











TLV62568A, TLV62569A

#### SLVSE95B - APRIL 2018 - REVISED MARCH 2020

# TLV6256xA 1-A, 2-A Step Down Converter with Forced PWM in SOT563 Package

#### **Features**

- Forced PWM to reduce output voltage ripple
- Up to 95% efficiency
- Low  $R_{DS(ON)}$  switches: 100 m $\Omega$  / 60 m $\Omega$
- 2.5-V to 5.5-V input voltage range
- Adjustable output voltage from 0.6 V to VIN
- 100% duty cycle for lowest dropout
- 1.5-MHz typical switching frequency
- Power good output
- Over current protection
- Internal soft startup
- Thermal shutdown protection
- Available in SOT563 package
- Pin-to-pin compatible with TLV62568, TLV62569
- Create a custom design with the WEBENCH® **Power Designer**

# Applications

- General purpose point-of-load (POL) supply
- STB & DVR
- IP network camera
- Wireless router
- Solid state drive (SSD) enterprise

# 3 Description

The TLV62568A, TLV62569A devices synchronous step-down buck DC-DC converters optimized for high efficiency and compact solution size. The device integrates switches capable of delivering an output current up to 2 A. At the whole load range, the device operates in pulse width modulation (PWM) mode with 1.5-MHz switching frequency. In shutdown, the current consumption is reduced to less than 2 μA.

An internal soft start circuit limits the inrush current during startup. Other features like over current protection, thermal shutdown protection and power good are built-in. The device is available in a SOT563 package.

#### Device Information(1)

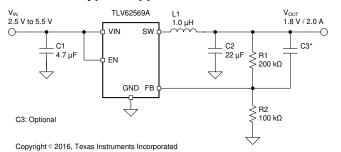
PART NUMBER	PACKAGE	BODY SIZE (NOM)
TLV62568ADRL		
TLV62568APDRL	COTEGO (C)	1 60 mm v 1 60 mm
TLV62569ADRL	SOT563 (6)	1.60 mm x 1.60 mm
TLV62569APDRL		

(1) For all available packages, see the orderable addendum at the end of the datasheet.

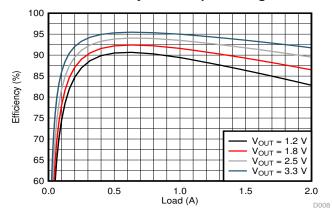
### **Device Comparison**

PART NUMBER	OUTPUT CURRENT	FUNCTION
TLV62568ADRL	1 A	-
TLV62568APDRL	IA	Power Good
TLV62569ADRL	2 A	-
TLV62569APDRL		Power Good

### **Typical Application Schematic**



### Efficiency at 5-V Input Voltage





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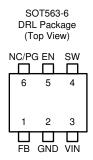
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# 4 Revision History

Cł	Changes from Revision A (May 2018) to Revision B	Page
•	Changed Power Good pin sink current capability from 1 mA to 2 mA	
Cł	Changes from Original (April 2018) to Revision A	Page
•	Changed status from Advance Information to Production Data	



# 5 Pin Configuration and Functions



#### **Pin Functions**

	SOT563-6			
NAME	PIN NUMBER	I/O/PWR	DESCRIPTION	
FB	1	I	Feedback pin for the internal control loop. Connect this pin to an external feedback divider.	
GND	2	PWR	Ground pin.	
VIN	3	PWR	Power supply voltage input.	
sw	4	PWR	Switch pin connected to the internal FET switches and inductor terminal. Connect the inductor of the output filter to this pin.	
EN	5	I	Device enable logic input. Logic high enables the device, logic low disables the device and turns it into shutdown. Do not leave floating.	
PG	6	0	Power good open drain output pin for TLV62569APDRL. The pull-up resistor should not be connected to any voltage higher than 5.5V. If it's not used, leave the pin floating.	
NC	6	-	No connection pin for TLV62569ADRL. The pin can be connected to the output or the ground for enhancing thermal performance. Or leave it floating.	

