# Digital Signal Processing

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## Chapter 4 Discrete-Time System

### **Chapter 4 Discrete-Time Systems**

- 4.1 Discrete-Time System Examples
- 4.2 Classification of Discrete-Time Systems
- 4.3 Impulse and Step Responses
- 4.4 Time-Domain Characterization of LTI Discrete-Time Systems
- 4.5 Simple Interconnection Schemes
- 4.6 Finite-Dimensional LTI Discrete-Time Systems
- 4.7 Classification of LTI Discrete-Time Systems
- 4.8 Frequency-Domain Representations of LTI Discrete-Time Systems
- 4.9 Phase and Group Delays



### § 4.1 Discrete-Time Systems

• Accumulator:

$$y[n] = \sum_{\ell=-\infty}^{n} x[\ell] = \sum_{\ell=-\infty}^{n-1} x[\ell] + x[n] = y[n-1] + x[n]$$
  
递归表达式

$$y[n] = \sum_{\ell=-\infty}^{-1} x[\ell] + \sum_{\ell=0}^{n} x[\ell]$$
  
=  $y[-1] + \sum_{\ell=0}^{n} x[\ell], n \ge 0$ 

输入输出关系式

§ 4.1 Discrete-Time Systems

• M-point Moving-Average System-

$$y[n] = \frac{1}{M} \sum_{k=0}^{M-1} x[n-k]$$

- Use: smoothing random variations in data
- In most applications, the data x[n] is a bounded sequence, so M-point average y[n] is also bounded
- If there is no bias in the measurements, an improved estimate of the noisy data is obtained by simply increasing M



### § 4.1 Discrete-Time Systems

$$y[n] = \frac{1}{M} \sum_{k=0}^{M-1} x[n-k]$$

• A more efficient implementation

$$= \frac{1}{M} \left( \sum_{\ell=1}^{M} x[n-\ell] + x[n] - x[n-M] \right) (\mathfrak{B} \equiv \mathfrak{K} \not{\mathfrak{B}} \ell = \ell - 1)$$

$$= \frac{1}{M} \left( \sum_{\ell=0}^{M-1} x[n-1-\ell] + x[n] - x[n-M] \right)$$

$$= \frac{1}{M} \left( \sum_{\ell=0}^{M-1} x[n-1-\ell] \right) + \frac{1}{M} \left( x[n] - x[n-M] \right)$$

$$= y[n-1] + \frac{1}{M} \left( x[n] - x[n-M] \right)$$

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### § 4.2 Classification of Discrete-Time Systems

- Linear System
- Shift-Invariant System
- Linear Time-Invariant System
- Causal System
- Stable System
- Passive and Lossless Systems(无源和无损)



• **Definition** - If y<sub>1</sub>[n] is the output due to an input x<sub>1</sub>[n] and y<sub>2</sub>[n] is the output due to an input x<sub>2</sub>[n] then for an input

 $x[n]=ax_1[n]+bx_2[n]$ 

the output is given by

 $y[n]=ay_1[n]+by_2[n]$ 

 Above property must hold for any arbitrary constants a and b and for all possible inputs x<sub>1</sub>[n] and x<sub>2</sub>[n]

### § 4.2.1 Linear Systems

Example 4.3-Property of the Accumulator

$$y_1[n] = \sum_{\ell=-\infty}^n x_1[\ell]$$
  $y_2[n] = \sum_{\ell=-\infty}^n x_2[\ell]$ 

 $x[n] = \alpha x_1[n] + \beta x_2[n]$ 

$$y[n] = \sum_{\ell=-\infty}^{n} (\alpha x_1[\ell] + \beta x_2[\ell])$$

$$= \alpha \sum_{\ell=-\infty}^{n} x_1[\ell] + \beta \sum_{\ell=-\infty}^{n} x_2[\ell] = \alpha y_1[n] + \beta y_2[n]$$

