
EE3230 Lecture 8: Sequential Circuit Design

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Outline

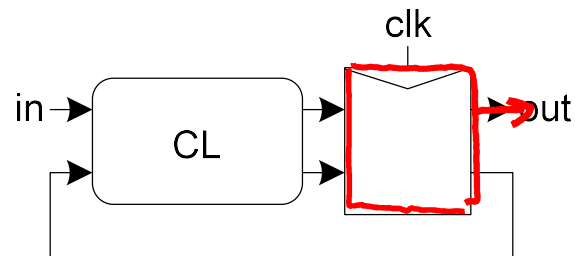
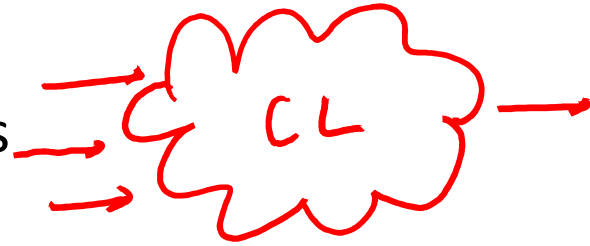
- Sequencing Methods
- Max/Min Delay, Clock Skew, Timing Borrowing
- Sequencing Element Design

Outline

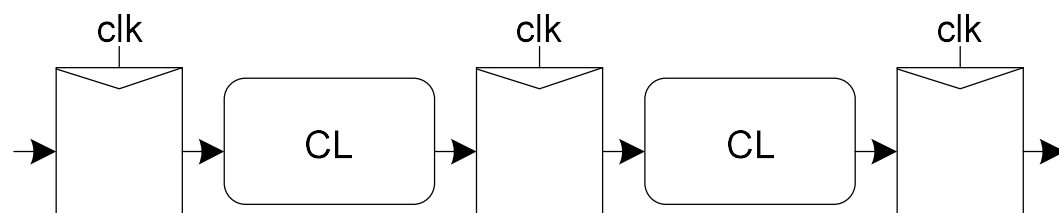
- **Sequencing Methods**
- Max/Min Delay, Clock Skew, Timing Borrowing
- Sequencing Element Design

Sequencing

- Combinational logic
 - Outputs depend on current inputs
- Sequential logic
 - Outputs depend on current and previous inputs
 - Previous, current, and future separated
 - Memory elements required
 - Called states or tokens
 - Examples: Finite-State Machine (FSM), pipeline



Finite State Machine

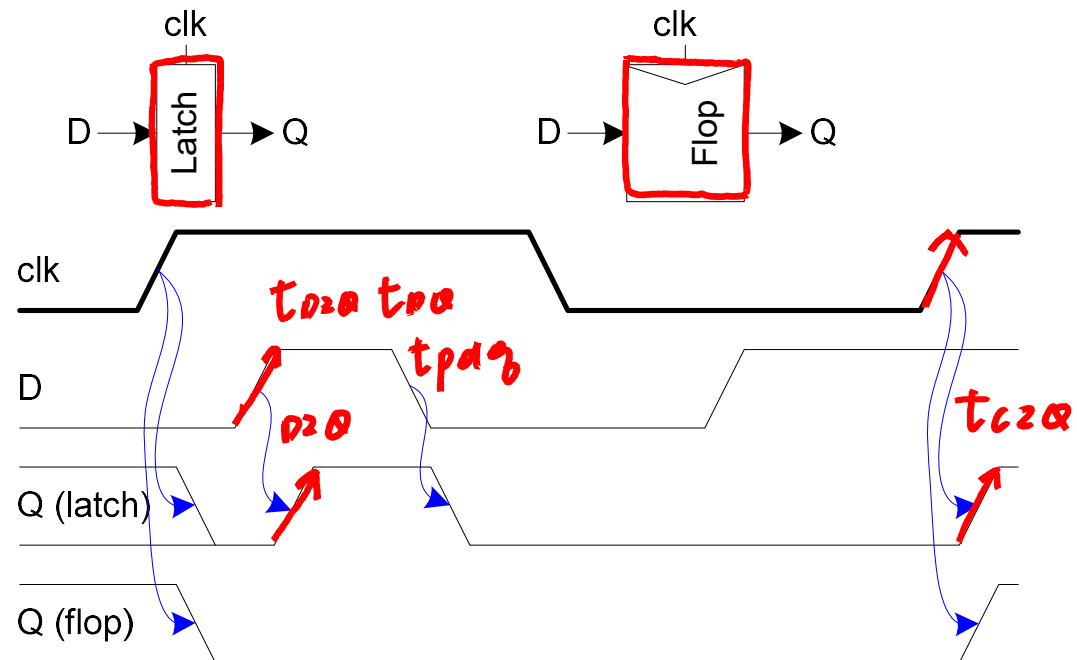


Pipeline



Sequencing Elements

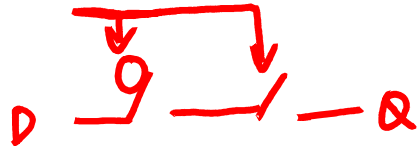
- **Latch:** level-sensitive
 - a.k.a. transparent latch, D latch, etc.
- **Flip-flop:** edge-triggered
 - a.k.a. master-slave flip-flop, D flip-flop, D register, etc.

- Timing diagram
 - Transparent
 - Opaque
 - Edge-triggered



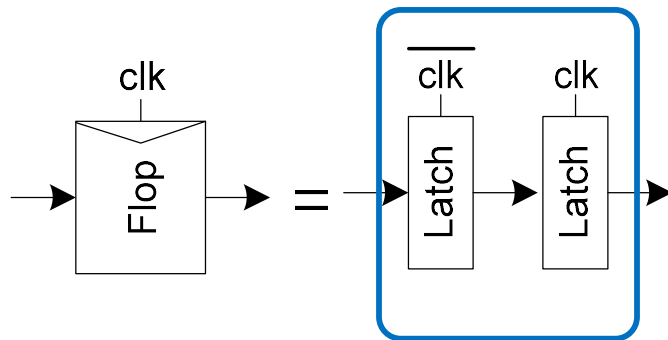
Latch vs. Flip-Flops

- **Latch: level-sensitive** 
 - Passes input D to Q when the clock is high – transparent mode
 - Input D sampled on the falling edge of the clock is held stable when clock is low – hold mode
- **Flip-flop: edge-triggered** 
 - Samples input D on a clock transition
 - Can be either positive edge-triggered or negative edge-triggered
- ★ ~~Built using two latches (master-slave flip-flops)~~



