

State Space Methods

Lecture 2: controllability and state feedback

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A continuous time system

$$\dot{x}(t) = Ax(t) + Bu(t), \quad x(0) = 0$$

is said to be *controllable* iff for any $\xi \in \mathbb{R}^n$ there exists u(t) such that for some T > 0, $x(T) = \xi$.



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A discrete time system

$$x(k+1) = Ax(k) + Bu(k), \quad x(0) = 0$$

is said to be *controllable* iff for any $\xi \in \mathbb{R}^n$ there exists $(u(0), u(1), \ldots)$ such that for some N > 0, $x(N) = \xi$.



We consider the discrete time system

$$x(k+1) = Ax(k) + Bu(k), \quad x(0) = 0$$

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We consider the discrete time system

$$x(k+1) = Ax(k) + Bu(k), \quad x(0) = 0$$

$$x(1) = Bu(0) x(2) = ABu(0) + Bu(1) x(3) = A^{2}Bu(0) + ABu(1) + Bu(2) \vdots x(n) = A^{n-1}Bu(0) + ... + ABu(n-2) + Bu(n-1)$$



Writing the equation

$$x(n) = A^{n-1}Bu(0) + \ldots + ABu(n-2) + Bu(n-1)$$

in matrix form we obtain:



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$$x(n) = \begin{pmatrix} B & AB & \dots & A^{n-1}B \end{pmatrix} \begin{pmatrix} u(n-1) \\ u(n-2) \\ \vdots \\ u(0) \end{pmatrix}$$



Writing the equation

$$x(n) = A^{n-1}Bu(0) + \ldots + ABu(n-2) + Bu(n-1)$$

in matrix form we obtain:

$$x(n) = \underbrace{\begin{pmatrix} B & AB & \dots & A^{n-1}B \end{pmatrix}}_{\text{Controllability matrix}} \begin{pmatrix} u(n-1) \\ u(n-2) \\ \vdots \\ u(0) \end{pmatrix}$$

When is $x(n) = \xi$ solvable for any $\xi \in \mathbb{R}^n$?



Theorem

A system

$$\Sigma : \begin{cases} \dot{x}(t) = Ax(t) + Bu(t) \text{ (continuous time)} \\ x(k+1) = Ax(k) + Bu(k) \text{ (discrete time)} \end{cases}$$

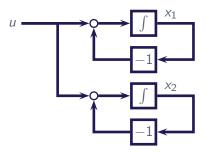
where $x \in \mathbb{R}^n$, $u \in \mathbb{R}^m$, is controllable if and only if

$$\operatorname{rank}(B \ AB \ \dots \ A^{n-1}B) = n$$

For m = 1 this reduces to

$$\det \begin{pmatrix} B & AB & \dots & A^{n-1}B \end{pmatrix} \neq 0$$

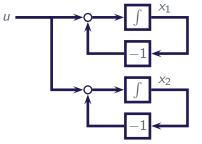




State space equations:

$$\left\{ \begin{array}{l} \dot{x}_1 = -x_1 + u \\ \dot{x}_2 = -x_2 + u \end{array} \right\}$$





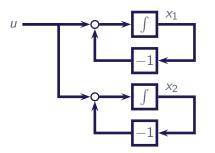
State space equations:

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State space equations in matrix form:

$$\begin{pmatrix} x_1 \\ x_2 \end{pmatrix} = \begin{pmatrix} -1 & 0 \\ 0 & -1 \end{pmatrix} \begin{pmatrix} x_1 \\ x_2 \end{pmatrix} + \begin{pmatrix} 1 \\ 1 \end{pmatrix} u$$



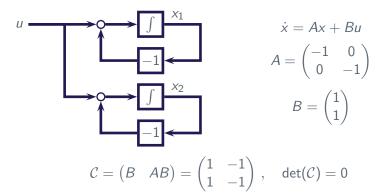


$$\dot{x} = Ax + Bu$$

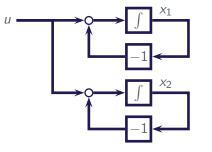
$$A = \begin{pmatrix} -1 & 0 \\ 0 & -1 \end{pmatrix}$$

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$$\dot{x} = Ax + Bu$$

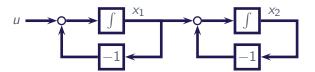
$$A = \begin{pmatrix} -1 & 0 \\ 0 & -1 \end{pmatrix}$$

$$B = \begin{pmatrix} 1 \\ 1 \end{pmatrix}$$

$$\mathcal{C} = egin{pmatrix} B & AB \end{pmatrix} = egin{pmatrix} 1 & -1 \\ 1 & -1 \end{pmatrix} \,, \quad \det(\mathcal{C}) = 0$$