—Linear system



### § 4.2.1 Linear Systems

**Example 4.3-For Accumulator with a causal input** 

$$y_{1}[n] = y_{1}[-1] + \sum_{\ell=0}^{n} x_{1}[\ell] \qquad y_{2}[n] = y_{2}[-1] + \sum_{\ell=0}^{n} x_{2}[\ell]$$
$$x[n] = \alpha x_{1}[n] + \beta x_{2}[n]$$
$$y[n] = y[-1] + \sum_{\ell=0}^{n} (\alpha x_{1}[\ell] + \beta x_{2}[\ell])$$

$$= y[-1] + \alpha \sum_{\ell=0}^{n} x_1[\ell] + \beta \sum_{\ell=0}^{n} x_2[\ell]$$

 $? = \alpha y_1[n] + \beta y_2[n]$ 



## § 4.2.1 Linear Systems

$$\begin{aligned} \alpha y_1[n] + \beta y_2[n] \\ &= \alpha \bigg( y_1[-1] + \sum_{\ell=0}^n x_1[\ell] \bigg) + \beta \bigg( y_2[-1] + \sum_{\ell=0}^n x_2[\ell] \bigg) \\ &= \big( \alpha y_1[-1] + \beta y_2[-1] \big) + \bigg( \alpha \sum_{\ell=0}^n x_1[\ell] + \beta \sum_{\ell=0}^n x_2[\ell] \bigg) \end{aligned}$$

Thus,

 $y[n] = \alpha y_1[n] + \beta y_2[n]$  |*if*  $y[-1] = \alpha y_1[-1] + \beta y_2[-1]$ 

This condition cannot be satisfied unless accumulator is initially at rest with zero initial condition



§ 4.2.2 Shift-Invariant System



• For a shift-invariant system, if y<sub>1</sub>[n] is the response to an input x<sub>1</sub>[n], then the response to an input

 $\mathbf{x}[\mathbf{n}] = \mathbf{x}_1[\mathbf{n} - \mathbf{n}_0]$ 

is simply

 $y[n]=y_1[n-n_0]$ 

where n<sub>0</sub> is any positive or negative integer

- The above relation must hold for any arbitrary input and its corresponding output
- The above property is called time-invariance property, or shift-invariant proterty

### § 4.2.2 Shift-Invariant System



- Time-invariance property ensures that for a specified input, the output is independent of the time the input is being applied
  - Example Consider the up-sampler

$$x[n] \longrightarrow \uparrow L \longrightarrow x_u[n]$$

with an input-output relation given by

$$x_u[n] = \begin{cases} x[n/L], & n = 0, \pm L, \pm 2L, \dots \\ 0, & \text{otherwise} \end{cases}$$



#### • An example of the up-sampling operation



- For an input  $x_1[n] = x[n n_o]$  the output  $x_{1,u}[n]$  is given by  $x_{1,u}[n] = \begin{cases} x_1[n/L], & n = 0, \pm L, \pm 2L, \dots, \\ 0, & \text{otherwise} \end{cases}$  $= \begin{cases} x[(n - Ln_o)/L], & n = 0, \pm L, \pm 2L, \dots, \\ 0, & \text{otherwise} \end{cases}$
- However from the definition of the up-sampler  $x_u[n-n_o]$   $=\begin{cases} x[(n-n_o)/L], & n=n_o, n_o \pm L, n_o \pm 2L, ...., \\ 0, & \text{otherwise} \end{cases}$  $\neq x_{1,u}[n]$
- Hence, the up-sampler is a time-varying system



#### 例题: 判断线性性和时不变性

- y(n)=2x(n)+5 非线性、时不变
- y(n)=x<sup>2</sup>(n) 非线性、时不变
- y(n)=nx(n)
- $y(n)=x(n-n_0)$

- 线性、时变
- 线性、时不变





$$y(n) = x(n - n_0)$$

 $x_{1}[n] \Rightarrow y_{1}[n] = x_{1}[n - n_{0}],$   $x_{2}[n] \Rightarrow y_{2}[n] = x_{2}[n - n_{0}],$   $ax_{1}[n] + bx_{2}[n] \Rightarrow y[n] = ax_{1}[n - n_{0}] + bx_{2}[n - n_{0}]$  $y[n] = ay_{1}[n] + by_{2}[n]$ 

