

## General Aptitude

**Q. No. 1 – 5 Carry One Mark Each**

1. An apple costs Rs. 10. An onion costs Rs. 8.  
Select the most suitable sentence with respect to grammar and usage.

(A) The price of an apple is greater than an onion.  
(B) The price of an apple is more than onion.  
(C) The price of an apple is greater than that of an onion.  
(D) Apples are more costlier than onions.

**Key:** (C)

2. The Buddha said, "Holding on to anger is like grasping a hot coal with the intent of throwing it at someone else; you are the one who gets burnt."

Select the word below which is closest in meaning to the word underlined above.

(A) burning (B) igniting (C) clutching (D) flinging

**Key:** (C)

3. M has a son Q and a daughter R. He has no other children. E is the mother of P and daughter-in-law of M. How is P related to M?

(A) P is the son-in-law of M. (B) P is the grandchild of M.  
(C) P is the daughter-in law of M. (D) P is the grandfather of M.

**Key:** (B)

4. The number that least fits this set: (324, 441, 97 and 64) is \_\_\_\_\_.

(A) 324 (B) 441 (C) 97 (D) 64

**Key:** (C)

**Exp:**  $324 = 18^2$ ;  $441 = 21^2$ ;  $64 = 8^2$  but  $97 \neq x^2$  for any positive integer

i.e. 97 is odd man out

5. It takes 10 s and 15 s, respectively, for two trains travelling at different constant speeds to completely pass a telegraph post. The length of the first train is 120 m and that of the second train is 150 m. The magnitude of the difference in the speeds of the two trains (in m/s) is \_\_\_\_\_.

(A) 2.0 (B) 10.0 (C) 12.0 (D) 22.0

**Key:** (A)

**Exp:**  $\text{Speed} = \frac{\text{length}}{\text{time}} \Rightarrow \text{length} = \text{speed} \times \text{time}$

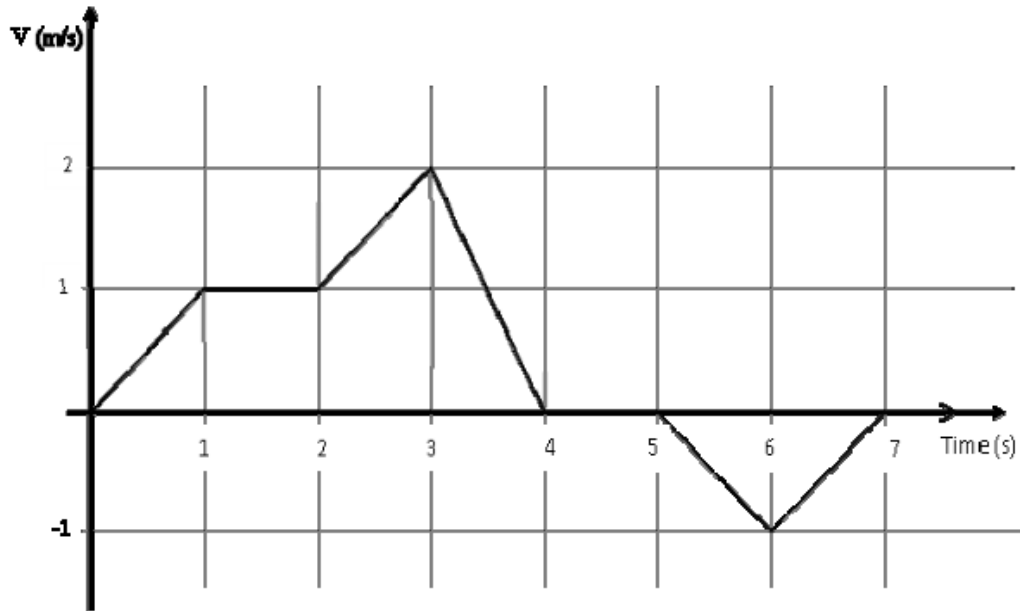
$$120 = 10 \times s_1 \Rightarrow s_1 = 12$$

$$150 = 15 \times s_2 \Rightarrow s_2 = 10$$

$$|s_1 - s_2| = 2$$

**Q. No. 6 – 10 Carry Two Marks Each**

6. The velocity  $V$  of a vehicle along a straight line is measured in m/s and plotted as shown with respect to time in seconds. At the end of the 7 seconds, how much will the odometer reading increase by (in m)?



(A) 0

(B) 3

(C) 4

(D) 5

**Key:** (D)

7. The overwhelming number of people infected with rabies in India has been flagged by the World Health Organization as a source of concern. It is estimated that inoculating 70% of pets and stray dogs against rabies can lead to a significant reduction in the number of people infected with rabies.

Which of the following can be logically inferred from the above sentences?

- (A) The number of people in India infected with rabies is high.  
(B) The number of people in other parts of the world who are infected with rabies is low.  
(C) Rabies can be eradicated in India by vaccinating 70% of stray dogs  
(D) Stray dogs are the main sources of rabies worldwide.

**Key:** (A)

8. A flat is shared by four first year undergraduate students. They agreed to allow the oldest of them to enjoy some extra space in the flat. Manu is two months older than Sravan, who is three months younger than Trideep. Pavan is one month older than Sravan. Who should occupy the extra space in the flat?

- (A) Manu                      (B) Sravan                      (C) Trideep                      (D) Pavan

**Key:** (C)

9. Find the area bounded by the lines  $3x+2y=14$ ,  $2x-3y=5$  in the first quadrant.

- (A) 14.95                      (B) 15.25                      (C) 15.70                      (D) 20.35

**Key:** (B)

**Exp**  $A = \left(\frac{14}{3}, 0\right)$

$B = (0, 7)$

$C = \left(\frac{5}{2}, 0\right)$

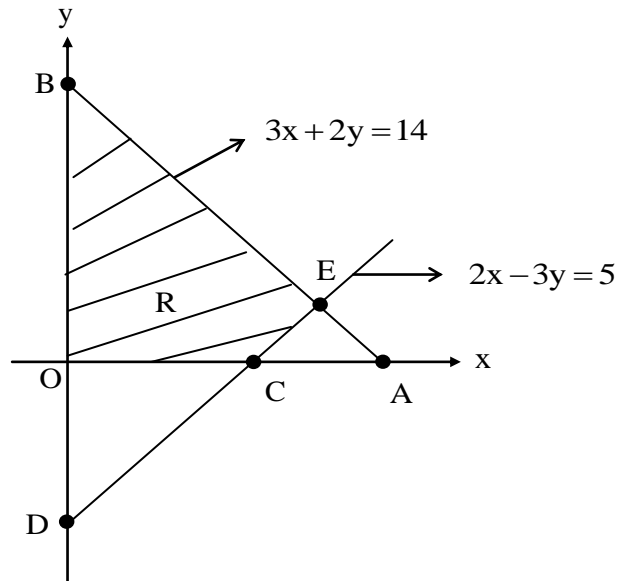
$D = \left(0, \frac{-5}{3}\right)$

$E = (4, 1)$

Required area is area of

$\Delta OAB$  – area of  $\Delta CEA$

$$= \frac{1}{2} \left( \frac{14}{3} \right) (7) - \frac{1}{2} \left( \frac{13}{6} \right) (1) = 15.25 \text{ sq.units}$$



10. A straight line is fit to a data set  $(\ln x, y)$ . This line intercepts the abscissa at  $\ln x = 0.1$  and has a slope of  $-0.02$ . What is the value of  $y$  at  $x = 5$  from the fit?

(A)  $-0.030$

(B)  $-0.014$

(C)  $0.014$

(D)  $0.030$

**Key:** (A)

**Exp:**  $y = a + bx$ , where  $x = \ln x$  and  $\frac{-a}{b} = 0.1$ ,  $b = -0.02$

$$= a - 0.02(x) \Rightarrow a = 0.002$$

$$= 0.002 - 0.02(x)$$

$$\text{at } x = 5, y = 0.002 - 0.02(1.609) = -0.03018$$

$$\approx -0.030$$

### Instrumentation Engineering

**Q. No. 1 – 25 Carry One Mark Each**

1. A straight line of the form  $y = mx + c$  passes through the origin and the point  $(x, y) = (2, 6)$ . The value of  $m$  is \_\_\_\_\_.

**Key:** 3

**Exp:**  $y = mx + c$  passing through  $(0, 0) \Rightarrow 0 = 0 + c \Rightarrow c = 0$

$y = mx + c$  passing through  $(2, 6) \Rightarrow 6 = 2m \Rightarrow m = 3$

2.  $\lim_{n \rightarrow \infty} (\sqrt{n^2 + n} - \sqrt{n^2 + 1})$  is \_\_\_\_\_.

**Key:** 0.5

**Exp:**  $\lim_{n \rightarrow \infty} (\sqrt{n^2 + n} - \sqrt{n^2 + 1}) \times \frac{\sqrt{n^2 + n} + \sqrt{n^2 + 1}}{\sqrt{n^2 + n} + \sqrt{n^2 + 1}}$

$$= \lim_{n \rightarrow \infty} \frac{n^2 + n - n^2 - 1}{\sqrt{n^2 + n} + \sqrt{n^2 + 1}} = \lim_{n \rightarrow \infty} \frac{(1 - \frac{1}{n})}{\sqrt{1 + \frac{1}{n}} + \sqrt{1 + \frac{1}{n^2}}} = \frac{1}{2}$$

3. A voltage  $V_1$  is measured 100 times and its average and standard deviation are 100 V and 1.5 V respectively. A second voltage  $V_2$ , which is independent of  $V_1$ , is measured 200 times and its average and standard deviation are 150 V and 2 V respectively.  $V_3$  is computed as:  $V_3 = V_1 + V_2$ . Then the standard deviation of  $V_3$  in **volt** is \_\_\_\_\_.