 $rank(C) = 1 < 2 \Longrightarrow uncontrollable$





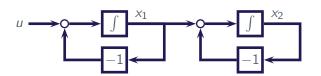
State equations:

$$\left\{ \begin{array}{ll} \dot{x}_1 & = & -x_1 + u \\ \dot{x}_2 & = & -x_2 + x_1 \end{array} \right\}$$

State space model in matrix form:

$$\begin{pmatrix} \dot{x}_1 \\ x_2 \end{pmatrix} = \begin{pmatrix} -1 & 0 \\ 1 & -1 \end{pmatrix} \begin{pmatrix} x_1 \\ x_2 \end{pmatrix} + \begin{pmatrix} 1 \\ 0 \end{pmatrix} u$$



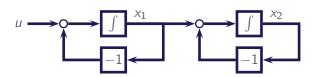


$$A = \begin{pmatrix} -1 & 0 \\ 1 & -1 \end{pmatrix} , \quad B = \begin{pmatrix} 1 \\ 0 \end{pmatrix}$$

Controllability analysis

$$C = \begin{pmatrix} B & AB \end{pmatrix} = \begin{pmatrix} 1 & -1 \\ 0 & 1 \end{pmatrix}$$



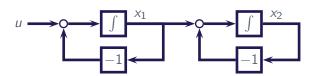


$$A = \begin{pmatrix} -1 & 0 \\ 1 & -1 \end{pmatrix} , \quad B = \begin{pmatrix} 1 \\ 0 \end{pmatrix}$$

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eq 0$$





$$A = \begin{pmatrix} -1 & 0 \\ 1 & -1 \end{pmatrix} , \quad B = \begin{pmatrix} 1 \\ 0 \end{pmatrix}$$

Controllability analysis

$$\mathcal{C} = egin{pmatrix} B & AB \end{pmatrix} = egin{pmatrix} 1 & -1 \ 0 & 1 \end{pmatrix} \,, \quad \det(\mathcal{C}) = 1
eq 0$$

$$rank(C) = 2 \Longrightarrow controllable$$

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Any controllable *single input* system can be written in the form:

$$\dot{x}_c = A_c x_c + B_c u, \quad x_c \in \mathbb{R}^n, \ u \in \mathbb{R}$$

where

$$A_c = \left(\begin{array}{c|c} & a^T \\ \hline & I_{n-1} & 0_{(n-1)\times 1} \end{array}\right) , \quad B_c = \left(\begin{array}{c|c} 1 \\ \hline 0_{(n-1)\times 1} \end{array}\right)$$

and where $a \in \mathbb{R}^{n \times 1}$, $a^T = \begin{pmatrix} a_1 & a_2 & \dots & a_n \end{pmatrix}$. It can be shown that

$$\det(\lambda I - A_c) = \lambda^n - a_1 \lambda^{n-1} - \dots - a_n$$



For n = 3 the controllable canonical form becomes:

$$A_c = \begin{pmatrix} a_1 & a_2 & a_3 \\ \hline 1 & 0 & 0 \\ 0 & 1 & 0 \end{pmatrix}, B_c = \begin{pmatrix} 1 \\ \hline 0 \\ 0 \end{pmatrix}$$

which is indeed controllable:

$$C_c = \begin{pmatrix} B_c & A_c B_c & A_c^2 B_c \end{pmatrix} = \begin{pmatrix} 1 & a_1 & a_1^2 + a_2 \\ 0 & 1 & a_1 \\ 0 & 0 & 1 \end{pmatrix}$$

 $det(C) = 1 \neq 0 \Longrightarrow$ system is controllable.



Given a state space model of a controllable system:

$$\dot{x} = Ax + Bu$$
, $x \in \mathbb{R}^n$, $u \in \mathbb{R}$

we wish to find a basis transformation $x = Tx_c$, such that:

$$\dot{x}_c = A_c x_c + B_c u, \quad x_c \in \mathbb{R}^n, u \in \mathbb{R}$$

where $A_c = T^{-1}AT$ and $B_c = T^{-1}B$, is in controllable canonical form.



