

Article

Enhancing Wind Energy Harvesting Using Passive Turbulence Control Devices

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Abstract: Aiming to predict the performance of galloping piezoelectric energy harvesters, a theoretical model is established and verified by experiments. The relative error between the model and experimental results is 5.3%. In addition, the present model is used to study the AC output characteristics of the piezoelectric energy harvesting system under passive turbulence control (PTC), and the influence of load resistance on the critical wind speed, displacement, and output power under both strong and weak coupling are analyzed from the perspective of electromechanical coupling strength, respectively. The results show that the critical wind speed initially increases and then decreases with increasing load resistance. For weak and critical coupling cases, the output power firstly increases and then decreases with the increase of the load resistance, and reaches the maximum value at the optimal load. For the weak, critical, and strong coupling cases, the critical optimal load is $1.1\text{ M}\Omega$, $1.1\text{ M}\Omega$, and $3.0\text{ M}\Omega$, respectively. Overall, the response mechanism of the presented harvester is revealed.

Keywords: galloping; passive turbulence control; electromechanical coupling strength; critical velocity

1. Introduction

In recent years, micro-electromechanical systems (MEMS) and wireless sensor networks (WSNs) have been developed rapidly in the industrial field. Traditional batteries for powering the above devices have some disadvantages. For example, they have a limited lifespan and are difficult to be replaced in some application areas. The technique of harvesting energy from environmental vibrations can effectively provide an energy supplement for MEMS and WSNs [1–5]. In addition, advanced nonlinear systems were invented to enhance broadband energy harvesting [6–10]. In ambient environments, fluid kinetic energy has the advantages of being renewable and clean, such as wind and water energy, can be efficiently harvested [11–15], and has sufficient potential for energy conversion [16–20].

In particular, wind energy harvesting technology, based on different physical mechanisms to excite nonlinear resonators, has been developed rapidly [21–25]. Liu et al. [26] presented a comprehensive review of state-of-the-art piezoelectric energy harvesting (PEH). Various key aspects to improve the overall performance of a PEH device are discussed in detail. Joseph et al. [27] established the possibility of utilizing energy harvesting from mechanical vibrations to estimate extreme value responses of the host structure, and indicated that vibration energy harvesting, in its own right, has the potential to be used for extreme value analysis and estimates. Furthermore, this work is applied to health monitoring and assessment, and to the impact of exposure to the natural environment throughout

its life cycle. Piezoelectric or electromagnetic transducers attached to the resonators can convert the kinetic energy of wind into useful electric energy. It can work in hostile environments, which overcomes the limitations of large wind installations. A wide band vibrating piezoelectric energy harvester with adjustable frequency was presented by Yang and Zhang [28]. The working frequency of the piezoelectric energy harvester could be adjusted by a rotating joint beam. Their results showed that a multi-beam system broadened the working frequency range of the harvester and improved the energy harvesting efficiency. Ibrahim et al. [29] designed a turbine using the double multi-flow tube algorithm in QBlade software, which can harness 200 W output power at 0.582 m/s water velocity with a 5.5 m² swept area. Zhang et al. [30] studied the influence of the spring-damped rectangular cylinder on the cross-sectional aspect ratio of a shark fin and the hydrokinetic energy conversion by a numerical method, with results indicating that the converted power shows an increasing trend. As the aspect ratio decreased from 2 to 1/4, the maximum Flow Induced Motion (FIM) energy conversion efficiency also accordingly increased to 15.5%. Hu et al. [31,32] studied the performance of a cylinder galloping piezoelectric energy harvester with two small rod-shaped accessories installed on the main cylinder through experiments. The rod sections were round, triangular, and square. The results showed that by attaching these three rod types at the circumferential locations of 45° and 60°, power could be harvested continuously beyond the critical wind speed and was dramatically superior over both the plain circular cylinder and the cases with other attachment locations. In addition, they measured aerodynamic force coefficients and found that the circular cylinder with triangular rods at the location of 60° had larger transverse force coefficients than the other two rod type cases. Bernitsas et al. [33,34] discussed the basic concepts of Vortex Induced Vibration Aquatic Clean Energy (VIVACE) converters and gave the design of physical models and laboratory prototypes. Based on this, a mathematical model was established and the design parameters of various application scales were calculated. Sun et al. [35] experimentally studied a kind of nonlinear harvester and tested the energy conversion efficiency of fluid dynamics under passive turbulence control when the Reynolds number was $30,000 \leq Re \leq 120,000$. Results showed that the utilized power increased with the increase of nonlinearity at low flow velocity. He et al. [36] tested a new type of wind energy harvester. Results showed that the effective wind speed region was integrated with the galloping wind speed region, forming a single region with a high-power output of the harvester. Mutsuda et al. [37] developed a highly flexible galloping piezoelectric energy harvester and optimized the design of the whole system by considering aspect ratio, support system, initial tension, and other factors. Chang et al. [38] studied the effects of roughness location and surface coverage on vibration by the passive turbulence control (PTC) method with surface roughness selective distribution, and the range of Re was $3.0 \times 10^4 < Re < 1.2 \times 10^5$. The results showed that 16° roughness coverage is effective in the range inducing the reduced vortex-induced vibration (VIV), enhanced VIV, or galloping.