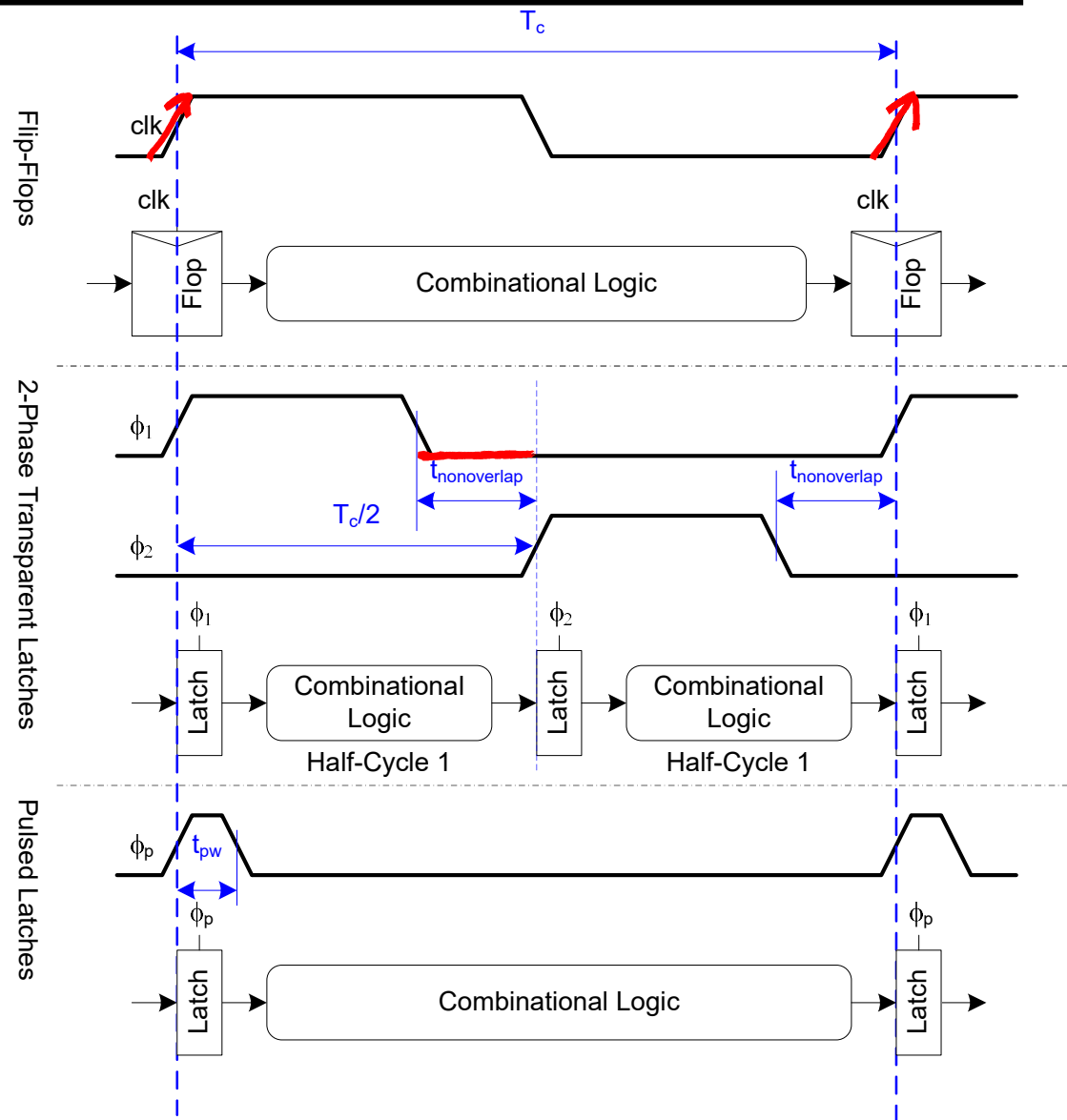
Sequencing Methods

- Flip-flops

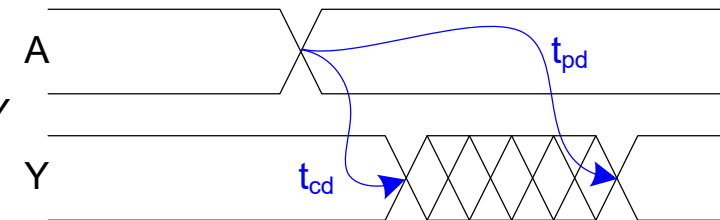
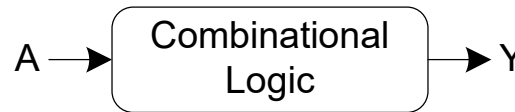


- 2-phase latches

- Pulsed latches

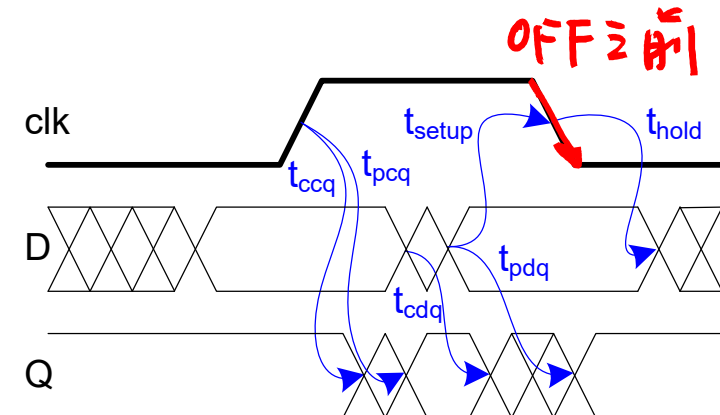
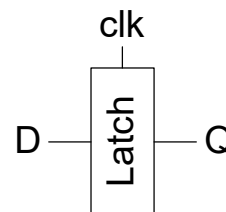
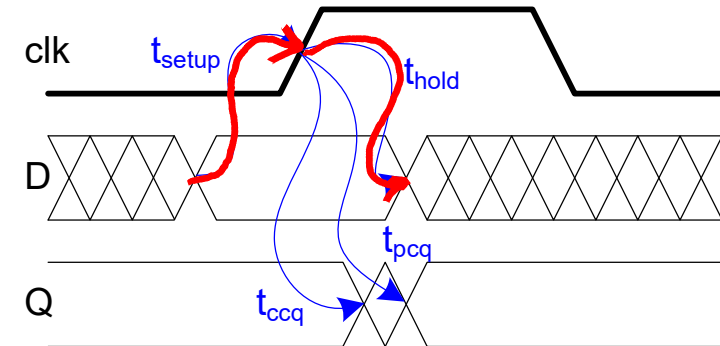
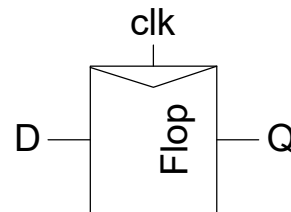


Timing Diagrams



t_{pd}	Logic Prop. Delay
t_{cd}	Logic Cont. Delay
t_{pcq}	Latch/Flop Clk-Q Prop Delay
t_{ccq}	Latch/Flop Clk-Q Cont. Delay
t_{pdq}	Latch D-Q Prop Delay
t_{cdq}	Latch D-Q Cont. Delay
t_{setup}	Latch/Flop Setup Time
t_{hold}	Latch/Flop Hold Time

* Contamination & Propagation delays



Sequencing Overhead

- Flip-flops are used to delay **fast tokens** so that they move through exactly one stage per cycle
- Inevitably add some delay to **slow tokens**
- Make circuits slower than just the logic delay
 - Called sequencing overhead
 - Some people call this clocking overhead
 - Inevitable side effect of maintaining sequence

Outline

- Sequencing Methods
- **Max/Min Delay, Clock Skew, Timing Borrowing**
- Sequencing Element Design

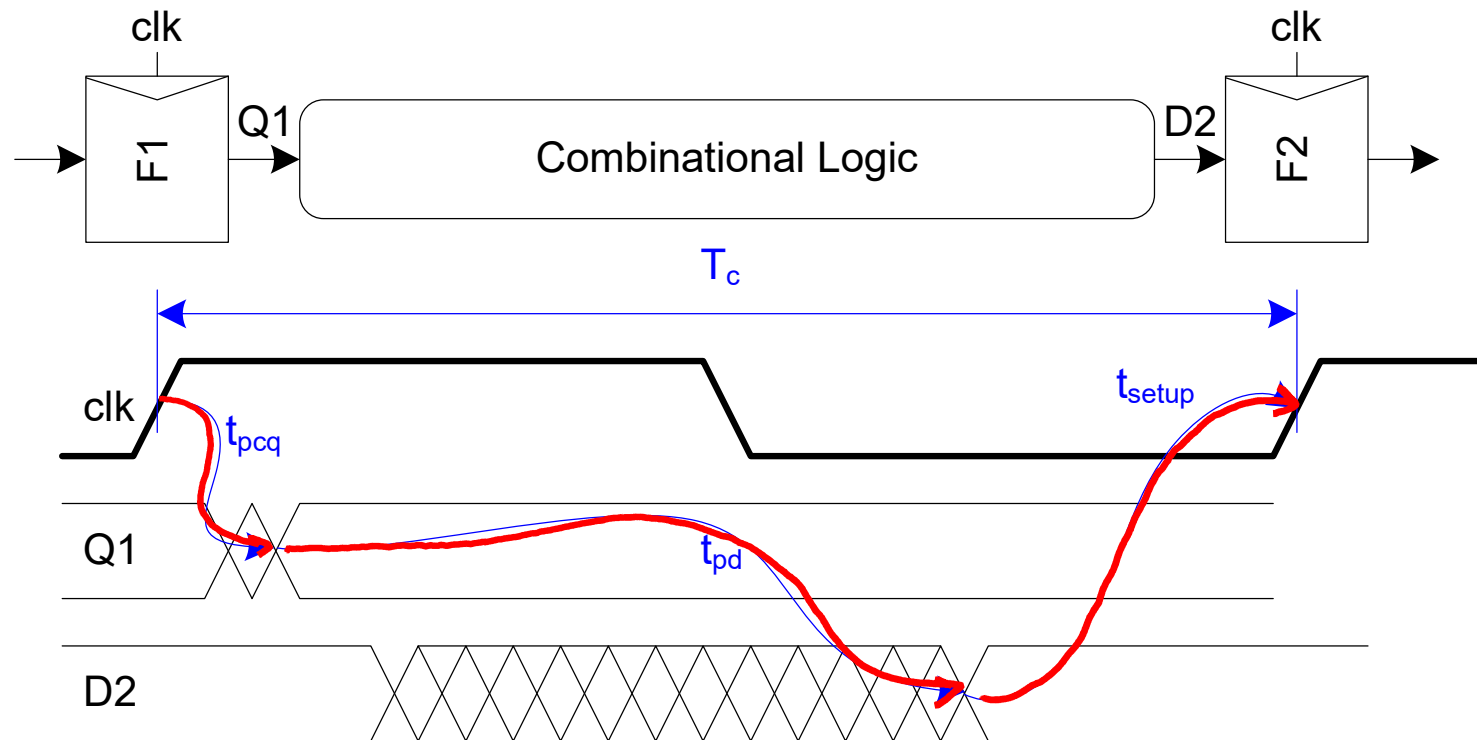
Max/Min Delay Constraints

- Max delay constraint
 - Combinational logic delay cannot be too late, or the receiving element would miss its setup time and sample the wrong value → **setup time failure**
- Min delay constraint
 - Contamination delay of logic cannot be too small, or the data can incorrectly propagate through one clock edge and corrupt the state → **hold-time failure, race condition**
 - Redesign for slower logic