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We can solve for \mathcal{T}^{-1} by rewriting these equations as

$$A_c T^{-1} = T^{-1} A$$
 and $B_c = T^{-1} B$



We consider n=3, and introduce the following notation for the rows of \mathcal{T}^{-1}

$$T^{-1} = \begin{pmatrix} s_1 \\ s_2 \\ s_3 \end{pmatrix}, \quad s_1, s_2, s_3 \in \mathbb{R}^{1 \times n}$$

Then we can rewrite the transformation equations $A_c T^{-1} = T^{-1} A$ and $T^{-1} B = B_c$ as:

$$\begin{pmatrix} a_1 & a_2 & a_3 \\ 1 & 0 & 0 \\ 0 & 1 & 0 \end{pmatrix} \begin{pmatrix} s_1 \\ s_2 \\ s_3 \end{pmatrix} = \begin{pmatrix} s_1 \\ s_2 \\ s_3 \end{pmatrix} A, \begin{pmatrix} s_1 \\ s_2 \\ s_3 \end{pmatrix} B = \begin{pmatrix} 1 \\ 0 \\ 0 \end{pmatrix}$$



Writing out these equations:

$$\begin{pmatrix} a_1 & a_2 & a_3 \\ 1 & 0 & 0 \\ 0 & 1 & 0 \end{pmatrix} \begin{pmatrix} s_1 \\ s_2 \\ s_3 \end{pmatrix} = \begin{pmatrix} s_1 \\ s_2 \\ s_3 \end{pmatrix} A, \begin{pmatrix} s_1 \\ s_2 \\ s_3 \end{pmatrix} B = \begin{pmatrix} 1 \\ 0 \\ 0 \end{pmatrix}$$



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$$\left\{ \left[s_1 = s_2 A \right] \right\}, \left\{ \right.$$



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$$\left\{ \begin{array}{c} s_1 = s_2 A \\ \boxed{s_2 = s_3 A} \end{array} \right\} \,, \left\{ \begin{array}{c} \end{array} \right\}$$



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Combining the equations

$$\left\{\begin{array}{c} s_1 = s_2 A \\ s_2 = s_3 A \end{array}\right\}, \left\{\begin{array}{c} s_1 B = 1 \\ s_2 B = 0 \\ s_3 B = 0 \end{array}\right\}$$

we obtain

$$s_3 (B \quad AB \quad A^2B) = \begin{pmatrix} 0 & 0 & 1 \end{pmatrix}$$

yielding the solution

$$s_3=\begin{pmatrix} 0 & 0 & 1 \end{pmatrix} \mathcal{C}^{-1}$$
, $s_2=s_3A$, $s_1=s_2A$ for nonsingular $\mathcal{C}=\begin{pmatrix} B & AB & A^2B \end{pmatrix}$.



We consider the system

$$\dot{x} = \begin{pmatrix} 2 & -3 \\ 4 & -5 \end{pmatrix} x + \begin{pmatrix} 2 \\ 3 \end{pmatrix} u$$

$$y = \begin{pmatrix} -3 & 2 \end{pmatrix} x$$

having the controllability matrix

$$\mathcal{C} = ig(B \quad AB ig) = ig(egin{matrix} 2 & -5 \ 3 & -7 \end{matrix} ig) \;, \quad \det(\mathcal{C}) = 1
eq 0$$



We compute the rows of T^{-1} by

$$s_2 = \begin{pmatrix} 0 & 1 \end{pmatrix} \mathcal{C}^{-1} = \begin{pmatrix} 0 & 1 \end{pmatrix} \begin{pmatrix} -7 & 5 \\ -3 & 2 \end{pmatrix} = \begin{pmatrix} -3 & 2 \end{pmatrix}$$

 $s_1 = s_2 A = \begin{pmatrix} -3 & 2 \end{pmatrix} \begin{pmatrix} 2 & -3 \\ 4 & -5 \end{pmatrix} = \begin{pmatrix} 2 & -1 \end{pmatrix}$

$$T^{-1} = \left(\begin{array}{c} \\ \end{array} \right)$$



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 $s_1 = s_2 A = \begin{pmatrix} -3 & 2 \end{pmatrix} \begin{pmatrix} 2 & -3 \\ 4 & -5 \end{pmatrix} = \begin{pmatrix} 2 & -1 \end{pmatrix}$

$$T^{-1} = \begin{pmatrix} 2 & -1 \\ -3 & 2 \end{pmatrix} \implies T = \begin{pmatrix} 2 & 1 \\ 3 & 2 \end{pmatrix}$$



