Lecture 3 Microstrip Line

- 1. Caculation of *RLGC* Parameters of TEM Transmission Lines
- 2. Parallel-plate Transmission Line
- 3. Microstrip Line
- 4. Transmission Line Calculators
- 5. Coding Example

1. Caculation of *RLGC* Parameters of TEM Transmission Lines

- TEM (transverse electromagnetic)
- Transverse field only: $E_z = H_z = 0$
- Number of conductors ≥ 2
- Cutoff frequency $f_c = 0$
- No dispersion (ideal case): $\beta = \frac{2\pi}{\lambda_g}$, $\lambda_g = \frac{\lambda_0}{\sqrt{\varepsilon_r}}$, $\alpha = 0$ or constant

Use of *RLGC* parameters

$$Z_0 = \sqrt{\frac{R + j\omega L}{G + j\omega C}}$$
 (characteristic impedance)

$$\gamma = \sqrt{(R + j\omega L)(G + j\omega C)} = \alpha + j\beta$$
 (propagation constant)

 α : attenuation constant

 β : phase constant

$$V(z) = V_0^+ e^{-\gamma z} + \Gamma(0) V_0^+ e^{\gamma z}$$

$$V(z) = V_0^+ e^{-\gamma z} + \Gamma(0) V_0^+ e^{\gamma z}$$
$$I(z) = \frac{V_0^+ e^{-\gamma z}}{Z_0} - \frac{\Gamma(0) V_0^+ e^{\gamma z}}{Z_0}$$

$$Z_{\rm in}(z) = \frac{V(z)}{I(z)}$$

Low-loss line

 $R << \omega L$ and $G << \omega C$

$$Z_0 = \sqrt{\frac{L}{C}} = \frac{1}{Y_0}$$
 (characteristic impedance)

 $\gamma = \alpha + j\beta$ (propagation constant)

$$\alpha = \alpha_c + \alpha_d$$
 (attenuation constant)

$$\alpha_c = \frac{R}{2Z_0}$$
 (attenuation constant due to conductor loss)

$$\alpha_d = \frac{G}{2Y_0}$$
 (attenuation constant due to dielectric loss)

$$\beta = \frac{2\pi}{\lambda_g} = \omega \sqrt{L}C \text{ (phase propagation constant)}$$

$$\lambda_g = \frac{v_p}{f}$$
 (guided wavelength)

$$v_p = \frac{1}{\sqrt{LC}}$$
 (phase velocity)

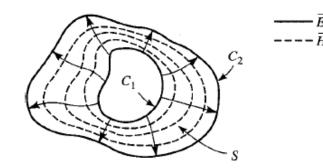
High-frequency formulas for *RLGC* parameter calculation

$$-R >> R_{\rm DC} \text{ for } f >> 1$$

$$R = \frac{R_s}{|I_o|^2} \int_{C_1 + C_2} \bar{H} \cdot \bar{H}^* dl \, \Omega/\text{m}. \qquad L = \frac{\mu}{|I_o|^2} \int_S \bar{H} \cdot \bar{H}^* ds \, \text{H/m}.$$

$$G = \frac{\omega \epsilon''}{|V_o|^2} \int_S \bar{E} \cdot \bar{E}^* ds \, \text{S/m}. \qquad C = \frac{\epsilon}{|V_o|^2} \int_S \bar{E} \cdot \bar{E}^* ds \, \text{F/m}.$$

$$G = \frac{\omega \epsilon''}{|V_o|^2} \int_S \bar{E} \cdot \bar{E}^* ds \text{ S/m.} \qquad C = \frac{\epsilon}{|V_o|^2} \int_S \bar{E} \cdot \bar{E}^* ds \text{ F/m.}$$



$$G = \frac{\omega \varepsilon''}{\varepsilon} C$$

$$R_s = \frac{1}{\sigma \delta} = \sqrt{\frac{\pi f \mu}{\sigma}}$$
 (surface resistance)

$$\delta = \sqrt{\frac{1}{\pi f \mu \sigma}}, \ \sigma = \frac{1}{\rho}$$
 (conductivity of metal)

$$\mu = \mu_0 \mu_r$$
 (permeability)

$$\varepsilon'' = \varepsilon_0 \varepsilon_r''$$
 (imaginary part of permittivity)

$$\varepsilon = \varepsilon_0 \varepsilon_r = \varepsilon_0 \varepsilon_r'$$
 (real part of permittivity)

