

# **ECE 7810 ASSIGNMENT REPORT**

## **(Solution of Fields by Num. Mtds I)**

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## QUESTION A

Transmission line equation:

$$\frac{\partial V(x,t)}{\partial x} + L \frac{\partial I(x,t)}{\partial t} + RI(x,t) = 0 \quad \text{--- (i)}$$

$$\frac{\partial I(x,t)}{\partial x} + C \frac{\partial V(x,t)}{\partial t} + GV(x,t) = 0 \quad \text{--- (ii)}$$

where  $L$  = inductance,  $C$  = capacitance  
 $V$  = voltage,  $I$  = current  
 $R$  = impedance,  $G$  = conductance

Equations (i) and (ii) can be written as:

$$C \frac{\partial V}{\partial t} = -GV - \frac{\partial I}{\partial x} \quad \text{--- (1)}$$

$$L \frac{\partial I}{\partial t} = -RI - \frac{\partial V}{\partial x} \quad \text{--- (2)}$$

Creating vector-matrix equation from (1) and (2):

$$\frac{\partial}{\partial t} \begin{bmatrix} V \\ I \end{bmatrix} = - \begin{bmatrix} \frac{G}{C} & \frac{1}{C} \frac{\partial}{\partial x} \\ \frac{1}{L} \frac{\partial}{\partial x} & \frac{R}{L} \end{bmatrix} \begin{bmatrix} V \\ I \end{bmatrix} \quad \text{--- (3)}$$

It can be conveniently expressed as a single equation:

$$\frac{\partial F}{\partial t} = \bar{S} F \quad \text{--- (4)}$$

where  $F$  = voltage-current vector  
 $\bar{S}$  = operator matrix

$$F = \begin{bmatrix} V & I \end{bmatrix}^T \quad \text{--- (5)}$$

$$\bar{S} = - \begin{bmatrix} \frac{G}{C} & \frac{1}{C} \frac{\partial}{\partial x} \\ \frac{1}{L} \frac{\partial}{\partial x} & \frac{R}{L} \end{bmatrix} \quad \text{--- (6)}$$

Equation (4) can be solved by finding the propagator matrix  $\bar{K}(x, t)$

$$\frac{\partial \bar{K}(x, t)}{\partial t} = \bar{S} \bar{K}(x, t) \quad \text{--- (7)}$$

with initial condition:

$$\lim_{t \rightarrow 0} \bar{K}(x, t) = \bar{I} \delta(x - x') \quad \text{--- (8)}$$

where  $\delta$  = Dirac delta function

$x'$  = initial position of the voltage

$\bar{I}$  = Identity matrix

$$\bar{K} = e^{\bar{S}t} \delta(x - x') \quad \text{--- (9)}$$

where Dirac delta function can be expressed in terms of Fourier integral:

$$\delta(x - x') = \frac{1}{2\pi} \int_{-\infty}^{\infty} e^{jk(x-x')} dk$$

$\therefore$  Equation (9) becomes, with exponential expressed in power series, as:

$$\bar{K} = \left(1 + \bar{S}t + \frac{\bar{S}^2 t^2}{2!} + \dots\right) \frac{1}{2\pi} \int_{-\infty}^{\infty} e^{jk(x-x')} dk \quad \text{--- (10)}$$

$$\bar{K} = \frac{1}{2\pi} \int_{-\infty}^{\infty} e^{\bar{S}t} \exp[jk(x-x')] dk \quad \text{--- (11)}$$

$$\bar{S}_f = - \begin{bmatrix} \frac{G}{C} & \frac{jk}{C} \\ \frac{jk}{L} & \frac{R}{L} \end{bmatrix} \quad \text{--- (12)}$$

The exponential  $e^{\bar{S}_f t}$  can be converted to a  $2 \times 2$  transition matrix  $\bar{A}$ , then the eigenvalue  $\lambda_1, \lambda_2$  of  $\bar{S}_f$  are found by extracting the roots of  $\det[\lambda I - \bar{S}_f] = 0$

$$\det \begin{bmatrix} \lambda + \frac{G}{C} & \frac{jK}{C} \\ \frac{jK}{L} & \lambda + \frac{R}{L} \end{bmatrix}$$

Giving that:

$$\lambda_1, \lambda_2 = -\frac{G}{2L} \left[ 1 \pm \sqrt{1 - \gamma^2} \right] - \frac{R}{2L} \left[ 1 \pm \sqrt{1 - \gamma^2} \right]$$

where  $\gamma = \frac{V_p R}{G}$   
 $\left( \frac{G}{2L} - \frac{R}{2L} \right)$

$V_p =$  velocity of propagation

Eigenvectors can be constructed by selecting one column of  $\text{adj}[\lambda_1 \mathbf{I} - \mathbf{S}_p]$  and one column of  $\text{adj}[\lambda_2 \mathbf{I} - \mathbf{S}_p]$ . Therefore, the respective first and second columns of the model matrix is given by:

$$\bar{M} = \begin{bmatrix} -Z_0 & -Z_0 \\ \frac{1 - \sqrt{1 - \gamma^2}}{j\gamma} & \frac{1 + \sqrt{1 - \gamma^2}}{j\gamma} \end{bmatrix}$$

where  $Z_0 = \sqrt{LC}$

$V_p = \frac{1}{\sqrt{LC}}$

$\bar{M} =$  model matrix

The model matrix and its inverse along with a diagonalized exponential eigenvalue matrix are combined to form the matrix product:

$$\bar{A} = e^{\bar{S}t} = \bar{M} \begin{bmatrix} e^{\lambda_1 t} & 0 \\ 0 & e^{\lambda_2 t} \end{bmatrix} \bar{M}^{-1}$$

$\Rightarrow$

$$A_{11} = e^{-\frac{R}{L}t} \left[ \frac{-b}{Z_p} \frac{\sin \left[ \frac{t V_p \sqrt{K^2 - K_*^2}}{\sqrt{K^2 - K_*^2}} \right] + \cos \left[ \frac{t V_p \sqrt{K^2 - K_*^2}}{\sqrt{K^2 - K_*^2}} \right]} \right]$$

$$A_{12} = -jkZ_0 e^{-at} \left[ \frac{\sin \left[ tV_p \sqrt{k^2 - k_*^2} \right]}{\sqrt{k^2 - k_*^2}} \right]$$

$$A_{21} = \frac{A_{12}}{Z_0}$$

$$A_{22} = e^{-at} \left[ \frac{b}{V_p} \frac{\sin \left[ tV_p \sqrt{k^2 - k_*^2} \right]}{\sqrt{k^2 - k_*^2}} + \cos \left[ tV_p \sqrt{k^2 - k_*^2} \right] \right]$$

$$\text{where } k_* = \frac{G}{2cV_p} - \frac{R}{2cV_p}, \quad a = \frac{1}{2} \left( \frac{G}{c} + \frac{R}{c} \right), \quad b = \frac{1}{2} \left( \frac{G}{c} - \frac{R}{c} \right)$$