# 6 Specifications

# 6.1 Absolute Maximum Ratings

over operating temperature range (unless otherwise noted)(1)

		MIN	MAX	UNIT
	VIN, EN, PG	-0.3	6	
Voltage <sup>(2)</sup>	SW (DC)	-0.3	$V_{IN} + 0.3$	V
voitage	SW (AC, less than 10ns) <sup>(3)</sup>	-3.0	9	V
	FB	-0.3	3	
$T_{J}$	Junction temperature	-40	150	°C
T <sub>stg</sub>	Storage temperature	-65	150	°C

<sup>(1)</sup> Stresses beyond those listed under Absolute Maximum Rating may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under Recommended Operating Condition. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

# 6.2 ESD Ratings

V.— — •	011 105 Kumigo						
			VALUE	UNIT			
V	Electrostatio discharge	Human body model (HBM), per ANSI/ESDA/JEDEC JS-001, all pins <sup>(1)</sup>	±2000	N/			
V <sub>(ESD)</sub>	Electrostatic discharge	Charged device model (CDM), per JEDEC specification JESD22-C101, all pins <sup>(2)</sup>	±500	V			

(1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.

(2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

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<sup>(2)</sup> All voltage values are with respect to network ground terminal.

<sup>(3)</sup> While switching.



# 6.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

		MIN	NOM MAX	UNIT
V <sub>IN</sub>	Input voltage	2.5	5.5	V
$V_{OUT}$	Output voltage	0.6	$V_{IN}$	V
I <sub>OUT</sub>	Output current	0	2	Α
TJ	Junction temperature	-40	125	°C

# 6.4 Thermal Information

		TLV62568Ax,	TLV62568Ax, TLV62569Ax			
	THERMAL METRIC (1)	JEDEC (DRL)	EVM (DRL)	UNIT		
		6 PINS	6 PINS			
$R_{\theta JA}$	Junction-to-ambient thermal resistance	142.8	124.8	°C/W		
$R_{\theta JC(top)}$	Junction-to-case (top) thermal resistance	51.1	n/a <sup>(2)</sup>	°C/W		
$R_{\theta JB}$	Junction-to-board thermal resistance	28.9	n/a <sup>(2)</sup>	°C/W		
$\Psi_{JT}$	Junction-to-top characterization parameter	1.4	1.6	°C/W		
$\Psi_{JB}$	Junction-to-board characterization parameter	28.7	23.1	°C/W		

<sup>(1)</sup> For more information about traditional and new thermal metrics, see the Semiconductor and IC Package Thermal Metrics application report.

### 6.5 Electrical Characteristics

 $V_{IN}$  = 5.0 V,  $T_J$  = 25 °C, unless otherwise noted

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
SUPPLY						
I <sub>SD</sub>	Shutdown current into VIN pin	EN = 0 V		0.01	2	μA
\/	Under voltage lock out	V <sub>IN</sub> falling		2.3	2.45	V
$V_{UVLO}$	under voltage lock out hysteresis			100		mV
<b>-</b>	The average objected according	T <sub>J</sub> rising		150		00
$T_{JSD}$	Thermal shutdown	T <sub>J</sub> falling		130		°C
LOGIC IN	TERFACE					
V <sub>IH</sub>	High-level input voltage at EN pin	2.5 ≤ V <sub>IN</sub> ≤ 5.5	1.2			V
V <sub>IL</sub>	Low-level input voltage at EN pin	2.5 ≤ V <sub>IN</sub> ≤ 5.5			0.4	V
t <sub>SS</sub>	Soft startup time	From EN high to 95% of VOUT nominal		0.9		ms
	Dawar and threat ald	V <sub>FB</sub> rising, referenced to V <sub>FB</sub> nominal		95%		
$V_{PG}$	Power good threshold	V <sub>FB</sub> falling, referenced to V <sub>FB</sub> nominal		90%		
$V_{PG,OL}$	Low-level output voltage at PG pin	I <sub>SINK</sub> = 1 mA			0.4	V
I <sub>PG,LKG</sub>	Input leakage current into PG pin	V <sub>PG</sub> = 5 V		100		nA
t <sub>PG,DLY</sub>	Power good delay time	V <sub>FB</sub> falling		40		μs

<sup>(2)</sup> Not applicable to an EVM.