**Key:** 2.5

**Exp:**  $\text{Var}(V_3) = \text{Var}(V_1 + V_2) = \text{Var}(V_1) + \text{Var}(V_2) = 2.25 + 4 = 6.25$

Standard deviation of  $V_3 = +\sqrt{\text{Var}(V_3)} = +\sqrt{6.25} = 2.5$

4. The vector that is **NOT** perpendicular to the vectors  $(i + j + k)$  and  $(i + 2j + 3k)$  is \_\_\_\_.

(A)  $(i - 2j + k)$  (B)  $(-i + 2j - k)$  (C)  $(0i + 0j + 0k)$  (D)  $(4i + 3j + 5k)$

**Key:** (D)

**Exp:** We know that if  $\vec{a} \cdot \vec{b} = 0$  then  $\vec{a}$  and  $\vec{b}$  are perpendicular

Verify options (a), (b), (c) are perpendicular Option (d) is not perpendicular

5. In the neighborhood of  $z = 1$ , the function  $f(z)$  has a power series expansion of the form  $f(z) = 1 + (1-z) + (1-z)^2 + \dots$

Then  $f(z)$  is

(A)  $\frac{1}{z}$  (B)  $\frac{-1}{z-2}$  (C)  $\frac{z-1}{z+1}$  (D)  $\frac{1}{2z-1}$

**Key:** (A)

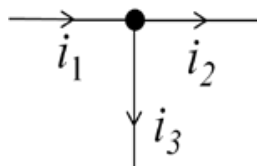
**Exp:**  $1 + (1-z) + (1-z)^2 + \dots$

put  $w = 1 - z$ , then the given series becomes

$= 1 + w + w^2 + w^3 + \dots$

$= \frac{1}{1-w} = \frac{1}{1-(1-z)} = \frac{1}{z}$

6. Three currents  $i_1$ ,  $i_2$  and  $i_3$  meet at a node as shown in the figure below. If  $i_1 = 3\cos(\omega t)$  ampere,  $i_2 = 4\sin(\omega t)$  ampere and  $i_3 = I_3 \cos(\omega t + \theta)$  ampere, the value of  $I_3$  in **ampere** is \_\_\_\_.



**Key:** 5

**Exp:** By KCL  $i_1(t) = i_2(t) + i_3(t)$

$$\Rightarrow i_3(t) = i_1(t) - i_2(t)$$

By phasor  $I_3 = \bar{I}_1 - \bar{I}_2$

$$= [3 \angle 0] - [4 \angle -90^\circ] = 5 \angle 53.13$$

$$\Rightarrow i_3(t) = 5 \cos(\omega t + 53.13)$$

So by comparison  $I_3 = 5$ .

7. An air cored coil has a Q of 5 at a frequency of 100 kHz. The Q of the coil at 20 kHz (neglecting skin effect) will be \_\_\_\_\_.

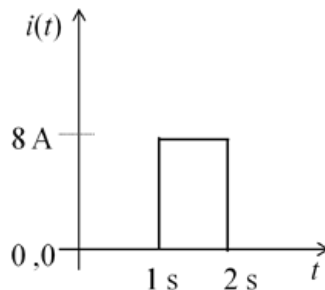
**Key:** 1

**Exp:**  $Q = \frac{\omega L}{R}$

at 100kHz,  $5 = \frac{\omega L}{R} = \frac{2\pi \times 100 \times 10^3 \times L}{R} \Rightarrow \frac{L}{R} = 7.9577 \times 10^{-6}$

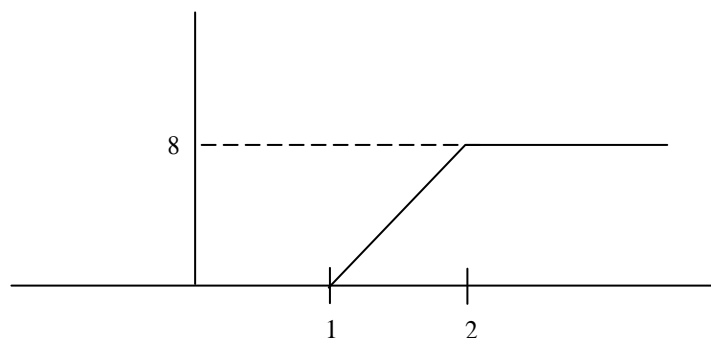
at 20kHz,  $Q = \frac{\omega L}{R} = 2\pi \times 20 \times 10^3 \times 7.9577 \times 10^{-6} = 1$

8. A current  $i(t)$  shown in the figure below is passed through a 1 F capacitor that had zero initial charge. The voltage across the capacitor for  $t > 2$  s in volt is \_\_\_\_\_.



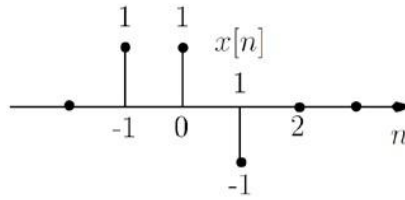
**Key:** 8

**Exp:**  $V_c = \frac{1}{C} \int i dt = \frac{1}{1} \int 8[u(t-1) - u(t-2)] dt = 8[r(t-1) - r(t-2)]$



$V_c(t > 2) = 8V$

9. The signal  $x[n]$  shown in the figure below is convolved with itself to get  $y[n]$ . The value of  $y[-1]$  is \_\_\_\_\_.



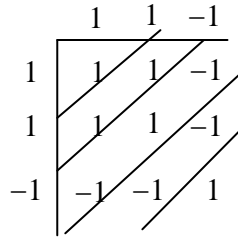
**Key:** 2

**Exp:**  $x(n) = \{1, 1, -1\} \quad (-1 \leq n \leq 1)$

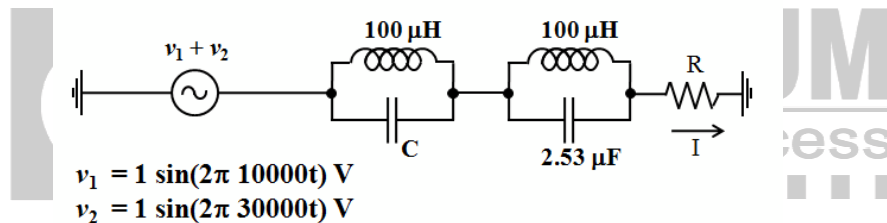
↑

$$y(n) = x(n) * x(n) = \{1, 2, -1, -2, 1\}$$

$$y(-1) = 2$$



10. In the circuit shown below  $(v_1 + v_2) = [1 \sin(2\pi 10000t) + 1 \sin(2\pi 30000t)]$  V. The RMS value of the current through the resistor R will be minimum if the value of the capacitor C in **microfarad** is \_\_\_\_\_.



**Key:** 0.28 to 0.283

11. If  $X(s)$ , the Laplace transform of signal  $x(t)$  is given by  $X(s) = \frac{(s+2)}{(s+1)(s+3)}$ , then the value of  $x(t)$  as  $t \rightarrow \infty$  is \_\_\_\_\_.

**Key:** 0

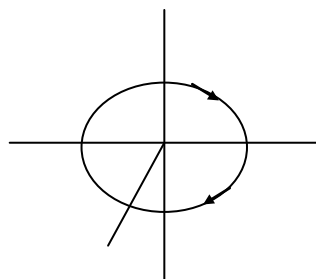
**Exp:**  $\lim_{t \rightarrow \infty} f(t) = \lim_{s \rightarrow 0} s \frac{s+2}{(s+1)(s+3)} = 0$

12. The number of times the Nyquist plot of  $G(s) = \frac{s-1}{s+1}$  will encircle the origin clockwise is \_\_\_\_\_.

**Key:** 1

**Exp:**  $G(s) = \frac{s-1}{s+1}$

$$|G(s)| = \frac{\sqrt{1+\omega^2}}{\sqrt{1+\omega^2}} = 1$$



$$|G(s)| = 180 - 2 \tan^{-1}(\omega)$$

$$G(s) = 1|180^\circ - 2 \tan^{-1}(\omega)$$

$$\rightarrow G(0) = 1|180^\circ$$

$$G(\infty) = 1|0^\circ$$

$$G(1) = 1|90^\circ$$

Using  $G(0)$ ,  $G(1)$ ,  $G(\infty)$  information, the plot will look like

$\rightarrow$  Hence it encircle the origin only 1 time in clockwise direction

13. The value of  $a_0$  which will ensure that the polynomial  $s^3 + 3s^2 + 2s + a_0$  has roots on the left half of the  $s$ -plane is

(A) 11

(B) 9

(C) 7

(D) 5

**Key:** (D)

**Exp:**

$s^3$	1	2
$s^2$	3	$a_0$
$s^1$	$\frac{6-a_0}{3}$	
$s^0$	$a_0$	

For Stability  $\frac{6-a_0}{3} > 0 \Rightarrow a_0 < 6$



14. The input  $i(t) = 2 \sin(3t + \pi)$  is applied to a system whose transfer function  $G(s) = \frac{8}{(s+10)^2}$ .

The amplitude of the output of the system is\_.