Yan et al. [39] theoretically studied a galloping energy harvester with a bluff body of the triangular cross-section using the quasi-steady state approximation method to model the aerodynamic load. They concluded that the maximum power was obtained when the lateral displacement was the smallest with the change in load resistance. On the other hand, Mutsuda and Tabesh et al. [40,41] established a quasi-steady state vibration model of a piezoelectric cantilever beam under the condition of small deflection vibration. In addition, they applied the frequency domain analysis method to solve the electromechanical coupling equation by Laplace transform (for a linear system), and put forward the concepts of electric damping and electric stiffness. Barrero et al. [42] studied the influence of different cross-section shapes and mechanical parameters of the harvester on the energy harvesting efficiency by establishing a single-degree-of-freedom (SDOF) model, and they theoretically analyzed the potential possibility of galloping energy harvesting. Wang et al. [43–45] solved a fluid–mechanical–electrical coupling model of flow-induced vibration energy harvesting systems using a concentrated parameter method and the quasi-steady state theory. Ramasur and Sirohi [46,47] developed a harvester in which an equilateral triangular cross-section of the resistive fluid is connected to two piezoelectric cantilevers. When the wind speed is 4.92 m/s, the maximum power can reach 53 mW. Zhao and

Dai et al. [48–50] used wind tunnel experiment tests and circuit simulations to derive explicitly the analytical solution of steady-state mechanical and electrical responses of galloping-based energy harvesters with the synchronized switch harvesting on inductor (SSH) interface. However, the above research of vibration energy harvesting mostly uses a pure resistive load circuit and seldom involves high strength electromechanical coupling of complex nonlinear load circuits. In many cases, strong and weak coupling exist side by side, such as in AC circuits, high-speed motors, elevators, wireless sensors, and some high-strength detection devices. Energy harvesting technologies can be used in conjunction with working sensors or other microelectronic devices. Recently, some researchers have explored the strong and weak coupling effects based on galloping energy harvesting. Liao et al. [51] analyzed the maximum output power of a piezoelectric energy harvester. They proposed that the total power limit could be obtained by tuning the energy harvesting circuit, and they defined three different types of coupling strength by analyzing the closed expression between the mechanical damping and the effective electromechanical coupling coefficient. Shu and Lien [52] studied the optimal output power of system rectifier piezoelectric devices. They proposed design guidelines for devices with a large coupling coefficient and quality factor. In addition, Renno et al. [53] studied the maximum possible power of the harvester with an R-L (resistive load and inductor) circuit interface. Liang and Liao [54] experimentally studied the output power of the harvester under base excitations, which was 200% higher than that with a standard circuit.

In this paper, the behavior of galloping piezoelectric energy harvesters with two 60° PTC structures (GEH-PTC) is experimentally and theoretically studied. A theoretical model is provided. This model is used to analyze the energy harvesting efficiency of the harvester under the condition of strong and weak coupling load circuits, and the response of amplitude, power, and bandwidth. The analysis and verification of the models are established. In addition, the performance of the two energy harvesters are compared. Furthermore, the paper analyzes the influence of wind speed and external load on the performance of the GEH-PTC under the strong and weak coupling condition.

2. Physical Mathematical Model

2.1. Physical and Mathematical Model

Galloping vibration can be a kind of bending vibration of the cantilever-based structure caused by the self-excitation of airflows [55–59]. For the VIV vibration transforming the galloping vibration, Hu et al. [31,32] studied the influence of attached rods with cross-sections and equipping angles on the transformation from vortex-induced vibration to galloping vibration, and they explored the influence of attack angle on transverse force coefficients. Inspired by Hu et al. [31,32], Bernitsas et al. [33,34], and our previous study [13], the physical model in our paper uses quite a different structure called a Passive Turbulence Control (PTC) device, which is made of sanding belts with very coarse surface. In addition, this paper focuses on the influence of different electrical circuits components, including electromechanical coupling strengths, wind speeds, and load resistances on the performance of the harvester.

Considering a curved section structure mounted on a flexible support, the surface along the axial direction of 60° is symmetrical with two belts (to induce turbulence conditions for PTC). A previous study has proved that 60° Y-shaped attachments can arouse galloping phenomenon [13] and VIV is caused by the vortex shedding effect, while resonance occurs when the vortex shedding frequency is close to the natural frequency of the vibratory system. In other words, VIV is mainly caused by the vortex shedding effect. Equipping a Passive Turbulence Control (PTC) device on the bluff body can affect the formation of the vortex street behind the bluff body, and different flow-induced motions (FIM) can be observed depending primarily on the circumferential location of the two strips [60,61]. Furthermore, galloping can be aroused when a negative nonlinear aerodynamic damping appears for the equipment of the PTC in this work. In this study, two compact belts are designed to assist the passive turbulence control to improve the performance of the wind energy harvester, as shown in

Figure 1. As shown in Figure 1a, wind speed U is perpendicular to the cylindrical section. Based on the Euler-Bernoulli beam theory, Kirchhoff's law, piezoelectric effect, and galloping self-induced vibration, the aero-electromechanical coupling model of the energy harvester is able to be obtained in a distributed parameter model, which can be included into a lumped parameter model form [13,62]. The structure shown in Figure 1a is simplified to an equivalent Single-degree-of-freedom (SDOF) mass-damping-spring (M - C - K) model in Figure 1b. The cantilever beam is equivalent to a spring oscillator, and α is the angle of attack. Therefore, the governing equations can be expressed as:

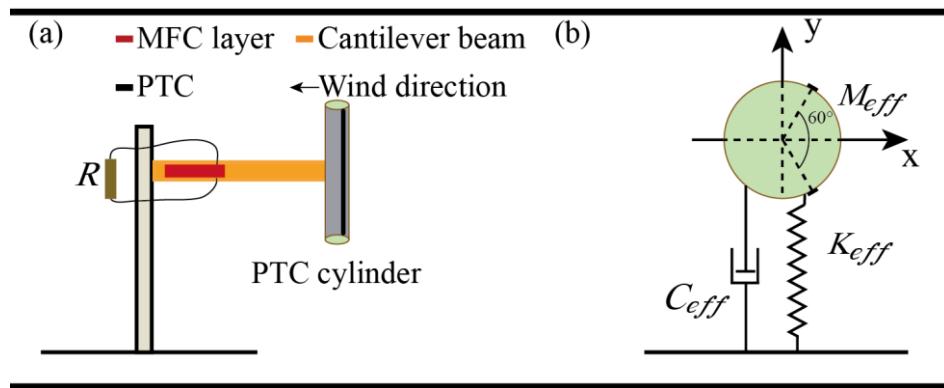


Figure 1. The Galloping based energy harvester with PTC(GEH-PTC): (a) The structural diagram; (b) the equivalent model.

$$M_{eff}\ddot{y}(t) + C_{eff}\dot{y}(t) + K_{eff}y(t) + \theta V(t) = F_y(t) \quad (1)$$

$$\frac{V(t)}{R} + C_p \dot{V}(t) - \theta \dot{y}(t) = 0 \quad (2)$$

where M_{eff} is the equivalent mass. $M_{eff} = (33/140)m_1 + m_2$, (m_1 and m_2 are the mass of the piezoelectric beam and the bluff body, respectively). $C_{eff} = 2\zeta\omega_n M_{eff}$ is the equivalent mechanical damping, where ζ is the damping ratio. $K_{eff} = \omega_n^2 M_{eff}$ is the equivalent stiffness, where ω_n is the natural frequency. $F_y(t)$ is the aerodynamic force. $y(t)$ is the displacement in the y direction. $V(t)$ represents output voltage across the load resistance R . θ is the electromechanical coupling coefficient, which can be obtained by:

$$\theta = \sqrt{(\omega_{oc}^2 - \omega_{sc}^2)M_{eff}C_p} \quad (3)$$

where ω_{oc} and ω_{sc} are open circuit and short circuit frequencies, respectively. C_p is the piezoelectric capacitance. In order to analyze the performance of the harvester under the condition of strong and weak couplings, the dimensionless quantity k should be defined (k represents the electromechanical coupling strength). In the load circuit interface, Liao et al. [51] expressed the electromechanical coupling strength as $k^2 = \frac{\theta^2}{C_p K_{eff}}$, and three different coupling strengths are respectively defined as follows:

$$k^2 < 4\zeta + 4\zeta^2 \quad (4)$$

$$k^2 = 4\zeta + 4\zeta^2 \quad (5)$$

$$k^2 > 4\zeta + 4\zeta^2 \quad (6)$$

The three expressions in Equations (4)–(6) represent weak coupling, critical coupling, and strong coupling, respectively. Assuming that the wind speed remains constant, under quasi-static conditions, $F_y(t)$ can be expressed as [63].

$$F_y(t) = \frac{1}{2}\rho SU^2 \left[\sum_{i=1,2,\dots} a_i \left(\frac{\dot{y}(t)}{U} \right)^i \right] \quad (7)$$

where ρ is the air density. S is the area of the bluff body perpendicular to the wind direction, U is the wind speed, a_i is the empirical coefficient related to the shape of the bluff body section. $\sum_{i=1,2,\dots} a_i \left[\frac{\dot{y}(t)}{U} \right]^i$ is the aerodynamic coefficient, it is expressed as $C_{F_y(t)}$, where $\frac{\dot{y}(t)}{U} = \tan\alpha$. The galloping stability of the system depends on whether the aerodynamic coefficient and angle of attack are in the same direction. The vibration is stable [64] when $\frac{\partial C_{F_y(t)}}{\partial \alpha} > 0$. Combining Equations (1) and (7), then Equation (1) can be rewritten as:

$$M_{eff}\ddot{y}(t) + K_{eff}y(t) + \theta V(t) + \left[C_{eff} - \frac{1}{2}a_1\rho SU - \frac{1}{2}a_3\rho S \frac{\dot{y}^2(t)}{U} \right] \dot{y}(t) = 0 \quad (8)$$

For $i = 1$, the damping term of the system is $C_{eff} - 1/2a_1\rho SU$, and the damping term is linear. For $i > 1$, the damping term exists in a nonlinear form. Supposing the critical wind speed of the bluff body is U_{cr} . When $U < U_{cr}$, the damping term is $C_{eff} - 1/2a_1\rho SU > 0$, the nonlinear term can be neglected, at this time the damping term inhibits the vibration of the bluff body. Critical wind speed U_{cr} is expressed as [56]:

$$U_{cr} = \frac{R_1 + \frac{RN^2}{1+R_1^2\omega_n^2C_p^2}}{\frac{1}{2}\rho Sa_1} \quad (9)$$

where R and N are load resistance and ideal transformer ratio, respectively. R_1 is the equivalent component resistance. As wind speed increases, the bluff body begins to vibrate when $U = U_{cr}$, and the damping term becomes negative. At this time, the nonlinear damping term starts to maintain the vibration of the bluff body. For a complex alternating circuit, the critical wind speed of the GEH-PTC can be determined by circuit simulation. Taking into account the rotation of the bluff body, the revised aerodynamic coefficient is given as:

$$C_{F_y(t)} = \sum_{i=1,2,\dots} a_i \left[\frac{\dot{y}(t)}{U} + y'(t) \right]^i \quad (10)$$

where the angle of attack is $\tan\alpha = \frac{\dot{y}(t)}{U} + y'(t)$ ($\lambda = \frac{y'(t)}{\dot{y}(t)}$). The coefficient of the rotation angle and the lateral displacement at the free end of the cantilever beam is $\lambda = 1.5/L$ (L is the length of the free segment of the cantilever beam) [65].

