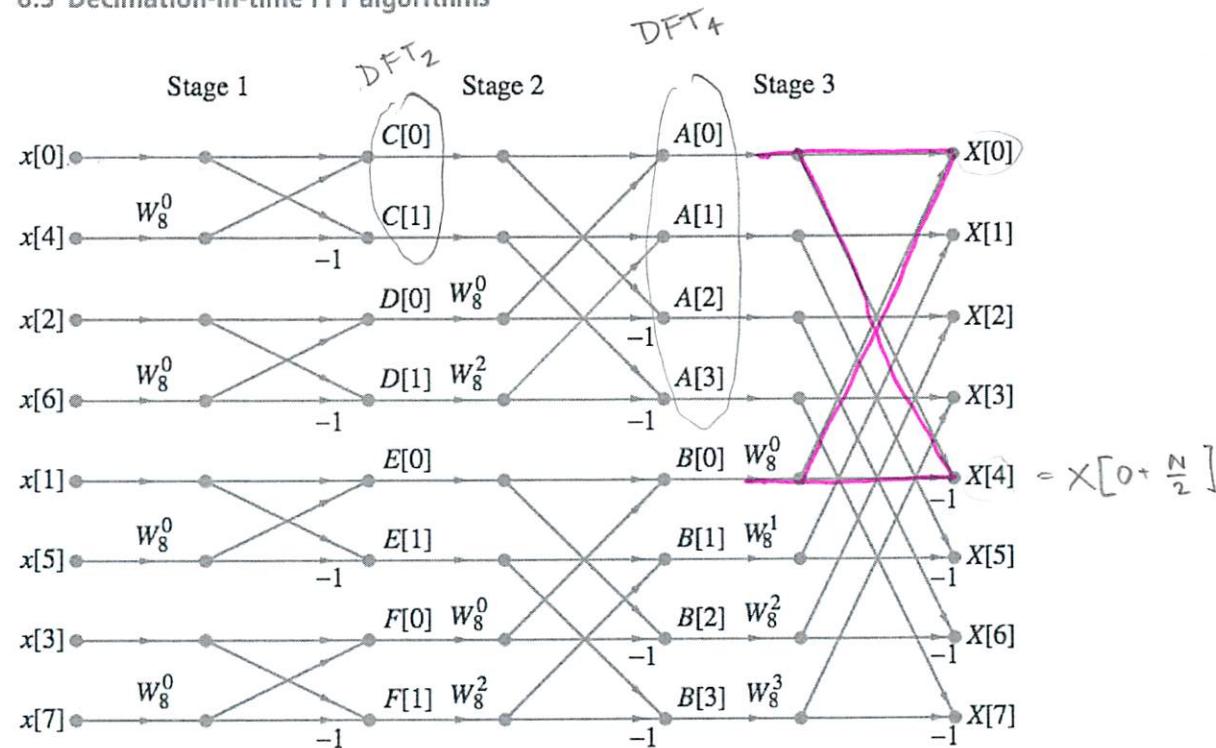
$$\{C[k]\}_{n=0}^{\frac{N}{4}} = \text{DFT}_{\frac{N}{4}} \{x[0], x[4], \dots, x[N-4]\}$$

$$\{D[k]\}_{n=0}^{\frac{N}{4}} = \text{DFT}_{\frac{N}{4}} \{x[2], x[6], \dots, x[N-2]\}$$



Ex. $N = 8$

8.3 Decimation-in-time FFT algorithms



Number of operations?

For $N = 2^v$, we

have $v = \log_2 N$

stages

N multiply-accumulate operations per stage

$\Rightarrow N \log_2 N$ operations!

Figure 8.6 Flow graph of 8-point decimation-in-time FFT algorithm using the butterfly computation shown in Figure 8.4. The trivial twiddle factor $W_8^0 = 1$ is shown for the sake of generality.

Observations:

- Decimation in time : separating even and odd time indices (decimation in frequency separates even/odd frequency indices)
- Radix-2 we break DFT_N into two $DFT_{N/2}$
- inverse FFT can be derived in similar way