Max Delay: Flip-Flops

$$t_{pd} \leq T_c - \underbrace{(t_{\text{setup}} + t_{pcq})}_{\text{sequencing overhead}}$$

$$t_{cq} + t_p + t_{\text{setup}} < T$$



$\geq t_{\text{hold}} - t_{\text{ccq}}$

CK

$t_{\text{ccq}} + t_{\text{p}} > t_{\text{hold}}$



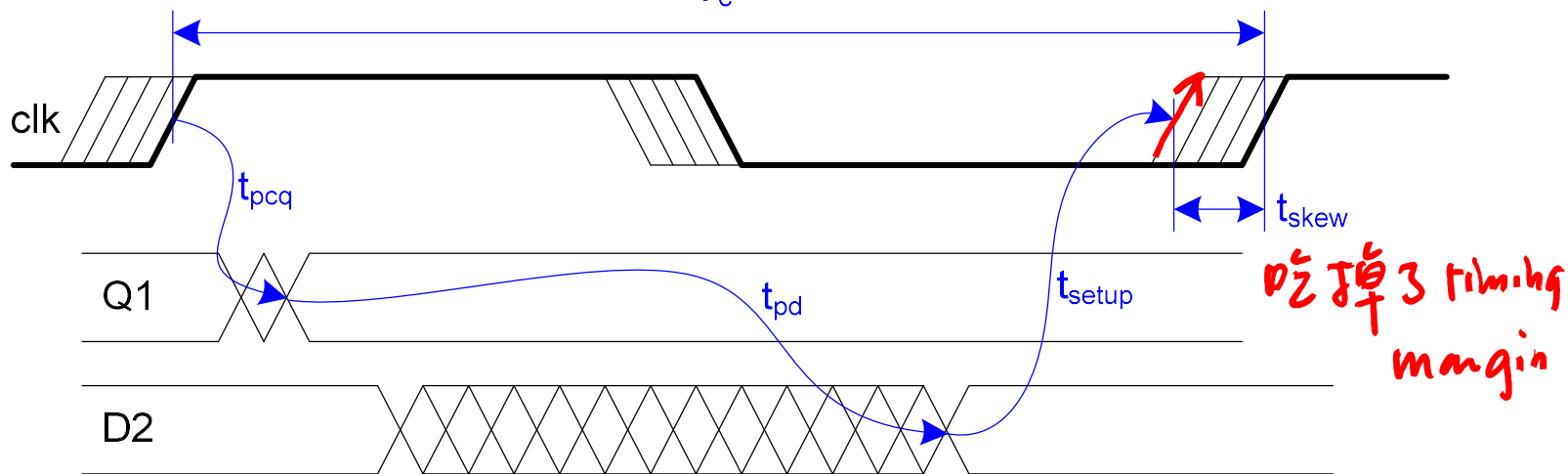
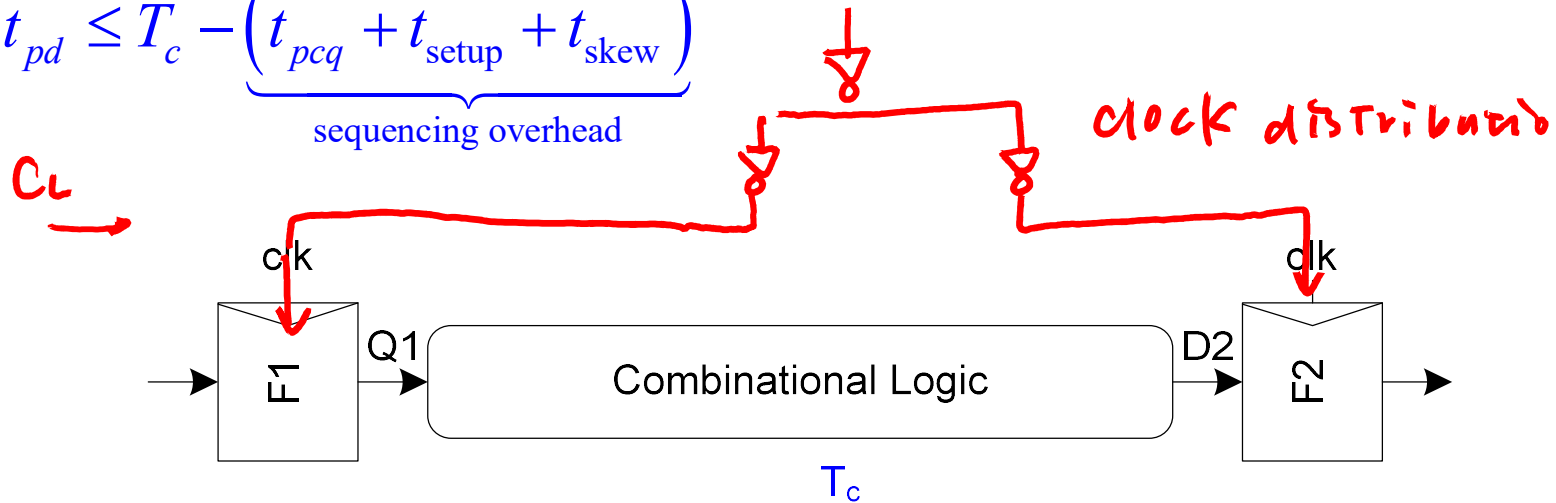
Clock Skew

ideally

- We have assumed zero clock skew
- Clock signals have uncertainty in arrival time

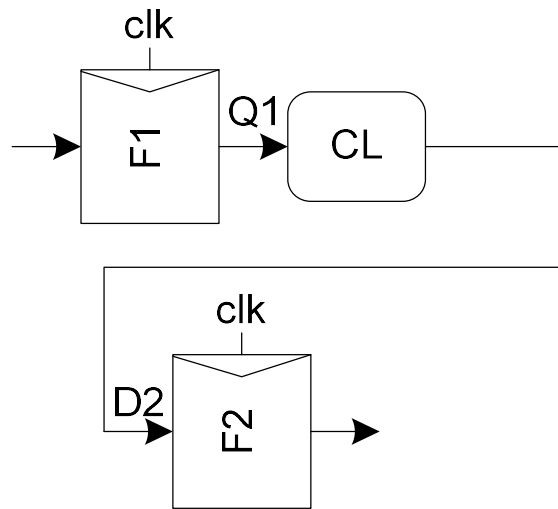
Skew: Flip-Flops: Max Delay

$$t_{pd} \leq T_c - \underbrace{(t_{pcq} + t_{setup} + t_{skew})}_{\text{sequencing overhead}}$$