- All-frequency formulas for *RLGC* parameter calculation
- *RLGC* parameters from DC to THz

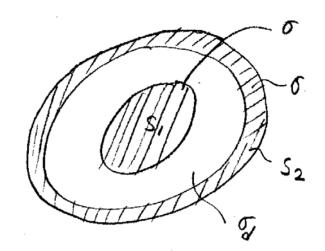
$$R'\cong R_0+\sqrt{f}R$$

$$K = K_0 + \sqrt{f} K$$

$$L' \cong L + \frac{R}{2\pi\sqrt{f}}$$

$$G' \cong G_0 + G$$

$$C' = C$$



$$R_0 = \frac{1}{\sigma S_1} + \frac{1}{\sigma S_2}$$
 (DC resistance), $S_1 \& S_2$ (conductor cross sectional area)

$$G_0 = \frac{\sigma_d}{\mathcal{E}}C$$
, σ_d (conductivity of medium filling the line)

Air-filled transmission line

$$Z_{0,\text{air}} = \sqrt{\frac{L_0}{C_0}}, \ L_0 C_0 = \mu_0 \varepsilon_0, \ \mu_0 \varepsilon_0 = 1/c^2, \ c = 3 \times 10^8 \text{ m/s (speed of light in vacuum)}$$

$$Z_{0,\text{air}} = \frac{\sqrt{\mu_0 \varepsilon_0}}{C_0} = \frac{1}{cC_0}$$

$$Z_{0,\text{air}} = \frac{\sqrt{\mu_0 \varepsilon_0}}{C_0} = \frac{1}{cC_0}$$

$$Z_{0,\text{air}} = \frac{L_0}{\sqrt{\mu_0 \varepsilon_0}} = cL_0$$

$$C_0 = \frac{1}{cZ_{0,\text{air}}}$$

$$L_0 = \frac{Z_{0,\text{air}}}{c}$$

Uniformly dielectric filled line

- *RLGC* extraction from Z_0 , ε_r , α_c , α_d

$$Z_{0,\text{air}} = \sqrt{\frac{L}{C}}, \ LC = \mu_0 \varepsilon_0 \varepsilon_r, \ \mu_0 \varepsilon_0 = 1/c^2$$

$$Z_0 = \frac{Z_{0,\text{air}}}{\sqrt{\varepsilon_r}} = \frac{\sqrt{\varepsilon_r}}{cC}$$

$$Z_0 = \frac{Z_{0,\text{air}}}{\sqrt{\varepsilon_r}} = \frac{\sqrt{\varepsilon_r}}{cC}$$

$$Z_0 = \frac{Z_{0,\text{air}}}{\sqrt{\varepsilon_r}} = \frac{L_0}{\sqrt{\mu_0 \varepsilon_0 \varepsilon_r}} = \frac{cL_0}{\sqrt{\varepsilon_r}}$$

$$C = \frac{\varepsilon_r}{cZ_{0,\text{air}}}, \quad L = \frac{Z_{0,\text{air}}}{c} = L_0$$

$$\alpha = \frac{1}{2} \left(\frac{R}{Z_0} + \frac{G}{Y_0} \right) = \alpha_c + \alpha_d \rightarrow R = 2\alpha_c Z_0, G = 2\alpha_d Y_0$$

$$Y_0 = 1/Z_0$$

Partially dielectric filled line

- *RLGC* extraction from Z_0 , ε_{re} , α_c , α_d

$$Z_{0,\text{air}} = \sqrt{\frac{L}{C}}, \ LC = \mu_0 \varepsilon_0 \varepsilon_{re}, \ \mu_0 \varepsilon_0 = 1/c^2$$