The propagator equation expressed in terms of matrix  $\bar{A}$  becomes:

$$\bar{K} = \frac{1}{2\pi} \int_{-\infty}^{\infty} \begin{bmatrix} A_{11} & A_{12} \\ A_{21} & A_{22} \end{bmatrix} \exp[jk\bar{x}] dk$$

$$\text{where } \bar{x} = x - x'$$

Integrating each integral above yields:

$$\bar{K} = \begin{bmatrix} K_{11} & K_{12} \\ K_{21} & K_{22} \end{bmatrix}$$

$$K_{11} = \frac{e^{-at}}{2} \left[ \delta(\bar{x} + tV_p) + \delta(\bar{x} - tV_p) - \frac{b}{V_p} J_0 \left[ \frac{b}{V_p} \sqrt{\bar{x}^2 - (tV_p)^2} \right] + \right. \\ \left. b t J_1 \left[ \frac{b}{V_p} \sqrt{\bar{x}^2 - (tV_p)^2} \right] \right]$$

$$K_{12} = -Z_0 \frac{e^{-at}}{2} \left[ \delta(\bar{x} + tV_p) - \delta(\bar{x} - tV_p) - \bar{x} \frac{b}{V_p} J_1 \left[ \frac{b}{V_p} \sqrt{\bar{x}^2 - (tV_p)^2} \right] \right]$$

$$K_2 = \frac{K}{Z_0^2}$$

$$K_2 = \frac{e^{-\alpha t}}{2} \left[ \delta(x+tv_p) + \delta(x-tv_p) + \frac{b}{v_p} J_0 \left[ \frac{b}{v_p} \sqrt{x^2 - (tv_p)^2} \right] + bt J_1 \left[ \frac{b}{v_p} \sqrt{x^2 - (tv_p)^2} \right] \frac{1}{\sqrt{x^2 - (tv_p)^2}} \right] \quad (13)$$

Propagation equation ~~for~~ <sup>with</sup> initial voltage and current can be expressed as:

$$\begin{bmatrix} V(x,t) \\ I(x,t) \end{bmatrix} = \int_{-\infty}^{\infty} \tilde{K} \cdot \begin{bmatrix} V(x') \\ I(x') \end{bmatrix} dx' \quad (14)$$

Insert (13) into (14):

$$\begin{aligned} V(x,t) = \frac{e^{-\alpha t}}{2} & \left[ V(x+tv_p) + V(x-tv_p) - Z_0 [I(x+tv_p) - I(x-tv_p)] \right. \\ & + bt \int_{x-tv_p}^{x+tv_p} J_1 \left[ \frac{b}{v_p} \sqrt{x'^2 - (tv_p)^2} \right] \frac{1}{\sqrt{x'^2 - (tv_p)^2}} V(x') dx' \\ & - \frac{b}{v_p} \int_{x-tv_p}^{x+tv_p} J_0 \left[ \frac{b}{v_p} \sqrt{x'^2 - (tv_p)^2} \right] V(x') dx' \\ & \left. + Z_0 \frac{b}{v_p} \int_{x-tv_p}^{x+tv_p} \tilde{x} J_1 \left[ \frac{b}{v_p} \sqrt{x'^2 - (tv_p)^2} \right] \frac{1}{\sqrt{x'^2 - (tv_p)^2}} I(x') dx' \right] \quad (15) \end{aligned}$$

$$\begin{aligned}
 I(x,t) = \frac{e^{-\alpha t}}{2} & \left[ I(x+tv_p) + I(x-tv_p) - \frac{1}{z_0} [V(x+tv_p) - V(x-tv_p)] \right. \\
 & + bt \int_{x-tv_p}^{x+tv_p} \frac{J_1 \left[ \frac{b}{v_p} \sqrt{x^2 - (tv_p)^2} \right]}{\sqrt{x^2 - (tv_p)^2}} I(x') dx' \\
 & + \frac{b}{v_p} \int_{x-tv_p}^{x+tv_p} J_0 \left[ \frac{b}{v_p} \sqrt{x^2 - (tv_p)^2} \right] I(x') dx' \\
 & \left. + \frac{1}{z_0} \frac{b}{v_p} \int_{x-tv_p}^{x+tv_p} \frac{J_1 \left[ \frac{b}{v_p} \sqrt{x^2 - (tv_p)^2} \right]}{\sqrt{x^2 - (tv_p)^2}} V(x') dx' \right]
 \end{aligned}$$

(15) and (16) can be integrated using Simpson's rule <sup>and properties of Bessel's functions</sup> to give: (17)

$$\begin{aligned}
 V(x,t) = \frac{e^{-\alpha t}}{2} & \left[ \left[ 1 - \frac{(bt)^2}{3} + \frac{(bt)^2}{6} \right] [V(x+tv_p) + V(x-tv_p)] \right. \\
 & - z_0 \left[ 1 + \frac{(bt)^2}{6} \right] [I(x+tv_p) - I(x-tv_p)] \\
 & \left. + \frac{4}{3} (bt) [I_1(bt) - I_0(bt)] V(x) \right] \quad \text{--- (17)}
 \end{aligned}$$

$$\begin{aligned}
 I(x,t) = \frac{e^{-\alpha t}}{2} & \left[ \left[ 1 + \frac{(bt)^2}{3} + \frac{(bt)^2}{6} \right] [I(x+tv_p) + I(x-tv_p)] \right. \\
 & - \frac{1}{z_0} \left[ 1 + \frac{(bt)^2}{6} \right] [V(x+tv_p) - V(x-tv_p)] \\
 & \left. + \frac{4}{3} (bt) [I_1(bt) + I_0(bt)] I(x) \right] \quad \text{--- (18)}
 \end{aligned}$$

where  $I_0$  = 0 order modified Bessel function of the first kind

$I_1$  = 1<sup>st</sup> order modified Bessel function of the first kind

Equations (17) and (18) are for lossy transmission line.