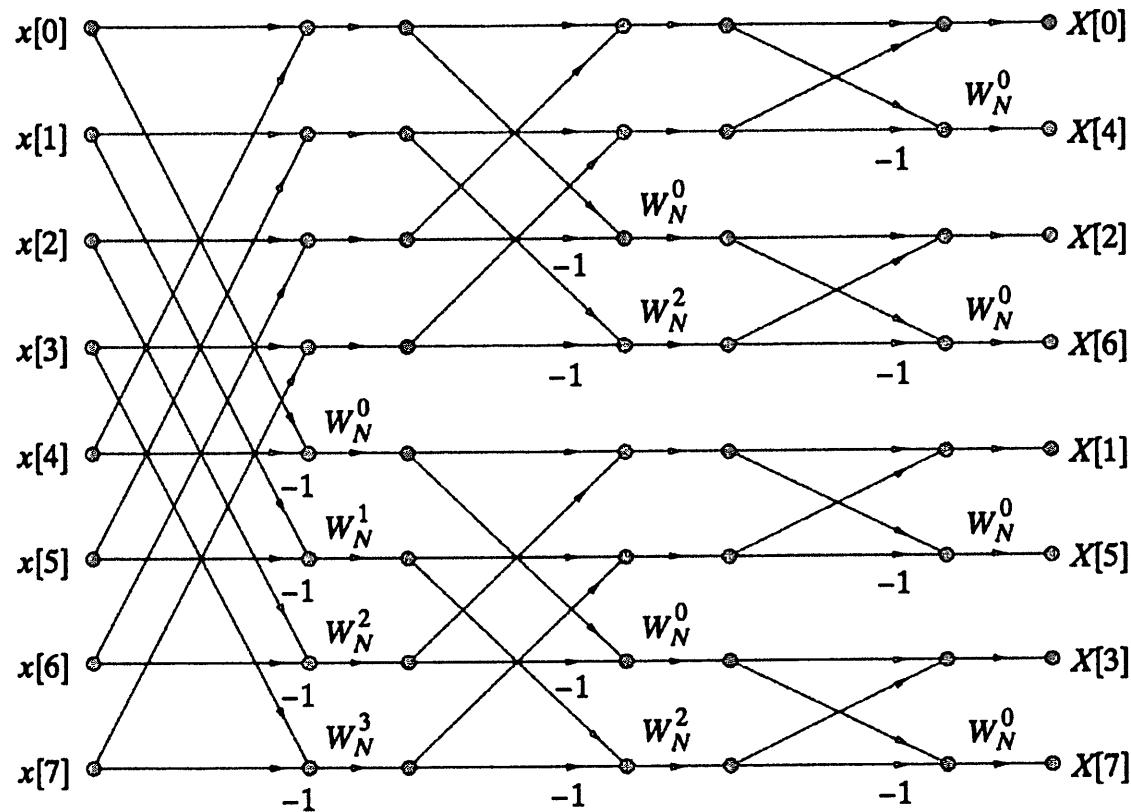
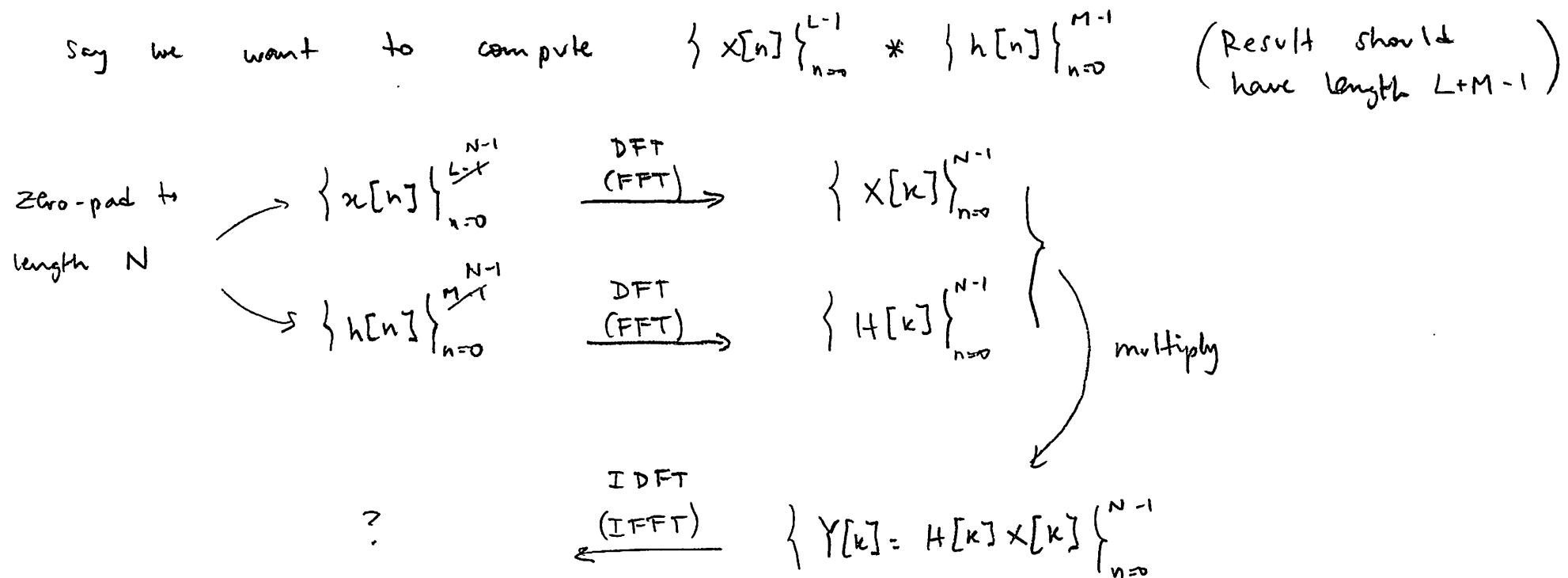


Figure 8.13 Flow graph for the decimation-in-frequency 8-point FFT algorithm. The input sequence is in natural order and the output sequence in bit-reversed order.

Fast convolution using FFT



DFT property: $x[n] \circledast_N h[n] \xleftrightarrow{\text{DFT}} X[k] H[k]$

$$\sum_{m=0}^{N-1} x[m] h[n-m]$$

→ this is not what we wanted !

How can we get (linear) convolution from circular convolution?