

EECE423-01: 현대제어이론

Modern Control Theory

Chapter 6: Controllability

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◆ The main topics of this chapter are

1. Concept for Controllability
2. Conditions for Controllability
3. Similarity Transform and Stabilizability

1. Concepts for Controllability

◆ Motivation of controllability

For the LTI system

$$\begin{cases} \frac{dx(t)}{dt} = Ax(t) + Bu(t) \\ y(t) = Cx(t) + Du(t) \end{cases}, \quad x(t) \in \mathbb{R}^n, \quad u(t) \in \mathbb{R}^k, \quad y(t) \in \mathbb{R}^m$$

let us consider the following statement:

Is there a control input $u(t)$ that move $x(t)$ from any initial state x_0 to any other final state x_1 in a finite time interval?

◆ What is a controllability?

If there exist $s \geq 0$ and $u(t)$ ($0 \leq t \leq s$) that move $x(t)$ from $x(0) = x_0$ to $x(s) = x_1$ for arbitrary $x_0, x_1 \in \mathbb{R}^n$, the LTI system

$$\begin{cases} \frac{dx(t)}{dt} &= Ax(t) + Bu(t) \\ y(t) &= Cx(t) + Du(t) \end{cases}$$

is called a *controllable*. Otherwise, the LTI system is called an *uncontrollable*.

2. Conditions for Controllability

◆ Necessary and sufficient condition of controllability

The LTI system $\begin{cases} \frac{dx(t)}{dt} &= Ax(t) + Bu(t) \\ y(t) &= Cx(t) + Du(t) \end{cases}$ is controllable

if and only if the following controllability matrix has rank n .

$$U_c = [B \quad AB \quad \cdots \quad A^{n-1}B]$$

Based on this, we also say that the pair (A, B) is *controllable*.

Proof: (1. necessary condition)

The solution $x(t)$ at $t = s$ is given by

$$x(s) = e^{As} \left(x(0) + \int_0^s e^{-A\tau} Bu(\tau) d\tau \right)$$

and we can obtain

$$e^{-As}x(s) - x(0) = \int_0^s e^{-A\tau} Bu(\tau) d\tau$$

Here, it readily follows from Cayley-Hamilton Theorem that

$$e^{At} = q_1(t)I + q_2(t)A + \cdots + q_n(t)A^{n-1}$$

where $q_i(t)$ is an adequately defined scalar function.

If we define $h_i = \int_0^s q_i(-\tau)u(\tau)d\tau$,

$$\begin{aligned} e^{-As}x(s) - x(0) &= \int_0^s e^{-A\tau}Bu(\tau)d\tau \\ &= Bh_1 + \cdots + A^{n-1}Bh_n = U_c \begin{bmatrix} h_1 \\ \vdots \\ h_n \end{bmatrix} \end{aligned}$$

To put it another way,

$$e^{-As}x(s) - x(0) = U_c \begin{bmatrix} h_1 \\ \vdots \\ h_n \end{bmatrix}$$

To exist h_1, \dots, h_n (and thus $u(t)$ ($0 \leq t \leq s$)) that satisfy

$$e^{-As}x(s) - x(0) = U_c \begin{bmatrix} h_1 \\ \vdots \\ h_n \end{bmatrix}$$

for arbitrary $x(0) = x_0$ and $x(s) = x_1$ in \mathbb{R}^n ,

U_c should have rank n .

(2. sufficient condition)

If $\text{rank}(U_c) = n$, the controllability gramian

$$W_s := \int_0^s e^{At} B B^T e^{A^T t} dt$$

is nonsingular for an arbitrary $s > 0$ (the proof will be provided later).

If we consider for some $s > 0$ the control input defined as

$$u(t) = B^T e^{A^T(s-t)} W_s^{-1} (-e^{As} x_0 + x_1), \quad 0 \leq t \leq s$$

then we obtain the following:

$$\begin{aligned} x(s) &= e^{As} x_0 + \int_0^s e^{A(s-\tau)} B B^T e^{A^T(s-\tau)} d\tau \cdot W_s^{-1} (-e^{As} x_0 + x_1) \\ &= e^{As} x_0 - e^{As} x_0 + x_1 = x_1. \end{aligned}$$

(W_s is a nonsingular matrix)

Assume that $|W_s| = 0$ for some $s > 0$. Then, there exists a nonzero v such that $W_s v = 0$. Hence,

$$v^T W_s v = \int_0^s v^T e^{A\tau} B B^T e^{A^T \tau} v d\tau = \int_0^s \|B^T e^{A^T \tau} v\|_2^2 d\tau = 0$$

