

Fourier Representation of Signals and LTI systems

- Read Introduction

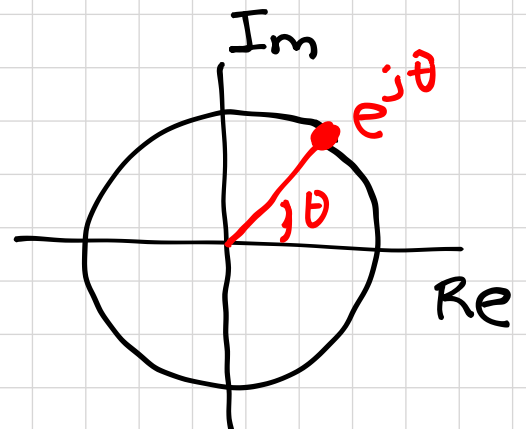
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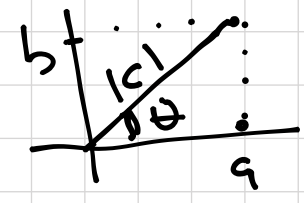
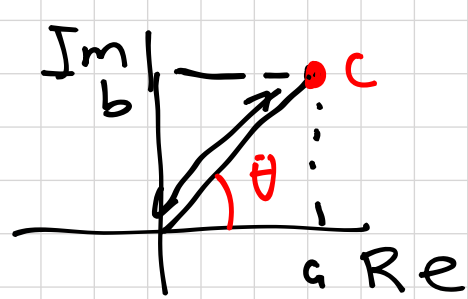
* Representing signals as weighted superposition of complex sinusoids.

* Euler's Formula

$$e^{j\theta} = \cos \theta + j \sin \theta$$



* Polar Form



$$\begin{cases} c = a + j b \\ |c| = \sqrt{a^2 + b^2} : \text{Magnitude} \\ \arg\{c\} = \theta = \arctan\left(\frac{b}{a}\right) : \text{Phase} \\ \rightarrow c = |c| e^{j \arg\{c\}} \quad \text{Polar form} \end{cases}$$

Complex Sinusoids and Frequency Response of LTI Systems.

DT

$$x[n] \rightarrow \mathcal{H} \rightarrow y[n] = \mathcal{H}\{x[n]\}$$

$h[n]$ is the impulse response
 $h[n] = \mathcal{H}\{\delta[n]\}$

For a given input, $x[n]$

$$y[n] = \mathcal{H}\{x[n]\} = x[n] * h[n] = h[n] * x[n]$$

$$y[n] = \sum_{k=-\infty}^{+\infty} h[k] x[n-k]$$

* Let $x[n] = e^{j\Omega n}$ Ω : frequency.

Then the output

$$y[n] = \mathcal{H}\{e^{j\Omega n}\} = \sum_{k=-\infty}^{+\infty} h[k] e^{j\Omega(n-k)}$$

$$= e^{j\Omega n} \sum_{k=-\infty}^{+\infty} h[k] \cdot e^{-j\Omega k}$$

$$\text{Let's define } \underbrace{H(e^{j\Omega})}_{\text{Frequency Response}} = \sum_{k=-\infty}^{+\infty} h[k] \cdot e^{-j\Omega k}$$

Frequency Response

- Not a function of time, but FREQUENCY.

$$y[n] = \underline{H(e^{j\Omega})} \cdot \underline{e^{j\Omega n}} = \mathcal{H}\{e^{j\Omega n}\}$$

$$\underbrace{e^{j\Omega n}} \rightarrow \mathcal{H} \rightarrow \underbrace{H(e^{j\Omega})}_{\text{Frequency Response}} \cdot \underbrace{e^{j\Omega n}}$$

