1

Fourier Series

EE3900: Linear Systems and Signal Processing

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1. Periodic Function

Let

$$x(t) = A_0 |\sin(2\pi f_0 t)| \tag{1.1}$$

1.1 Plot x(t)

Solution: Download the following Python code that plots Fig. 1.1.

wget https://github.com/MouktikaCherukupalli/fourier/blob/main/codes/1.1.py

Run the code by executing

python 1.1.py

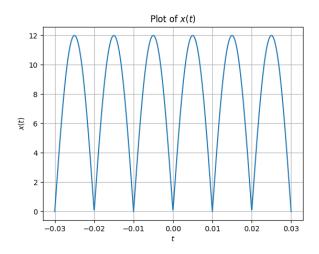


Fig. 1.1. Plot of x(t)

1.2 Show that x(t) is periodic and find its period **Solution:** Since x(t) is the absolute value of a sinusoidal function, it is periodic, which is also evident from the plot

Consider $x(t + \frac{1}{2f_0})$

$$x\left(t + \frac{1}{2f_0}\right) = A_0 \left| \sin\left(2\pi f_0 \left(t + \frac{1}{2f_0}\right)\right) \right|$$
 (1.2)

$$= A_0 \left| \sin \left(2\pi f_0 t + \pi f_0 \right) \right| \tag{1.3}$$

$$= A_0 \left| (-1)^{f_0} \sin \left(2\pi f_0 t \right) \right| \tag{1.4}$$

$$= A_0 |\sin(2\pi f_0 t)| \tag{1.5}$$

$$= x(t) \tag{1.6}$$

Therefore, x(t) is periodic with period $\frac{1}{2f_0}$

2. Fourier Series

Consider $A_0 = 12$ and $f_0 = 50$ for all numerical calculations

2.1 If

$$x(t) = \sum_{k=-\infty}^{\infty} c_k e^{j2\pi k f_0 t}$$
 (2.1)

show that

$$c_k = f_0 \int_{-\frac{1}{2f_0}}^{\frac{1}{2f_0}} x(t)e^{-j2\pi k f_0 t} dt$$
 (2.2)

Solution:

$$x(t)e^{-j2\pi nf_0t} = \sum_{k=-\infty}^{\infty} c_k e^{-j2\pi(n-k)f_0t}$$

$$\implies \int_{-\frac{1}{2f_0}}^{\frac{1}{2f_0}} x(t)e^{-j2\pi nf_0t} dt = \sum_{k=-\infty}^{\infty} c_k \int_{-\frac{1}{2f_0}}^{\frac{1}{2f_0}} e^{-j2\pi(n-k)f_0t} dt$$
(2.3)

But

$$\int_{-\frac{1}{2f_0}}^{\frac{1}{2f_0}} e^{-j2\pi(n-k)f_0t} dt = \begin{cases} \frac{1}{f_0} & k = n\\ 0 & k \neq n \end{cases}$$

$$= \frac{1}{f_0} \delta(n-k)$$
(2.5)

$$\sum_{k=-\infty}^{\infty} c_k \int_{-\frac{1}{2f_0}}^{\frac{1}{2f_0}} e^{-j2\pi(n-k)f_0 t} dt = \sum_{k=-\infty}^{\infty} c_k \frac{1}{f_0} \delta(n-k)$$

$$= \frac{1}{f_0} c_n * \delta(n) \quad (2.8)$$

$$= \frac{1}{f_0} c_n \quad (2.9)$$

Therefore

$$c_k = f_0 \int_{-\frac{1}{2f_0}}^{\frac{1}{2f_0}} x(t)e^{-j2\pi kf_0t} dt$$
 (2.10)