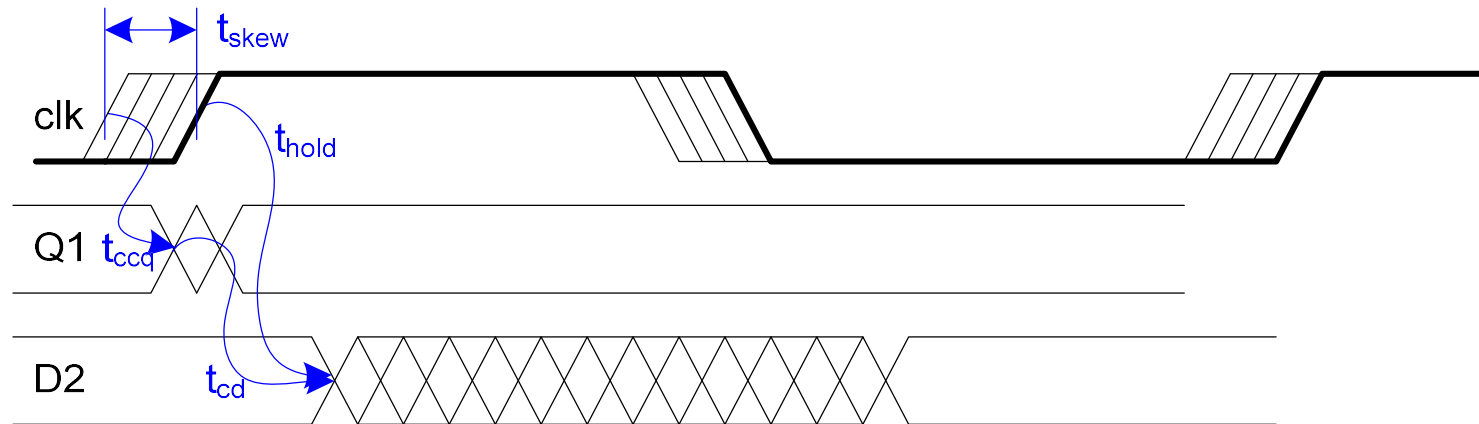


Skew: Flip-Flops: Min Delay

$$t_{cd} \geq t_{\text{hold}} - t_{ccq} + t_{\text{skew}}$$



$$t_{\text{cq}} + t_p > t_{\text{skew}} + t_{\text{hold}}$$



Max Delay: 2-Phase Latches

$$t_{pd} = t_{pd1} + t_{pd2}$$

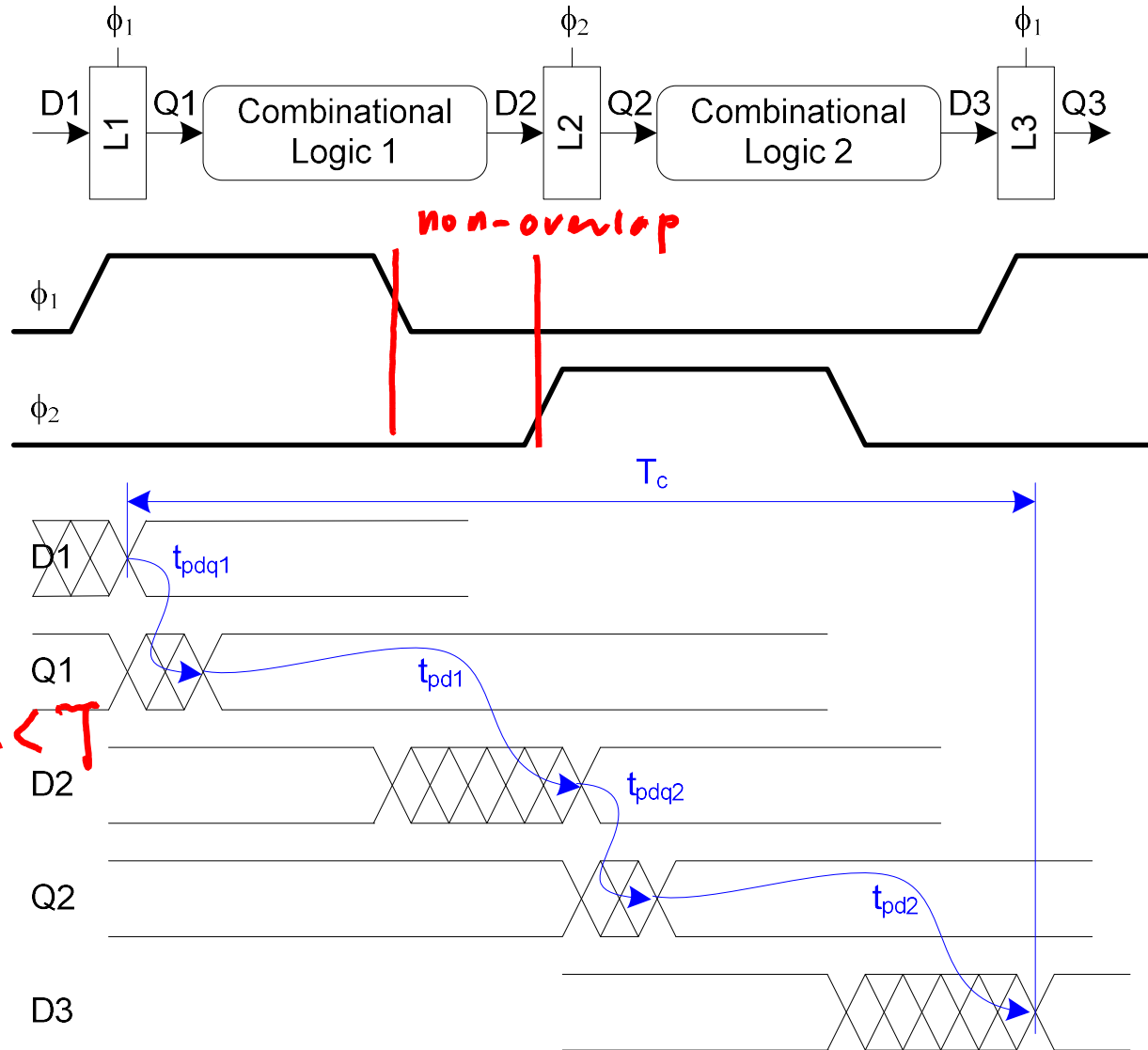
$$\leq T_c - \underbrace{(2t_{pdq})}_{\text{sequencing overhead}}$$