**Key:** 0.1467

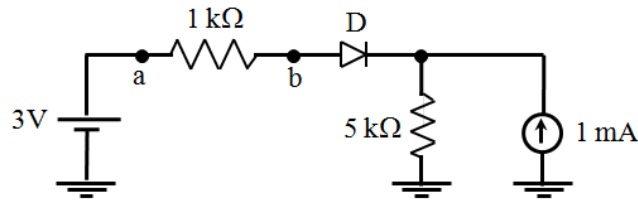
**Exp:**  $G(s) = \frac{8}{(s+10)^2} = \frac{8}{s^2 + 20s + 100} = \frac{8}{(100 - \omega^2) + j(20\omega)}$

$$G(s) = \frac{8}{\sqrt{(100 - \omega^2)^2 + 3600}} \angle -\tan^{-1} \frac{60}{91} = \frac{8}{109} \angle -33.4^\circ$$

$$y(t) = \left( 2 \times \frac{8}{109} \right) \sin(3t + \pi - 33.4^\circ) = 0.1467 \sin(3t + 146.6^\circ)$$

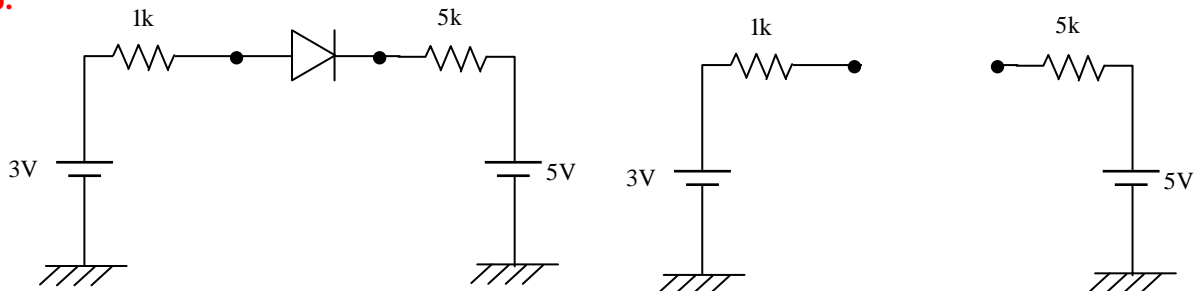
So amplitude is 0.1467.

15. The diode D used in the circuit below is ideal. The voltage drop  $V_{ab}$  across the  $1\text{ k}\Omega$  resistor in volt is \_\_\_\_\_.



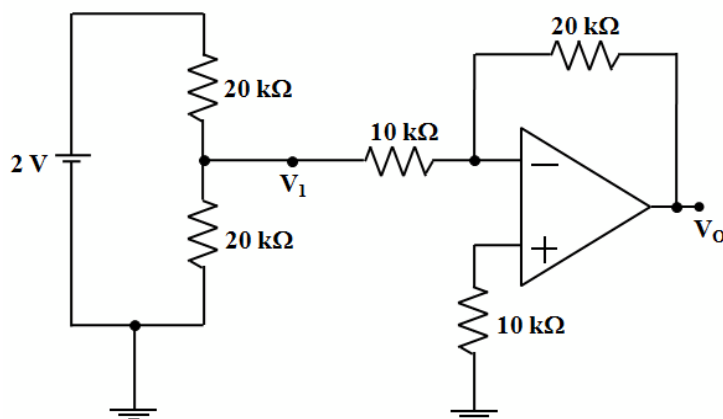
**Key:** 0

**Exp:**



Since  $V_P < V_N$  diode is open circuit and no current flow through  $1\text{ k}\Omega$ , So  $V_{1\text{k}\Omega} = 0\text{V}$ .

16. In the circuit given below, the op-amp is ideal. The output voltage  $V_O$  in volt is \_\_\_\_\_.



**Key:** -1

**Exp:**

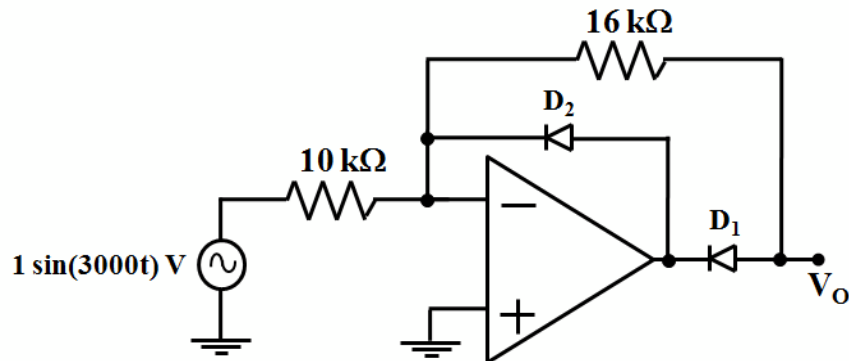
$$\frac{V_x - 2}{20} + \frac{V_x}{20} + \frac{V_x}{10} = 0 \Rightarrow V_x \left( \frac{1}{20} + \frac{1}{20} + \frac{1}{10} \right) = \frac{2}{20} \Rightarrow V_x (1+1+2) = 2$$

$$\Rightarrow V_x = \frac{1}{2}$$

$$V_o = \frac{-R_f}{R} V_x = -\left( \frac{20}{10} \right) \frac{1}{2} = -1\text{V}$$



17. In the circuit given below, the diodes  $D_1$  and  $D_2$  have a forward voltage drop of 0.6 V. The op-amp used is ideal. The **magnitude** of the negative peak value of the output  $V_O$  in **volt** is \_\_\_\_\_.



**Key:** 1.6

18. The Boolean expression  $XY + (X' + Y')Z$  is equivalent to

(A)  $XYZ' + X'Y'Z$  (B)  $X'Y'Z' + XYZ$  (C)  $(X+Z)(Y+Z)$  (D)  $(X'+Y')(Y'+Z)$

**Key:** (C)

**Exp:**  $F = XY + \bar{X}Z + \bar{Y}Z$

The min term of F are

X	Y	-	$\bar{X}$	-	Z	-	$\bar{Y}$	Z
= 1	1	0	0	0	1	0	0	1
1	1	1	0	1	1	1	0	1

$$F = \sum m(1, 3, 5, 6, 7)$$

If we go for option B

$$F = (x+z)(y+z) = z + xy$$

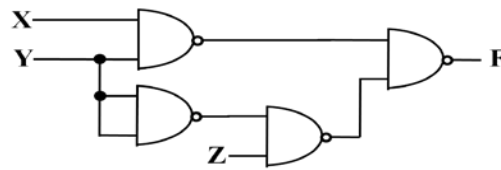
→ Its minterms are

-	-	Z		X	Y	-
0	0	1		1	1	0
0	1	1		1	1	1
1	0	1				
1	1					

$$F = \sum m(1, 3, 5, 6, 7)$$

Since minterms are same these two functions are equal

19. In the digital circuit given below, F is



- (A)  $XY + Y\bar{Z}$       (B)  $XY + \bar{Y}Z$       (C)  $\bar{X}\bar{Y} + Y\bar{Z}$       (D)  $XZ + \bar{Y}$

**Key:** (B)

**Exp:** From the circuit

$$F = \overline{XY \cdot YZ} = XY + \bar{Y}Z$$

20. A 3 ½ digit DMM has an accuracy specification of  $\pm 1\%$  of full scale (accuracy class 1). A reading of 100.0 mA is obtained on its 200 mA full scale range. The worst case error in the reading in **milliampere** is  $\pm$  \_\_\_\_\_.

**Key:** 2

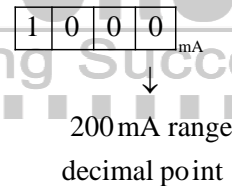
**Exp:** Since all information are given in mA unit assume the scale in mA unit.

→ Since it is given that error is  $\pm 1\%$  of full scale

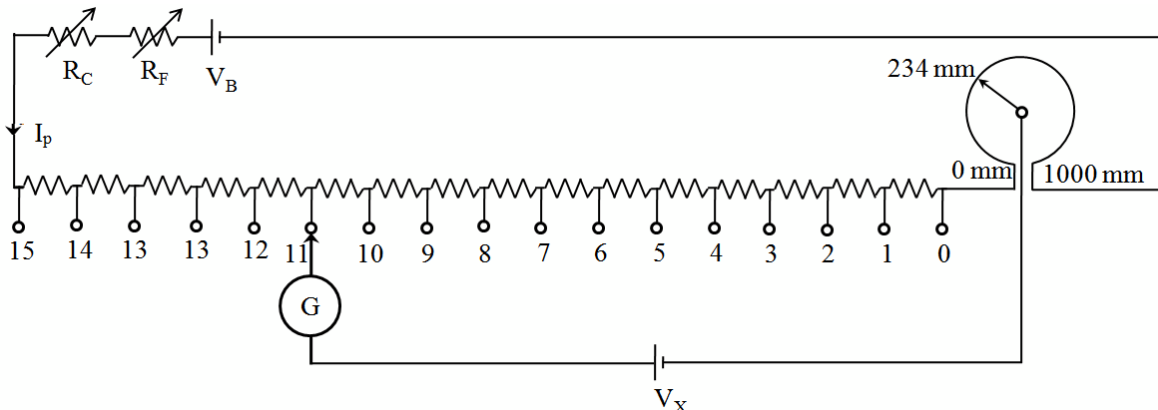
So, error =  $\pm 1\%$  of 200mA =  $\pm 2$ mA.