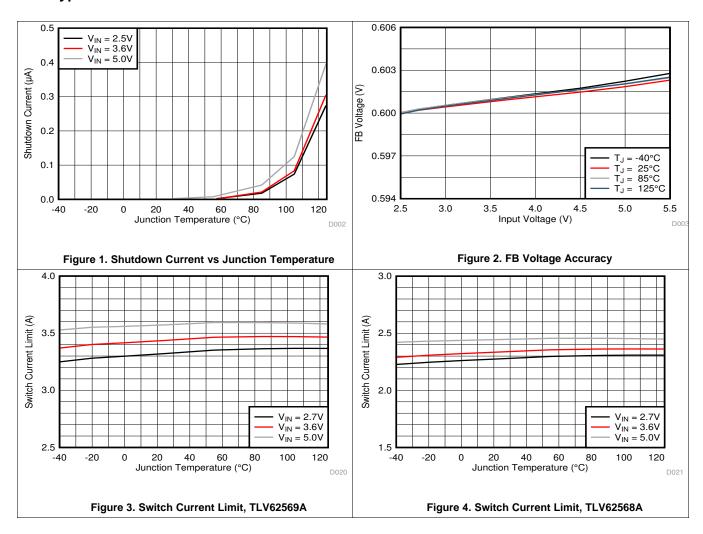


# **Electrical Characteristics (continued)**

 $V_{IN}$  = 5.0 V,  $T_J$  = 25 °C, unless otherwise noted

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
OUTPUT					·	
V <sub>FB</sub>	Feedback regulation voltage		0.588	0.6	0.612	V
I <sub>FB</sub>	Input leakage current into FB pin	V <sub>FB</sub> = 0.6 V		10		nA
_	High-side FET on resistance			100		0
R <sub>DS(on)</sub>	Low-side FET on resistance			60		mΩ
	High-side FET current limit	TLV62569A, TLV62569AP	3			А
ILIM		TLV62568A, TLV62568AP	2			
f <sub>SW</sub>	Switching frequency			1.5		MHz

# 6.6 Typical Characteristics





### 7 Detailed Description

#### 7.1 Overview

The device is a high-efficiency synchronous step-down converter. The device operates with an adaptive off time with peak current control scheme. The device operates at typically 1.5-MHz frequency pulse width modulation (PWM) . Based on the  $V_{\text{IN}}/V_{\text{OUT}}$  ratio, a simple circuit sets the required off time for the low-side MOSFET. It makes the switching frequency relatively constant regardless of the variation of input voltage, output voltage, and load current.

### 7.2 Functional Block Diagrams

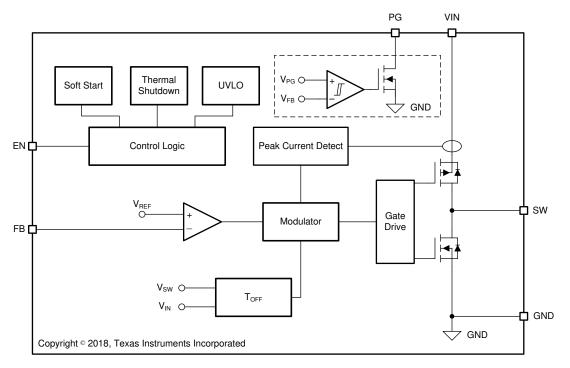


Figure 5. TLV62569A Functional Block Diagram

#### 7.3 Feature Description

### 7.3.1 100% Duty Cycle Low Dropout Operation

The device offers a low input-to-output voltage differential by entering 100% duty cycle mode. In this mode, the high-side MOSFET switch is constantly turned on and the low-side MOSFET is switched off. The minimum input voltage to maintain output regulation, depending on the load current and output voltage, is calculated as:

$$V_{IN(MIN)} = V_{OUT} + I_{OUT} \times (R_{DS(ON)} + R_L)$$

where

R<sub>DS(ON)</sub> = High side FET on-resistance

#### 7.3.2 Soft Startup

After enabling the device, internal soft startup circuitry ramps up the output voltage which reaches nominal output voltage during a startup time. This avoids excessive inrush current and creates a smooth output voltage rise slope. It also prevents excessive voltage drops of primary cells and rechargeable batteries with high internal impedance.



### **Feature Description (continued)**

The device is able to start into a pre-biased output capacitor. The converter starts with the applied bias voltage and ramps the output voltage to its nominal value.

#### 7.3.3 Switch Current Limit

The switch current limit prevents the device from high inductor current and drawing excessive current from a battery or input voltage rail. Excessive current might occur with a heavy load or shorted output circuit condition. The device adopts the peak current control by sensing the current of the high-side switch. Once the high-side switch current limit is reached, the high-side switch is turned off and low-side switch is turned on to ramp down the inductor current with an adaptive off-time.

