

3.8 Fitting Harmonic Functions — The FFT

For this section of the practical you will find the Signals and Systems textbook and hand-out *Sampled Signals, the DTFT and the DFT* useful. In an experimental situation we often measure a time-series which we want to Fourier transform to obtain a spectrum, or a frequency-dependent quantity which we wish to inverse Fourier transform in order to retrieve a time-dependent quantity. This requires us to perform a Discrete Fourier Transform (DFT), which is usually calculated using the Fast Fourier Transform (FFT) algorithm (Brigham). The DFT (and therefore the FFT) work on a finite number N data samples, and return the same number of transform samples. The FFT algorithm places restrictions on the possible values of N , and the FFT algorithm used by the Excel FFT tool restricts us to $N = 2^n$, i.e. N must be an integer power of 2 (e.g. 64, 512, 2048).

3.8.1 Exercise: The Spectrum of a Truncated and Sampled Signal

1. What is the Fourier transform of the signal:

$$x_1(t) = \cos(\omega_1 t) + \sin(\omega_2 t)$$

where $\omega_1 = 2\pi \times 10^3$ rad/s and $\omega_2 = 4\pi \times 10^3$ rad/s?

2. What is the Fourier transform of the signal if its is forced to be causal and is truncated after 16 ms, i.e. the signal:

$$x_2(t) = x_1(t) \times [u(t) - u(t - 16 \times 10^{-3})] = x_1(t) \times \Pi\left(\frac{t - 8 \times 10^{-3}}{16 \times 10^{-3}}\right)$$

3. Use an Excel spreadsheet to generate 512 samples of the signal $x_2(t)$ using a sampling interval of $T = 3.125 \times 10^{-5}$ s, i.e. generate:

$$x_2[n] = x_2(nT) \quad n = 0 \cdots 511$$

Plot this signal on suitable axes.

4. Use the Excel FFT tool to obtain the DFT $X[k]$ of the sampled signal. Notice that the returned array is in cartesian complex form, with the index k running from 0 through to $N - 1$. Convert the returned array to polar form and plot the magnitude on appropriate axes. The equivalent frequency spacing of the returned DFT samples is $\Delta\omega = 2\pi/(NT)$. Use this relationship to scale the frequency axis appropriately. Recall that the DFT is a finite set of samples of the DTFT, which in turn is related to the aliased CTFT spectrum, i.e.:

$$X[k] \approx \frac{1}{NT} X(\omega)|_{\omega=2\pi k/(NT)}$$

You must now reorganize the returned FFT samples $X[k]$ to give a more physically realistic appearance to the spectrum (Recall that the DFT is periodic with a period N , i.e. $X[k] = X[k + N]$). Thus, the 512 samples are $X[0] \cdots X[511]$, of which the first 256 are $X[0] \cdots X[255]$. The remaining 256 samples $X[256] \cdots X[511]$ must now be copied to positions $X[-256] \cdots X[-1]$. Note that this places the “DC” component corresponding to $k = 0$ at position 257. The point $X[256]$ (now at position 1) is called the “folding point” Plot $X[-256] \cdots X[255]$ vs ω .

5. Repeat the above exercise using a sampling period $T = 3.0 \times 10^{-5}$.

Plot $X[-256] \cdots X[255]$ vs ω . Can you explain the difference in the resulting DFT spectrum? Which signal is more likely to be obtained in a real experiment?

3.8.2 Exercise: The Aliased Spectrum of a Signal

1. What is the Fourier transform of the signal:

$$x_3(t) = \Pi\left(\frac{t - 25 \times 10^{-3}}{50 \times 10^{-3}}\right)$$

2. What is the Fourier transform of this signal after impulse train sampling with a sampling period of $T = 10^{-3}$ s, i.e.:

$$x_4(t) = x_3(t) \times \sum_{\ell=-\infty}^{+\infty} \delta(t - \ell \times 10^{-3})$$

3. Use an Excel spreadsheet to generate and plot 512 samples of the signal $x_3(t)$ using a sampling period of $T = 10^{-3}$ s, i.e.:

$$x_4[n] = x_3(nT) \quad n = 0 \cdots 511$$

4. Use Excel to find the DFT spectrum $X(\omega)$, $k = 0 \cdots 511$, re-order to give a 512 point spectrum centred near zero, plotted against ω , $k = -256, \cdots 0, \cdots 255$, and compare this DFT spectrum with the analytic spectrum of the signal $x_3(t)$. Why are there differences?

3.8.3 Exercise: The Inverse DFT

1. What is the inverse Fourier Transform of the signal:

$$X_5(\omega) = \Pi\left(\frac{\omega}{4\pi \times 10^3}\right)$$

2. Use an Excel spreadsheet to generate 512 samples of the function $X_5(\omega)$, which might be considered to be a measured frequency response of an electronic filter, i.e.:

$$X[k] = X_5(k\Delta\omega) \quad k = -\frac{N}{2} \cdots \frac{N}{2} - 1$$

Use a frequency sampling interval of $\Delta\omega = 40\pi$ rad/s. Notice that this choice of N and $\Delta\omega$ implies a total frequency coverage of:

$$-\left(\frac{N}{2}\right) \times \Delta\omega \leq \omega \leq +\left(\frac{N}{2} - 1\right) \times \Delta\omega$$

Note $k = 0$ will be the 257th sample. Re-order the samples prior to inverse DFT processing so that the index k runs from 0 to $N - 1$. Ensure that the samples are Hermitian to ensure that the transformed function is purely real.

3. Use the Excel inverse DFT tool to determine and plot $x_5(t)$, which is the impulse response for the hypothetical filter, usually denoted $h(t)$.
Compare the DFT result with the expected analytic result (Be careful to use the correct frequency width to generate the analytic result).
4. As an optional exercise, include a linear phase factor in the hypothetical discrete frequency response and see how it affects the computed impulse response.

3.9 References

- J[B23] Bevington, P.R. & Robinson, D.K., *Data Reduction and Error Analysis for the Physical Sciences*, 1992, McGraw-Hill.
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- J[O2] Brigham, E.O., *The Fast Fourier Transform*, 1974, Prentice-Hall.
- DD1[O3] Oppenheim, A.V. & Willsky, A.S., *Signals and Systems*, 1997, Prentice-Hall.