This implies that $B^T e^{A^T \tau} v \equiv 0$ ($0 \leq \tau \leq s$). For $f(\tau) := B^T e^{A^T \tau} v$,

we obtain $f(0)v = 0$, $\frac{df(\tau)}{d\tau}v|_{\tau=0} = 0, \dots, \frac{d^{n-1}f(\tau)}{d\tau^{n-1}}v|_{\tau=0} = 0$, and thus

$$\begin{bmatrix} B^T \\ \vdots \\ B^T (A^T)^{n-1} \end{bmatrix} v = U_c^T v = 0 \quad (v \neq 0). \text{ This contradicts } \text{rank}(U_c) = n.$$

◆ Example

The LTI system $\dot{x} = \begin{bmatrix} 2 & 0 & -1 \\ 3 & -1 & -1 \\ 2 & -1 & 1 \end{bmatrix} x + \begin{bmatrix} 1 & 1 \\ 0 & 2 \\ 0 & 1 \end{bmatrix} u$

is controllable, since the controllability matrix

$$U_c = [B \quad AB \quad A^2B] = \begin{bmatrix} 1 & 1 & 2 & 1 & 2 & 1 \\ 0 & 2 & 3 & 0 & 1 & 2 \\ 0 & 1 & 2 & 1 & 3 & 3 \end{bmatrix}$$

has rank 3.

◆ Properties of controllability

For the LTI system
$$\begin{cases} \frac{dx(t)}{dt} &= Ax(t) + Bu(t) \\ y(t) &= Cx(t) + Du(t) \end{cases},$$

the followings are equivalent.

- (a) The pair (A, B) is controllable.
- (b) The controllability matrix $U_c = [B \quad AB \quad \cdots \quad A^{n-1}B]$ has rank n .
- (c) The controllability gramian $W_s := \int_0^s e^{At} B B^T e^{A^T t} dt$ is positive definite for every $s > 0$.

Proof: The equivalence between (a) and (b) together with the assertion (b) \Rightarrow (c) have already been shown. We show (c) \Rightarrow (b).

Suppose that the controllability matrix U_c is not of full rank.

Then, there exists a nonzero vector $v \in \mathbb{R}^n$ such that

$$v^T A^k B = 0 \text{ for } k = 0, 1, \dots, n-1.$$

It readily follows from Cayley-Hamilton theorem that

$$v^T A^k B = 0 \text{ for all } k \in \mathbb{N}.$$

Hence, $v^T e^{At} B = 0, \quad \forall t \geq 0.$

Therefore, $v^T W_s = 0$ and this contradicts (c).

3. Similarity Transform and Stabilizability

◆ Motivation of similarity transform

The LTI system $\begin{cases} \frac{dx(t)}{dt} = Ax(t) + Bu(t) \\ y(t) = Cx(t) + Du(t) \end{cases}$ is uncontrollable



$$\text{rank}(U_c) = r < n, \text{ where } U_c = [B \quad AB \quad \cdots \quad A^{n-1}B]$$

What if the system is uncontrollable?

→ Similarity transform may play an important role.

◆ Review of similarity transform

Two continuous-time LTI systems

$$\begin{cases} \frac{dx}{dt} = Ax + Bu \\ y = Cx + Du \end{cases} \quad \longleftrightarrow \quad \begin{cases} \frac{d\tilde{x}}{dt} = \tilde{A}\tilde{x} + \tilde{B}u \\ y = \tilde{C}\tilde{x} + \tilde{D}u \end{cases}$$

Algebraically equivalent

are called *algebraically equivalent* if there exists a nonsingular matrix T such that the followings hold:

$$\tilde{A} := TAT^{-1}, \quad \tilde{B} := TB, \quad \tilde{C} := CT^{-1}, \quad \tilde{D} := D$$

The corresponding map $\tilde{x} = Tx$ is called a similarity transform.

◆ Controllable decomposition

Suppose that the LTI system
$$\begin{cases} \frac{dx(t)}{dt} &= Ax(t) + Bu(t) \\ y(t) &= Cx(t) + Du(t) \end{cases}$$

is uncontrollable, i.e.,

$$\text{rank}(U_c) = r < n, \text{ where } U_c = [B \quad AB \quad \cdots \quad A^{n-1}B]$$

Then, there exists a similarity transform matrix T such that the following assertions are true.

(a) The transformed pair has the form

$$\tilde{A} = TAT^{-1} = \begin{bmatrix} \tilde{A}_{11} & \tilde{A}_{12} \\ 0 & \tilde{A}_{22} \end{bmatrix}, \quad \tilde{B} = TB = \begin{bmatrix} \tilde{B}_1 \\ 0 \end{bmatrix},$$

where $\tilde{A}_{11} \in \mathbb{R}^{r \times r}$ and $\tilde{B}_1 \in \mathbb{R}^{r \times k}$.

(b) The pair $(\tilde{A}_{11}, \tilde{B}_1)$ is controllable.