2.2. Solution and Verification of the Model

The influence of the load at the interface of the circuit on the natural frequency, the system damping and the critical wind speed of the harvester can be determined by linear analysis of electromechanical coupling problems. The following state variables are introduced:

$$X = \begin{bmatrix} x_1 \\ x_2 \\ x_3 \end{bmatrix} = \begin{bmatrix} y(t) \\ \dot{y}(t) \\ V(t) \end{bmatrix} \quad (11)$$

Therefore, the matrix form of Equations (1) and (2) can be expressed as:

$$\dot{X} = \begin{cases} x_2 \\ -\frac{C_{eff}}{M_{eff}}x_2 - \frac{K_{eff}}{M_{eff}}x_1 - \frac{\theta}{M_{eff}}x_3 + \frac{\rho S U^2}{2M_{eff}} \\ \left(a_1 \left(\frac{x_2}{U} + \lambda x_1 \right) + a_3 \left(\frac{x_2}{U} + \lambda x_1 \right)^3 \right) \\ \frac{\theta}{C_p}x_2 - \frac{x_3}{R C_p} \end{cases} \quad (12)$$

MATLAB program solver “ODE45.m” was used to obtain the analytical solution of Equations (9) and (12). The aerodynamic parameters and the fitting results of linear and nonlinear coefficients a_1 and a_3 in the matrix equation are shown in Table 1. Positive nonlinear mechanical damping is introduced into the harvester. The amplitude is self-limited, so that the harvester undergoes steady-state limit cycle oscillation at each wind speed.

Table 1. Values of system aerodynamic parameters and a_1, a_3 .

U	λ/K_{eff}	$(-1/2)\rho S U^2$	a_1	a_3
1.276	0.62	-0.0038	2.4	-35
1.413	0.62	-0.0046	2.4	-28
1.550	0.62	-0.0055	2.3	-21
1.687	0.62	-0.0066	2.4	-24
1.824	0.62	-0.0077	2.6	-24
1.961	0.62	-0.0089	2.8	-23
2.098	0.62	-0.0101	2.9	-20
2.235	0.62	-0.0115	3.0	-20
2.372	0.62	-0.0130	3.0	-19
2.509	0.62	-0.0145	3.0	-19
2.646	0.62	-0.0161	3.0	-19

In addition, the resistance value ranges from $100\text{ K}\Omega$ to $10\text{ M}\Omega$. The electromechanical coupling coefficient is $\theta = 2.0 \times 10^{-5}\text{ N}\cdot\text{V}^{-1}$. The piezoelectric plates are made of fiber MFC (Macro Fiber Composite). Capacitance is $C_p = 15.7\text{ nF}$. The other values of the relevant parameters are listed in Table 2.

Table 2. Parameters of the harvester.

Material and Physical Properties	Symbol	Numerical Value
The effective length of the cantilever beam/m	L_a	0.14
Bluff body diameter/m	d	0.032
The length of the bluff/m	L	0.12
Equivalent mass/kg	M_{eff}	5.15×10^{-3}
width of the PTC / m	L_b	0.004
The system damping/ $\text{N}\cdot\text{s}\cdot\text{m}^{-1}$	C_{eff}	7.6×10^{-3}
The system stiffness/ $\text{N}\cdot\text{m}^{-1}$	K_{eff}	17.21
Damping ratio	ζ	0.013
Natural frequency/Hz	f_n	9.2

In order to verify the theoretical model of the harvester, a T-shaped device was designed, as shown in Figure 2. The device was installed in a circular wind tunnel with a diameter of 400 mm and was connected to a linear resistance load with wires. The honeycomb region in the middle of the wind tunnel is the stable wind section. The piezoelectric plate was $35 \times 10 \times 0.3\text{ mm}^3$ in size and was placed on the top of the aluminum cantilever beam. The width of the PTC was 3 mm and the length was the same as the bluff body. Simple harmonics produced by the bluff body can be captured by a digital oscilloscope (ISDS220B) and the vibration is measured by a laser sensor (Panasonic, hg-c1400, Kadoma shi, Japan).