For lossless transmission line:

$R, G, \alpha, \beta = 0$ , in equations (17) and (18) becomes:

$$V(x, t) = \frac{1}{2} \left[ [V(x + v_p t) + V(x - v_p t)] - z_0 [I(x + v_p t) - I(x - v_p t)] \right]$$

$$I(x, t) = \frac{1}{2} \left[ [I(x + v_p t) + I(x - v_p t)] - \frac{1}{z_0} [V(x + v_p t) - V(x - v_p t)] \right]$$



## QUESTION B

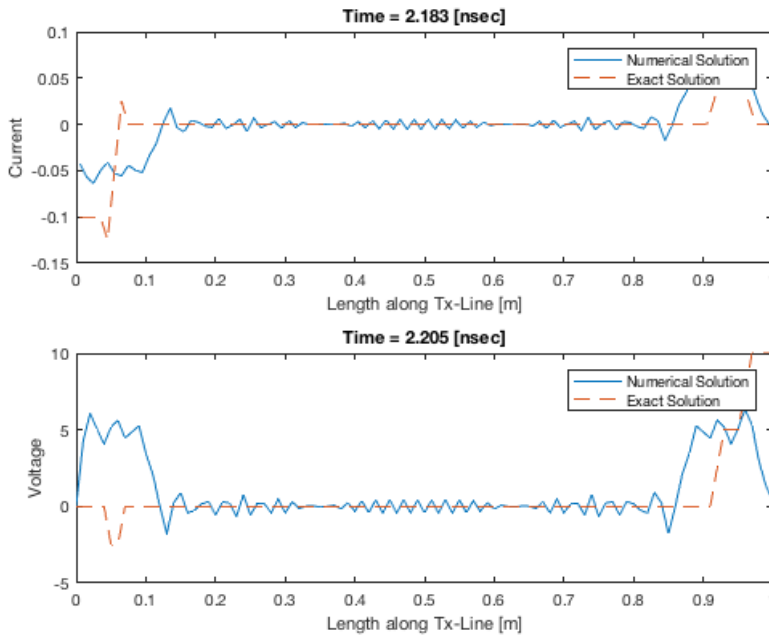
This question was implemented in the file named “**fdtdLosslessVsExact.m**”. The exact solution was implemented as derived in the solution to the “Question A” above. The code is commented and documented.

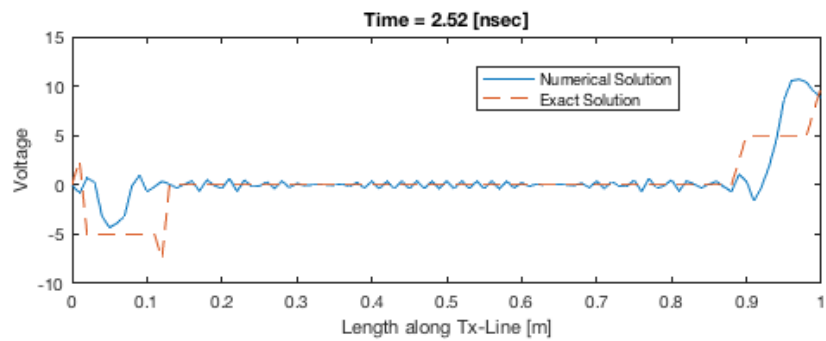
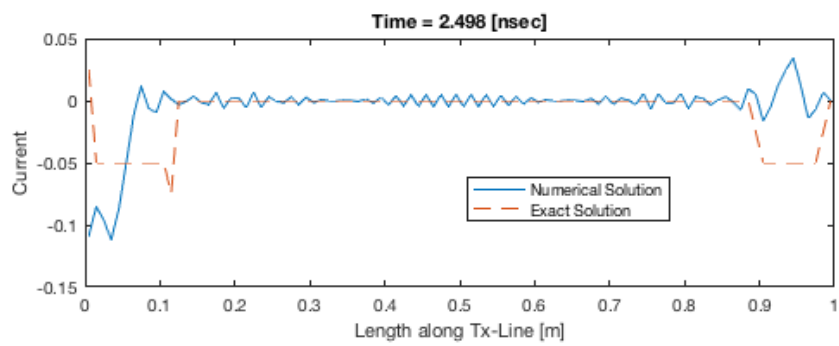
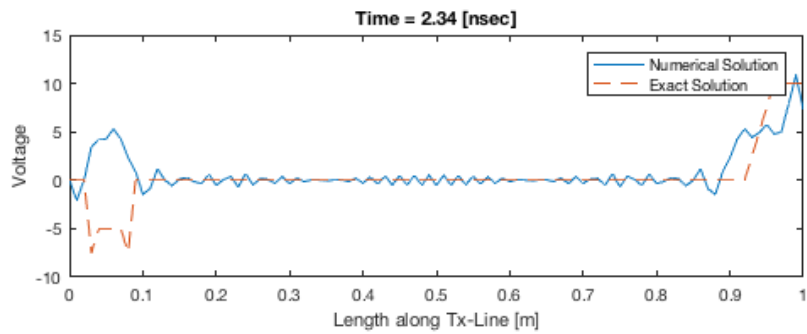
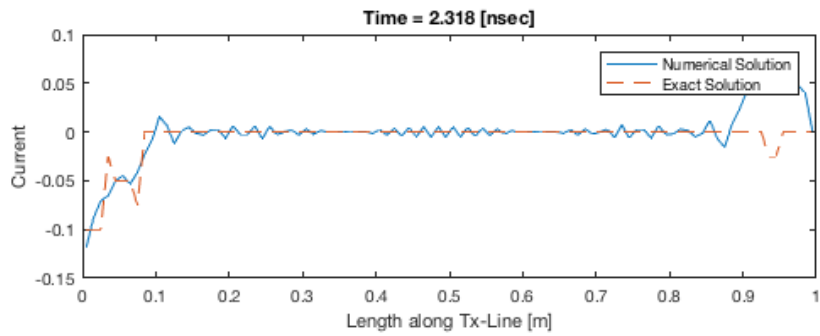
The following results were obtained:

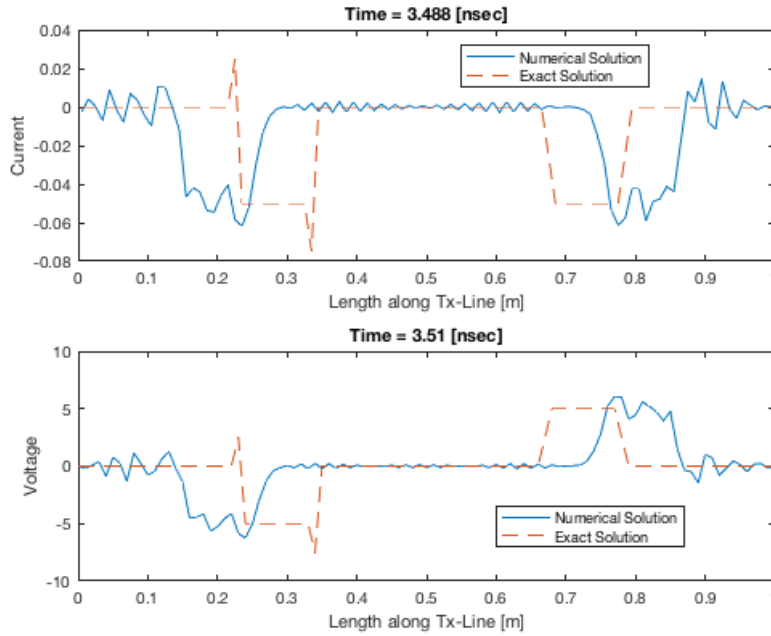
Given that  $C=50\text{pF/m}$  and  $L=0.5\mu\text{H/m}$ , the velocity of propagation obtained,  $c_0$  was