$$Z_0 = \frac{Z_{0,\text{air}}}{\sqrt{\varepsilon_{re}}} = \frac{\sqrt{\varepsilon_{re}}}{cC}$$

$$Z_0 = \frac{Z_{0,\text{air}}}{\sqrt{\varepsilon_{re}}} = \frac{\sqrt{\varepsilon_{re}}}{cC}$$

$$Z_0 = \frac{Z_{0,\text{air}}}{\sqrt{\varepsilon_{re}}} = \frac{L_0}{\sqrt{\mu_0 \varepsilon_0 \varepsilon_{re}}} = \frac{cL_0}{\sqrt{\varepsilon_{re}}}$$

$$C = \frac{\varepsilon_{re}}{cZ_{0,air}}, \quad \boxed{L = \frac{Z_{0,air}}{c} = L_0}$$

$$\alpha = \frac{1}{2} \left(\frac{R}{Z_0} + \frac{G}{Y_0} \right) = \alpha_c + \alpha_d \rightarrow R = 2\alpha_c Z_0, G = 2\alpha_d Y_0$$

$$Y_0 = 1/Z_0$$

Constants in *RLGC* calculation

$$\omega = 2\pi f$$

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$$\mu = \mu_0 = 4\pi \times 10^{-7} \text{ H/m (non-magnetic dielectric materials)}$$

$$\varepsilon' = \varepsilon_0 \varepsilon_r \text{ (real part of complex permittivity)}$$

$$\varepsilon' = \varepsilon_0 \varepsilon_r$$
 (real part of complex permittivity)

$$\varepsilon_0 = 8.854 \times 10^{-12} \text{ F/m (permittivity of vacuum)}$$

 ε_r : dielectric constant

 $\varepsilon'' = \varepsilon_0 \varepsilon_r \tan \delta$ (imaginary part of complex permittivity)

 $\tan \delta$: loss tangent

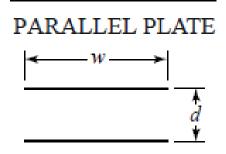
$$R_s = \frac{1}{\sigma \delta} = \sqrt{\frac{\pi f \mu}{\sigma}}$$
 (surface resistance)

 σ : conductivity (S/m), $\sigma = 5.8 \times 10^7$ S/m (copper)

$$\delta = \sqrt{\frac{1}{\pi f \,\mu \sigma}} \,\,(\text{m}) : \text{skin depth}$$

2. Parallel-plate Transmission Lines

Formula summary



$$\frac{\mu d}{w}$$

$$\frac{\epsilon' w}{d}$$

$$\frac{2R_s}{w}$$

$$\frac{\omega \epsilon'' w}{d}$$

Formula derivation

$$V = -E_y h$$
, $I = H_x w$

$$Z = \frac{V}{I} = \frac{-E_y h}{H_x w} = \eta \, \frac{h}{w}$$

$$E_y = -\eta H_X$$

Torificial derivation
$$V = -E_y h, \quad I = H_x w$$

$$Z = \frac{V}{I} = \frac{-E_y h}{H_x w} = \eta \frac{h}{w}$$

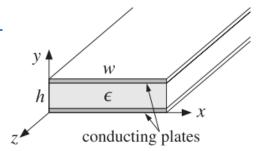
$$E_y = -\eta H_x.$$

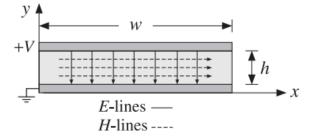
$$P_T = \frac{1}{2\eta} |E_y|^2 (wh) = \frac{1}{2\eta} \frac{V^2}{h^2} wh = \frac{1}{2\eta} \frac{w}{h} V^2 = \frac{1}{2Z} V^2 = \frac{1}{2} ZI^2$$

$$L' = \mu \frac{h}{w}, \quad C' = \epsilon \frac{w}{h}$$

$$P'_{\text{loss}} = 2\frac{1}{2}R_s|H_x|^2w = \frac{1}{w}R_sI^2$$

$$\alpha_c = \frac{P'_{\rm loss}}{2P_T} = \frac{R'}{2Z} = \frac{R_s}{wZ} = \frac{R_s}{h\eta}$$