### 7.3.4 Under Voltage Lockout

To avoid mis-operation of the device at low input voltages, under voltage lockout is implemented that shuts down the device at voltages lower than  $V_{\text{UVLO}}$  with  $V_{\text{HYS\_UVLO}}$  hysteresis.

#### 7.3.5 Thermal Shutdown

The device enters thermal shutdown once the junction temperature exceeds the thermal shutdown rising threshold,  $T_{JSD}$ . Once the junction temperature falls below the falling threshold, the device returns to normal operation automatically.

# 7.4 Device Functional Modes

### 7.4.1 Enabling/Disabling the Device

The device is enabled by setting the EN input to a logic High. Accordingly, a logic Low disables the device. If the device is enabled, the internal power stage starts switching and regulates the output voltage to the set point voltage. The EN input must be terminated and should not be left floating.

#### 7.4.2 Power Good

The TLV62568AP and TLV62569AP have a power good output. The PG pin goes high impedance once the output is above 95% of the nominal voltage, and is driven low once the output voltage falls below typically 90% of the nominal voltage. The PG pin is an open-drain output and is specified to sink up to 2 mA. The power good output requires a pull-up resistor connecting to any voltage rail less than 5.5 V. The PG signal can be used for sequencing of multiple rails by connecting it to the EN pin of other converters. Leave the PG pin unconnected when not used.

Table 1. PG Pin Logic

DEVICE CONDITIONS		LOGIC STATUS		
	DEVICE CONDITIONS		LOW	
Enable	EN = High, V <sub>FB</sub> ≥ V <sub>PG</sub>	V		
Enable	$EN = High, V_{FB} \le V_{PG}$		$\checkmark$	
Shutdown	EN = Low		√	
Thermal Shutdown	$T_{J} > T_{JSD}$		√	
UVLO	$1.4 \text{ V} < \text{V}_{\text{IN}} < \text{V}_{\text{UVLO}}$		√	
Power Supply Removal	V <sub>IN</sub> ≤ 1.4 V	V		



# Application and Implementation

#### NOTE

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

# 8.1 Application Information

The following section discusses the design of the external components to complete the power supply design for several input and output voltage options by using typical applications as a reference.

# 8.2 Typical Application

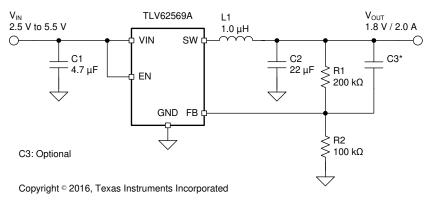


Figure 6. TLV62569A 1.8-V Output Application

#### 8.2.1 Design Requirements

For this design example, use the parameters listed in Table 2 as the input parameters.

**Table 2. Design Parameters** 

DESIGN PARAMETER	EXAMPLE VALUE				
Input voltage	2.5 V to 5.5 V				
Output voltage	1.8 V				
Maximum output current	2.0 A				

Table 3 lists the components used for the example.

**Table 3. List of Components** 

REFERENCE	DESCRIPTION	MANUFACTURER (1)
C1	4.7 μF, Ceramic Capacitor, 10 V, X7R, size 0805, GRM21BR71A475KA73L	Murata
C2	22 μF, Ceramic Capacitor, 6.3 V, X7T, size 0805, GRM21BD70J226ME44	Murata
L1	1.0 µH, Power Inductor, size 4mmx4mm, XAL4020-102ME	Coilcraft
R1,R2,R3	Chip resistor,1%,size 0603	Std.
C3	Optional, 10 pF if it is needed	Std.

(1) See Third-party Products Disclaimer

# 8.2.2 Detailed Design Procedure

# 8.2.2.1 Custom Design With WEBENCH® Tools

Click here to create a custom design using the TLV62569A device with the WEBENCH® Power Designer.



- Start by entering the input voltage (V<sub>IN</sub>), output voltage (V<sub>OUT</sub>), and output current (I<sub>OUT</sub>) requirements.
- 2. Optimize the design for key parameters such as efficiency, footprint, and cost using the optimizer dial.
- 3. Compare the generated design with other possible solutions from Texas Instruments.