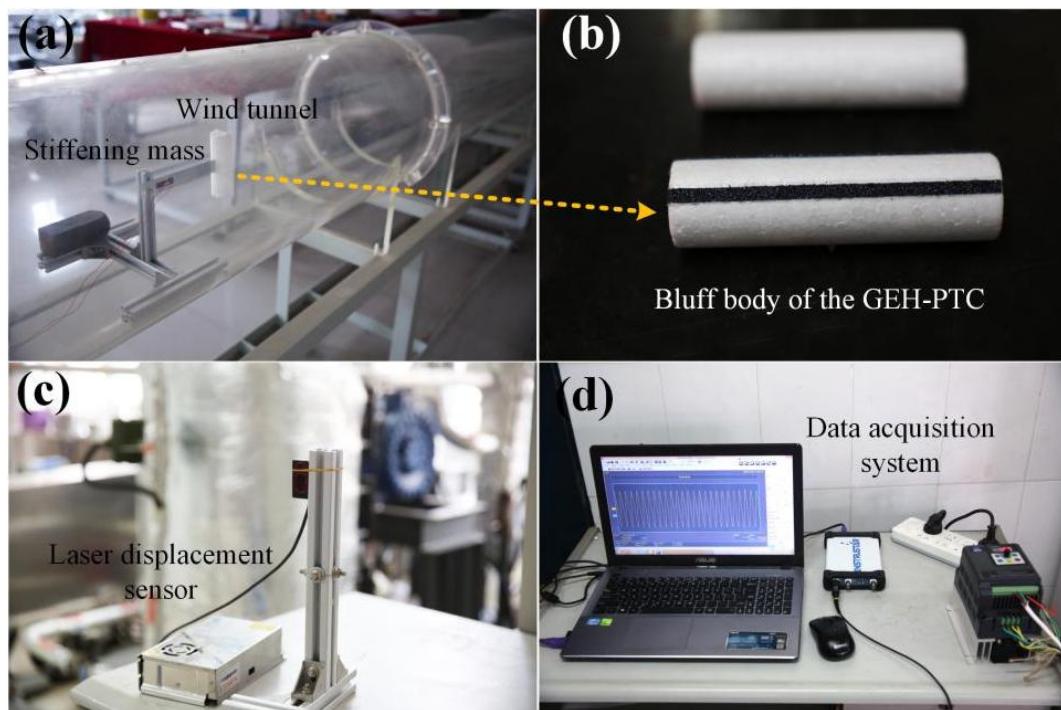


Figure 2. Experimental setup: (a) the wind tunnel with the bluff body with attachments (PTC), which is made of (b) sanding belts with a very coarse surface, (c) the laser sensor, and (d) the data acquisition system.

The theoretical model was verified by experimental results as shown in Figure 3, where Figure 3a shows the effect of load resistance R on the critical wind speed U_{cr} . It can be found that for the galloping energy harvester, U_{cr} firstly increases and then decreases with the increase of R .

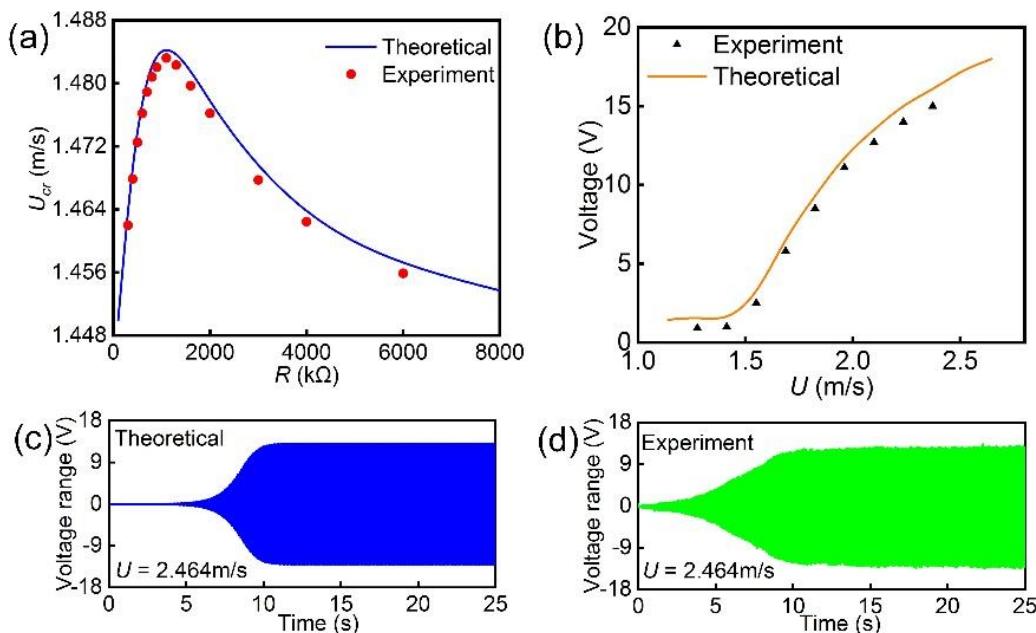


Figure 3. Verification of the theoretical model: (a) the critical wind speed U_{cr} versus R ; (b) the output voltage versus R ; (c,d) are respectively theoretical and experimental time-domain output voltage range of the GEH-PTC at $U = 2.464$ m/s and $R = 1.1$ M Ω .

The model maximizes at the optimum load of $1.1 \text{ M}\Omega$, $U_{crmax} = 1.485 \text{ m/s}$. Figure 3b is the output voltage amplitude along with the wind speed in the open circuit condition. The voltage increases along with increases in the wind speed, which is a common characteristic of the galloping energy harvester. Figure 3c,d are respectively the time-domain output voltage range of the GEH-PTC with $R = 1.1 \text{ M}\Omega$ at $U = 2.464 \text{ m/s}$. The stable theoretical and experimental voltage amplitudes are 13.2 V and 12.5 V, respectively. Figure 3 shows that both the critical wind speed and the output voltage are slightly smaller than the theoretical values when the wind speed is very small, but the divergence is small. In general, the results from the theoretical model are very close to the experimental results, proving its accuracy.

3. Performance of the GEH-PTC

In this section, we experimentally compare two different harvesters (their cylinders have the same mass). Figure 4a,b are respectively the experimental output voltage and vibration frequency along with the wind speed under the open circuit condition. From Figure 4a, the GEH-PTC and the vortex-induced vibration energy harvester (VIVEH) with an ordinary smooth cylinder respectively represent two completely different vibration forms, which are galloping vibration and vortex-induced vibration. Galloping vibration shows that the output voltage increases along with increases in wind speed, while vortex-induced vibration shows that the output voltage firstly increases and then decreases with an increase in wind speed. Figure 4b further analyzes the difference between them based on the vibration frequency.