→ So if it measures 100mA then the reading will be in the range  $(100 \pm 2)$ mA

→ the worst source error is  $\pm 2$ mA.



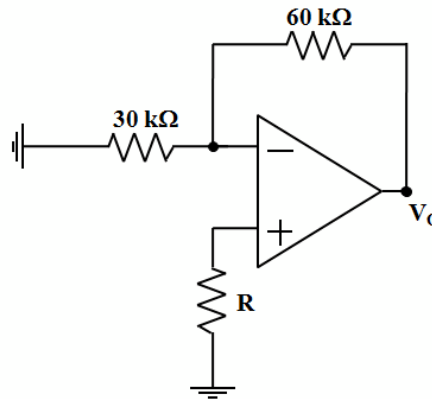
21. A dc potentiometer, shown in figure below, is made by connecting fifteen  $10\Omega$  resistors and a  $10\Omega$  slide wire of length 1000 mm in series. The potentiometer is standardized with the current  $I_p = 10.0000$  mA. Balance for an unknown voltage is obtained when the dial is in position 11 (11 numbers of the fixed  $10\Omega$  resistor are included) and the slide wire is on the 234th mm position. The unknown voltage (up to four decimal places) in **volt** is \_\_\_\_\_.



**Key:** 1.1234

**Exp:**  $V_x = (10 \times 10^{-3}) \left[ 11 \times 10 + 234 \text{mm} \times \frac{10}{1000 \text{mm}} \right] = 1.1234 \text{V}$

22. In the circuit given below, each input terminal of the op-amp draws a bias current of 10 nA. The effect due to these input bias currents on the output voltage  $V_o$  will be zero, if the value of R chosen in **kilo-ohm** is \_\_\_\_\_.



**Key:** 20

**Exp:**  $R = (60/30) = 20 \text{k}\Omega$

23. A piezo-electric type pressure sensor has a sensitivity of 1 mV/kPa and a bandwidth of 300 Hz to 300 kHz. For a constant (dc) pressure of 100 kPa, the steady state output of the sensor in **millivolt** is \_\_\_\_\_.

**Key:** 0

**Exp:** Piezoelectric transducer produces output for changing input, but here input is constant, So output is 0.

24. The signal  $m(t) = \cos(\omega_m t)$  is SSB (single side-band) modulated with a carrier  $\cos(\omega_c t)$  to get  $s(t)$ . The signal obtained by passing  $s(t)$  through an ideal envelope detector is

- (A)  $\cos(\omega_m t)$  (B)  $\sin(\omega_m t)$   
(C)  $\cos(\omega_m t) + \sin(\omega_m t)$  (D) 1

**Key:** (D)

**Exp:**  $\delta(t) = X_{SSB}(t)$

$X_{SSB}(t) = m(t) \cos \omega_c t \pm \hat{m}(t) \sin \omega_c t$

$\hat{m}(t) = \text{hilbert transform of } m(t)$

$m(t) = \cos \omega_m t$

$\hat{m}(t) = \sin \omega_m t$

$\delta(t) = \cos \omega_m t \cos \omega_c t + \sin \omega_m t \sin \omega_c t = \cos(\omega_c \pm \omega_m) t$

$|\delta(t)| = 1$

25. Let  $s(t) = \text{rect}\left(\frac{t-3}{3}\right)$  be a signal passed through an AWGN (additive white Gaussian noise) channel with noise power spectral density (PSD)  $\frac{N_0}{2}$  to get  $v(t)$ . If  $v(t)$  is passed through a matched-filter that is matched to  $s(t)$ , then output signal-to noise ratio (SNR) of the matched-filter
- (A)  $\frac{1}{N_0}$                       (B)  $\frac{2}{N_0}$                       (C)  $\frac{3}{N_0}$                       (D)  $\frac{4}{N_0}$

**Key:** (B)

**Q. No. 26 – 55 carry Two Marks Each**

26. Let  $f: [-1, 1] \rightarrow \mathbb{R}$ , where  $f(x) = 2x^3 - x^4 - 10$ . The minimum value of  $f(x)$  is \_\_\_\_\_.

**Key:** -13

**Exp:**  $f(x) = 2x^3 - x^4 - 10$

$$f'(x) = 6x^2 - 4x^3$$

$$f''(x) = 12x - 12x^2$$

$$f'(x) = 0 \Rightarrow 6x^2 - 4x^3 = 0$$

$$x^2(6 - 4x) = 0$$

$$\Rightarrow x = 0, x = \frac{3}{2} \text{ are stationary points}$$

$$f''(0) = 0 \text{ neither maxima nor minima at } x = 0$$

$$x = \frac{3}{2} \notin [-1, 1]$$

$$\therefore \text{minimum of } f(x) = \text{minimum} \{f(-1), f(1)\}$$

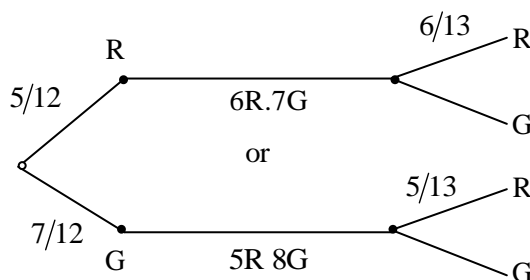
$$= \text{minimum} \{-13, -9\} = -13$$

27. An urn contains 5 red and 7 green balls. A ball is drawn at random and its colour is noted. The ball is placed back into the urn along with another ball of the same colour. The probability of getting a red ball in the next draw is

- (A)  $\frac{65}{156}$                       (B)  $\frac{67}{156}$                       (C)  $\frac{79}{156}$                       (D)  $\frac{89}{156}$

**Key:** (A)

**Exp:**  $\frac{5}{12} \times \frac{6}{13} + \frac{7}{12} \times \frac{5}{13} = \frac{65}{12 \times 13} = \frac{65}{156}$



28. Consider the matrix  $A = \begin{pmatrix} 2 & 1 & 1 \\ 2 & 3 & 4 \\ -1 & -1 & -2 \end{pmatrix}$  whose eigenvalues are 1, -1 and 3. Then Trace of  $(A^3 - 3A^2)$  is \_\_\_\_\_.

**Key:** -6

**Exp:** eigen values of  $A^3 - 3A^2$  corresponding to -1, 1, 3 are -4, -2, 0 respectively.

$$\therefore \text{Trace of } (A^3 - 3A^2) = -4 - 2 + 0 = -6$$

29. The relationship between the force  $f(t)$  and the displacement  $x(t)$  of a spring-mass system (with a mass  $M$ , viscous damping  $D$  and spring constant  $K$ ) is

$$M \frac{d^2 x(t)}{dt^2} + D \frac{dx(t)}{dt} + Kx(t) = f(t).$$

$X(s)$  and  $F(s)$  are the Laplace transforms of  $x(t)$  and  $f(t)$  respectively. With  $M = 0.1$ ,  $D = 2$ ,  $K = 10$  in

appropriate units, the transfer function  $G(s) = \frac{X(s)}{F(s)}$  is

(A)  $\frac{10}{s^2 + 20s + 100}$

(B)  $\frac{1}{s^2 + 20s + 100}$

(C)  $\frac{10s^2}{s^2 + 20s + 100}$

(D)  $\frac{s}{s^2 + 20s + 100}$

**Key:** (A)

**Exp:**  $M \frac{d^2 x(t)}{dt^2} + D \frac{dx(t)}{dt} + kx(t) = f(t)$

Taking Laplace both side of above we get

$$\frac{X(s)}{F(s)} = \frac{1}{Ms^2 + DS + k}$$

Using  $M=0.1$ ,  $D=2$  and  $k=10$  then the function becomes

$$\frac{1}{0.1s^2 + 2s + 10} = \frac{10}{s^2 + 20s + 100}$$

30. The value of the integral  $\frac{1}{2\pi j} \oint_C \frac{z^2 + 1}{z^2 - 1} dz$  where  $z$  is a complex number and  $C$  is a unit circle with center at  $1+0j$  in the complex plane is \_\_\_\_\_.

**Key:** 1

**Exp:** Given  $\frac{1}{2\pi j} \oint_C \frac{z^2 + 1}{z^2 - 1} dz = \frac{1}{2\pi j} \oint_C \frac{z^2 + 1}{(z-1)(z+1)} dz$

Poles are  $z = 1, -1$

Given C is  $(x-1)^2 + y^2 = 1$

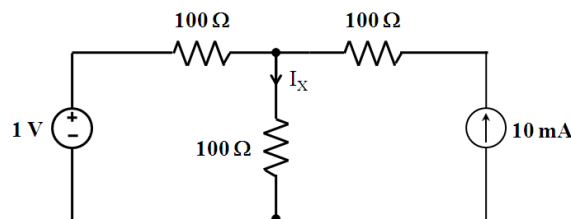
Clearly 1 lies inside of C and -1 outside of C

$$\left[ \operatorname{Res} f(z) \right]_{z=1} = \lim_{z \rightarrow 1} (z-1) \frac{z^2+1}{(z-1)(z+1)} = 1$$

$\therefore$  By Cauchy's Residue theorem

$$\frac{1}{2\pi j} \oint_C \frac{z^2+1}{z^2-1} dz = \frac{1}{2\pi j} \times 2\pi j \times 1 = 1$$

31. The current  $I_X$  in the circuit given below in **milliampere** is \_\_\_\_\_.



**Key:** 10

**Exp:**  $(I_X - 10)\text{mA}$

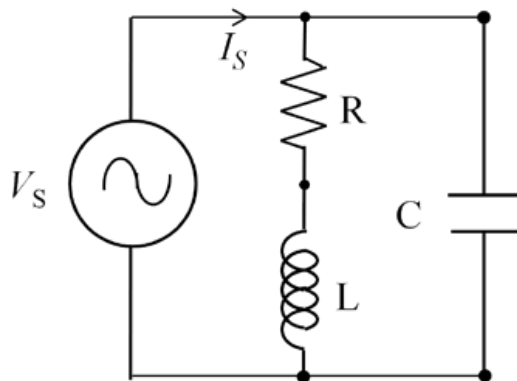
Diagram showing current flow: A horizontal line with an arrow pointing right. A vertical line goes down from it, labeled  $I_X \text{ (mA)}$ . Another horizontal line continues to the right, with an arrow pointing left towards the junction, labeled  $10\text{mA}$ .