$$c_0 = \frac{1}{\sqrt{LC}} = 200000000 \text{ m/s}$$

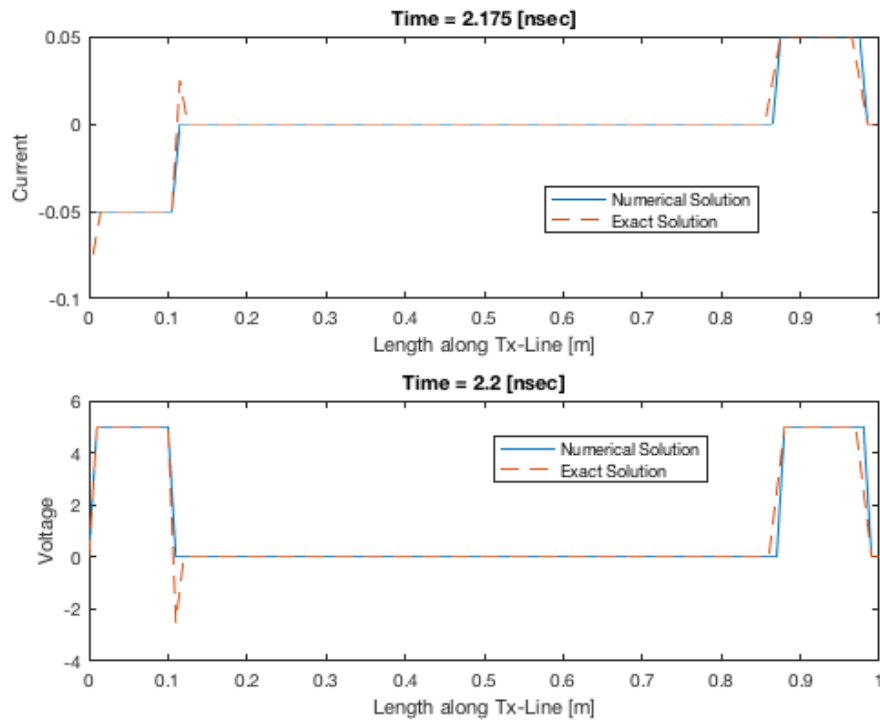
**Case 1:** For Courant limit,  $\text{CFL} = 0.9$ , the following graphs were obtained at voltage times: 2.2, 2.3, 2.5 and 3.5ns

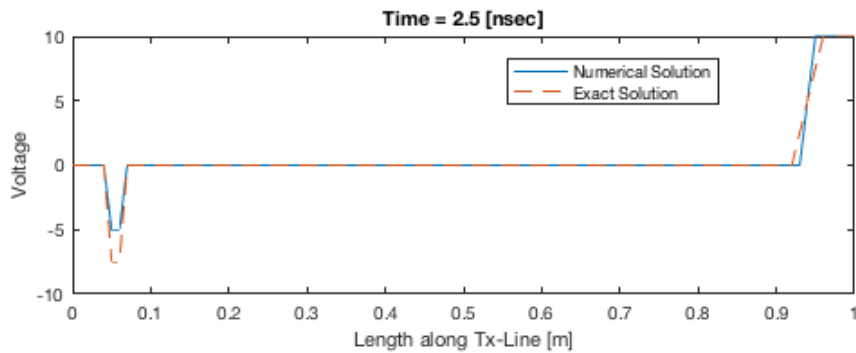
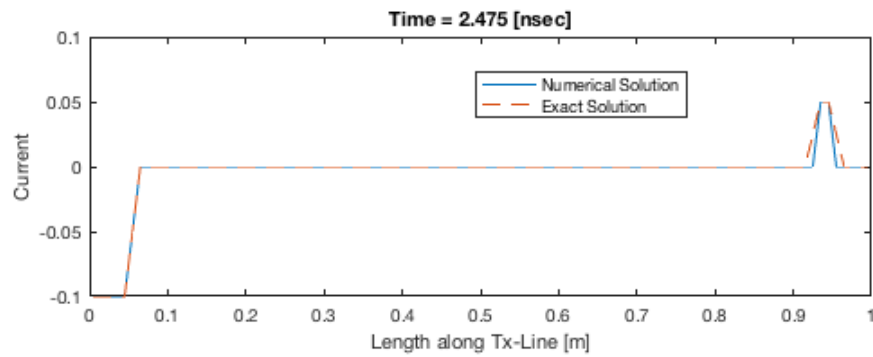
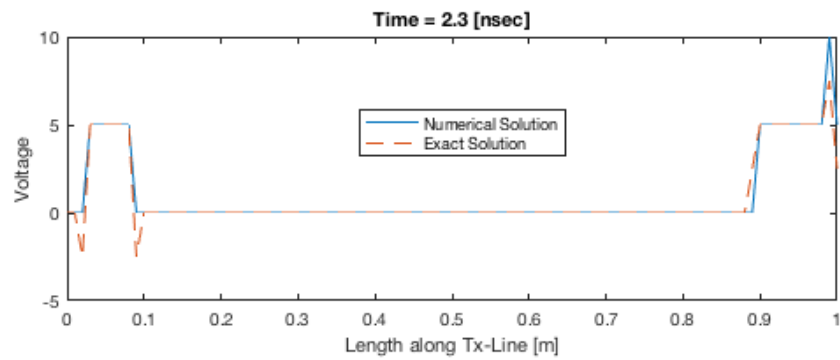
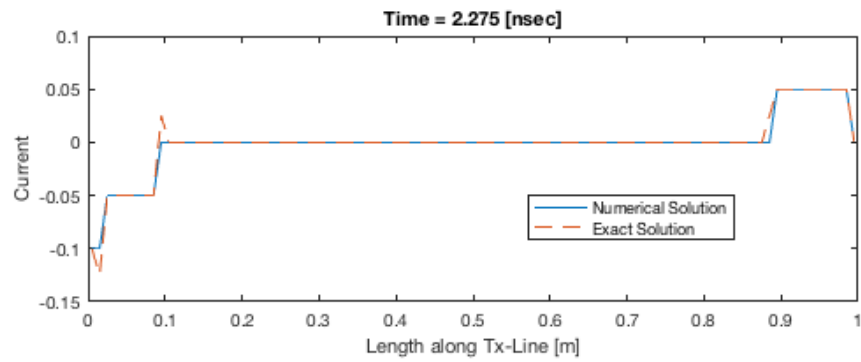