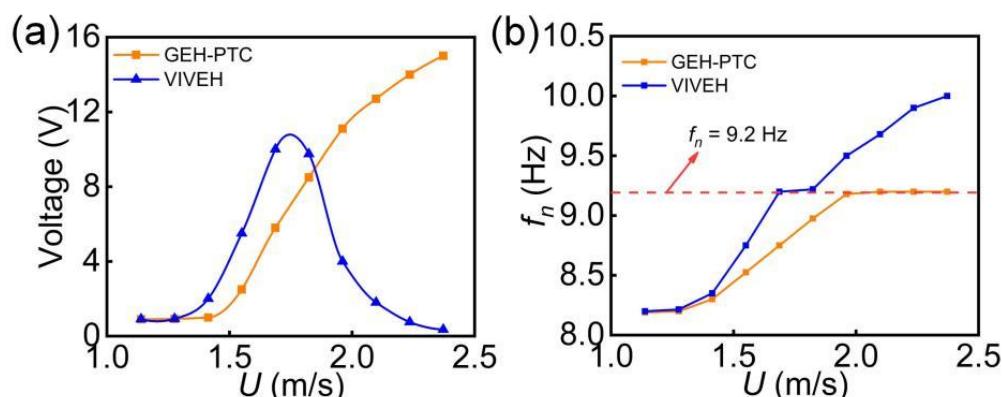


Figure 4. Experimental comparison of the GEH-PTC and the vortex-induced vibration energy harvester (VIVEH): (a) output voltage; (b) vibration frequency.

The natural frequency of both two harvesters is 9.2 Hz. When the wind speed is small, the vibration frequency of the GEH-PTC and the VIVEH differs greatly from the natural frequency, therefore, it is difficult to generate resonant vibration. With the increase in wind speed, the vibration frequency of the two energy harvesters is close to the natural frequency. Their effective speed ranges are $U > 1.9 \text{ m/s}$ and $1.67\text{--}1.83 \text{ m/s}$, respectively. Therefore, the GEH-PTC has a much wider effective speed range than the VIVEH. In other words, the GEH-PTC can efficiently harvest energy once the wind speed is larger than the critical wind speed. More importantly, the design of the GEH-PTC is very convenient for changing to different PTCs, which may change its dynamic characteristics and benefit wind energy harvesting. Figure 5 compares the output power of two harvesters with the optimal load $R_{opt} = 1.1 \text{ M}\Omega$. It can be found that the maximum power of the GEH-PTC and the VIVEH is 0.058 mW and 0.03 mW when $U_1 = 2.372 \text{ m/s}$ and $U_2 = 1.687 \text{ m/s}$, respectively. The output power of the GEH-PTC is improved by 93% compared with the VIVEH.

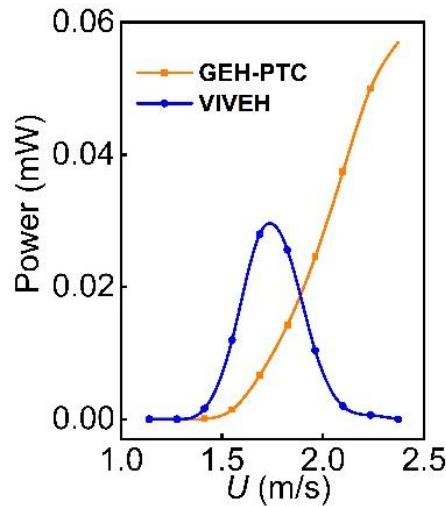


Figure 5. The output power comparison of the GEH-PTC and the VIVEH in experiment.

4. Effect of the Coupling Strength

In this section, the influence of the electromechanical coupling strength, wind speed, and the load resistance on the output voltage and the response displacement of the harvester will be numerically analyzed. The parameters of the energy harvester are the same as those in the last section. The damping ratio is calculated by free decay experiments $\zeta = 0.013$. $k^2/(4\zeta + 4\zeta^2) = 0.029, 1, 2.9$, respectively

4.1. Effect of Coupling Strength

Figure 6 shows the output power of the GEH-PTC along with the load resistance with three different electromechanical coupling strengths. Figure 6a,b respectively represent the weak coupling ($k^2/(4\zeta + 4\zeta^2) = 0.029$) and the critical coupling ($k^2/(4\zeta + 4\zeta^2) = 1$), and the wind speed range is 2.098–2.646 m/s. When the electromechanical coupling strength presents weak coupling and critical coupling, the output power firstly increases and then decreases with the increase of external load, and it reaches the maximum value at the optimal load resistance. The optimal load resistance of the weak coupling and the critical coupling is respectively 1.1 MΩ and 3.0 MΩ, which is very close to the theoretical one ($R_{opt} = 1/\omega_n C_P$) [64]. When the wind speed is 2.646 m/s, the peak output power in the above two cases reaches 0.15 mW and 0.5 mW, respectively.