sequencing overhead

in average

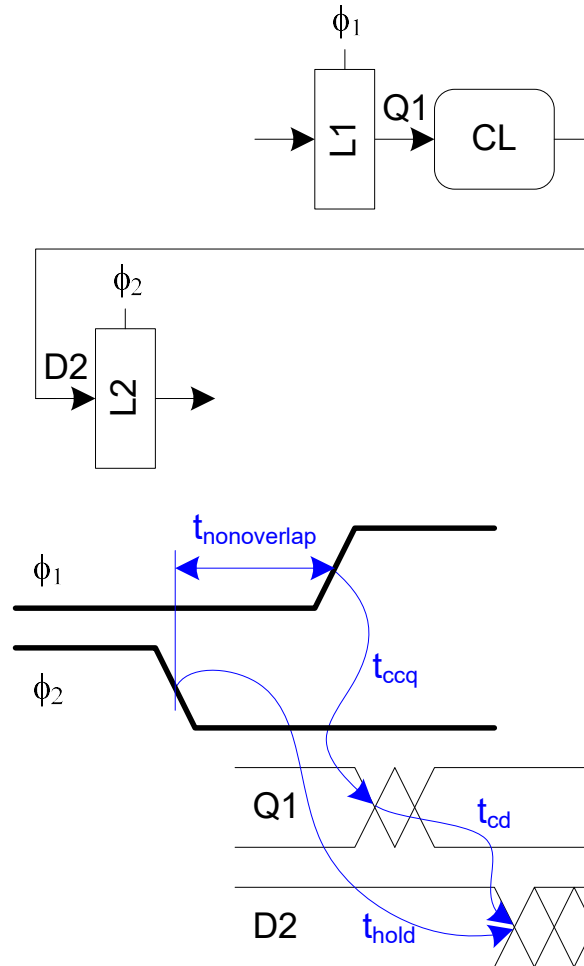
$$t_{D1 \rightarrow D3} < T$$

$$t_{D2 \rightarrow D1} + t_{p1} + t_{D2 \rightarrow D3} + t_{p2} < T$$



Min Delay: 2-Phase Latches

$$t_{cd1}, t_{cd2} \geq t_{\text{hold}} - t_{ccq} - t_{\text{nonoverlap}}$$



- Hold time error reduced by non-overlap compared to flip-flops

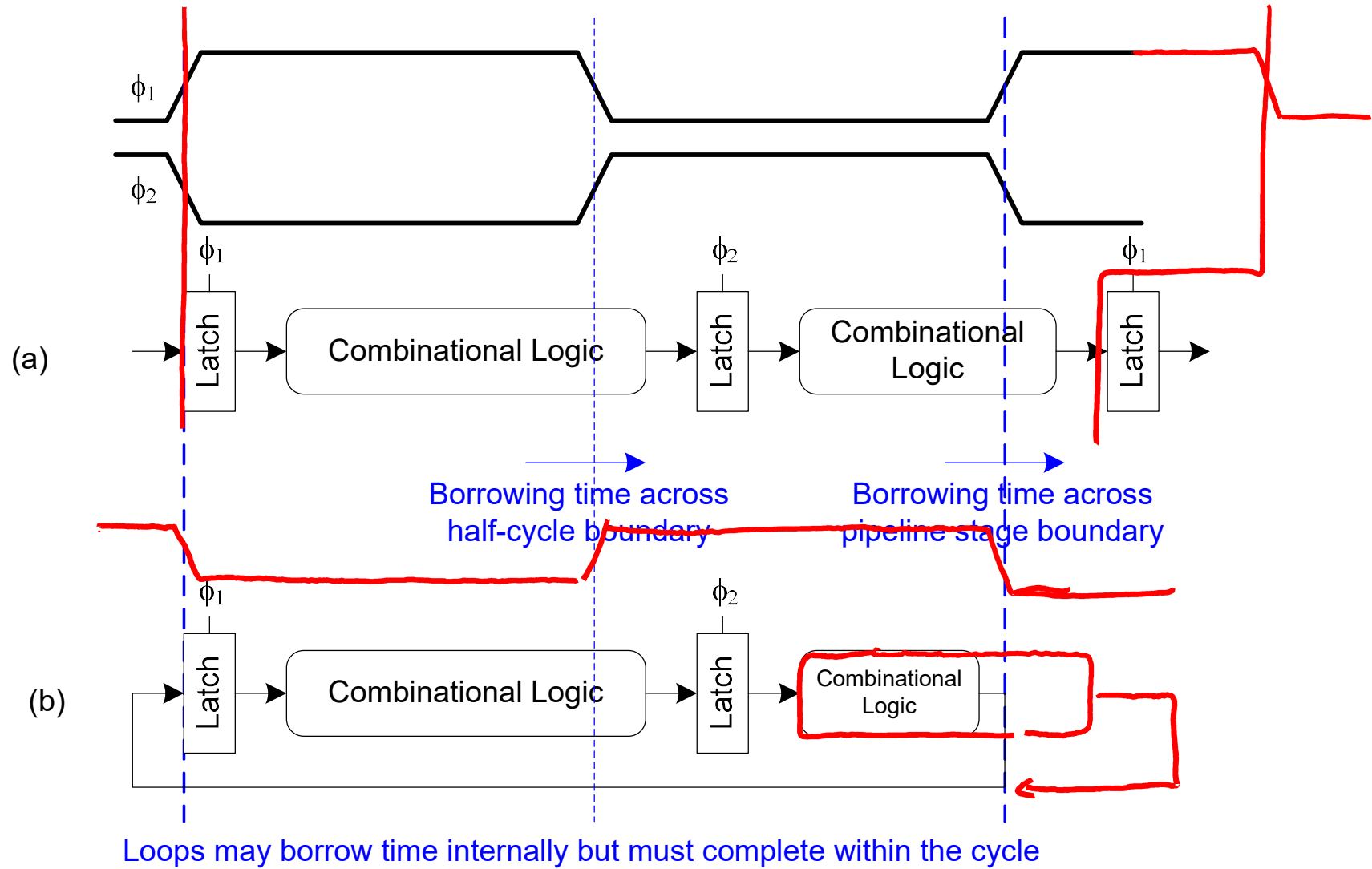
→ Internal race condition of flip-flops

$$t_{\text{nonoverlap}} + t_{ccq} + t_{cd} \geq t_{\text{hold}}$$

Timing Borrowing

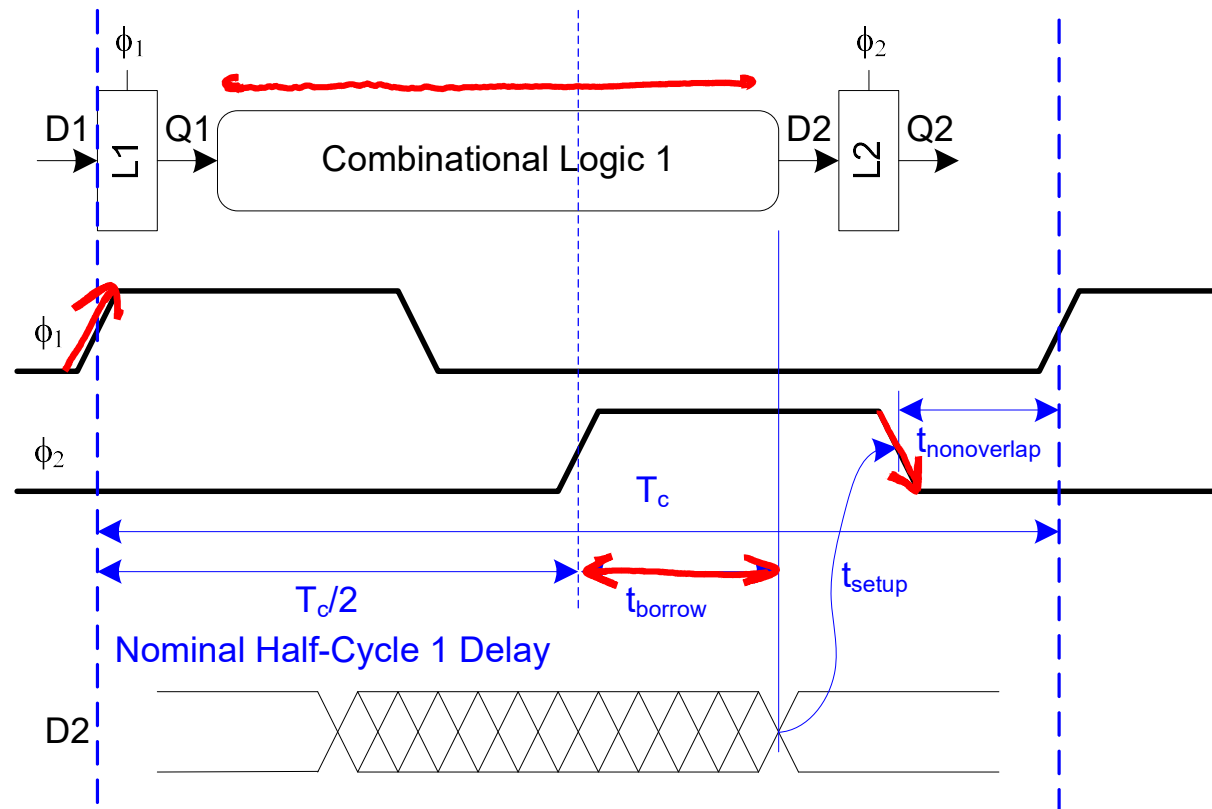
- In flip-flop systems
 - Data launch on one rising edge
 - Must setup before next rising edge
 - If data arrive late → system fails
 - If data arrive early → system fails
 - **Flip-flop systems have hard edges**
- In latched-based systems
 - Data can pass through latches while transparent
 - Long logic delay can borrow time into next cycle
 - As long as each loop completes in time

Time Borrowing Example



How Much Borrowing?

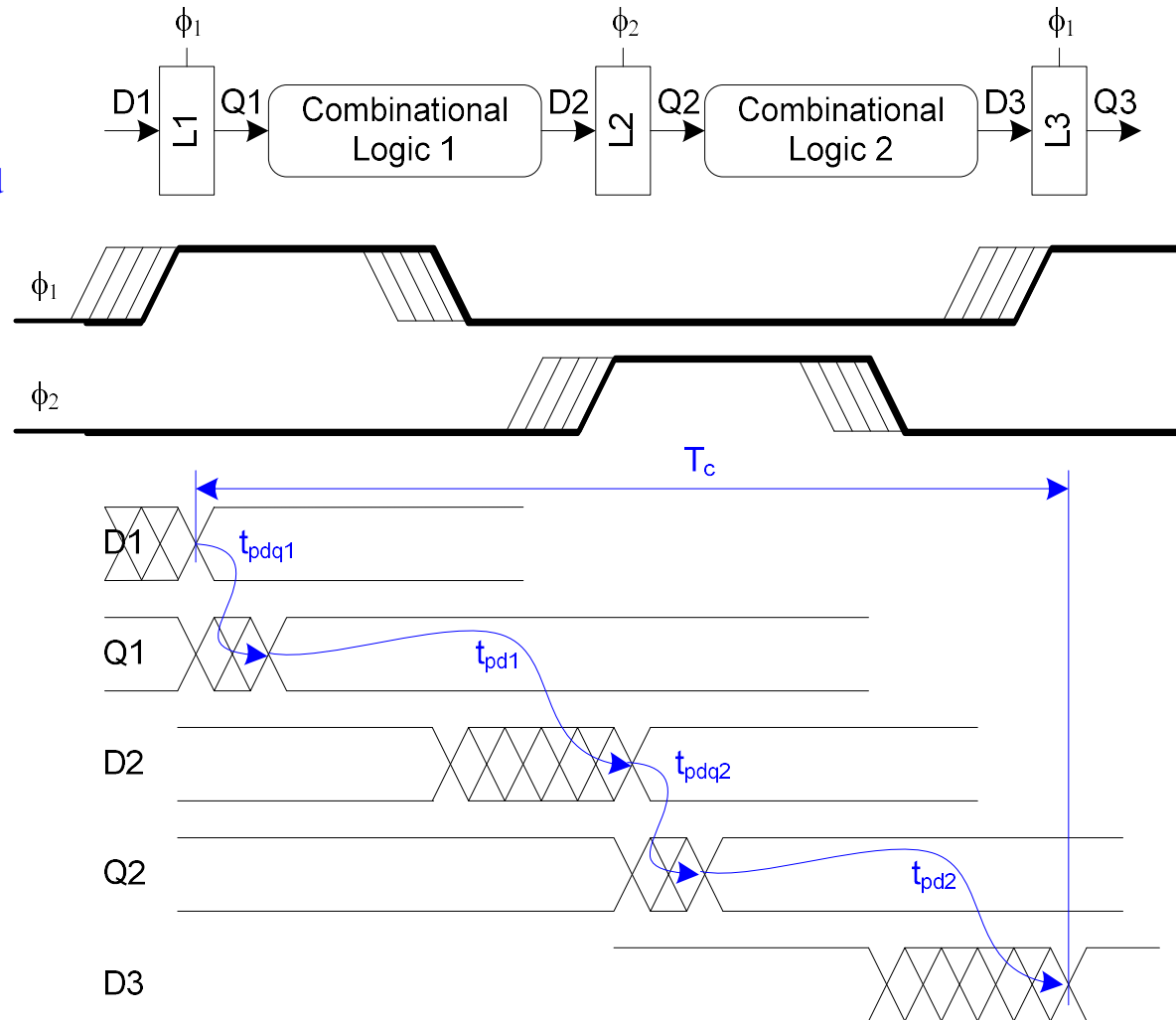
$$t_{\text{borrow}} \leq \frac{T_c}{2} - (t_{\text{setup}} + t_{\text{nonoverlap}})$$