The WEBENCH Power Designer provides a customized schematic along with a list of materials with real-time pricing and component availability.

In most cases, these actions are available:

- Run electrical simulations to see important waveforms and circuit performance
- Run thermal simulations to understand board thermal performance
- Export customized schematic and layout into popular CAD formats
- Print PDF reports for the design, and share the design with colleagues

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### 8.2.2.2 Setting the Output Voltage

An external resistor divider is used to set output voltage according to Equation 2.

When sizing R2, in order to achieve low current consumption and acceptable noise sensitivity, use a maximum of 200 k $\Omega$  for R2. Larger currents through R2 improve noise sensitivity and output voltage accuracy but increase current consumption.

$$V_{OUT} = V_{FB} \times \left(1 + \frac{R1}{R2}\right) = 0.6V \times \left(1 + \frac{R1}{R2}\right)$$
 (2)

A feed forward capacitor, C3 improves the loop bandwidth to make a fast transient response (shown in Figure 24). A 10-pF capacitance is recommended for R2 of 100-k $\Omega$  resistance. A more detailed discussion on the optimization for stability vs. transient response can be found in SLVA289.

### 8.2.2.3 Output Filter Design

The inductor and output capacitor together provide a low-pass filter. To simplify this process, Table 4 outlines possible inductor and capacitor value combinations. Checked cells represent combinations that are proven for stability by simulation and lab test. Further combinations should be checked for each individual application.

Table 4. Matrix of Output Capacitor and Inductor Combinations

V [V]	L [µH] <sup>(1)</sup>	C <sub>OUT</sub> [μF] <sup>(2)</sup>							
V <sub>OUT</sub> [V]	L լµпյ` /	4.7	10	22	47	100			
0.6 ≤ V <sub>OUT</sub> < 1.2	1				+				
1.2 ≤ V <sub>OUT</sub>	1			++(3)	+				

- (1) Inductor tolerance and current de-rating is anticipated. The effective inductance can vary by +20% and -30%.
- (2) Capacitance tolerance and bias voltage de-rating is anticipated. The effective capacitance can vary by +20% and -50%.
- (3) This LC combination is the standard value and recommended for most applications.

#### 8.2.2.4 Inductor Selection

The main parameters for inductor selection is inductor value and then saturation current of the inductor. To calculate the maximum inductor current under static load conditions, Equation 3 is given:

$$I_{L,MAX} = I_{OUT,MAX} + \frac{\Delta I_L}{2}$$

$$\Delta I_L = V_{OUT} \times \frac{1 - \frac{V_{OUT}}{V_{IN}}}{L \times f_{SW}}$$

where:

- I<sub>OUT,MAX</sub> is the maximum output current
- ΔI<sub>L</sub> is the inductor current ripple



- f<sub>SW</sub> is the switching frequency
- L is the inductor value

It is recommended to choose a saturation current for the inductor that is approximately 20% to 30% higher than  $I_{L,MAX}$ . In addition, DC resistance and size should also be taken into account when selecting an appropriate inductor.

### 8.2.2.5 Input and Output Capacitor Selection

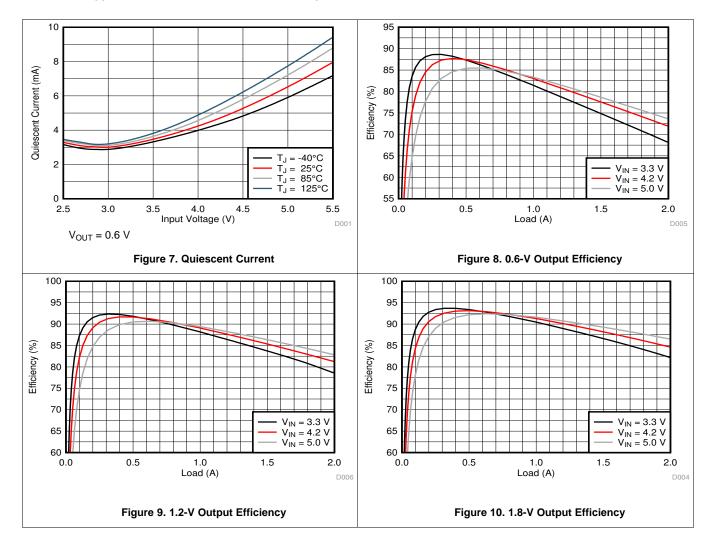
The architecture of the device allows use of tiny ceramic-type output capacitors with low equivalent series resistance (ESR). These capacitors provide low output voltage ripple and are thus recommended. To keep its resistance up to high frequencies and to achieve narrow capacitance variation with temperature, it is recommended to use X7T or X5R dielectric.