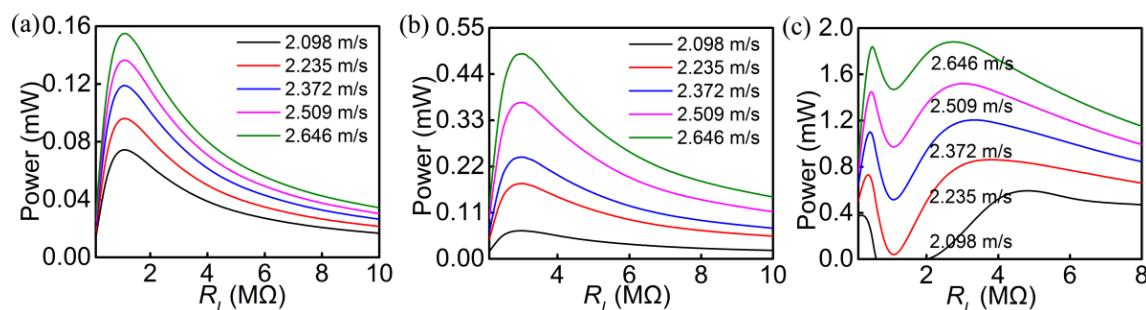


Figure 6. Numerical output power versus load resistance: (a) Weak coupling $k^2/(4\zeta + 4\zeta^2) = 0.029$; (b) critical coupling $k^2/(4\zeta + 4\zeta^2) = 1$; (c) strong coupling $k^2/(4\zeta + 4\zeta^2) = 2.9$.

For the strong coupling ($k^2/(4\zeta + 4\zeta^2) = 2.9$), results are shown in Figure 6c. At each wind speed, there will be a trough between the two power peaks, and the corresponding resistance value of the trough is the optimal load resistance. Therefore, based on Figure 6, we can guess that the coupling strength is an important factor affecting the output power. The output power increases with an

increase in electromechanical coupling strength. However, when the coupling strength is higher than the critical coupling strength, the output power does not increase further, and the power reaches saturation. The change of output power along with the wind speed is shown in Figure 7, and the resistance is in range of 0.5–8.0 MΩ. As a whole, it can be observed that under the three coupling strengths power increases with increasing wind speed. However, at a higher wind speed, the power increase rate gradually decreases, which is consistent with the experimental results. In addition, as shown in Figure 7c, when the electromechanical coupling strength presents as strong coupling, the power increase rate under small resistance ($R = 0.5$ MΩ) is obviously greater than that under large resistance ($R = 8.0$ MΩ), which can be seen from the intersection of bold solid lines. Different from strong coupling, weak coupling (Figure 7a) and critical coupling (Figure 7b) do not show such a phenomenon of solid line crossing, and the power reaches the maximum at the optimal load (Figure 7a: $R_{opt} = 1.1$ MΩ; Figure 7b: $R_{opt} = 3.0$ MΩ), which can also be reflected in Figure 6.

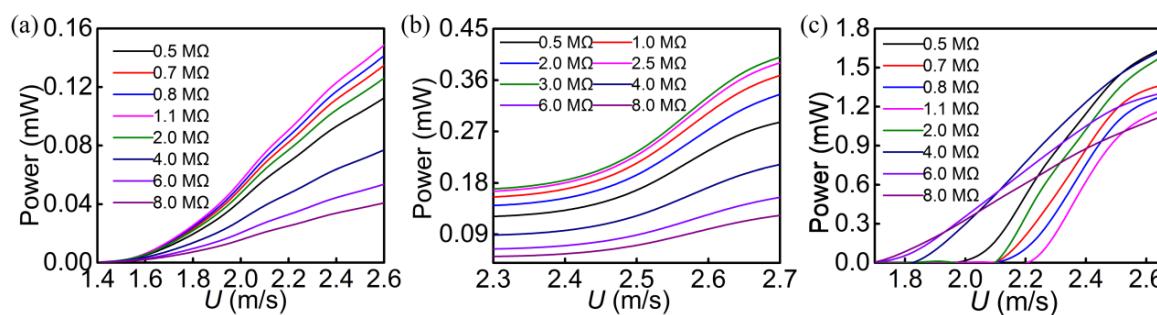


Figure 7. Numerical output power versus the wind speed: (a) Weak coupling $k^2/(4\zeta + 4\zeta^2) = 0.029$; (b) critical coupling $k^2/(4\zeta + 4\zeta^2) = 1$; (c) strong coupling $k^2/(4\zeta + 4\zeta^2) = 2.9$.

4.2. Effect of the Coupling Strength on Displacement

The response displacement of the GEH-PTC with three coupling strengths is shown in Figure 8. It is found that the response displacement increases along with increasing coupling strength and wind speed. The maximum displacement the GEH-PTC with the weak, critical, and strong couplings reached 6 mm, 10 mm, and 13.5 mm, respectively. Meanwhile, the corresponding wind speed is 1.824 m/s, 2.372 m/s, 2.646 m/s, respectively. Compared with the weak coupling, the displacement in the strong coupling case increases by 125%. However, increases in coupling strength makes the response displacement curve of the GEH-PTC with the optimal load appear as a ‘depression’ under each wind speed, which is obvious in the case of strong coupling. For the critical coupling (Figure 8b) and the strong coupling (Figure 8c), the response displacement curve even breaks when the wind speed is 1.687 m/s and 2.098 m/s due to the ‘sag’ phenomenon. At this time, the load resistance ranges are 2–4 MΩ and 0.5–2 MΩ, respectively. Galloping phenomenon does not occur when the load resistance is in these ranges. Therefore, it is called a ‘suppressed load’. When the load resistance in the three coupling cases are 1.1 MΩ, 3.0 MΩ, and 1.1 MΩ, respectively, the galloping suppression is the most obvious. On the contrary, this value is shown as the optimal load in Figure 6a,b.