By KVL,  $1 = [100(I_X - 10) + 100I_X]10^{-3}$

$$1000 = 200I_X - 1000$$

$$I_X = 10\text{mA}$$

32. In the circuit shown below,  $V_S = 101\angle 0^\circ\text{V}$ ,  $R = 10\Omega$  and  $\omega L = 100\Omega$ . The current  $I_S$  is in phase with  $V_S$ . The magnitude of  $I_S$  in **milliampere** is \_\_\_\_\_.



**Key:** 100

**Exp:** In phase means circuit is under resonance and the admittance seen by source must be real i.e. imaginary part of  $Y_{eq} = 0$

$$\Rightarrow Y_{eq} = \frac{1}{R + j\omega L} + \frac{1}{(1/j\omega C)}$$

$$= \frac{R - j\omega L}{R^2 + (\omega L)^2} + j\omega C$$

$$\rightarrow \text{Real}(Y) = \frac{R}{R^2 + (\omega L)^2} = \frac{10}{100 + (100)^2} = \frac{1}{1010}$$

$$I = VY = \frac{101}{1010} = 0.1 = 100\text{mA}$$

33. A symmetrical three-phase three-wire RYB system is connected to a balanced delta-connected load. The RMS values of the line current and line-to-line voltage are 10 A and 400 V respectively. The power in the system is measured using the two wattmeter method. The first wattmeter connected between R-line and Y-line reads zero. The reading of the second wattmeter (connected between B-line and Y-line) in watt is \_\_\_\_\_.

**Key:** 3464.1

**Exp:**  $W_1 = V_L I_L \cos(30 - \phi)$

$$W_2 = V_L I_L \cos(30 + \phi)$$

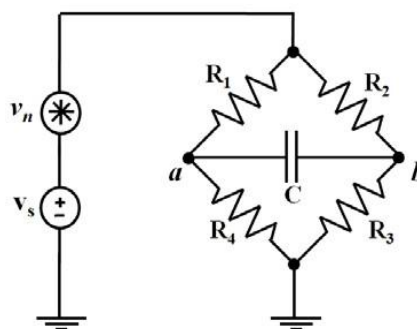
When  $\cos \phi = 0.5$

$$\phi = 60$$

$$W_2 = V_L I_L \cos(30 + 60) = V_L I_L \cos 90 = 0$$

$$\therefore W_1 = 10 \times 400 \cos(30 - 60) = 3464.1\text{W}$$

34. In the strain gauge bridge circuit given below,  $R_1 = R_3 = R(1 - x)$  and  $R_2 = R_4 = R(1 + x)$ , where  $R$  is  $350\Omega$ . The voltage sources  $v_s$  and  $v_n$  represent the dc excitation and the undesired noise/interference, respectively. The value of capacitor  $C$  in **microfarad** that is required to ensure that the output across  $a$  and  $b$  is low-pass filtered with a cutoff frequency of 150 Hz is \_\_\_\_\_.



**Key:** 3.02

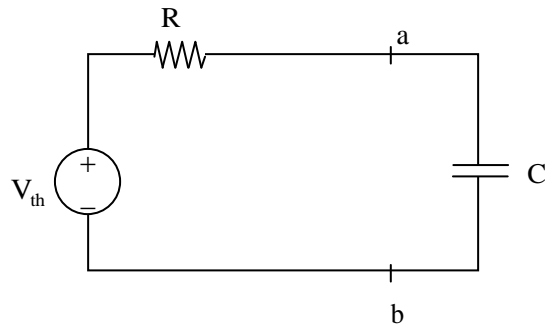
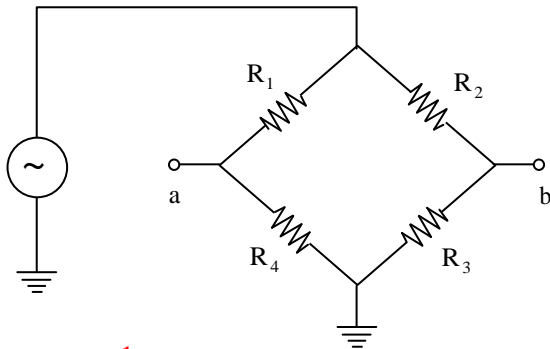
**Exp:** For DC excitation,  $C$  is open

$$R_1 R_3 = R_2 R_4$$

$$[R(1 - x)]^2 = [R(1 + x)]^2 \Rightarrow x = 0$$

For AC excitation

Across a and,  $R_{th} = R_1 \parallel R_4 + R_2 \parallel R_3 = R$

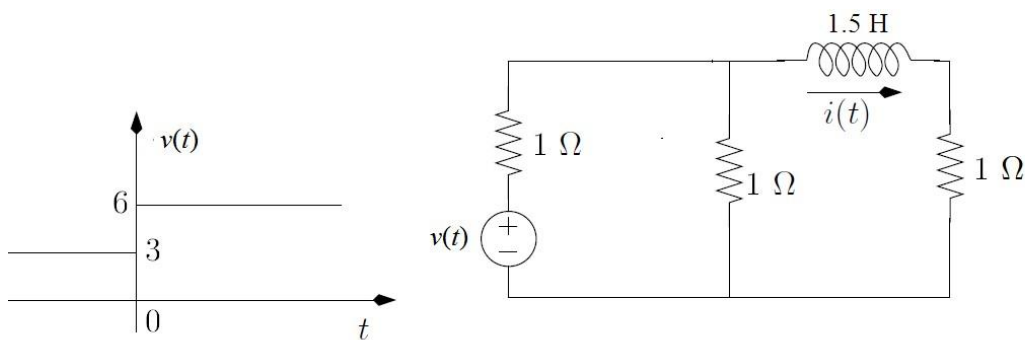


$$f = \frac{1}{2\pi R_C}$$

$$150 = \frac{1}{2\pi \times 350 \times C}$$

$$C = 3.02 \mu F$$

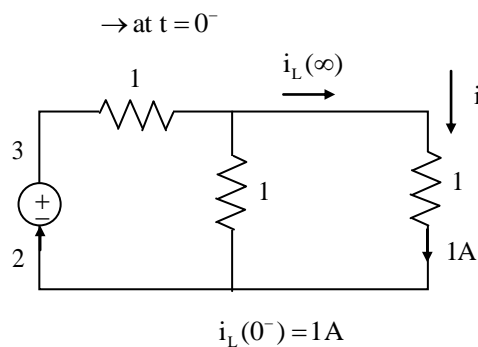
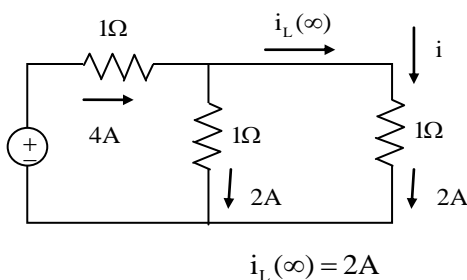
35. The voltage  $v(t)$  shown below is applied to the given circuit.  $v(t) = 3V$  for  $t < 0$  and  $v(t) = 6V$  for  $t > 0$ . The value of current  $i(t)$  at  $t = 1s$ , in **ampere** is \_\_\_\_\_.



**Key:** 1.632

**Exp:**  $i_L(t) = i_L(\infty) + [i_L(0^-) - i_L(\infty)]e^{-\frac{t}{\tau}}$

at  $t = \infty$ , supply is 6V



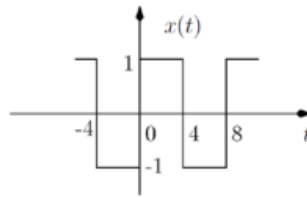
$$\rightarrow \tau = \frac{L}{R_{in}} = \frac{1.5}{1.5} = 1$$

$$\rightarrow i_L(t) = 2 - e^{-t}$$

$$i_L(t) = 2 - e^{-1} = 1.632$$



36. For the periodic signal  $x(t)$  shown below with period  $T = 8$  s, the power in the 10<sup>th</sup> harmonic is



- (A) 0      (B)  $\frac{1}{2} \left( \frac{2}{10\pi} \right)^2$       (C)  $\frac{1}{2} \left( \frac{4}{10\pi} \right)^2$       (D)  $\frac{1}{2} \left( \frac{4}{5\pi} \right)^2$

**Key:** (A)

**Exp:** The given square wave satisfy odd and half wave symmetry so it does not have any eigen harmonic. Since 10<sup>th</sup> harmonic amplitude is 0. So Re also 0

37. The fundamental period  $N_0$  of the discrete-time sinusoid  $x[n] = \sin\left(\frac{301}{4}\pi n\right)$  is \_\_\_\_\_.