2.2 Find c_k for (1.1)

Solution:

$$c_k = f_0 \int_{-\frac{1}{2f_0}}^{\frac{1}{2f_0}} A_0 \left| \sin \left(2\pi f_0 t \right) \right| e^{-J2\pi k f_0 t} dt \quad (2.11)$$

$$c_k = f_0 \int_{-\frac{1}{2f_0}}^0 A_0 \left(-\sin(2\pi f_0 t) \right) e^{-J2\pi k f_0 t} dt$$
$$+ f_0 \int_0^{\frac{1}{2f_0}} A_0 \left(\sin(2\pi f_0 t) \right) e^{-J2\pi k f_0 t} dt \quad (2.12)$$

$$c_k = f_0 \int_0^{\frac{1}{2f_0}} A_0 \sin(2\pi f_0 u) e^{J2\pi k f_0 u} dt$$
$$+ f_0 \int_0^{\frac{1}{2f_0}} A_0 \sin(2\pi f_0 t) e^{-J2\pi k f_0 t} dt \quad (2.13)$$

$$c_k = f_0 \int_0^{\frac{1}{2f_0}} A_0 \sin(2\pi f_0 t) \left(e^{j2\pi k f_0 t} + e^{-j2\pi k f_0 t} \right) dt$$
(2.14)

$$= f_0 A_0 \int_0^{\frac{1}{2f_0}} 2\sin(2\pi f_0 t) \cos(2\pi k f_0 t) dt$$
(2.15)

 $= f_0 A_0 \int_0^{\frac{1}{2f_0}} \left\{ \sin \left(2\pi (1+k) f_0 t \right) + \sin \left(2\pi (1-k) f_0 t \right) \right\} dt$

$$(2.16)$$

$$= \frac{f_0 A_0}{2\pi f_0} \left[\frac{1 - (-1)^{1+k}}{1+k} + \frac{1 - (-1)^{1-k}}{1-k} \right]$$
(2.18)
$$= \left(1 + (-1)^k \right) \frac{A_0}{2\pi} \left[\frac{1}{1+k} + \frac{1}{1-k} \right]$$
(2.19)

$$= \left(1 + (-1)^k\right) \frac{A_0}{\pi (1 - k^2)} \tag{2.20}$$

Therefore

$$c_k = \begin{cases} \frac{2A_0}{\pi(1-k^2)} & k \text{ is even} \\ 0 & k \text{ is odd} \end{cases}$$
 (2.21)

2.3 Verify (1.1) using Python

Solution: Download the following Python code that plots Fig. 3.8.

wget https://github.com/MouktikaCherukupalli /fourier/blob/main/codes/2.3.py

Run the code by executing

python 2.3.py

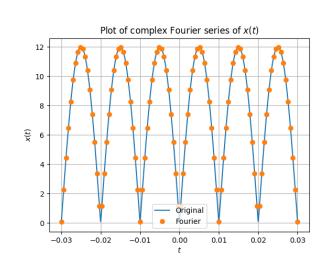


Fig. 2.3. Plot of x(t) along with its complex Fourier series expansion

2.4 Show that

$$x(t) = \sum_{k=0}^{\infty} (a_k \cos(2\pi k f_0 t) + b_k \sin(2\pi k f_0 t))$$
(2.22)

and obtain the formulae for a_k and b_k

$$x(t) = \sum_{k=-\infty}^{\infty} c_k e^{j2\pi k f_0 t}$$

$$= c_0 + \sum_{k=-\infty}^{\infty} c_k e^{j2\pi k f_0 t} + c_{-k} e^{-j2\pi k f_0 t}$$
(2.23)

(2.23)

Thus

$$x(t) = c_0 + \sum_{k=1}^{\infty} (c_k + c_{-k}) \cos(2\pi k f_0 t)$$
$$+ \sum_{k=1}^{\infty} J(c_k - c_{-k}) \sin(2\pi k f_0 t) \quad (2.25)$$

Therefore

$$a_k = \begin{cases} c_0 & k = 0 \\ c_k + c_{-k} & k > 0 \end{cases}$$
 (2.26)

$$b_k = j(c_k - c_{-k}) \quad k \ge 0 \tag{2.27}$$

2.5 Find a_k and b_k for (1.1)

Solution:

$$a_0 = c_0 = \frac{2A_0}{\pi} \tag{2.28}$$

For k > 0, if k is odd

$$a_k = 0 + 0 = 0 \tag{2.29}$$

and if k is even

$$a_k = \frac{2A_0}{\pi(1 - k^2)} + \frac{2A_0}{\pi(1 - k^2)} = \frac{4A_0}{\pi(1 - k^2)}$$
(2.30)