Skew: 2-Phase Latches: Max Delay

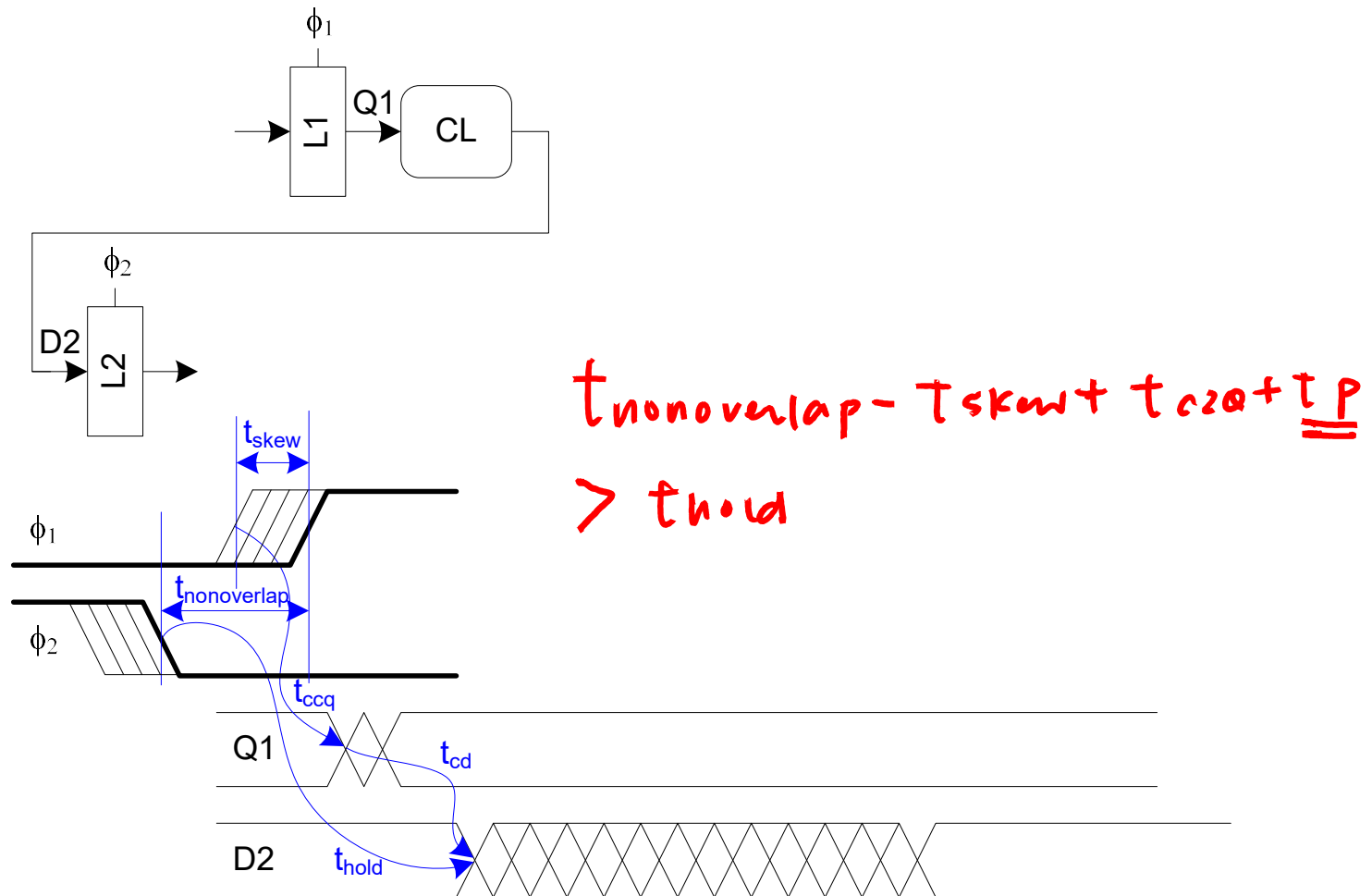
$$t_{pd} \leq T_c - \underbrace{(2t_{pdq})}_{\text{sequencing overhead}}$$

- Latch-based designs are **skew-tolerant**



Skew: 2-Phase Latches: Min Delay

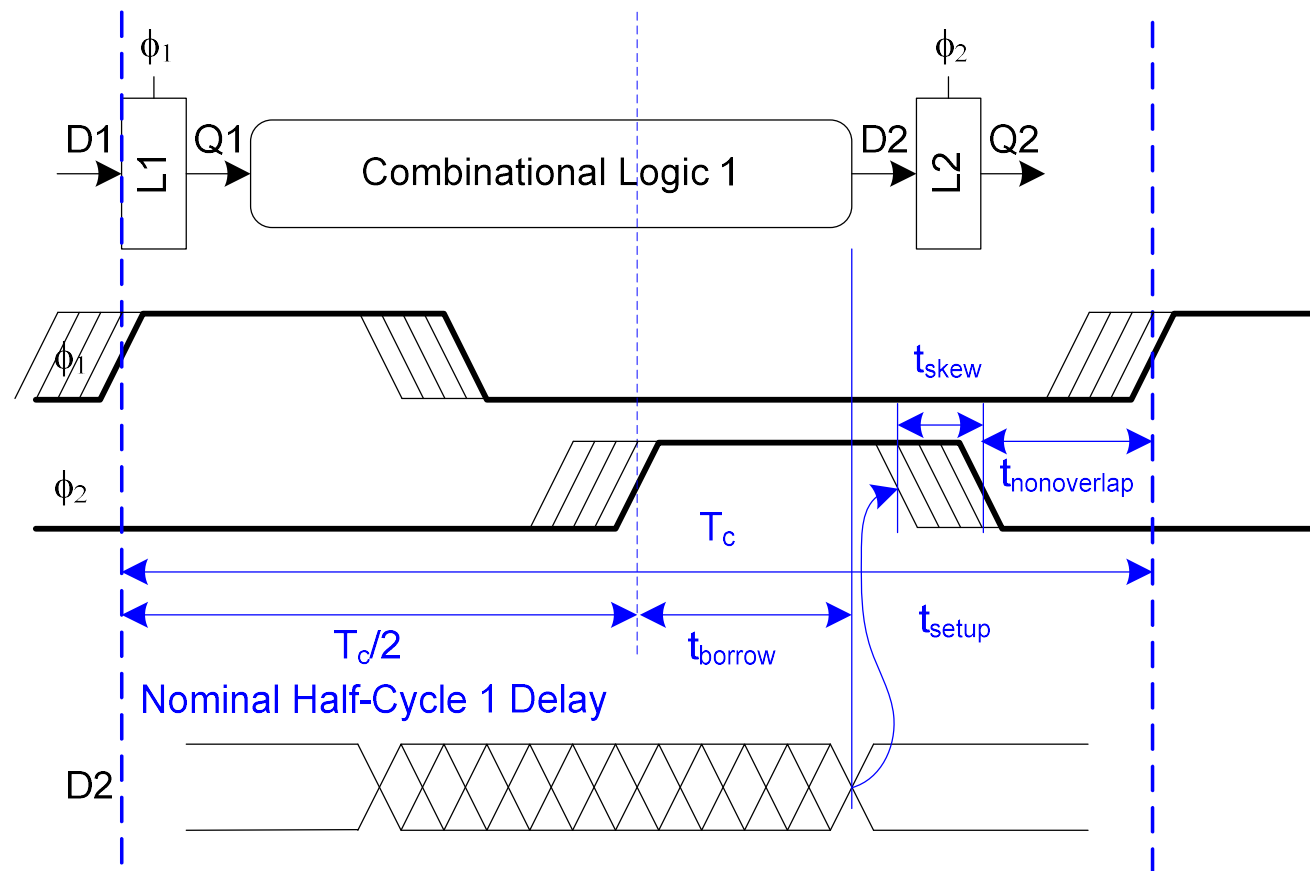
$$t_{cd1}, t_{cd2} \geq t_{\text{hold}} - t_{ccq} - t_{\text{nonoverlap}} + t_{\text{skew}}$$



Skew: How Much Borrowing?

$$t_{\text{borrow}} \leq \frac{T_c}{2} - (t_{\text{setup}} + t_{\text{nonoverlap}} + t_{\text{skew}})$$