The input capacitor is the low impedance energy source for the converter that helps provide stable operation. A low ESR multilayer ceramic capacitor is recommended for best filtering. For most applications,  $4.7-\mu F$  input capacitance is sufficient; a larger value reduces input voltage ripple.

The device is designed to operate with an output capacitor of 22 µF to 47 µF, as outlined in Table 4.

# 8.2.3 Application Performance Curves

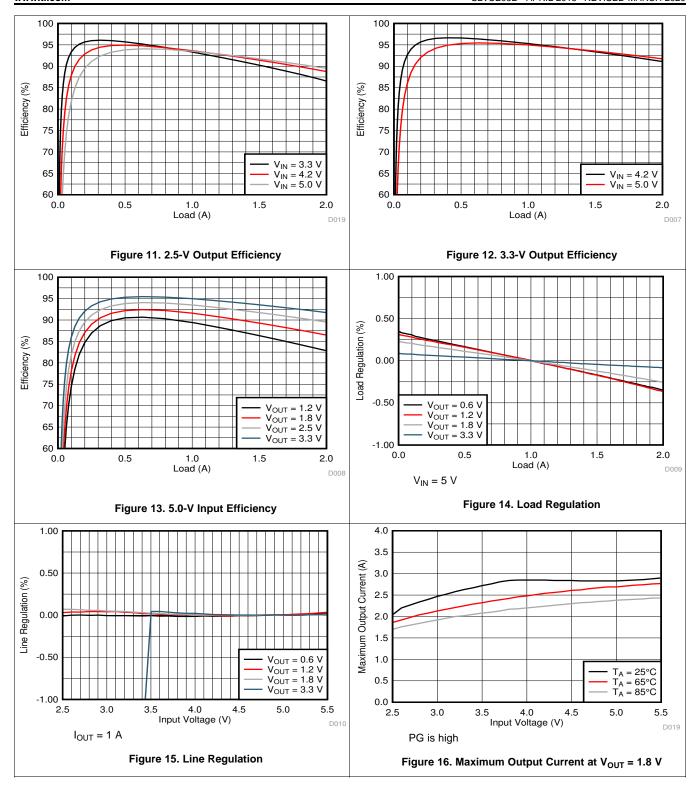
V<sub>IN</sub> = 5 V, V<sub>OUT</sub> = 1.8 V, T<sub>A</sub> = 25 °C, external components shown in Table 3, unless otherwise noted.



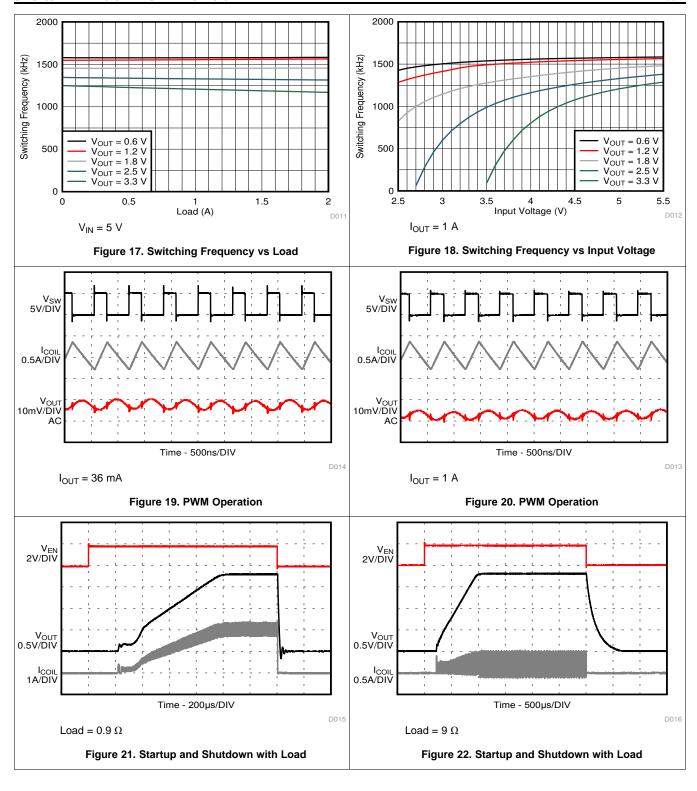
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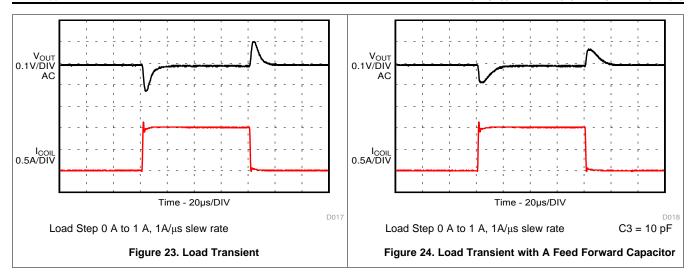