For odd or even k, $c_k = c_{-k}$ always

$$b_k = 0 \quad \forall k \ge 0 \tag{2.31}$$

Therefore

$$a_{k} = \begin{cases} \frac{2A_{0}}{\pi} & k = 0\\ \frac{4A_{0}}{\pi(1-k^{2})} & k = 2m, m \in \mathbb{N}\\ 0 & \text{otherwise} \end{cases}$$
 (2.32)

$$b_k = 0 \qquad k \ge 0 \tag{2.33}$$

2.6 Verify (2.22) using Python

Solution: Download the following Python code that plots Fig. 2.6.

wget https://github.com/MouktikaCherukupalli/fourier/blob/main/codes/2.6.py

Run the code by executing

python 2.6.py

3. Fourier Transform

3.1

$$\delta(t) = 0 \qquad t \neq 0 \tag{3.1}$$

$$\int_{-\infty}^{\infty} \delta(t) dt = 1$$
 (3.2)

3.2 The Fourier Transform of g(t) is

$$G(f) = \int_{-\infty}^{\infty} g(t)e^{-j2\pi ft} dt$$
 (3.3)

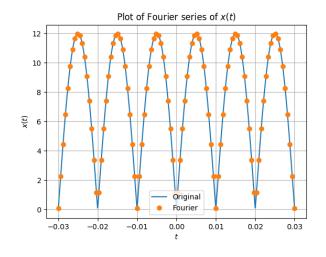


Fig. 2.6. Plot of x(t) along with its Fourier series expansion

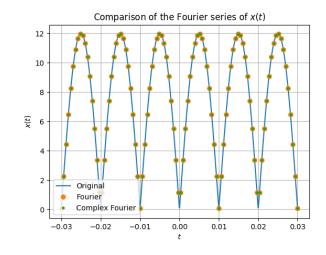


Fig. 2.6. Comparison of the Fourier series of x(t)

3.3 Show that

$$g(t-t_0) \stackrel{\mathcal{F}}{\longleftrightarrow} G(f)e^{-j2\pi ft_0}$$
 (3.4)

Solution:

$$g(t - t_0) \stackrel{\mathcal{F}}{\longleftrightarrow} \int_{-\infty}^{\infty} g(t - t_0) e^{-j2\pi f t} dt \qquad (3.5)$$

$$= \int_{-\infty}^{\infty} g(u)e^{-j2\pi f(u+t_0)} du$$
 (3.6)

$$= e^{-j2\pi f t_0} \int_{-\infty}^{\infty} g(u) e^{-j2\pi f u} \, du \quad (3.7)$$

$$=G(f)e^{-j2\pi ft_0} (3.8)$$

3.4 Show that

$$G(t) \stackrel{\mathcal{F}}{\longleftrightarrow} g(-f)$$
 (3.9)

Solution:

$$G(t) \stackrel{\mathcal{F}}{\longleftrightarrow} \int_{-\infty}^{\infty} G(t)e^{-j2\pi ft} dt$$
 (3.10)

But

$$g(t) = \int_{-\infty}^{\infty} G(f)e^{j2\pi ft} df \qquad (3.11)$$

$$= \int_{-\infty}^{\infty} G(u)e^{j2\pi ut} du \qquad (3.12)$$

$$\implies g(-f) = \int_{-\infty}^{\infty} G(u)e^{-j2\pi uf} du \qquad (3.13)$$
$$= \mathcal{F} \{G(t)\} \qquad (3.14)$$