This is called a *controllable decomposition*.

proof: (a) Let v_1, v_2, \dots, v_r be linearly independent columns of the controllability matrix U_c . We complete them by $n - r$ linearly independent vectors $v_{r+1}, v_{r+2}, \dots, v_n$ such that the matrix

$$Q = \begin{bmatrix} v_1 & v_2 & \cdots & v_n \end{bmatrix}$$

is nonsingular and show that $T = Q^{-1}$ has the desired property.

Because of the Cayley-Hamilton theorem, each vector Av_1, \dots, Av_r can be written as a linear combination of v_1, \dots, v_r . Hence, there is an $r \times r$ matrix \tilde{A}_{11} such that

$$[Av_1 \quad \cdots \quad Av_r] = [v_1 \quad \cdots \quad v_r] \tilde{A}_{11}$$

and we can write with certain matrices $\tilde{A}_{12}, \tilde{A}_{22}$

$$\begin{aligned} AQ = AT^{-1} &= [Av_1 \quad \cdots \quad Av_r \quad Av_{r+1} \quad \cdots \quad Av_n] \\ &= [v_1 \quad \cdots \quad v_r \quad v_{r+1} \quad \cdots \quad v_n] \begin{bmatrix} \tilde{A}_{11} & \tilde{A}_{12} \\ 0 & \tilde{A}_{22} \end{bmatrix} \\ &= T^{-1} \begin{bmatrix} \tilde{A}_{11} & \tilde{A}_{12} \\ 0 & \tilde{A}_{22} \end{bmatrix} \Rightarrow TAT^{-1} = \begin{bmatrix} \tilde{A}_{11} & \tilde{A}_{12} \\ 0 & \tilde{A}_{22} \end{bmatrix}. \end{aligned}$$

Similarly, it is possible to represent each column of B as a linear combination of v_1, \dots, v_r . Thus, there is matrix \tilde{B}_1 such that

$$\begin{aligned} B &= \begin{bmatrix} v_1 & \cdots & v_r \end{bmatrix} \tilde{B}_1 = \begin{bmatrix} v_1 & \cdots & v_r & v_{r+1} & \cdots & v_n \end{bmatrix} \begin{bmatrix} \tilde{B}_1 \\ 0 \end{bmatrix} \\ &= T^{-1} \begin{bmatrix} \tilde{B}_1 \\ 0 \end{bmatrix} \Rightarrow TB = \begin{bmatrix} \tilde{B}_1 \\ 0 \end{bmatrix}. \end{aligned}$$

This completes the proof of (a).

(b) It readily follows that

$$\begin{aligned}
TU_c &= T \begin{bmatrix} B & AB & \cdots & A^{n-1}B \end{bmatrix} \\
&= \begin{bmatrix} TB & TAT^{-1}TB & \cdots & (TAT^{-1})^{n-1}TB \end{bmatrix} \\
&= \begin{bmatrix} \tilde{B}_1 & \tilde{A}_{11}\tilde{B}_1 & \cdots & \tilde{A}_{11}^{r-1}\tilde{B}_1 & \cdots & \tilde{A}_{11}^{n-1}\tilde{B}_1 \\ 0 & 0 & \cdots & 0 & \cdots & 0 \end{bmatrix}
\end{aligned}$$

Because of the Cayley-Hamilton theorem, for every $l \geq r$,

\tilde{A}_{11}^l is a linear combination of $I, \tilde{A}_{11}, \dots, \tilde{A}_{11}^{r-1}$. Thus,

$$\text{rank}(TU_c) = \text{rank} \left(\begin{bmatrix} \tilde{B}_1 & \tilde{A}_{11}\tilde{B}_1 & \cdots & \tilde{A}_{11}^{r-1}\tilde{B}_1 \end{bmatrix} \right) = \text{rank}(U_c) = r.$$

This completes the proof of (b).

◆ Interpretation of controllable decomposition

$$\begin{cases} \frac{d\tilde{x}(t)}{dt} &= \tilde{A}\tilde{x}(t) + \tilde{B}u(t) \\ y(t) &= \tilde{C}\tilde{x}(t) + \tilde{D}u(t) \end{cases}, \quad \tilde{A} = \begin{bmatrix} \tilde{A}_{11} & \tilde{A}_{12} \\ 0 & \tilde{A}_{22} \end{bmatrix}, \quad \tilde{B} = \begin{bmatrix} \tilde{B}_1 \\ 0 \end{bmatrix}$$

- (a) The system is divided into the controllable part $(\tilde{A}_{11}, \tilde{B}_1)$ and the uncontrollable part $(\tilde{A}_{22}, 0)$.
- (b) The controllable part can always be stabilized by an adequate controller.
- (c) The uncontrollable part cannot be affected by a control at all.
- (d) The system can only be stabilized when the uncontrollable part is stable.