# 9 Power Supply Recommendations

The power supply to the TLV62569A must have a current rating according to the supply voltage, output voltage and output current.

# 10 Layout

# 10.1 Layout Guidelines

The PCB layout is an important step to maintain the high performance of the TLV62569A device.

- The input/output capacitors and the inductor should be placed as close as possible to the IC. This keeps the
  power traces short. Routing these power traces direct and wide results in low trace resistance and low
  parasitic inductance.
- The low side of the input and output capacitors must be connected properly to the power GND to avoid a GND potential shift.
- The sense traces connected to FB are signal traces. Special care should be taken to avoid noise being induced. Keep these traces away from SW nodes.
- GND layers might be used for shielding.

### 10.2 Layout Example

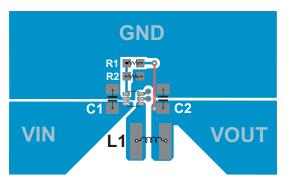


Figure 25. TLV62569APDRL Layout



#### 10.3 Thermal Considerations

Implementation of integrated circuits in low-profile and fine-pitch surface-mount packages typically requires special attention to power dissipation. Many system-dependent issues such as thermal coupling, airflow, convection surfaces, and the presence of other heat-generating components affect the power dissipation limits of a given component.

Two basic approaches for enhancing thermal performance are listed below:

- Improving the power dissipation capability of the PCB design
- · Introducing airflow in the system

For more details on how to use the thermal parameters, see the application notes: Thermal Characteristics Application Notes SZZA017 and SPRA953.

# 11 Device and Documentation Support

### 11.1 Device Support

### 11.1.1 Third-Party Products Disclaimer

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#### 11.1.2 Development Support

#### 11.1.2.1 Custom Design With WEBENCH® Tools

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Get more information about WEBENCH tools at www.ti.com/WEBENCH.

### 11.2 Documentation Support

### 11.2.1 Related Documentation

- Texas Instruments, Semiconductor and IC Package Thermal Metrics Application Report
- Texas Instruments, Thermal Characteristics of Linear and Logic Packages Using JEDEC PCB Designs Application Report

### 11.3 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on ti.com. In the upper right corner, click on *Alert me* to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

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#### 11.4 Support Resources

TI E2E™ support forums are an engineer's go-to source for fast, verified answers and design help — straight from the experts. Search existing answers or ask your own question to get the quick design help you need.

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#### 11.5 Trademarks

E2E is a trademark of Texas Instruments. WEBENCH is a registered trademark of Texas Instruments.

# 11.6 Electrostatic Discharge Caution



These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

### 11.7 Glossary

SLYZ022 — TI Glossary.

This glossary lists and explains terms, acronyms, and definitions.