3.5 Find the Fourier transform of $\delta(t)$ Solution:

$$\delta(t) \stackrel{\mathcal{F}}{\longleftrightarrow} \int_{-\infty}^{\infty} \delta(t) e^{-j2\pi f t} dt$$
 (3.15)

$$= e^{-j2\pi ft} \Big|_{t=0} \tag{3.16}$$

$$=1 \tag{3.17}$$

3.6 Find the Fourier transform of $e^{-j2\pi f_0 t}$ **Solution:**

$$\delta(t) \stackrel{\mathcal{F}}{\longleftrightarrow} 1$$
 (3.18)

$$\implies \delta(t - f_0) \stackrel{\mathcal{F}}{\longleftrightarrow} e^{-j2\pi f f_0} \tag{3.19}$$

$$\implies e^{-j2\pi t f_0} \stackrel{\mathcal{F}}{\longleftrightarrow} \delta(-f - f_0) \tag{3.20}$$

$$\therefore e^{-j2\pi t f_0} \stackrel{\mathcal{F}}{\longleftrightarrow} \delta(f + f_0) \tag{3.21}$$

3.7 Find the Fourier transform of $\cos(2\pi f_0 t)$ Solution:

$$\cos(2\pi f_0 t) = \frac{e^{j2\pi f_0 t} + e^{-j2\pi f_0 t}}{2}$$
 (3.22)

$$\implies \cos(2\pi f_0 t) \stackrel{\mathcal{F}}{\longleftrightarrow} \frac{\delta(f - f_0) + \delta(f + f_0)}{2}$$
(3.23)

3.8 Find the Fourier transform of x(t) and plot it. Verify using Python

Solution:

$$x(t) = \sum_{k=-\infty}^{\infty} c_k e^{j2\pi k f_0 t}$$
 (3.24)

$$\mathcal{F}\left\{x(t)\right\} = \sum_{k=-\infty}^{\infty} c_k \mathcal{F}\left\{e^{j2\pi k f_0 t}\right\}$$
 (3.25)

$$=\sum_{k=-\infty}^{\infty}c_k\delta(f-kf_0)$$
 (3.26)

$$= \frac{2A_0}{\pi} \sum_{k \text{ is even}} \frac{\delta(f - kf_0)}{1 - k^2}$$
 (3.27)

Download the following Python code that plots Fig. ??.

wget https://github.com/MouktikaCherukupalli/fourier/blob/main/codes/3.8.py

Run the code by executing

python 3.8.py

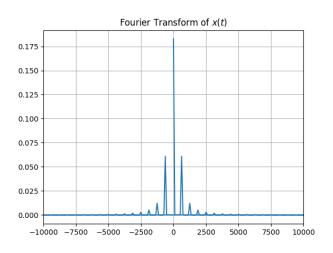


Fig. 3.8. Plot of the Fourier transform of x(t)

3.9 Show that

$$rect(t) \stackrel{\mathcal{F}}{\longleftrightarrow} sinc(f)$$
 (3.28)

Verify using Python

Solution:

$$\operatorname{rect}(t) = \begin{cases} 1 & |t| \le \frac{1}{2} \\ 0 & \text{otherwise} \end{cases}$$
 (3.29)

Its Fourier transform is given by

$$\operatorname{rect}(t) \stackrel{\mathcal{F}}{\longleftrightarrow} \int_{-\infty}^{\infty} \operatorname{rect}(t) e^{-j2\pi f t} dt$$
 (3.30)

$$= \int_{-\frac{1}{2}}^{\frac{1}{2}} e^{-j2\pi ft} dt$$
 (3.31)

$$=\frac{e^{-j\pi f}-e^{j\pi f}}{-j2\pi f}$$
(3.32)

$$=\frac{\sin \pi f}{\pi f} \tag{3.33}$$

$$= \operatorname{sinc}(f) \tag{3.34}$$

Download the following Python code that plots Fig. 3.9.

wget https://github.com/MouktikaCherukupalli/fourier/blob/main/codes/3.9.py

Run the code by executing

python 3.9.py

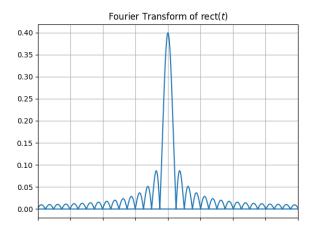


Fig. 3.9. Plot of the Fourier transform of rect(t)