讓你第一個寄存器時間變少



Clock Skew

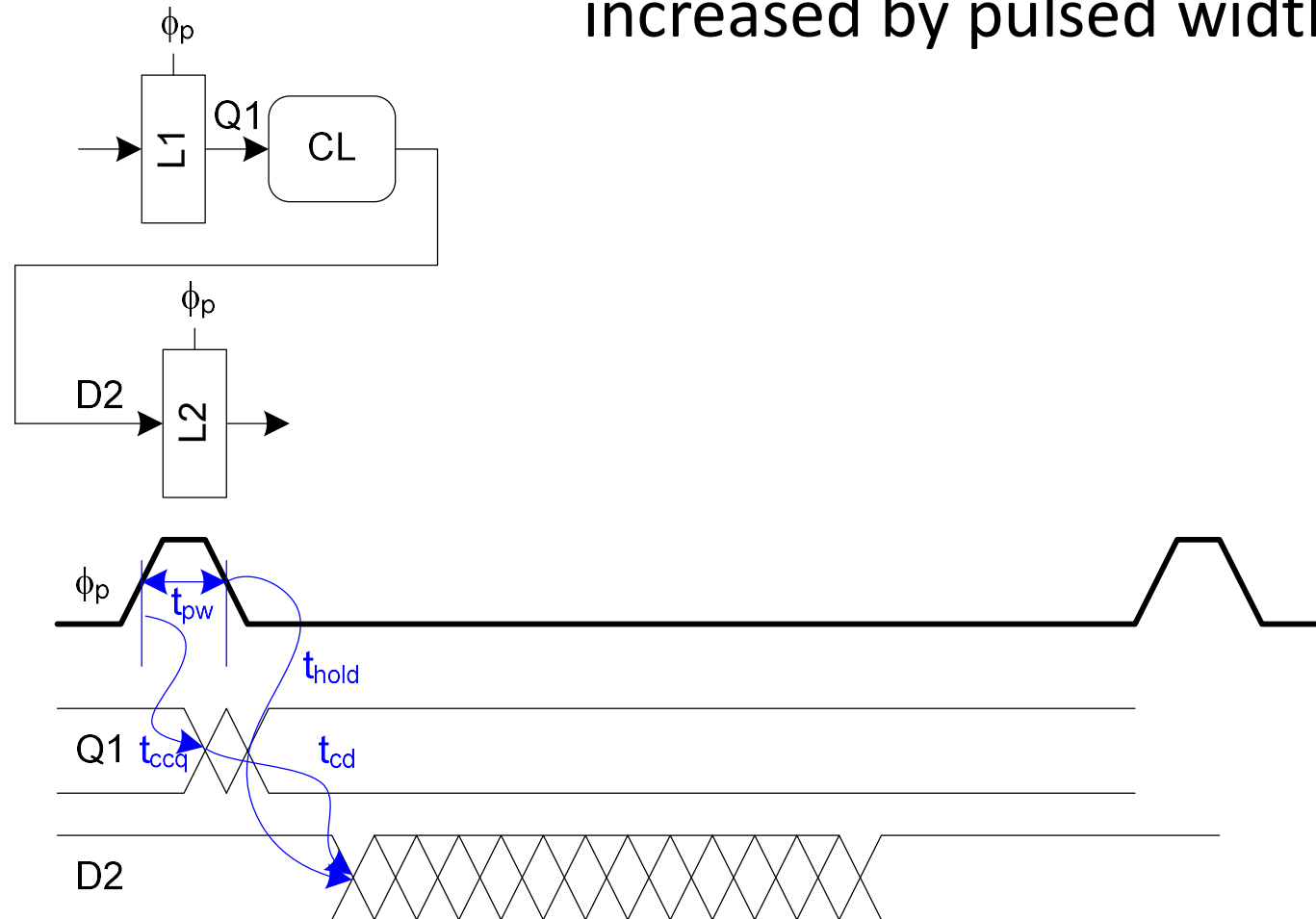
- We have assumed zero clock skew
- Clock signals have uncertainty in arrival time
→ Negative impact on timing margin
 - Decrease setup time margin
(effectively increase maximum delay)
 - Decrease hold time margin
(effectively decrease contamination delay)
 - Decrease timing borrowing

$$t_{pd} \leq T_c - \underbrace{\max(t_{pdq}, t_{pcq} + t_{\text{setup}})}_{\text{sequencing overhead}} - t_{pw} \quad \phi_p$$


Min Delay: Pulsed Latches

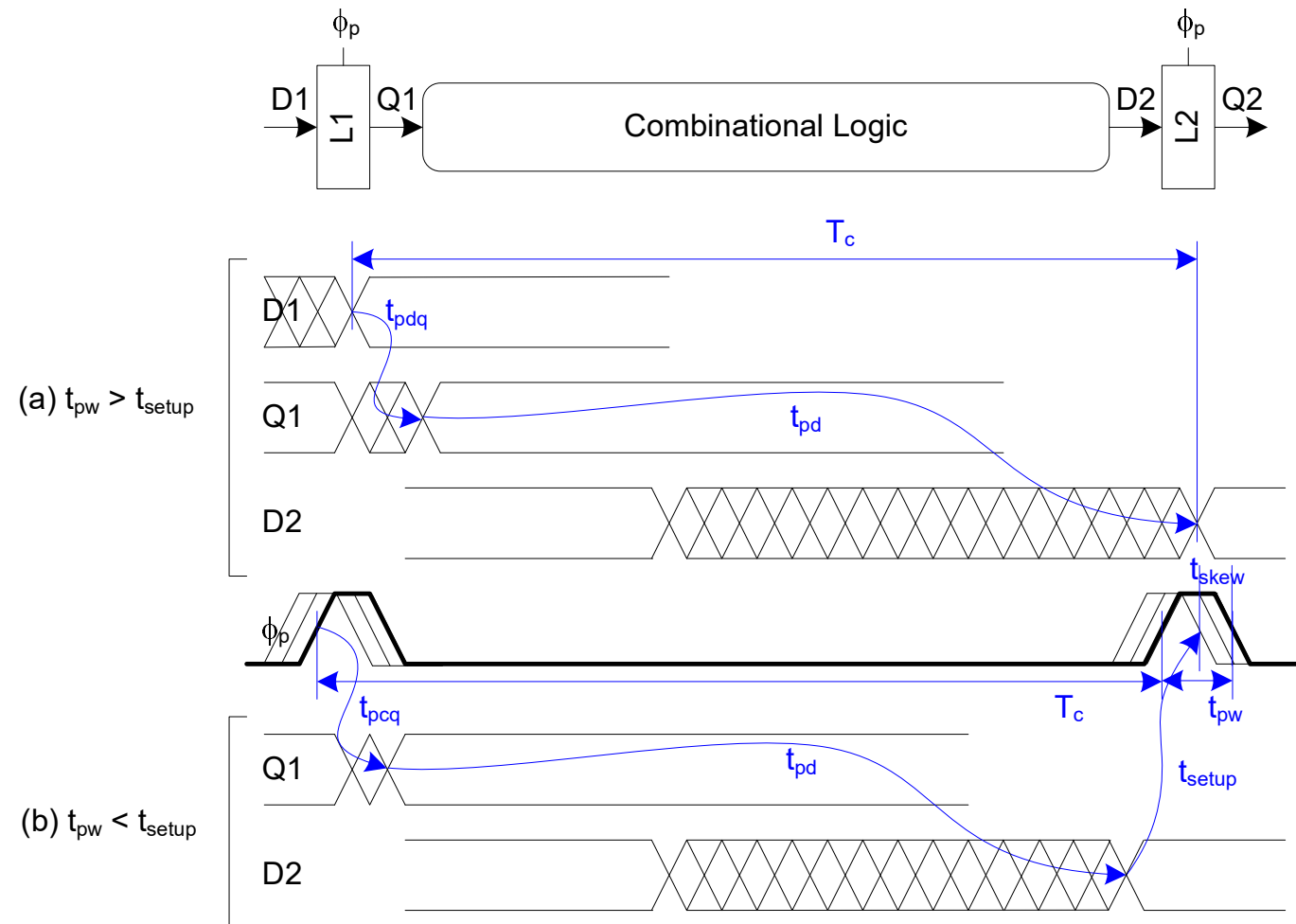
$$t_{cd} \geq t_{\text{hold}} - t_{ccq} + t_{pw}$$