**Key:** 8

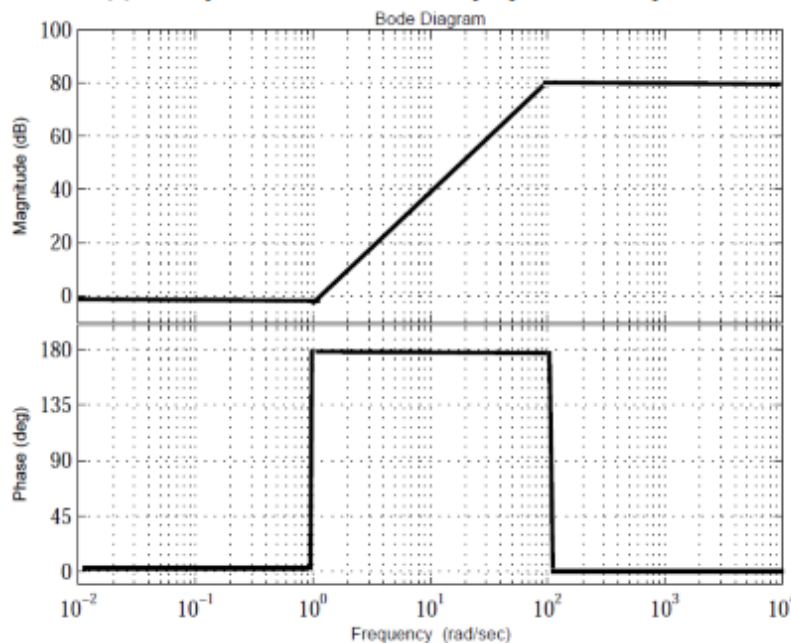
**Exp:** In discrete case  $\omega_0 N = 2\pi m \Rightarrow N = \frac{2\pi}{\omega_0} m$

Where m is the smallest positive integer that makes integer.

$$\rightarrow N = 2\pi \times \frac{4}{301\pi} \times m = \left[ \frac{8}{301} m \right]$$

If  $m = 301$ , then  $N = 8$

38. The transfer function  $G(s)$  of a system which has the asymptotic Bode plot shown below is



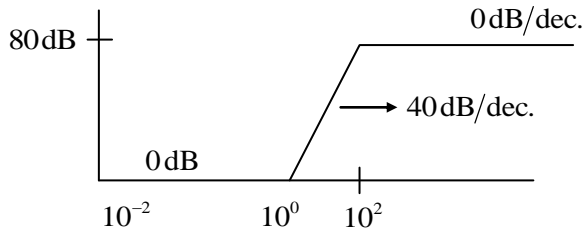
- (A)  $10^4 \frac{(s-1)^2}{(s+100)^2}$       (B)  $10^4 \frac{(s+1)^2}{(s+100)^2}$       (C)  $10^4 \frac{(s+1)}{(s+100)^2}$       (D)  $10^4 \frac{(s-1)^2}{(s-100)^2}$

**Key:** (B)

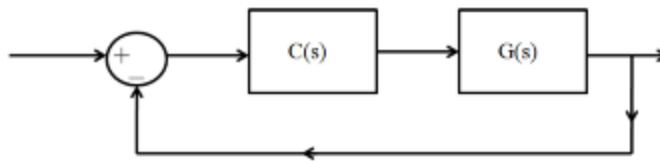
**Exp:**  $TF = k \cdot \frac{\left(1 + \frac{s}{1}\right)^2}{\left(1 + \frac{s}{100}\right)^2}$

$$= \frac{10^4 (s+1)^2}{(s+100)^2}$$

$20 \log_{10} k = 0 \quad k = 1$



39. For the feedback system given below, the transfer function  $G(s) = \frac{1}{(s+1)^2}$ . The system **CANNOT** be stabilized with



(A)  $C(s) = 1 + \frac{3}{s}$

(B)  $C(s) = 3 + \frac{7}{s}$

(C)  $C(s) = 3 + \frac{9}{s}$

(D)  $C(s) = \frac{1}{s}$

**Key:** (C)

**Exp:** The characteristic equation of system is  $1 + G(s)C(s) = 0$

$$\Rightarrow 1 + \frac{C(s)}{s^2 + 2s + 1} = 0$$

$$\Rightarrow s^2 + 2s + 1 + C(s) = 0$$

if we take  $C(s) = 3 + \frac{9}{s}$  then

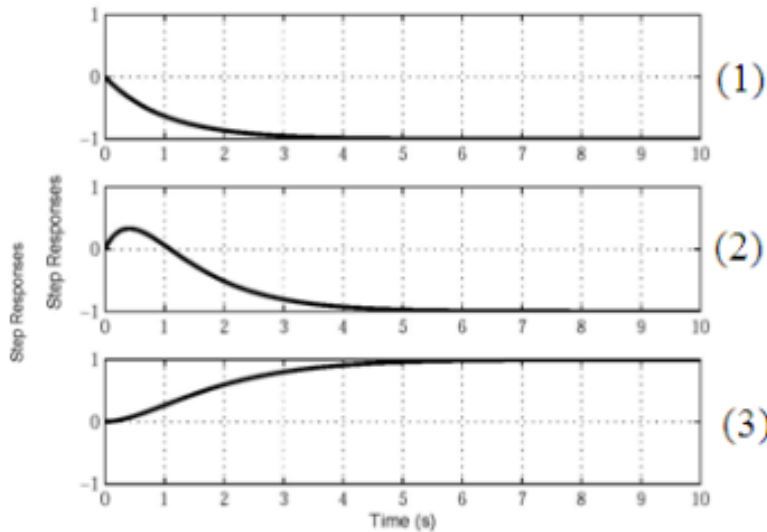
$$s^2 + 2s + 1 + 3 + \frac{9}{s} = 0$$

$$\Rightarrow s^3 + 2s^2 + 4s + 9 = 0$$

$$\begin{array}{c|cc} s^3 & 1 & 4 \\ s^2 & 2 & 9 \\ s^1 & -1/2 & \\ s^0 & 9 & \end{array}$$

So system is unstable, remaining options gives stable.

40. Match the unit-step responses (1), (2) and (3) with the transfer functions  $P(s)$ ,  $Q(s)$  and  $R(s)$ , given below.

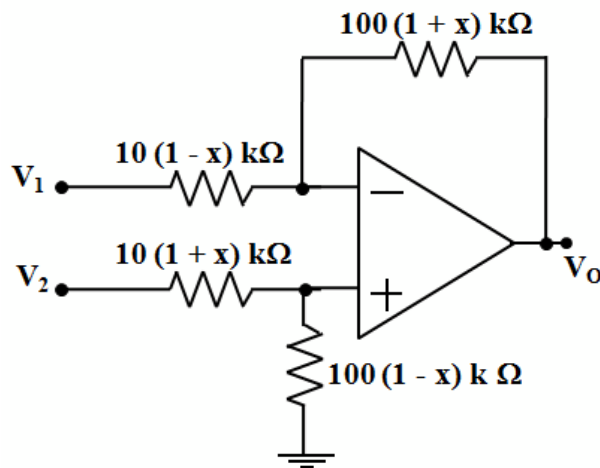


$P(s) = \frac{-1}{(s+1)}$
$Q(s) = \frac{2(s-1)}{(s+10)(s+2)}$
$R(s) = \frac{1}{(s+1)^2}$

- (A)  $P(s)-(3)$ ,  $Q(s)-(2)$ ,  $R(s)-(1)$       (B)  $P(s)-(1)$ ,  $Q(s)-(2)$ ,  $R(s)-(3)$   
(C)  $P(s)-(2)$ ,  $Q(s)-(1)$ ,  $R(s)-(3)$       (D)  $P(s)-(1)$ ,  $Q(s)-(3)$ ,  $R(s)-(2)$

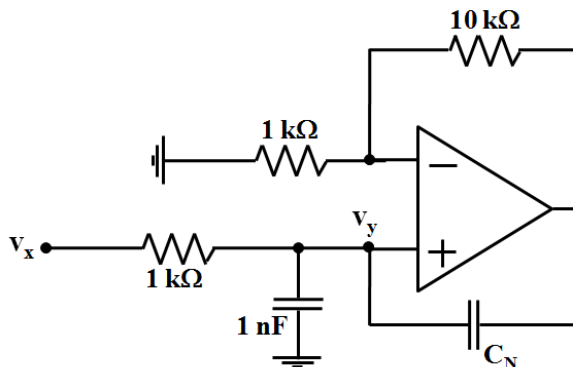
**Key:** (B)

41. An ideal op-amp is used to realize a difference amplifier circuit given below having a gain of 10. If  $x = 0.025$ , the CMRR of the circuit in **dB** is \_\_\_\_\_.



**Key:** 40 to 41

42. In the circuit given below, the op-amp is ideal. The input  $v_x$  is a sinusoid. To ensure  $v_y = v_x$ , the value of  $C_N$  in **picoFarad** is \_\_\_\_\_.