3.10 Find the Fourier transform of sinc (*t*). Verify using Python

Solution:

$$\operatorname{rect}(t) \stackrel{\mathcal{F}}{\longleftrightarrow} \operatorname{sinc}(f)$$
 (3.35)

$$\implies \operatorname{sinc}(t) \stackrel{\mathcal{F}}{\longleftrightarrow} \operatorname{rect}(-f)$$
 (3.36)

$$= rect(f) \tag{3.37}$$

Download the following Python code that plots Fig. 3.10.

wget https://github.com/MouktikaCherukupalli/fourier/blob/main/codes/3.10.py

Run the code by executing

python 3.10.py

4. Filter

4.1 Find H(f) which transforms x(t) to DC 5 V **Solution:** Since we want a DC output, the filter we need is a low-pass filter that only lets the zero frequency component pass through, i.e., the amplitude of a frequency components with a frequency higher than the cutoff frequency f_c has to be zero

We can use a rectangular filter for this purpose

$$H(f) = k \operatorname{rect}\left(\frac{f}{2f_c}\right) = \begin{cases} k & |f| \le f_c \\ 0 & \text{otherwise} \end{cases}$$
 (4.1)

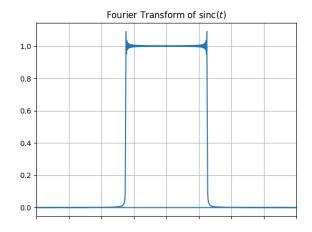


Fig. 3.10. Plot of the Fourier transform of sinc(t)

Now

$$H(0) = \frac{Y(0)}{X(0)} \tag{4.2}$$

where Y(k) and X(k) are the Fourier transforms of the output 5 V DC and the input signal respectively

$$k = \frac{5}{2A_0} = \frac{5\pi}{2A_0} \tag{4.3}$$

$$\therefore H(f) = \frac{5\pi}{2A_0} \operatorname{rect}\left(\frac{f}{2f_c}\right) \tag{4.4}$$

4.2 Find h(t)

Solution:

$$\sin(t) \stackrel{\mathcal{F}}{\longleftrightarrow} \operatorname{rect}(f) \qquad (4.5)$$

$$\implies \operatorname{sinc}(2f_c t) \stackrel{\mathcal{F}}{\longleftrightarrow} \frac{1}{2f_c} \operatorname{rect}\left(\frac{f}{2f_c}\right) \qquad (4.6)$$

$$\implies \frac{5\pi}{2A_0} 2f_c \operatorname{sinc}(2f_c t) \stackrel{\mathcal{F}}{\longleftrightarrow} \frac{5\pi}{2A_0} \operatorname{rect}\left(\frac{f}{2f_c}\right) \qquad (4.7)$$

$$\therefore h(t) = \frac{5\pi f_c}{A_0} \operatorname{sinc}(2f_c t) \tag{4.8}$$

4.3 Verify your result using through convolution **Solution:** Download the following Python code that plots Fig. 4.3.

wget https://github.com/MouktikaCherukupalli/fourier/blob/main/codes/4.3.py

Run the code by executing

python 4.3.py

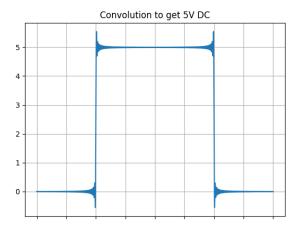


Fig. 4.3. Plot of the convolution of x(t) and h(t)

5. FILTER DESIGN

5.1 Design a Butterworth filter for H(f)Solution: The transfer function of a Butterworth filter is given by

$$|H_n(f)| = \frac{1}{\sqrt{1 + \left(\frac{f}{f_c}\right)^{2n}}}$$
 (5.1)

where n is the order of the filter and f_c is the cutoff frequency

Let the passband and stopband frequency thresholds be $50\,\mathrm{Hz}$ and $100\,\mathrm{Hz}$ and their corresponding attenuations be $-1\,\mathrm{dB}$ and $-5\,\mathrm{dB}$ respectively

$$A_p = 10\log_{10} |H_n(f_p)|^2 (5.2)$$

$$= -10\log_{10}\left(1 + \left(\frac{f_p}{f_c}\right)^{2n}\right)$$
 (5.3)

$$A_s = -10\log_{10}\left(1 + \left(\frac{f_s}{f_c}\right)^{2n}\right)$$
 (5.4)

$$\implies n = \frac{\log\left(\frac{10^{-\frac{A_p}{10}} - 1}{10^{-\frac{A_s}{10}} - 1}\right)}{2\log\left(\frac{f_p}{f_s}\right)} \approx 1.53$$
 (5.5)