$$x[n] \Rightarrow y[n] = x[n - n_0],$$
  

$$x[n-1] \Rightarrow y'[n] = x[n - n_0 - 1],$$
  

$$y[n-1] = x[(n-1) - n_0],$$
  

$$y'[n] = y[n-1]$$

### y(n)=ax(n)+b

$$x_{1}[n] \Rightarrow y_{1}[n] = ax_{1}[n] + b,$$
  

$$x_{2}[n] \Rightarrow y_{2}[n] = ax_{2}[n] + b,$$
  

$$x_{1}[n] + x_{2}[n] \Rightarrow y[n] = ax_{1}[n] + ax_{2}[n] + b$$
  

$$y[n] \neq y_{1}[n] + y_{2}[n] = ax_{1}[n] + ax_{2}[n] + 2b$$

$$x[n] \Rightarrow y[n] = ax[n] + b,$$
  

$$x[n - n_0] \Rightarrow y'[n] = ax[n - n_0] + b,$$
  

$$y[n - n_0] = y'[n],$$



$$y(n) = x(2n)$$

解: 
$$y(n-D) = x[2(n-D)] = x(2n-2D)$$
  
 $y_D(n) = T[x(n-D)] = x(2n-D)$   
 $\Rightarrow \quad y(n-D) \neq y_D(n)$ 



### Definition

- In a causal system, the  $n_0$ -th output sample  $y(n_0)$ depends only on input samples x(n) for  $n \le n_0$ and does not depend on input samples for  $n > n_0$
- For a causal system, changes in output samples do not precede changes in the input samples

### § 4.2.3 Causal System



• Examples of noncausal systems:  $y[n] = x_u[n] + \frac{1}{2}(x_u[n-1] + x_u[n+1])$ 



### Definition

- There are various definitions of stability. We consider here the *bounded-input*, *bounded-output* (*BIBO*) stability
- If *y*(*n*) is the response to an input *x*(*n*) and if *x*(*n*) is bounded, i.e.

 $|x(n)| < B_x$ , for all values of n

then y(n) is bounded, i.e

 $|y(n)| < B_y$ , for all values of n

### § 4.2.4 Stable System

#### **Example: M-point Moving-Average filter**

 $|x[n]| < B_x$ 

$$|y[n]| = \left|\frac{1}{M}\sum_{k=0}^{M-1} x[n-k]\right| \le \frac{1}{M}\sum_{k=0}^{M-1} |x[n-k]| \le \frac{1}{M} \bullet M \bullet B_x \le B_x$$



- A discrete-time system is defined to be **passive** if, for every finite-energy input x[n], the output y[n] has, at most, the same energy, i.e.

$$\sum_{n=-\infty}^{\infty} \left| y[n] \right|^2 \le \sum_{n=-\infty}^{\infty} \left| x[n] \right|^2 < \infty$$

• For a lossless system, the above inequality is satisfied with an equal sign for every input

- Example Consider the discrete-time system defined by y[n]=αx[n-N] with N a positive integer
- Its output energy is given by

$$\sum_{n=-\infty}^{\infty} |y[n]|^2 = |\alpha|^2 \sum_{n=-\infty}^{\infty} |x[n]|^2$$

• Hence, it is a passive system if  $|\alpha| \le 1$ and is a lossless system if  $|\alpha| = 1$ 

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- The response of a discrete-time system to a unit sample sequence {δ[n]} is called the unit impulse response or simply, the impulse response, and is denoted by {h[n]}
- The response of a discrete-time system to a unit step sequence {u[n]} is called the unit step response or simply, the step response, and is denoted by {s[n]}

### § 4.3 Impulse and Step Responses

- <u>Example</u> The impulse response of the system
- $y[n] = \alpha_1 x[n] + \alpha_2 x[n-1] + \alpha_3 x[n-2] + \alpha_4 x[n-3]$ is obtained by setting  $x[n] = \delta[n]$  resulting in

$$h[n] = \alpha_1 \delta[n] + \alpha_2 \delta[n-1] + \alpha_3 \delta[n-2] + \alpha_4 \delta[n-3]$$

• The impulse response is thus a finite-length sequence of length 4 given by  $\{h[n]\} = \{\alpha_1, \alpha_2, \alpha_3, \alpha_4\}$ 