**Key:** 100

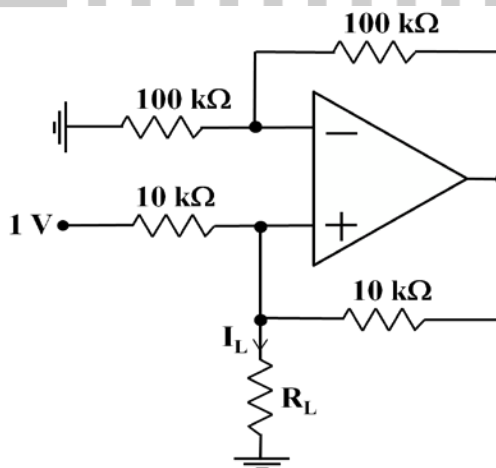
**Exp:** when  $V_x = V_y$

$$\frac{V_y}{10^{-9} S} + \frac{V_y - V_0}{C_N S} = 0$$

$$V_y (10^{-9} + C_N) = V_0 C_N = 10 C_N V_y$$

$$10 C_N = 10^{-9} \Rightarrow C_N = 10^{-10} = 100 \text{ pF}$$

43. In the circuit given below, the op-amp is ideal. The value of current  $I_L$  in **microampere** is \_\_\_\_\_.

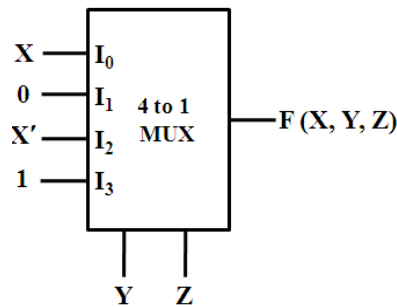


**Key:** 100

**Exp:** It is standard V to I converter, where  $[100k \times 10k] = [10k \times 100]$  i.e. the balanced bridge is formed so the current.

$$I_L = \frac{1}{10 \times 10^3} = 100 \mu A$$

44. A 4 to 1 multiplexer to realize a Boolean function  $F(X, Y, Z)$  is shown in the figure below. The inputs  $Y$  and  $Z$  are connected to the selectors of the MUX ( $Y$  is more significant). The canonical sum-of-product expression for  $F(X, Y, Z)$  is



- (A)  $\Sigma m(2, 3, 4, 7)$  (B)  $\Sigma m(1, 3, 5, 7)$  (C)  $\Sigma m(0, 2, 4, 6)$  (D)  $\Sigma m(2, 3, 5, 6)$

**Key:** (A)

**Exp:**  $F = \bar{x}\bar{y}z + \bar{o}yz + \bar{x}yz + 1.yz$

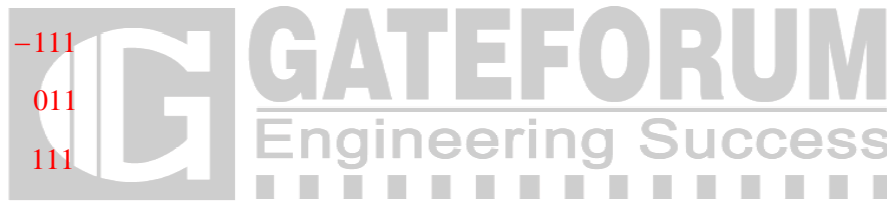
$$= \bar{x}\bar{y}z + \bar{x}yz + yz$$

$$= 100 \quad 100 \quad \bar{1}11$$

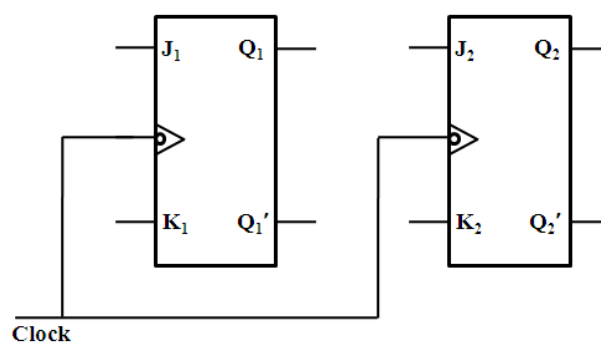
$$011$$

$$111$$

$$F(x, y, z) = \Sigma m = (2, 3, 4, 7)$$



45. A synchronous counter using two J - K flip flops that goes through the sequence of states:  $Q_1Q_2 = 00 \rightarrow 10 \rightarrow 01 \rightarrow 11 \rightarrow 00...$  is required. To achieve this, the inputs to the flip flops are



- (A)  $J_1 = Q_2, K_1 = 0; J_2 = Q_1', K_2 = Q_1$  (B)  $J_1 = 1, K_1 = 1; J_2 = Q_1, K_2 = Q_1$   
(C)  $J_1 = Q_2, K_1 = Q_2'; J_2 = 1', K_2 = 1$  (D)  $J_1 = Q_2', K_1 = Q_2; J_2 = Q_1, K_2 = Q_1'$

**Key:** (B)

**Exp:**

Present	State	Next	State	Flip flop Input			
$Q_1$	$Q_2$	$Q_1^+$	$Q_2^+$	$J_1$	$K_1$	$J_2$	$K_2$
0	0	1	0	1	X	0	X
1	0	0	1	X	1	1	X
0	1	1	1	1	X	X	0
1	1	0	0	X	1	X	1

From the column of  $J_1 K_1 J_2 K_2$

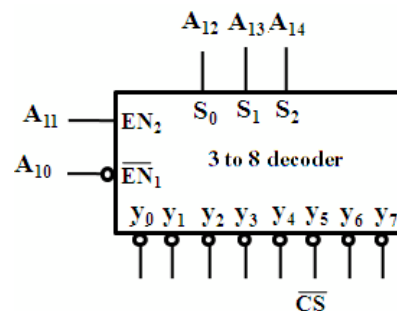
We can say  $J_1 = 1$

$K_1 = 1$

And  $T_2 = Q_1$

$K_2 = Q_1$

46. A 1 Kbyte memory module has to be interfaced with an 8-bit microprocessor that has 16 address lines. The address lines  $A_0$  to  $A_9$  of the processor are connected to the corresponding address lines of the memory module. The active low chip select  $\overline{CS}$  of the memory module is connected to the  $y_5$  output of a 3 to 8 decoder with active low outputs.  $S_0, S_1$ , and  $S_2$  are the input lines to the decoder, with  $S_2$  as the MSB. The decoder has one active low  $\overline{EN}_1$  and one active high  $EN_2$  enable lines as shown below. The address range(s) that gets mapped onto this memory module is (are)



- (A)  $3000_H$  to  $33FF_H$  and  $E000_H$  to  $E3FF_H$  (B)  $1400_H$  to  $17FF_H$   
(C)  $5300_H$  to  $53FF_H$  and  $A300_H$  to  $A3FF_H$  (D)  $5800_H$  to  $5BFF_H$  and  $D800_H$  to  $DBFF_H$

**Key:** (D)

**Exp:** → 1kB memory means 10 address lines  $A_9$  to  $A_0$

→ Since  $A_{15}$  line is missing it should be taken as don't care.

→ 5<sup>th</sup> output of decoder should be activated means  $A_{14} = 1; A_{13} = 0; A_{12} = 1$

→  $A_{11} = 1$  since active high enable

→  $A_{10} = 0$  since active low enable

	$A_{15}$	$A_{14}$	$A_{13}$	$A_{12}$	$A_{11}$	$A_{10}$	$A_9$	$A_8$	$A_7$	$A_6$	$A_5$	$A_4$	$A_3$	$A_2$	$A_1$	$A_0$
→		1	0	1	1	0	0	0	0	0	0	0	0	0	0	0
		1	0	1	1	0	1	1	1	1	1	1	1	1	1	1

→ If  $A_{15} = 0$  then the range is 5800 to 5BFF

→ If  $A_{15} = 1$  then the range is D800 to DBFF.

47. A coil is tested with a series type Q-meter. Resonance at a particular frequency is obtained with a capacitance of 110 pF. When the frequency is doubled, the capacitance required for resonance is 20 pF. The distributed capacitance of the coil in **pico farad** is \_\_\_\_\_.

**Key:** 10

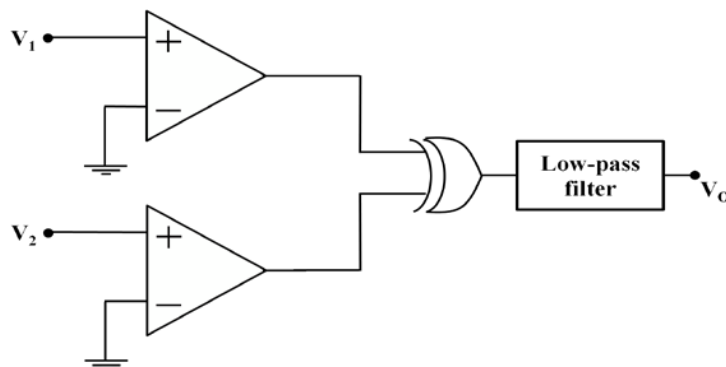
**Exp:**  $C_d = \frac{C_1 - n^2 C_2}{n^2 - 1} \Rightarrow \frac{110 - (4 \times 20)}{4 - 1} = 10 \text{ pf}$

here  $C_1 = 110 \text{ pf}$

$C_2 = 20 \text{ pf}$

$$\frac{f_2}{f_1} = n = 2$$

48. The comparators (output = '1', when input  $\geq 0$  and output = '0', when input  $< 0$ ), exclusive-OR gate and the unity gain low-pass filter given in the circuit are ideal. The logic output voltages of the exclusive-OR gate are 0 V and 5 V. The cutoff frequency of the low-pass filter is 0.1 Hz. For  $V_1 = 1 \sin(3000t + 36^\circ) \text{ V}$  and  $V_2 = 1 \sin(3000t) \text{ V}$ , the value of  $V_O$  in **volt** is \_\_\_\_\_.