Hence, we choose a 2nd order Butterworth filter with

$$f_c = \frac{f_p}{\left(10^{-\frac{A_p}{10}} - 1\right)^{\frac{1}{2n}}} \approx 77.74 \,\mathrm{Hz}$$
 (5.6)

5.2 Design a Chebyschev filter for H(f)

Solution: The transfer function of a Chebyshev filter is given by

$$|H_n(f)| = \frac{1}{\sqrt{1 + \epsilon^2 T_n^2 \left(\frac{f}{f_c}\right)}} \tag{5.7}$$

where ϵ is the ripple factor, f_c is the cutoff frequency and T_n is a Chebyshev polynomial of the n^{th} order

Assuming the same parameters as before along with a ripple of 0.1 dB, we get

$$\epsilon = \sqrt{10^{\frac{\delta}{10}} - 1} \approx 0.15 \tag{5.8}$$

Also, assume that $f_c = f_p \implies \frac{f_s}{f_c} > 1$

$$A_{s} = -10\log_{10}\left(1 + \epsilon^{2}T_{n}^{2}\left(\frac{f_{s}}{f_{c}}\right)\right) \quad (5.9)$$

$$\Longrightarrow T_n \left(\frac{f_s}{f_c} \right) = \frac{\sqrt{10^{-\frac{A_s}{10}} - 1}}{\epsilon} \tag{5.10}$$

$$\implies \cosh\left(n\cosh^{-1}\left(\frac{f_s}{f_c}\right)\right) = \frac{\sqrt{10^{-\frac{A_s}{10}} - 1}}{\epsilon}$$
(5.11)

Thus

$$n = \frac{\cosh^{-1}\left(\frac{\sqrt{10^{-\frac{A_s}{10}}-1}}{\epsilon}\right)}{\cosh^{-1}\left(\frac{f_s}{f_c}\right)} \approx 2.26$$
 (5.12)

Hence, we choose a 3rd order Chebyshev filter 5.3 Design a circuit for your Butterworth filter

Solution: Using the table of normalized Butterworth coefficients, we can see that for a 2nd order Butterworth filter

$$C_1 = 1.4142 \,\mathrm{F}$$
 (5.13)

$$L_2 = 1.4142 \,\mathrm{H}$$
 (5.14)

On denormalizing these values, we get

$$C_1' = \frac{C_1}{2\pi f_c} = 2.89 \,\text{mF}$$
 (5.15)

$$L_2' = \frac{L_2}{2\pi f_c} = 2.89 \,\text{mH}$$
 (5.16)

5.4 Design a circuit for your Chebyschev filter **Solution:** Using the table of normalized Chebyshev coefficients, we can see that for a 3rd order Chebyshev filter

$$C_1 = 1.4328 \,\mathrm{F}$$
 (5.17)

$$L_2 = 1.5937 \,\mathrm{H}$$
 (5.18)

$$C_3 = 1.4328 \,\mathrm{F}$$
 (5.19)

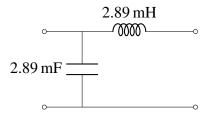


Fig. 5.3. 2nd order Butterworth filter circuit

On denormalizing these values, we get

$$C_1' = \frac{C_1}{2\pi f_c} = 4.56 \,\text{mF}$$
 (5.20)

$$L_2' = \frac{L_2}{2\pi f_c} = 5.07 \,\text{mH}$$
 (5.21)

$$C_3' = \frac{C_3}{2\pi f_c} = 4.56 \,\text{mF}$$
 (5.22)

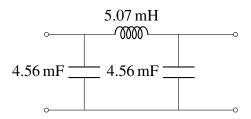


Fig. 5.4. 3rd order Chebyshev filter circuit