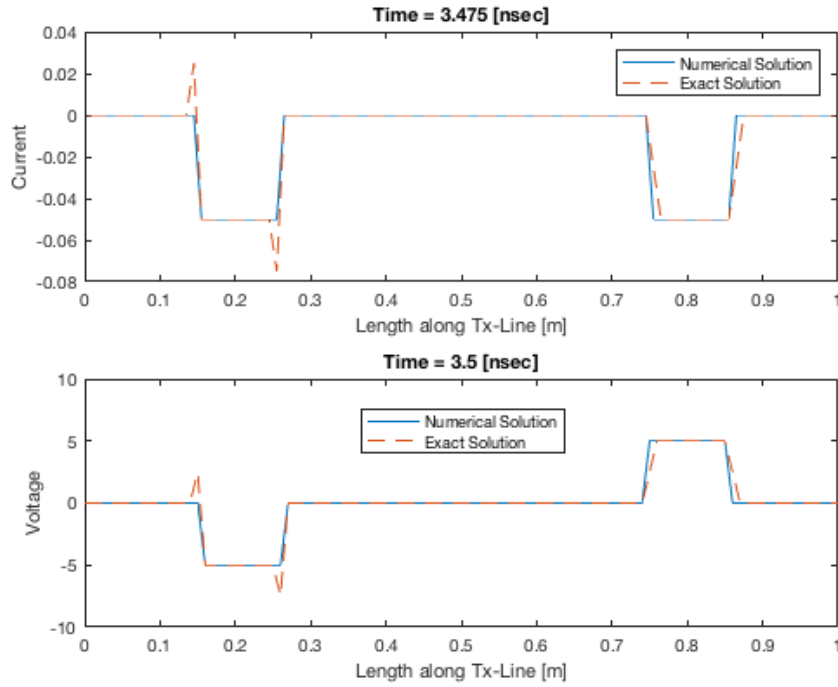




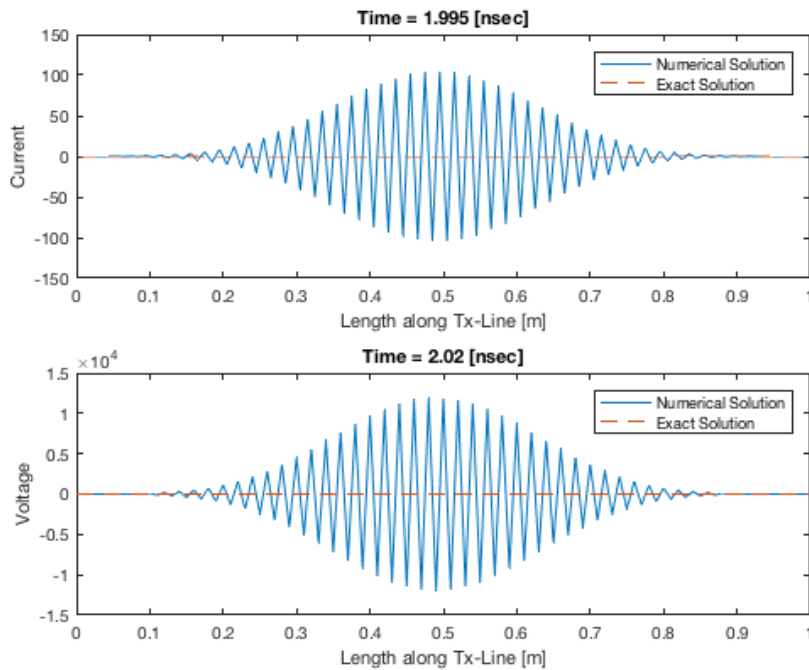
**Case 2:** For Courant limit,  $CFL = 1.0$ , the following graphs were obtained at voltage times: 2.2, 2.3, 2.5 and 3.5ns

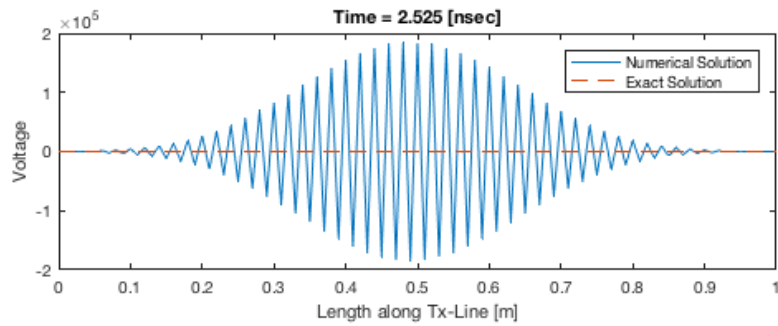
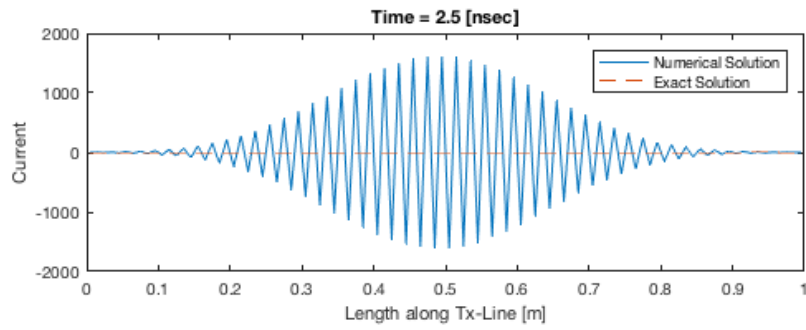
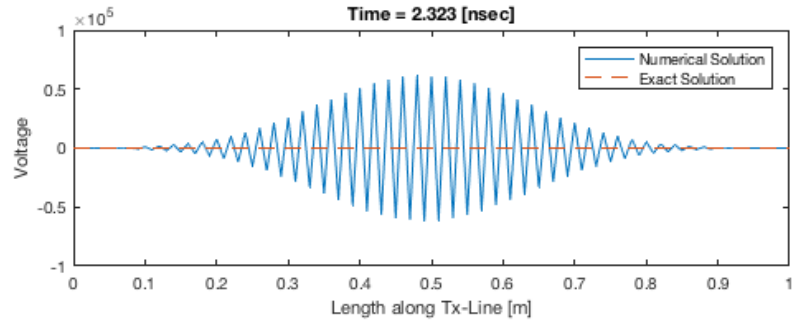
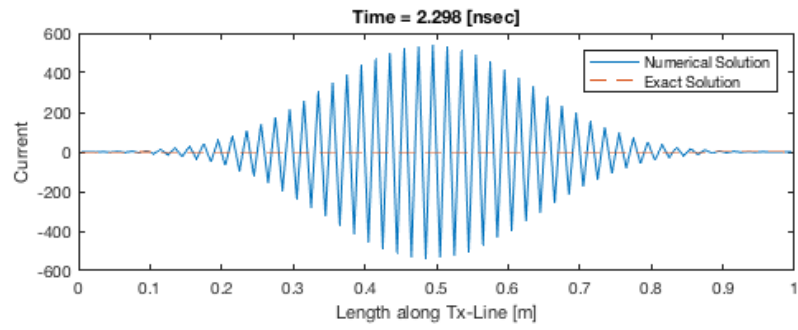