Eventually, we have

$$A_c = T^{-1}AT = \begin{pmatrix} 2 & -1 \\ -3 & 2 \end{pmatrix} \begin{pmatrix} 2 & -3 \\ 4 & -5 \end{pmatrix} \begin{pmatrix} 2 & 1 \\ 3 & 2 \end{pmatrix}$$
$$= \begin{pmatrix} -3 & -2 \\ 1 & 0 \end{pmatrix}$$



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$$B_c = T^{-1}B = \begin{pmatrix} 2 & -1 \\ -3 & 2 \end{pmatrix} \begin{pmatrix} 2 \\ 3 \end{pmatrix}$$



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$$B_c = T^{-1}B = \begin{pmatrix} 2 & -1 \\ -3 & 2 \end{pmatrix} \begin{pmatrix} 2 \\ 3 \end{pmatrix} = \begin{pmatrix} \frac{1}{0} \end{pmatrix}$$



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$$= \begin{pmatrix} -3 & -2 \\ 1 & 0 \end{pmatrix} \Rightarrow \det(\lambda I - A) = \lambda^2 + 3\lambda + 2$$

$$B_c = T^{-1}B = \begin{pmatrix} 2 & -1 \\ -3 & 2 \end{pmatrix} \begin{pmatrix} 2 \\ 3 \end{pmatrix} = \begin{pmatrix} \frac{1}{0} \end{pmatrix}$$



Eventually, we have

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$$= \begin{pmatrix} -3 & -2 \\ 1 & 0 \end{pmatrix} \Rightarrow \det(\lambda I - A) = (\lambda + 1)(\lambda + 2)$$

$$B_c = T^{-1}B = \begin{pmatrix} 2 & -1 \\ -3 & 2 \end{pmatrix} \begin{pmatrix} 2 \\ 3 \end{pmatrix} = \begin{pmatrix} \frac{1}{0} \end{pmatrix}$$

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For a state space model

$$\dot{x} = Ax + Bu$$

a state feedback is a feedback of the form

$$u = Fx$$



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$$\dot{x} = Ax + Bu$$

a state feedback is a feedback of the form

$$u = Fx$$

Combining these two equations, we obtain:

$$\dot{x} = Ax + BFx = (A + BF)x$$

Thus, the result of a state feedback is a system with a modified system matrix, and thus with modified poles.

For a single input system in companion form, a state feedback takes a particular simple form:

$$A_c = \begin{pmatrix} a_1 & a_2 & a_3 \\ \hline 1 & 0 & 0 \\ 0 & 1 & 0 \end{pmatrix}, B_c = \begin{pmatrix} 1 \\ \hline 0 \\ 0 \end{pmatrix}$$

Applying the feedback u = Fx with

$$F_c = \begin{pmatrix} f_1 & f_2 & f_3 \end{pmatrix}$$

We obtain:

$$A_c + B_c F_c = \begin{pmatrix} a_1 & a_2 & a_3 \\ \hline 1 & 0 & 0 \\ 0 & 1 & 0 \end{pmatrix} + \begin{pmatrix} 1 \\ \overline{0} \\ 0 \end{pmatrix} \begin{pmatrix} f_1 & f_2 & f_3 \end{pmatrix}$$
$$= \begin{pmatrix} a_1 + f_1 & a_2 + f_2 & a_3 + f_3 \\ \hline 1 & 0 & 0 \\ 0 & 1 & 0 \end{pmatrix}$$



Thus, the characteristic polynomium has been changed from

$$\det(\lambda I - A_c) = \lambda^n - a_1 \lambda^{n-1} - \ldots - a_n$$

to

$$\det(\lambda I - (A_c + B_c F_c)) = \lambda^n - (a_1 + f_1)\lambda^{n-1} - \dots - (a_n + f_n)$$

By choosing f_1, \ldots, f_n appropriately, any closed loop pole configuration can be obtained. This is known as *pole assignment*.



Let $A \in \mathbb{R}^{n \times n}$, $B \in \mathbb{R}^{n \times 1}$ be given.

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- 5. Compute resulting feedback gain $F = F_c T^{-1}$.



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5.
$$F = F_c T^{-1} = \begin{pmatrix} -6 & -18 \end{pmatrix} \begin{pmatrix} 2 & -1 \\ -3 & 2 \end{pmatrix} = \begin{pmatrix} 42 & -30 \end{pmatrix}$$