Example

Find R, L, G, C of the following lines with f = 10 GHz, $\varepsilon_r = 2$, $\tan \delta = 0.002$, $\sigma = 5 \times 107$ S/m,

1. Parallel-plate: w = 10 mm, d = 2 mm

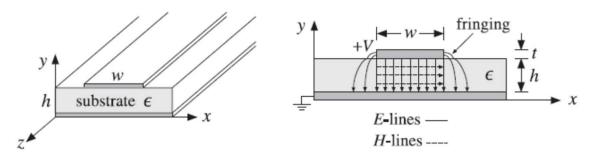
Example

Find Z_0 and γ of the following lines with f = 10 GHz, $\varepsilon_r = 2$, $\tan \delta = 0.002$, $\sigma = 5 \times 107$ S/m,

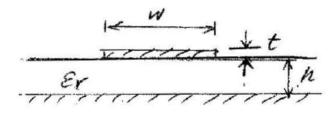
1. Parallel-plate: w = 10 mm, d = 2 mm

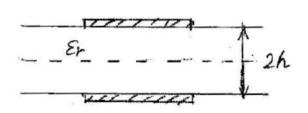
3. Microstrip Line

Microstrip structure



Microstrip vs parallel strip



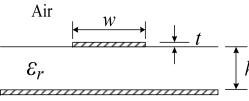


- Image theory

$$Z_{0,\text{microstrip}}(h) = \frac{1}{2}Z_{0,\text{parallel-strip}}(2h)$$

Microstrip formulas

$$u = w / h$$



$$Z_{0,\text{air}} = 30 \log \left\{ 1 + (4/u) \left[8/u + \sqrt{(8/u)^2 + \pi^2} \right] \right\} \text{ (air-filled microstrip line) [Wheeler(1977)]}$$

$$\varepsilon_{re} = \frac{\varepsilon_r + 1}{2} + \frac{\varepsilon_r - 1}{2} \frac{1}{\sqrt{1 + 12/u}} \text{ (effective dielectric constant) [Hammerstad(1975)]}$$

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 (effective dielectric constant) [Hammerstad(1975)]

$$Z_0 = \frac{Z_{0,\text{air}}}{\sqrt{\varepsilon_{re}}}$$
 (characteristic impedance of microstrip line)

$$R = \frac{2R_s}{w}$$
, $R_s = \sqrt{\frac{\pi f \mu}{\sigma}}$, $\alpha_c = \frac{R}{2Z_0}$ (attenuation due to conductor loss) (parallel-plate approximation)

$$G = (\omega \tan \delta) \frac{w}{h}$$
, $\alpha_d = \frac{G}{2Y_0}$ (attenuation due to dielectric loss) (parallel-plate approximation)

$$\alpha = \alpha_c + \alpha_d$$
 (total attenuation)

$$\beta = \frac{2\pi}{\lambda_g}, \ \lambda_g = \frac{\lambda_0}{\sqrt{\varepsilon_{re}}}$$

$$C = \frac{\varepsilon_{re}}{cZ_{0,air}}, L = \frac{Z_{0,air}}{c} = L_0$$

Example

Find R, L, G, C, Z_0 , γ of microstrip lines with f = 10 GHz, $\varepsilon_r = 2$, $\tan \delta = 0.002$, $\sigma = 5 \times 107$ S/m,

- 1. h = 0.580 mm, w = 2 mm
- 2. h = 0.787 mm, w = 0.2 mm
- 3. h = 1.27 mm, w = 4 mm

4. Transmission Line Calculator

CST Studio Suite

- Calculation module
- Draw transmission line structure, set up two wave ports, and do port analysis only.