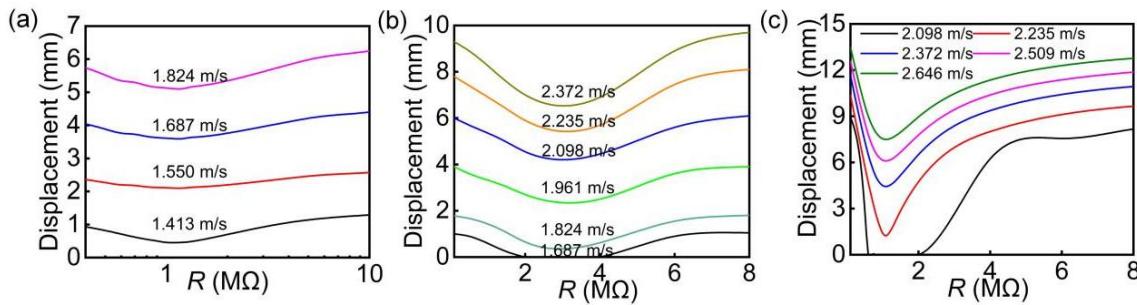


Figure 8. Variation of numerical displacement with the load resistance: (a) weak coupling $k^2/(4\zeta + 4\zeta^2) = 0.029$; (b) critical coupling $k^2/(4\zeta + 4\zeta^2) = 1$; (c) strong coupling $k^2/(4\zeta + 4\zeta^2) = 2.9$.

In Figure 9, the response displacement of the GEH-PTC with three coupling strengths are compared. The load resistances in each case are $0.5\text{ M}\Omega$, $0.7\text{ M}\Omega$, $1.1\text{ M}\Omega$, $2.0\text{ M}\Omega$, and $4.0\text{ M}\Omega$. The response displacement in the three cases also increases with increasing wind speed. When the wind speed is 2.6 m/s and the load resistance is $0.5\text{ M}\Omega$, the maximum displacement reaches 8 mm , 11 mm , and 14 mm for the weak, critical, and strong coupling cases, respectively. For a specific coupling strength, different resistance values correspond to different critical wind speeds U_{cr} . High load resistance leads to a high critical wind speed. However, the displacement corresponding to high resistance values will obviously decrease. For the critical (Figure 9a), weak (Figure 9b), and strong coupling (Figure 9c) cases, the critical wind speed range is respectively $1.92\text{--}1.97\text{ m/s}$, $1.94\text{--}2.04\text{ m/s}$, and $1.76\text{--}1.86\text{ m/s}$. The necessary condition for achieving galloping vibration is that the incoming wind speed is higher than the critical value.

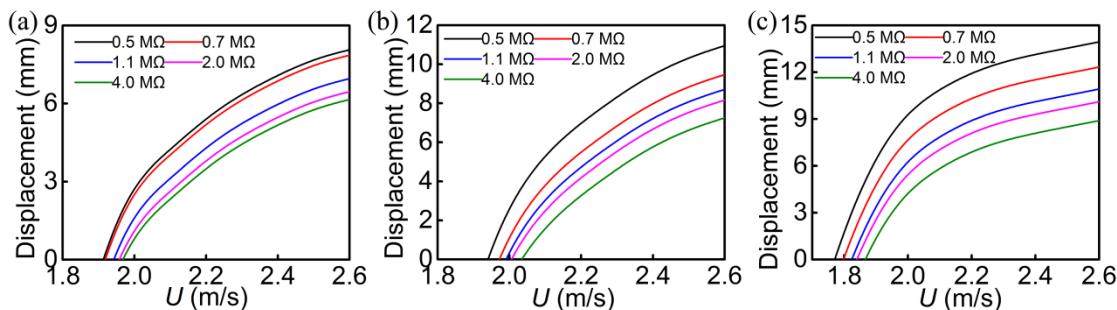


Figure 9. Numerical displacement versus wind speed: (a) Weak coupling $k^2/(4\zeta + 4\zeta^2) = 0.029$; (b) critical coupling $k^2/(4\zeta + 4\zeta^2) = 1$; (c) strong coupling $k^2/(4\zeta + 4\zeta^2) = 2.9$.

5. Conclusions

In this paper, the galloping piezoelectric energy harvester with two 60° PTC structures (GEH-PTC) is experimentally and theoretically studied. A theoretical model of the GEH-PTC is established and the accuracy of the model is verified by experiments. The effect of the load resistance on the critical wind speed (U_{cr}), the response displacement, and the output power of the GEH-PTC under strong and weak coupling conditions are analyzed from the perspective of the electromechanical coupling strength. Results show that U_{cr} increases firstly and then decreases along with increasing load resistance. Compared with the VIVEH in the studied wind range conditions, the maximum output power and voltage of the GEH-PTC increased by 93% and 66.7%, respectively. When the electromechanical coupling strength shows weak coupling and critical coupling, the output power increases firstly and then decreases with increased load resistance, and it reaches the maximum value at the optimal load, which is $1.1\text{ M}\Omega$ and $3.0\text{ M}\Omega$ for the former and the latter, respectively. At 2.646 m/s , the peak output power in the cases of the weak coupling and the critical coupling reached 0.15 mW and 0.5 mW , respectively. With increasing coupling strength and wind speed, the response displacement increases

obviously. Compared with the weak coupling case, the displacement in the strong coupling case increased by 125%. In future work, we will change the coupling strength by changing the piezoelectric materials in the experiment to obtain different coupling strengths. The key factor is the piezoelectric materials, specifically the arrangement mode, numbers, etc.

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