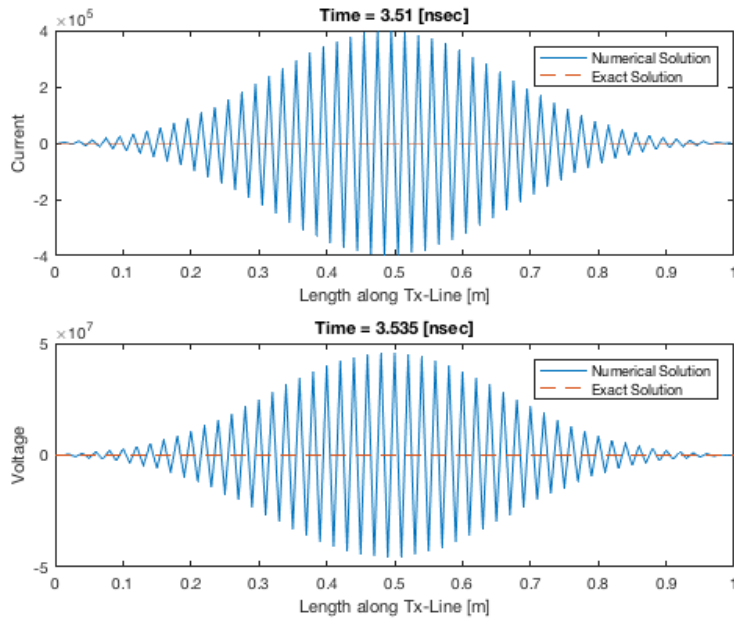




**Case 3:** For Courant limit,  $CFL = 1.01$ , the following graphs were obtained at voltage times: 2.2, 2.3, 2.5 and 3.5ns







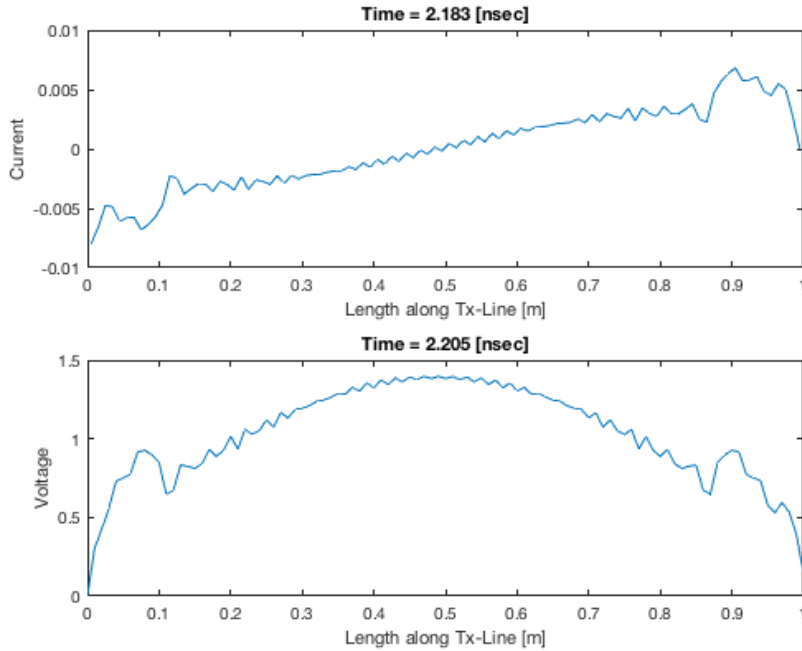
### Observation:

As shown above, when the Courant value was less than 1.0, precisely 0.9, there seems to be partial numerical dispersion; however, when CFL was 1.0 there was stability in the wave. With Courant number greater than 1.0 the dispersion was great in the wave high instability in the wave. Again, observing the exact solution also showed that the numerical solution tends to be closer to the exact solution.

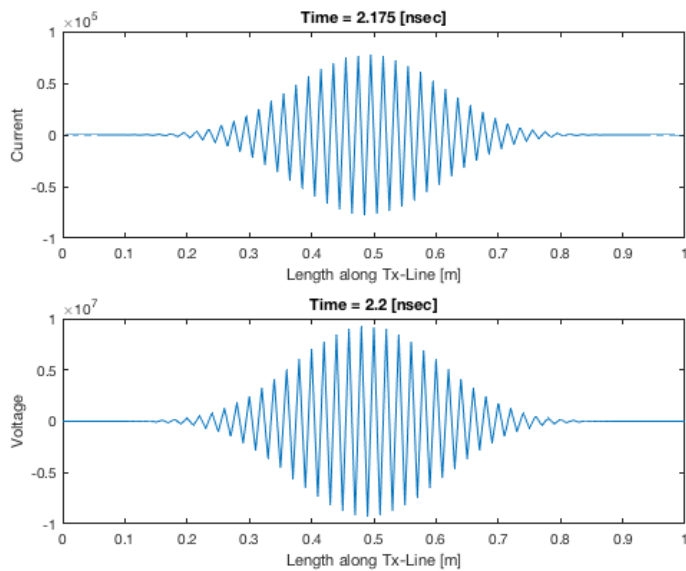
### QUESTION C

This question's implementation file is named "**fdtdLossy.m**". Different cases were tested based on the different Courant numbers given. This code is well implemented and documented to easy readability. The following results were obtained:

