

Fourier Representations

	Discrete Time	CT
Periodic	DT Fourier Series DFTS	Fourier Series FS
Non periodic	DT Fourier Transform DTFT	Fourier Transform FT

① Periodic Signals (Fourier Series)

— Representing Periodic Signals as weighted superposition of complex sinusoids.

— Each sinusoid in the representation must have the same period as the signal.

(Ex) $\underline{x(t)} = \underline{x_1(t)} + \underline{x_2(t)}$ T : period
 ω : frequency

$$x(t) = x_1(t + m \cdot T_1) + x_2(t + k \cdot T_2) \quad m, k \in \mathbb{Z}$$

$$T = m T_1 = k T_2 \Rightarrow \frac{m}{\omega_1} = \frac{k}{\omega_2} = \frac{1}{\omega} \quad T$$

$$m = \frac{\omega_1}{\omega} : \text{must be an integer}$$

$$k = \frac{\omega_2}{\omega} : \quad " \quad " \quad " \quad "$$

\therefore The fundamental frequency of $x_1(t)$ and $x_2(t)$ must be integer multiples of the fundamental frequency of $x(t)$

• If $x[n]$ is DT signal with period N
We want to represent $x[n]$ by

$$\hat{x}[n] = \sum_k A[k] \cdot \underbrace{e^{jk\Omega_0 n}}_{\text{sinusoid}}, \quad k \in \mathbb{Z}$$

Ω_0 : Fundamental frequency of $x[n]$

$\underbrace{k\Omega_0}_{\text{frequency of the } k^{\text{th}} \text{ sinusoid.}}$ } Integer multiple of Ω_0 ,

$\underbrace{A[k]}_{\text{Its weight}}$ | $e^{jk\Omega_0 n}$: k^{th} harmonic.

— Each of these sinusoids have the common period N

CT $x(t)$, $\omega_0 = \frac{2\pi}{T}$: fundamental frequency

We can represent $x(t)$ by

$$\hat{x}(t) = \sum_k A[k] e^{jk\omega_0 t}$$

$k\omega_0$: frequency of k^{th} sinusoid.

— $e^{jk\omega_0 t}$ is called the k^{th} harmonic
and $A[k]$ is its weight.

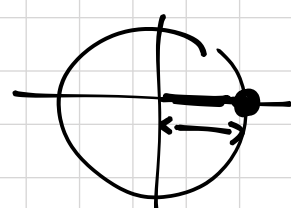
• How many harmonics should we use?

- DT Fourier Series

$e^{jk\Omega_0 n}$ is periodic if $N = \frac{2\pi}{\Omega_0}$

$$\begin{aligned} e^{j(\frac{N}{2}+k)\Omega_0 n} &= e^{j\frac{N}{2}\Omega_0 n} \cdot e^{jk\Omega_0 n} \\ &= \underline{e^{j2\pi n}} \cdot e^{jk\Omega_0 n} \\ &= 1 \cdot e^{jk\Omega_0 n} \end{aligned}$$

There are only N distinct complex sinusoids in the form of $e^{jk\Omega_0 n}$,
so we can only use N consecutive terms
• $k = 0 \dots (N-1)$



$$\begin{aligned} \hat{x}[n] &= \sum_{k=0}^{N-1} A[k] \cdot e^{jk\Omega_0 n} \\ &= \sum_{k=\langle N \rangle} A[k] \cdot e^{jk\Omega_0 n} \end{aligned}$$

The set of N consecutive terms is arbitrary, we can start from any value of k as long as we sum over \underline{N} values.

$$k = \underbrace{(m) \dots (N-1+m)}_N \quad m \in [0 \dots N]$$

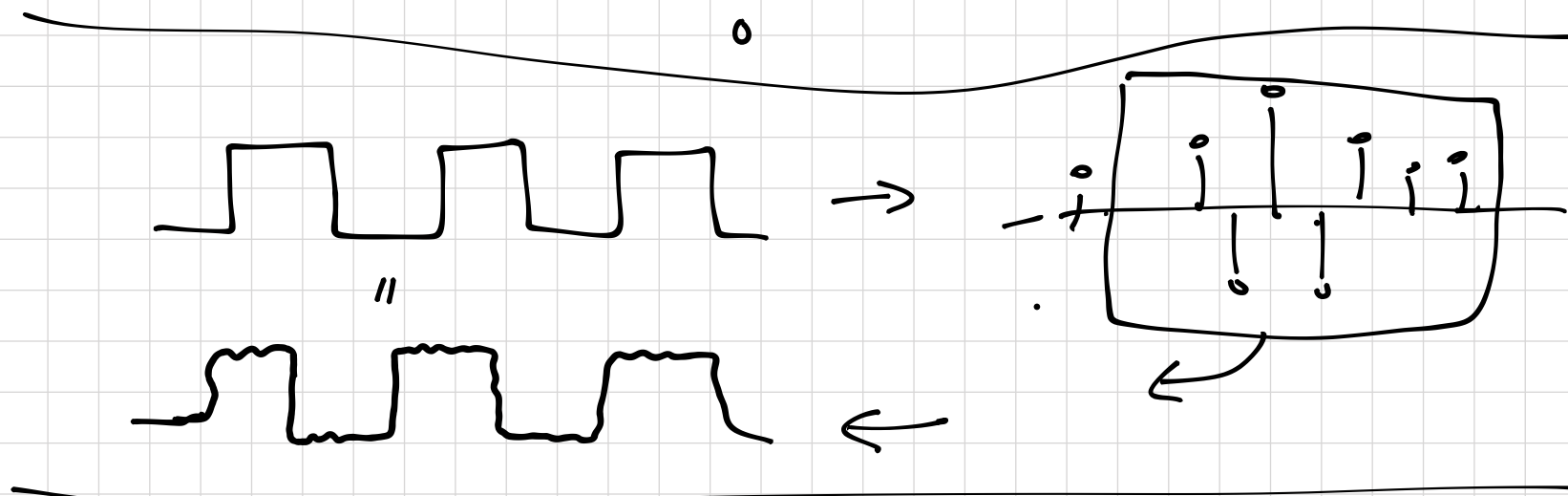
For FS (CT)

The CT complex sinusoids, $e^{jk\omega_0 t}$, with distinct frequencies $\underline{k\omega_0}$ are always distinct. So there could be an infinite number of distinct terms \therefore

$$\hat{x}(t) = \sum_{k=-\infty}^{+\infty} A[k] \cdot e^{jk\omega_0 t}$$

We want to find coefficients $A[k]$ such that $\hat{x}[n]$ and $\hat{x}(t)$ are good approximations to $x[n]$ and $x(t)$, respectively. We use a measure called MSE (mean-squared error) and we want to minimize this value.

$$\left. \begin{aligned} \text{DT} \quad \text{MSE} &= \frac{1}{N} \sum_{k=0}^{N-1} |x[n] - \hat{x}[n]|^2 \\ \text{CT} \quad \text{MSE} &= \frac{1}{T} \int_0^T |x(t) - \hat{x}(t)|^2 dt \end{aligned} \right\}$$



Nonperiodic Signals : FT

- There are no restrictions on the period of the complex sinusoids.

- The complex sinusoids will include a continuum of frequencies.

$$\left. \begin{aligned} \text{CT} : \quad \hat{x}(t) &= \frac{1}{2\pi} \int_{-\infty}^{+\infty} X(j\omega) e^{j\omega t} d\omega \\ \text{DT} : \quad \hat{x}[n] &= \frac{1}{2\pi} \int_{-\pi}^{\pi} X(e^{j\Omega}) \cdot e^{j\Omega n} d\Omega \end{aligned} \right\}$$