§ 4.3 Impulse and Step Responses

• <u>Example</u> - The impulse response of the discrete-time accumulator

$$y[n] = \sum_{\ell=-\infty}^{n} x[\ell]$$

is obtained by setting  $x[n] = \delta[n]$  resulting in

$$h[n] = \sum_{\ell = -\infty}^{n} \delta[\ell] = \mu[n]$$



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§ 4.4 Time-Domain Characterization of LTI Discrete-Time System

• Linear Time-Invariant (LTI) System -

A system satisfying both the linearity and the time-invariance property

- LTI systems are mathematically easy to analyze and characterize, and consequently, easy to design
- Highly useful signal processing algorithms have been developed utilizing this class of systems over the last several decades



- A consequence of the linear, time invariance property is that a LTI discrete time system is completely characterized by its impulse response
- Knowing the impulse response one can compute the output of the system for any arbitrary input

### § 4.4.1 Input-Output Relationship

Since h(n) is the response of input δ(n) and the system is time invariant, we have

$$\delta(n-k) \to h(n-k)$$

• Likewise, as the system is linear

$$x(k)\delta(n-k) \rightarrow x(k)h(n-k)$$

• Note that, *x*(*k*) is considered as a constant in this case

### § 4.4.1 Input-Output Relationship

• Taking advantage of the property of linear, we have

$$\sum_{k=-\infty}^{\infty} x(k)\delta(n-k) \to \sum_{k=-\infty}^{\infty} x(k)h(n-k)$$

• Eventually, the I-O relationship of an LTI system can be written as follows

$$y(n) = \sum_{k=-\infty}^{\infty} x(k)h(n-k) = \sum_{k=-\infty}^{\infty} h(k)x(n-k)$$





### **BIBO Stability Condition --**

- A discrete-time system is BIBO stable if the output sequence {y(n)} remains bounded for all bounded input sequence {x(n)}
- An LTI discrete-time system is BIBO stable if and only if its impulse response sequence {h(n)} is absolutely summable, i.e.

$$S = \sum_{n = -\infty}^{\infty} \left| h(n) \right| < \infty$$



### 4.4.3 Stability Condition

- Example Consider an LTI discrete-time system with an impulse response  $h[n] = (\alpha)^n \mu[n]$
- For this system

$$S = \sum_{n=-\infty}^{\infty} \left| \alpha^n \right| \mu[n] = \sum_{n=0}^{\infty} \left| \alpha \right|^n = \frac{1}{1 - \left| \alpha \right|} \quad \text{if } \left| \alpha \right| < 1$$

- Therefore  $S < \infty$  if  $|\alpha| < 1$  for which the system is BIBO stable
- If  $|\alpha| = 1$ , the system is not BIBO stable

### 4.4.4 Causality Condition

**Causality Condition** —

 An LTI discrete-time system is causal if and only if its impulse response {h(n)} is a causal sequence, i.e., h[n]=0, for all n<0.</li>

$$y[n] = \sum_{k=-\infty}^{\infty} x[n-k]h[k]$$

 $= \dots + x[n+2]h[-2] + x[n+1]h[-1] + x[n]h[0]$  $+ x[n-1]h[1] + x[n-2]h[2] + \dots$ 

### 4.4.4 Causality Condition



**Causality Condition** —

- A non-causal LTI discrete-time system with a finite-length impulse response can often be realized as a causal system by inserting an appropriate amount of delay.
- Clipping+delaying

### 4.4.4 Causality Condition

• <u>Example</u> - The discrete-time accumulator defined by

$$y[n] = \sum_{\ell=-\infty}^{n} \delta[\ell] = \mu[n]$$

is a causal system as it has a causal impulse response given by

$$h[n] = \sum_{\ell=-\infty}^{n} \delta[\ell] = \mu[n]$$

• 系统稳定性的时域充要条件:h(n)绝对可和, 即

$$S = \sum_{n = -\infty}^{\infty} \left| h[n] \right| < \infty$$

• 系统因果性的时域充要条件:

$$h[n] \equiv 0, n < 0$$



#### **Problems:**

# 4.3(b), 4.20(a), 4.23(a)(解卷积), 4.30(a)(互联结构), 4.67(第一问)

Matlab Exercises: M4.1