**Case 1:** For Courant limit,  $CFL = 0.9$ , the following graph was obtained at voltage time: 2.2ns

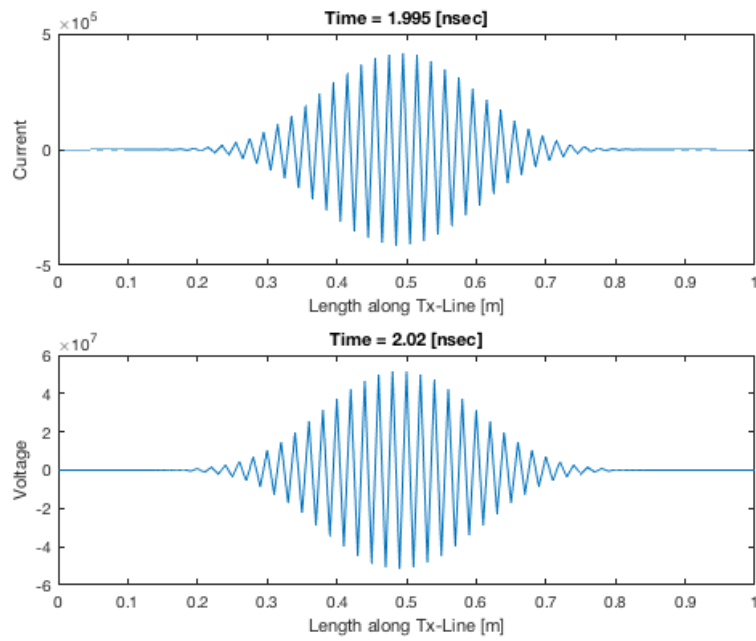


**Case 2:** For Courant limit,  $CFL = 1.0$ , the following graph was obtained at voltage time: 2.2ns





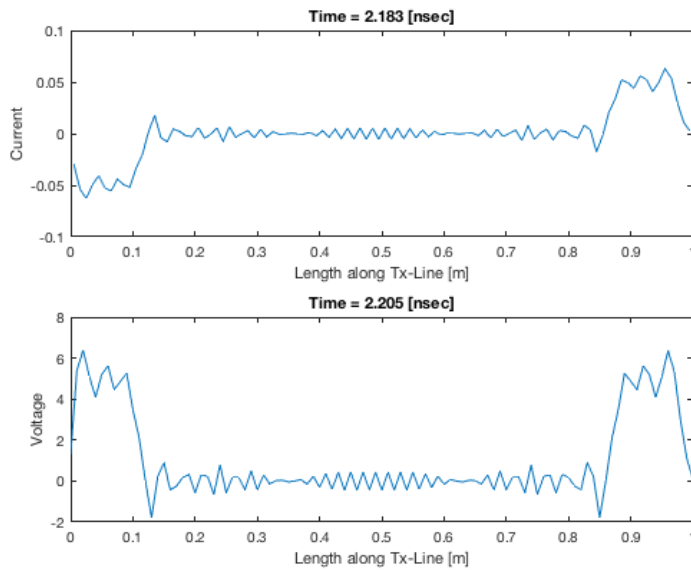
**Case 3:** For Courant limit, CFL = 1.01, the following graph was obtained at voltage time: 2.2ns



## QUESTION D

The implementation file for this question is named “**QuestionD.m**”. Like the previous codes written, this code is well documented. The following results were obtained for different Courant values.

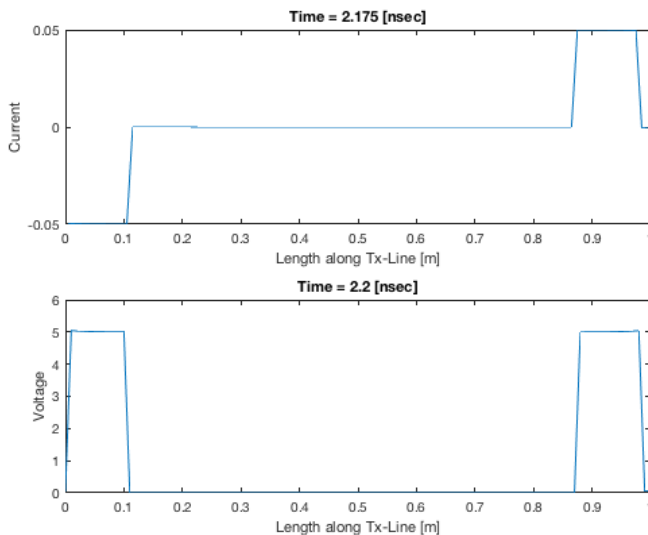
**Case 1:** For Courant limit, CFL = 0.9, the following graph was obtained at voltage time: 2.2ns



$$V_{t\_near} = 0.0798$$

$$V_{t\_far} = 0.0798$$

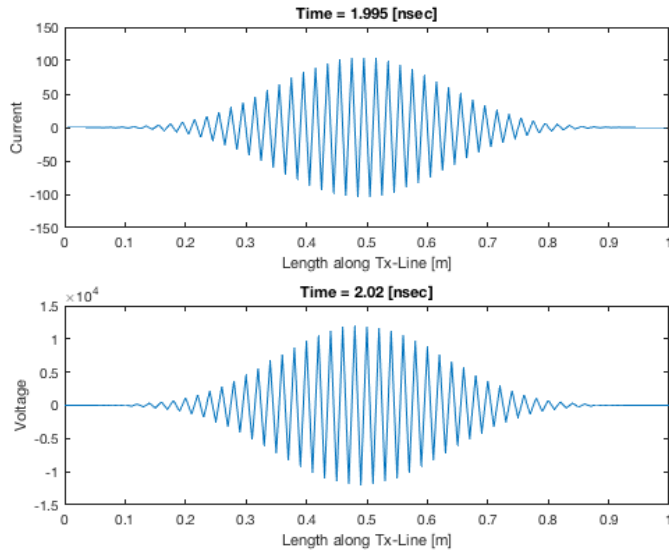
**Case 2:** For Courant limit, CFL = 1.0, the following graph was obtained at voltage time: 2.2ns