◆ Stabilizability

When the LTI system $\begin{cases} \frac{dx(t)}{dt} &= Ax(t) + Bu(t) \\ y(t) &= Cx(t) + Du(t) \end{cases}$ is uncontrollable,

this system is called a *stabilizable*, if the matrix \tilde{A}_{22} in the following normal form is stable.

$$\begin{cases} \frac{d\tilde{x}(t)}{dt} &= \tilde{A}\tilde{x}(t) + \tilde{B}u(t) \\ y(t) &= \tilde{C}\tilde{x}(t) + \tilde{D}u(t) \end{cases}, \quad \tilde{A} = \begin{bmatrix} \tilde{A}_{11} & \tilde{A}_{12} \\ 0 & \tilde{A}_{22} \end{bmatrix}, \quad \tilde{B} = \begin{bmatrix} \tilde{B}_1 \\ 0 \end{bmatrix}$$

◆ Popov-Belevitch-Hautus (PBH) test

The LTI system
$$\begin{cases} \frac{dx(t)}{dt} &= Ax(t) + Bu(t) \\ y(t) &= Cx(t) + Du(t) \end{cases}$$

(a) is controllable if and only if

$$\text{rank}([A - \lambda I \quad B]) = n \quad \text{for every} \quad \lambda \in \mathbb{C}.$$

(b) is stabilizable if and only if

$$\text{rank}([A - \lambda I \quad B]) = n \quad \text{for every} \quad \lambda \in \mathbb{C} \quad \text{with} \quad \text{Re}(\lambda) \geq 0.$$

proof: (a)–necessary condition

Let (A, B) is controllable and suppose that

$$\text{rank}([A - \lambda I \quad B]) < n.$$

Then there exists a nonzero vector v such that

$$v^T A = \lambda v^T \text{ and } v^T B = 0.$$

This implies that

$$v^T A^l = \lambda^l v^T \quad \text{for every } l \geq 1.$$

Thus,

$$v^T [B \quad AB \quad \dots \quad A^{n-1}B] = 0.$$

This is a contradiction to the assumed controllability.

(a)–sufficient condition

Suppose that (A, B) is uncontrollable and consider the decomposition

$$\begin{cases} \frac{d\tilde{x}(t)}{dt} &= \tilde{A}\tilde{x}(t) + \tilde{B}u(t) \\ y(t) &= \tilde{C}\tilde{x}(t) + \tilde{D}u(t) \end{cases}, \quad \tilde{A} = \begin{bmatrix} \tilde{A}_{11} & \tilde{A}_{12} \\ 0 & \tilde{A}_{22} \end{bmatrix}, \quad \tilde{B} = \begin{bmatrix} \tilde{B}_1 \\ 0 \end{bmatrix}$$

Then, for every eigenvalue λ of \tilde{A}_{22} , we see that

$$\text{rank} \left(\begin{bmatrix} \tilde{A} - \lambda I & \tilde{B} \end{bmatrix} \right) < n.$$

This means that

$$\text{rank} \left(\begin{bmatrix} A - \lambda I & B \end{bmatrix} \right) = \text{rank} \left(T^{-1} \begin{bmatrix} \tilde{A} - \lambda I & \tilde{B} \end{bmatrix} \begin{bmatrix} T & 0 \\ 0 & I \end{bmatrix} \right) < n.$$

(b)–necessary condition

Let (A, B) is stabilizable and suppose that

$$\text{rank}([A - \lambda I \quad B]) < n \quad \text{for a } \lambda \quad \text{with } \text{Re}(\lambda) \geq 0.$$

Because

$$\text{rank}([A - \lambda I \quad B]) = \text{rank}([\tilde{A} - \lambda I \quad \tilde{B}]) < n,$$

there exists a nonzero vector $v = [v_1^T \ v_2^T]^T$ such that

$$v_1^T \tilde{A}_{11} = \lambda v_1^T, \ v_1^T \tilde{B}_1 = 0 \text{ and } v_1^T \tilde{A}_{12} + v_2^T \tilde{A}_{22} = \lambda v_2^T$$

Since $(\tilde{A}_{11}, \tilde{B}_1)$ is controllable, $v_1 = 0$ and thus λ is an eigenvalue of \tilde{A}_{22} .

This is a contradiction to the assumed stabilizability.

(b)–sufficient condition

Suppose that (A, B) is not stabilizable.

Then, there exists an eigenvalue λ of \tilde{A}_{22} with $\operatorname{Re}(\lambda) \geq 0$.

By using this λ , we see that

$$\operatorname{rank} \begin{pmatrix} A - \lambda I & B \end{pmatrix} = \operatorname{rank} \begin{pmatrix} \tilde{A} - \lambda I & \tilde{B} \end{pmatrix} < n.$$