CT

$$x(t) \rightarrow \mathcal{H} \rightarrow y(t) = \mathcal{H}\{x(t)\}$$

$h(t)$: impulse response

Let $x(t) = e^{j\omega t}$, then the output

$$y(t) = \int_{-\infty}^{+\infty} h(\tau) x(t-\tau) d\tau$$

$$= \int_{-\infty}^{+\infty} h(\tau) e^{j\omega(t-\tau)} d\tau$$

$$= e^{j\omega t} \int_{-\infty}^{+\infty} h(\tau) e^{-j\omega \tau} d\tau$$

$$\text{Frequency Response : } H(j\omega) = \int_{-\infty}^{+\infty} h(\tau) e^{-j\omega \tau} d\tau$$

$$y(t) = \mathcal{H}\{e^{j\omega t}\} = e^{j\omega t} \cdot H(j\omega)$$

$$e^{j\omega t} \rightarrow \mathcal{H} \rightarrow e^{j\omega t} \cdot \underline{H(j\omega)}$$

Polar form

$$\boxed{\exp(\cdot) = e^{\cdot}}$$

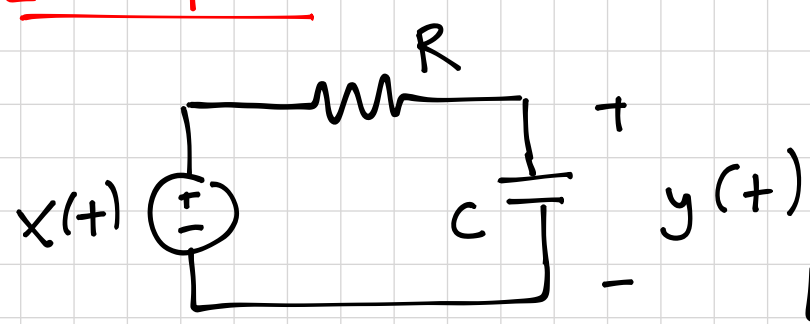
$$H(j\omega) = |H(j\omega)| \cdot \exp[j \cdot \arg\{H(j\omega)\}]$$

$$\underline{\mathcal{H}\{e^{j\omega t}\}} = H(j\omega) \cdot e^{j\omega t}$$

$$= \underline{|H(j\omega)|} \cdot \exp[j(\omega t + \underline{\arg\{H(j\omega)\}})]$$

$$\left[\begin{array}{l} |H(j\omega)| : \text{Magnitude Response} \\ \arg\{H(j\omega)\} : \text{Phase Response} \end{array} \right]$$

Example



$x(t)$: input voltage

$y(t)$: output voltage

let $\alpha = RC$

Impulse Response

$$h(t) = \frac{1}{\alpha} \cdot e^{-t/\alpha} u(t)$$

Frequency Response

$$H(j\omega) = \int_{-\infty}^{+\infty} h(z) e^{-j\omega z} dz$$

$$= \int_{-\infty}^{+\infty} \frac{1}{\alpha} e^{-z/\alpha} u(z) e^{-j\omega z} dz$$

$$= \frac{1}{\alpha} \int_0^{\infty} \exp\left\{-\left(j\omega + \frac{1}{\alpha}\right)z\right\} dz$$

$$= \frac{1}{\alpha} \cdot \frac{-1}{j\omega + 1/\alpha} \cdot \left[\exp\left\{-\left(j\omega + \frac{1}{\alpha}\right)z\right\} \right]_0^{\infty}$$