2. Microstrip Analysis/Synthesis Calculator: loss included

http://mcalc.sourceforge.net/

3. RFDH

Coaxial line calculator: http://www.rfdh.com/rfdb/coaxline.htm

Microstrip calculator: http://www.rfdh.com/rfdb/msline.htm

Inverted microstrip calculator: http://www.rfdh.com/rfdb/msline i.htm

Suspended microstrip calculator: http://www.rfdh.com/rfdb/msline_s.htm

Stripline calculator: http://www.rfdh.com/rfdb/sline.htm

Coplanar waveguide (CPW) calculator: http://www.rfdh.com/rfdb/cpw.htm

GCPW (CPW with ground) calculator: http://www.rfdh.com/rfdb/cpwg.htm

Coplnar strip (CPS) calculator: http://www.rfdh.com/rfdb/cps.htm

Slot line calculator: http://www.rfdh.com/rfdb/slotline.htm

5. Coding Example

Example

```
Microstrip:
Given f, w, h, \varepsilon_r, \tan \delta, \sigma
Find Z_0, \varepsilon_{re}, R, L, G, C, \alpha_c (dB/m), \alpha_d (dB/m), \alpha (dB/m), \beta, \lambda_{\sigma}
(Solution)
# Microwave Engingineering, Lecture 3 Python Code
# Microstrip calculation:
# Input: f, w, h, er, tand, sigma
# Output: ZO, ere, R, L, G, C, ac, ad, a, beta, lambda q
from math import *
pi=3.14159265; mu=4*pi*1e-7; v=3e8; e0=8.854e-12
while True:
  f=float(input('Frequency f (Hz, 1e9) ='))
  w=float(input('Strip width w (mm, 2) ='))
  h=float(input('Substrate thickness h (mm, 1) ='))
  er=float(input('Substrate dielectric constant er (4.3) ='))
  tand=float(input('Substrate loss tangent tand (0.02)='))
  sigma=float(input('Strip/groundplane conductivity sigma (S/m, 5.8e7)='))
  u=w/h
  z0air=30*log(1+4/u*(8/u+sgrt(u**2+pi**2)))
  ere=(er+1)/2+(er-1)/2/sqrt(1+12/u)
```

5. Coding Example

```
z0=z0air/sqrt(ere)
lambda 0=3e8/f
lambda_g=lambda_0/sqrt(ere)
beta=2*pi/lambda_q
rs=sqrt(pi*f*mu/sigma)
r=2*rs/(w*1e-3)
L=z0air/v
c=ere/(v*z0air)
g=2*pi*f*e0*er*tand*w/h
ac=r/(2*z0);ad=g/(2/z0);a=ac+ad
print('Z0 (ohm) = ',z0)
print(' ere =',ere)
print(' R (Ohm/m) = ',r)
print(' G (Ohm/m) = ',g)
print(' L (H/m) = ',L)
print(' C (F/m) = ',c)
print(' ac (dB/m) = ',ac*8.68)
print(' ad (dB/m) =',ad*8.68)
print(' a (dB/m) = ',a*8.68)
print(' beta (rad/m) =',beta)
print(' lambda_g (m) =',lambda_g)
```

5. Coding Example

```
1 1 1
Frequency f (Hz, 1e9) =
1e9
Strip width w (mm, 2) =
Substrate thickness h (mm, 1) =
Substrate dielectric constant er (4.3) =
4.3
Substrate loss tangent tand (0.02)=
0.02
Strip/groundplane conductivity sigma (S/m, 5.8e7)=
5.8e7
  Z0 \text{ (ohm)} = 46.43003113343705
  ere = 3.2736413804652247
 R (Ohm/m) = 8.250226487396457
  G (Ohm/m) = 0.0095685874951464
 L (H/m) = 2.800227207970832e-07
  C (F/m) = 1.2989586349544232e-10
  ac (dB/m) = 0.7711815409383731
  ad (dB/m) = 1.9281309984135606
  a (dB/m) = 2.6993125393519337
 beta (rad/m) = 37.894323997365646
  lambda q(m) = 0.16580808514849868
Frequency f (Hz, 1e9) =
1 1 1
```

Fin (End)