# 12 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

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9-Nov-2020

#### PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan	Lead finish/ Ball material	MSL Peak Temp	Op Temp (°C)	Device Marking (4/5)	Samples
TLV62568ADRLR	ACTIVE	SOT-5X3	DRL	6	3000	Green (RoHS & no Sb/Br)	Call TI   SN	Level-1-260C-UNLIM	-40 to 125	1BE	Samples
TLV62568ADRLT	ACTIVE	SOT-5X3	DRL	6	250	Green (RoHS & no Sb/Br)	Call TI   SN	Level-1-260C-UNLIM	-40 to 125	1BE	Sample
TLV62568APDRLR	ACTIVE	SOT-5X3	DRL	6	3000	Green (RoHS & no Sb/Br)	Call TI   SN	Level-1-260C-UNLIM	-40 to 125	1BF	Sample
TLV62568APDRLT	ACTIVE	SOT-5X3	DRL	6	250	Green (RoHS & no Sb/Br)	Call TI   SN	Level-1-260C-UNLIM	-40 to 125	1BF	Sample
TLV62569ADRLR	ACTIVE	SOT-5X3	DRL	6	3000	Green (RoHS & no Sb/Br)	Call TI   SN	Level-1-260C-UNLIM	-40 to 125	1BG	Sample
TLV62569ADRLT	ACTIVE	SOT-5X3	DRL	6	250	Green (RoHS & no Sb/Br)	Call TI   SN	Level-1-260C-UNLIM	-40 to 125	1BG	Sample
TLV62569APDRLR	ACTIVE	SOT-5X3	DRL	6	3000	Green (RoHS & no Sb/Br)	Call TI   SN	Level-1-260C-UNLIM	-40 to 125	1BH	Sample
TLV62569APDRLT	ACTIVE	SOT-5X3	DRL	6	250	Green (RoHS & no Sb/Br)	Call TI   SN	Level-1-260C-UNLIM	-40 to 125	1BH	Samples

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

(2) RoHS: TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

**Green:** TI defines "Green" to mean the content of Chlorine (CI) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

<sup>(3)</sup> MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

<sup>(4)</sup> There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.



# **PACKAGE OPTION ADDENDUM**

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(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

(6) Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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# **PACKAGE MATERIALS INFORMATION**

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# TAPE AND REEL INFORMATION





	Dimension designed to accommodate the component width
B0	Dimension designed to accommodate the component length
	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

# QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



#### \*All dimensions are nominal

Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TLV62568ADRLR	SOT-5X3	DRL	6	3000	180.0	8.4	2.0	1.8	0.75	4.0	8.0	Q3
TLV62568ADRLT	SOT-5X3	DRL	6	250	180.0	8.4	2.0	1.8	0.75	4.0	8.0	Q3
TLV62568APDRLR	SOT-5X3	DRL	6	3000	180.0	8.4	2.0	1.8	0.75	4.0	8.0	Q3
TLV62568APDRLT	SOT-5X3	DRL	6	250	180.0	8.4	2.0	1.8	0.75	4.0	8.0	Q3
TLV62569ADRLR	SOT-5X3	DRL	6	3000	180.0	8.4	2.0	1.8	0.75	4.0	8.0	Q3
TLV62569ADRLT	SOT-5X3	DRL	6	250	180.0	8.4	2.0	1.8	0.75	4.0	8.0	Q3
TLV62569APDRLR	SOT-5X3	DRL	6	3000	180.0	8.4	2.0	1.8	0.75	4.0	8.0	Q3
TLV62569APDRLT	SOT-5X3	DRL	6	250	180.0	8.4	2.0	1.8	0.75	4.0	8.0	Q3

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\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TLV62568ADRLR	SOT-5X3	DRL	6	3000	210.0	185.0	35.0
TLV62568ADRLT	SOT-5X3	DRL	6	250	210.0	185.0	35.0
TLV62568APDRLR	SOT-5X3	DRL	6	3000	210.0	185.0	35.0
TLV62568APDRLT	SOT-5X3	DRL	6	250	210.0	185.0	35.0
TLV62569ADRLR	SOT-5X3	DRL	6	3000	210.0	185.0	35.0
TLV62569ADRLT	SOT-5X3	DRL	6	250	210.0	185.0	35.0
TLV62569APDRLR	SOT-5X3	DRL	6	3000	210.0	185.0	35.0
TLV62569APDRLT	SOT-5X3	DRL	6	250	210.0	185.0	35.0



PLASTIC SMALL OUTLINE



### NOTES:

- All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
   This drawing is subject to change without notice.
   This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not
- exceed 0.15 mm per side.



PLASTIC SMALL OUTLINE



NOTES: (continued)

- 4. Publication IPC-7351 may have alternate designs.
- 5. Solder mask tolerances between and around signal pads can vary based on board fabrication site.



PLASTIC SMALL OUTLINE



NOTES: (continued)

- 6. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
- 7. Board assembly site may have different recommendations for stencil design.



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