$$= \frac{1}{\alpha} \cdot \frac{-1}{j\omega + 1/\alpha} [0 - 1]$$

$$\boxed{H(j\omega) = \frac{1/\alpha}{j\omega + 1/\alpha}}$$

magnitude Response $|H(j\omega)|$

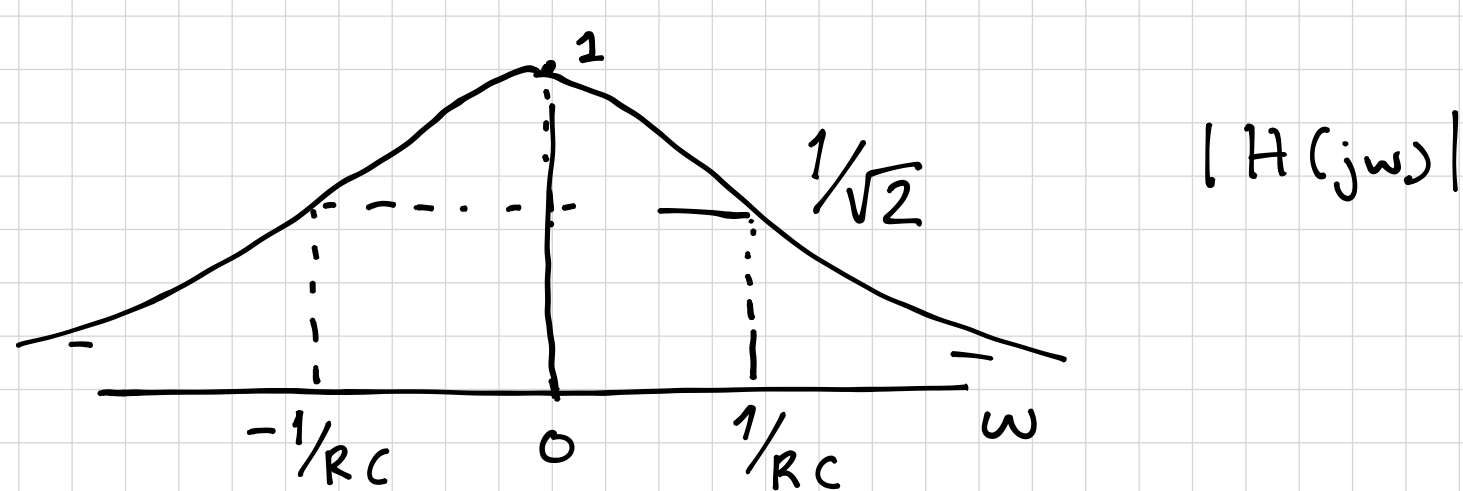
$$H(j\omega) = \frac{1/\alpha}{1/\alpha + j\omega} \cdot \left[\frac{1/\alpha - j\omega}{1/\alpha - j\omega} \right] = \frac{1}{\alpha} \cdot \frac{1/\alpha - j\omega}{\frac{1}{\alpha^2} - (-1 \cdot \omega^2)}$$

$$= \frac{1}{\alpha} \cdot \frac{1/\alpha - j\omega}{\left(\frac{1}{\alpha}\right)^2 + \omega^2}$$

$$= \frac{1/\alpha}{\left(\frac{1}{\alpha}\right)^2 + \omega^2} \left(\frac{1}{\alpha} - j\omega \right)$$

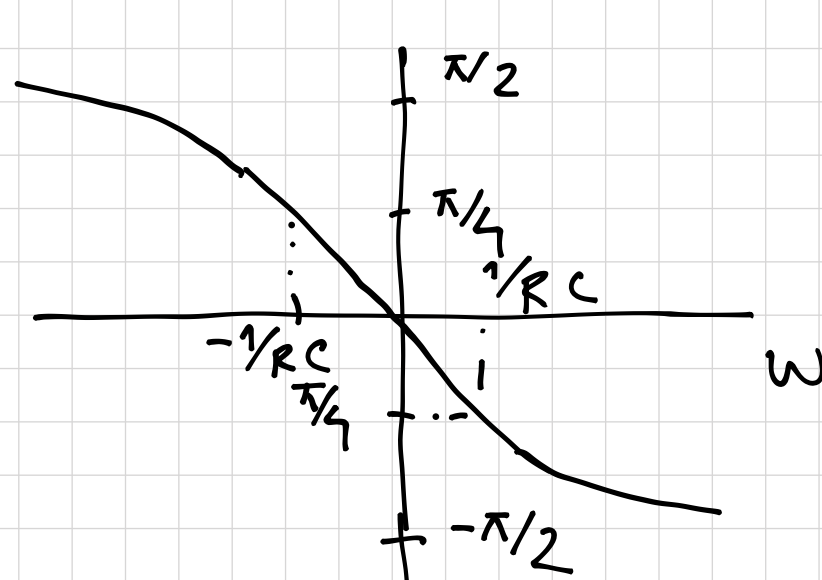
$$|H(j\omega)| = \frac{1/\alpha}{\left(\frac{1}{\alpha}\right)^2 + \omega^2} \sqrt{\left(\frac{1}{\alpha}\right)^2 + \omega^2}$$

$$= \frac{1/\alpha}{\sqrt{\frac{1}{\alpha^2} + \omega^2}} = \frac{1/RC}{\sqrt{\frac{1}{(RC)^2} + \omega^2}}$$