$$V_{t\_near} = 0.0773$$

$$V_{t\_far} = 0.0773$$

**Case 3:** For Courant limit, CFL = 1.01, the following graph was obtained at voltage time: 2.2ns



$$V_{t\_near} = 0.0215$$

$$V_{t\_far} = 0.0215$$

### Observation:

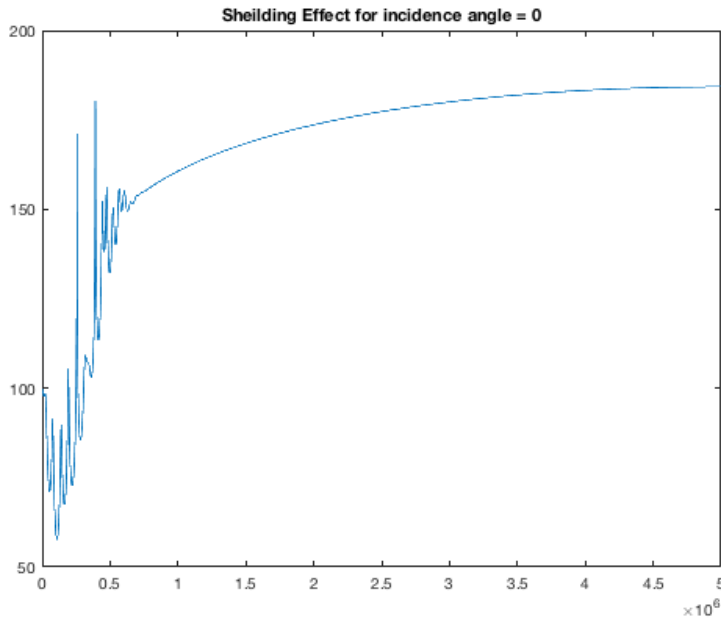
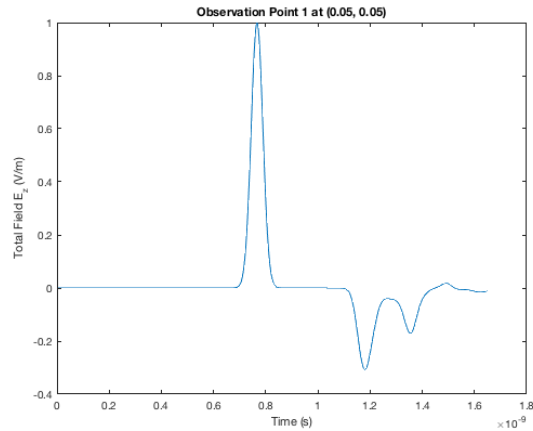
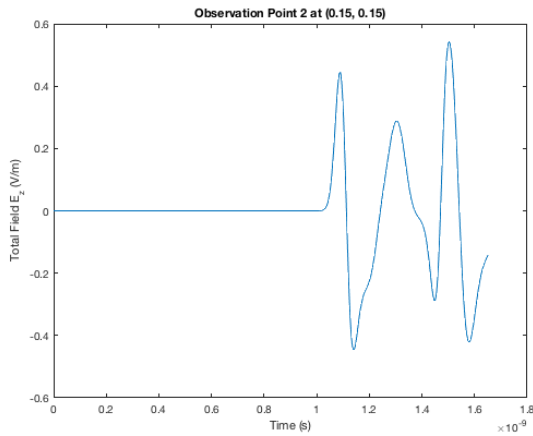
As shown above, when the Courant value was less than 1.0, precisely 0.9, there seems to be partial numerical dispersion; however, when CFL was 1.0 there was stability in the wave. With Courant number greater than 1.0 the dispersion was great in the wave high instability in the wave.

## QUESTION E

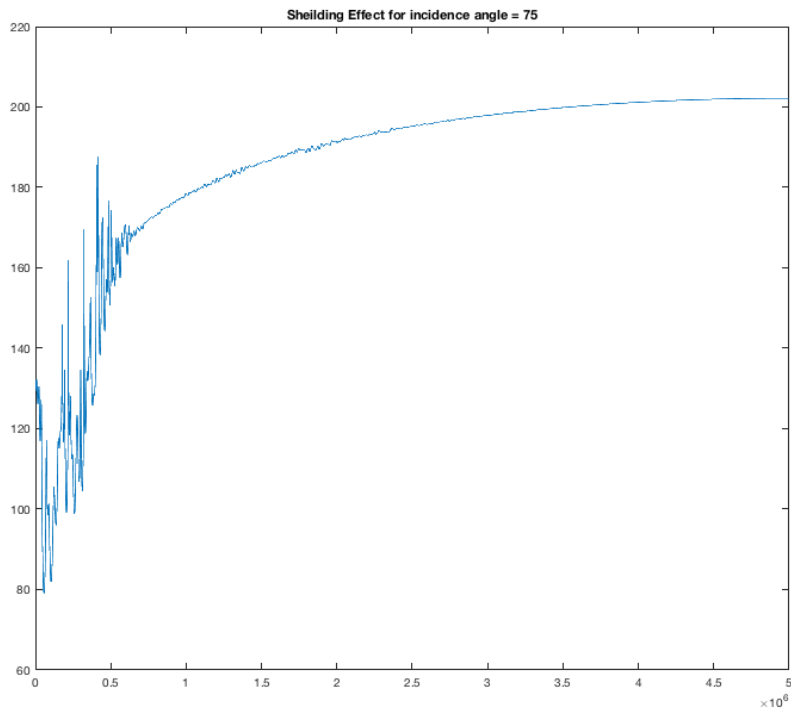
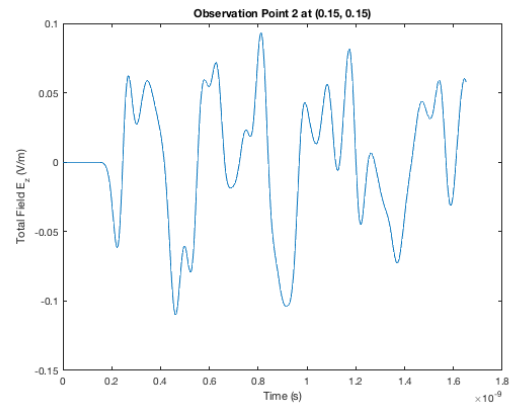
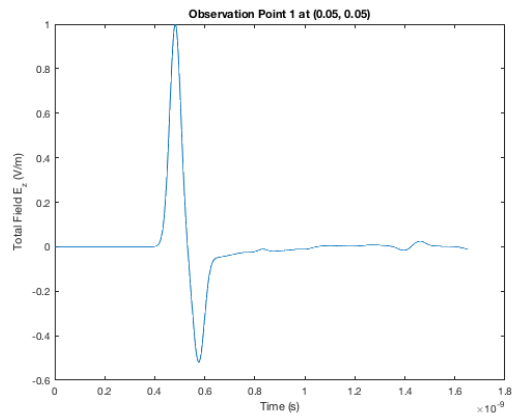
As contained in the question, the shielding was opened as given in the question as the observation points are two: one inside the shield and the other outside the shield.

For the TM case, the code was properly commented to show that the code is well understood. The following graphs were obtained from the two observation points for different incidence angles:

**Case 1:** For angle of incidence,  $\theta = 0$



**Case 2:** For angle of incidence,  $\theta = 75$



**Case 3:** For angle of incidence,  $\theta = 180$