**Key:** 1

49. A 200 mV full scale dual-slope 3 ½ digit DMM has a reference voltage of 100 mV and a first integration time of 100 ms. For an input of  $[100 + 10\cos(100\pi t)]$  mV, the conversion time (without taking the auto-zero phase time into consideration) in **millisecond** is \_\_\_\_\_.

**Key:** 200

**Exp:** In dual slope converter total conversion time

$$T = 1^{\text{st}} \text{ integration period} + 2^{\text{nd}} \text{ integration period}$$

$$= T_1 + T_2 = 100 \text{ msec} + T_2$$

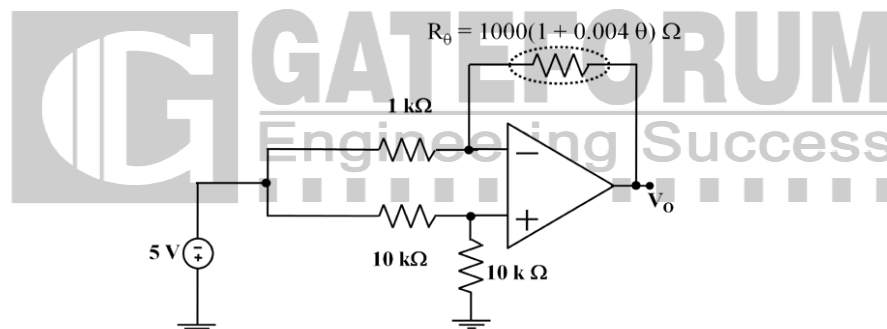
$$\text{To obtain } T_2 \text{ we can use } V_{\text{in}} T_1 = V_{\text{ref}} T_2$$

$$\Rightarrow 100 \text{ mV} \times 100 \text{ m.sec} = 100 \text{ mV} \times T_2$$

$$\Rightarrow T_2 = 100 \text{ msec}$$

$$\rightarrow T = 100 + 100 = 200 \text{ msec.}$$

50. In the circuit below, the op-amp is ideal and the sensor is an RTD whose resistance  $R_\theta = 1000(1 + 0.004\theta) \Omega$ , where  $\theta$  is temperature in  $^\circ\text{C}$ . The output sensitivity in **mV/ $^\circ\text{C}$**  is \_\_\_\_\_.



**Key:** 10

**Exp:**  $V_+ = \frac{V_{\text{in}}}{2} = V_-$

$$\frac{V_{\text{in}} - V_-}{1\text{k}} = \frac{V_- - V_o}{R_\theta} = \frac{V_- - V_o}{\left(1 + \frac{\theta}{250}\right)\text{k}}$$

$$\left(1 + \frac{\theta}{250}\right) \left(V_{\text{in}} - \frac{V_{\text{in}}}{2}\right) = \frac{V_{\text{in}}}{2} - V_o$$

$$\left(1 + \frac{\theta}{250}\right) \frac{V_{\text{in}}}{2} = \frac{V_{\text{in}}}{2} - V_o$$

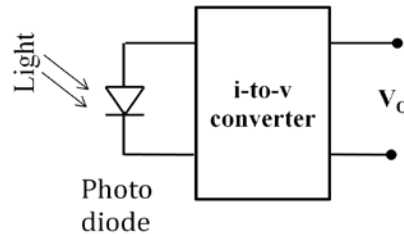
$$\frac{V_{\text{in}}}{2} + \frac{\theta \cdot V_{\text{in}}}{500} = \frac{V_{\text{in}}}{2} - V_o$$

$$V_{\text{in}} = -5\text{V}, \quad V_o = +\frac{0.5}{500} = +\frac{\theta}{100}$$

$$\frac{dV_o}{d\theta} = \frac{1}{100} = 10 \text{ mV}/^\circ\text{C}$$



51. The photo diode in the figure below has an active sensing area of  $10 \text{ mm}^2$ , a sensitivity of  $0.5 \text{ A/W}$  and a dark current of  $1 \mu\text{A}$ . The i-to-v converter has a sensitivity of  $100 \text{ mV}/\mu\text{A}$ . For an input light intensity of  $4 \text{ W/m}^2$ , the output  $V_O$  in **volt** is.



**Key:** 2

**Exp:**  $A = 10 \times 10^{-6} \text{ m}^2$

$$S = 0.5 \text{ A/W} = \frac{1}{2} \text{ A/W} \rightarrow 1 \text{ A} \rightarrow 2 \text{ W}$$

$$I = 1 \mu\text{A} \quad 1 \mu\text{A} \rightarrow 2 \mu\text{W}$$

$$S = 100 \text{ mV}/\mu\text{A}$$

$$I = 4 \text{ W} \times 10 \times 10^{-6} = 40 \mu\text{W}$$

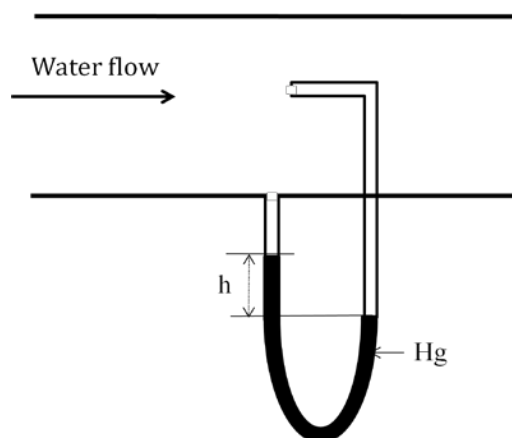
$$2 \mu\text{W} \rightarrow 1 \mu\text{A}$$

$$\therefore 40 \mu\text{W} \rightarrow 20 \mu\text{A}$$

$$1 \mu\text{A} \rightarrow 100 \text{ mV}$$

$$20 \mu\text{A} \rightarrow 2 \text{ V}$$

52. The velocity of flow of water (density  $1000 \text{ kg/m}^3$ ) in a horizontal pipe is measured using the Pitot tube shown below. The fluid in the U-tube manometer is mercury with a density of  $13534 \text{ kg/m}^3$ . Assume  $g = 9.81 \text{ m/s}^2$ . If the height difference ( $h$ ) is measured as  $94.1 \text{ mm}$ , the velocity of flow of water in **m/s** is \_\_\_\_\_.



**Key:** 4.81

**Exp:**  $V = \sqrt{2gh}$

$$h = x \left[ \frac{s_g}{s_0} - 1 \right] = 94.1 \times 10^{-3} \left[ \frac{13534}{1000} - 1 \right] = 1.18$$

$$V = \sqrt{2 \times 9.81 \times 1.18} = 4.81 \text{ m/sec}$$

53. The band gap in eV of a semiconductor material required to construct an LED that emits peak power at the wavelength of 620 nm is \_\_\_\_\_.

(Plank constant  $h = 4.13567 \times 10^{-15} \text{ eV s}$  and speed of light  $c = 2.998 \times 10^8 \text{ m/s}$ ).

**Key:** 2

**Exp:**  $E = \frac{hc}{\lambda} = \frac{4.13567 \times 10^{-15} \times 2.998 \times 10^8}{620 \times 10^{-9}} = 2 \text{ eV}$

54. The signal  $m(t) = \frac{\sin(100\pi t)}{100\pi t}$  is frequency modulated (FM) with an FM modulator of frequency deviation constant of 30 kHz/V. Using Carson's rule, the approximate bandwidth of the modulated wave in kilohertz is \_\_\_\_\_.

**Key:** 60

**Exp:**  $k_f = 30 \text{ KHz/V}$

$$(\Delta f)_{\max} = 30 \times 10^3 \times 1 = 30 \text{ KHz}$$

$$f_{\max} = 50 \text{ Hz}$$

$$\beta = \frac{30 \text{ KHz}}{50 \text{ Hz}} \gg 1$$

$$BW = 2(\Delta f + f_m) \approx 2(\Delta f)_{\max} = 60 \text{ kHz}$$

55. A signal  $m(t)$  varies from -3.5V to +3.5 V with an average power of 3 W. The signal is quantized using a midtread type quantizer and subsequently binary encoded. With the codeword of length 3, the signal to quantization noise ratio in dB is \_\_\_\_\_.

**Key:** 16.72

**Exp:**  $n = 3 \text{ bits}$

$$A_m = 3.5 \text{ V}$$

$$P_{\text{avg}} = 3.5 \text{ W}$$

$$SNR = \frac{\text{Signal power}}{\text{Noise power}}$$

$$\text{Quantization noise power} = \frac{\Delta^2}{12} = \frac{7^2}{12} = \frac{7^2}{L^2 \cdot 12} = \frac{7 \times 7}{64 \times 12}$$

$$L = 2^3 = 8, \Delta = \frac{2A_m}{L}$$

$$(SNR)_{\text{dB}} = 10 \log \left( \frac{0.3 \times 12 \times 64}{49} \right) = 16.72 \text{ dB}$$