Phase Response

$$\arg\{H(j\omega)\} = \arctan\left(\frac{-\omega}{1/\alpha}\right) = -\arctan(\omega RC)$$

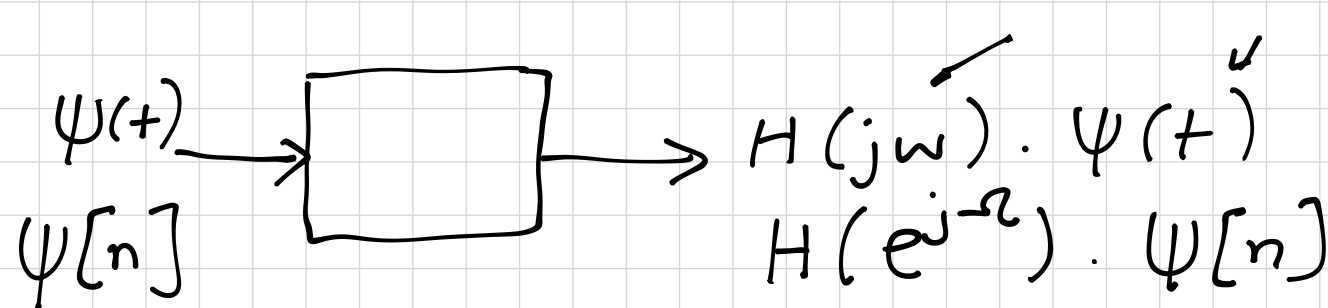


Eigenfunction

$\psi(t) = e^{j\omega t}$ is an eigenfunction of

the LTI system associated with the eigenvalue $\lambda = H(j\omega)$ because ψ satisfies the eigenvalue problem described by

$$\mathcal{H}\{\psi(t)\} = \lambda \cdot \psi(t)$$



* If \underline{e}_k is an eigenvector of a matrix \underline{A} with eigenvalue λ_k then $\underline{A}\underline{e}_k = \lambda_k \underline{e}_k$ */

→ By representing arbitrary signals as weighted superposition of eigenfunctions we transform convolution operation to multiplication ←

→ For example if $x(t)$ is a weighted sum of M complex sinusoids

$$x(t) = \sum_{k=1}^M a_k \cdot \exp\{j\omega_k t\}$$

If $e^{j\omega_k t}$ is eigenfunction of the system with eigenvalue $H(j\omega_k)$ then each term in $x(t)$, $a_k \cdot e^{j\omega_k t}$ produces an output $a_k H(j\omega_k) e^{j\omega_k t}$

$$\text{Output } y(t) = \sum_{k=1}^M a_k H(j\omega_k) e^{j\omega_k t}$$

- ex

$$y(t) = x(t-3)$$

If the input to this system is $x(t) = \underline{e^{j2t}}$ then the output

$$y(t) = e^{j2(t-3)} = \underline{e^{j2t}} \cdot \underline{e^{-j6}}$$

e^{j2t} is the eigenfunction associated with the eigenvalue $\underline{H(j2) = e^{-j6}}$

- we can show this by

$$y(t) = \underline{H(j\omega)} \cdot \underline{e^{j2t}}$$

Impulse Response $h(t) = \delta(t-3)$

$$\begin{aligned} \underline{H(j\omega)} &= \int_{-\infty}^{+\infty} h(z) \cdot e^{-j\omega z} dz \\ \text{Frequency response} &= \int_{-\infty}^{+\infty} \delta(t-3) e^{-j\omega z} dz \\ &= e^{-3j\omega} \\ \underline{H(j2)} &= e^{-3j2} = \underline{e^{-j6}} // \end{aligned}$$