- Hold-time error increased by pulsed width



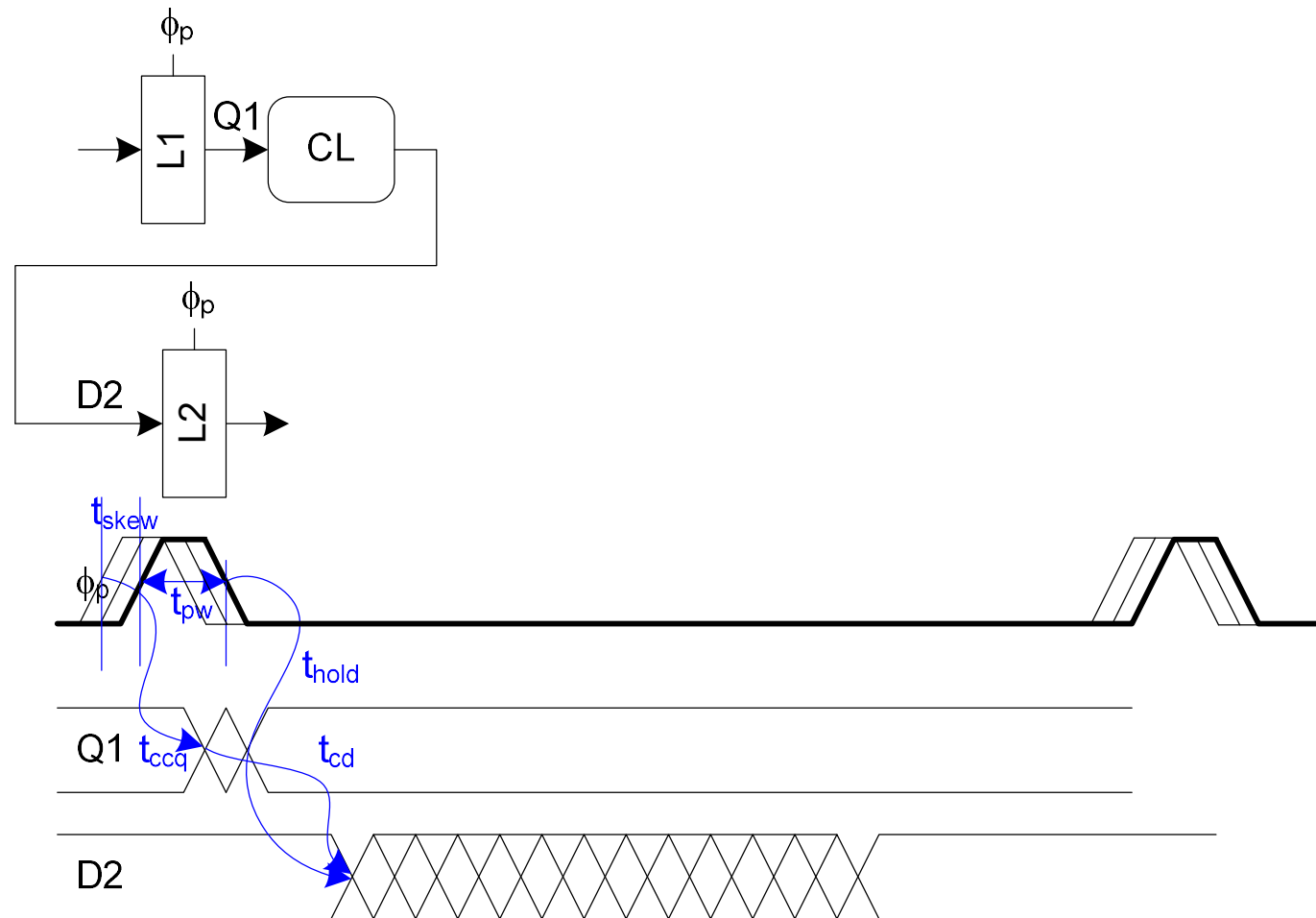
Skew: Pulsed Latches: Max Delay

$$t_{pd} \leq T_c - \underbrace{\max(t_{pdq}, t_{pcq} + t_{\text{setup}} - t_{pw} + t_{\text{skew}})}_{\text{sequencing overhead}}$$



Skew: Pulsed Latches: Min Delay

$$t_{cd} \geq t_{\text{hold}} + t_{pw} - t_{ccq} + t_{\text{skew}}$$



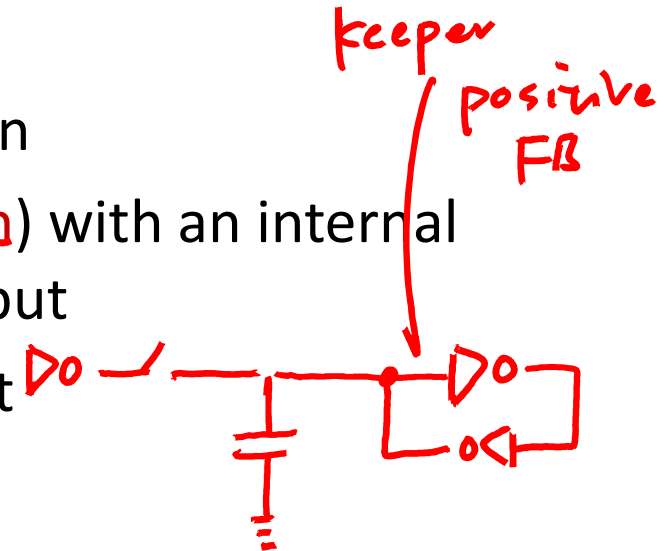
Outline

- Sequencing Methods
- Max/Min Delay, Clock Skew, Timing Borrowing
- **Sequencing Element Design**

Static vs. Dynamic Storage

- Static Storage

- Preserve state as long as power is on
- Use positive feedback (regeneration) with an internal connection between output and input
- Useful when updates are infrequent



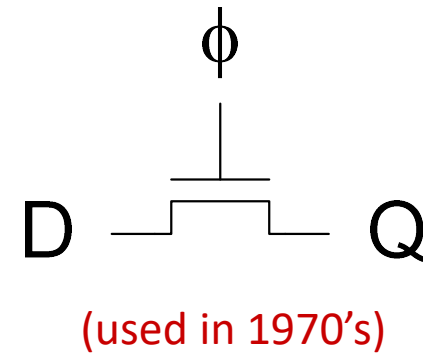
- Dynamic Storage

- Store state on parasitic capacitors
- Hold state only for short period of time (milli-seconds)
- Require periodic refresh
- Simpler design, higher speed, and lower power

meta stable
point
 $V_{ih} = V_{out} = 0.5 V_{DD}$

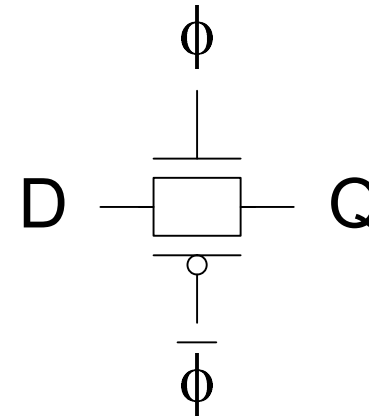
Latch Design (I)

- Pass-transistor latch
- Pros:
 - Tiny
 - Low load on clock signal
- Cons:
 - V_{th} drop (not rail-to-rail)
 - Non-restoring
 - Back-driving
 - Output noise sensitivity
 - Dynamic (float when opaque)
 - Diffusion input



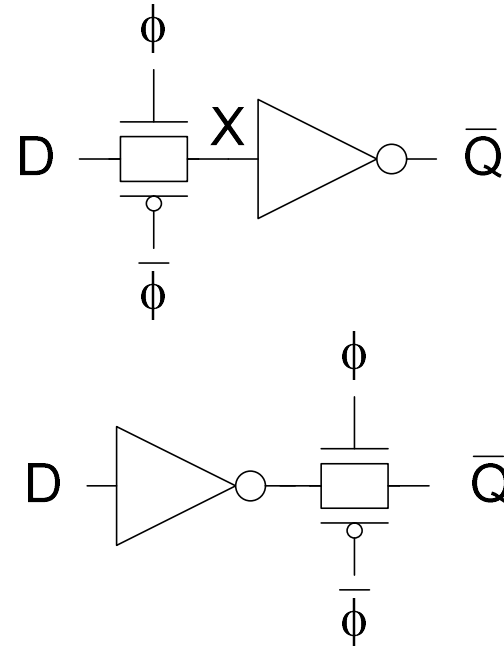
Latch Design (II)

- Transmission gate
- Pros:
 - No V_{th} drop
- Cons:
 - Inverted clock signal required



Latch Design (III)

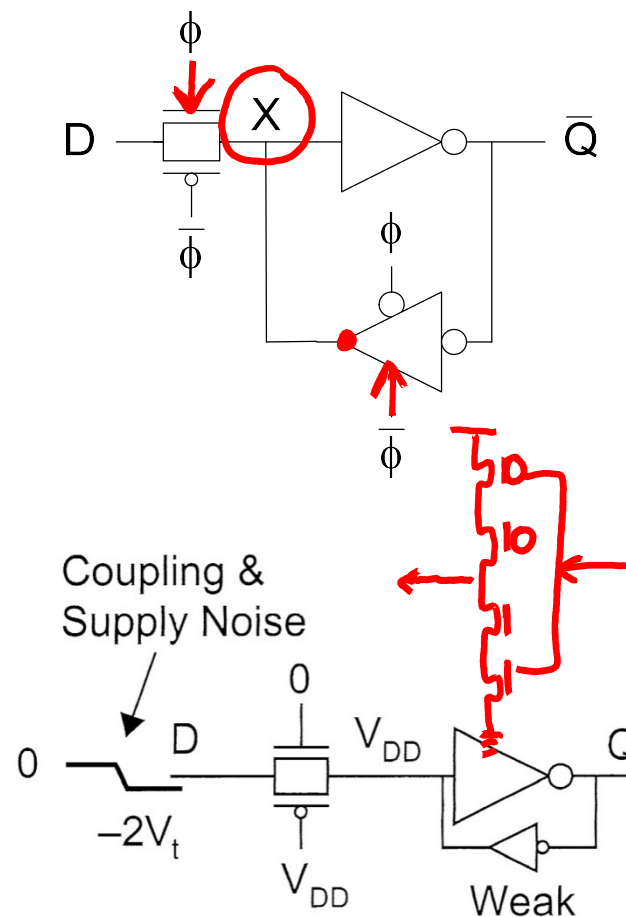
- Buffered version
- Pros:
 - Restoring
 - No back-driving that fixes either one of the followings
 - Output noise sensitivity
 - Diffusion input
- Cons:
 - Inverted output



$D \rightarrow \overline{DQ} \rightarrow \overline{DQ} \rightarrow Q$

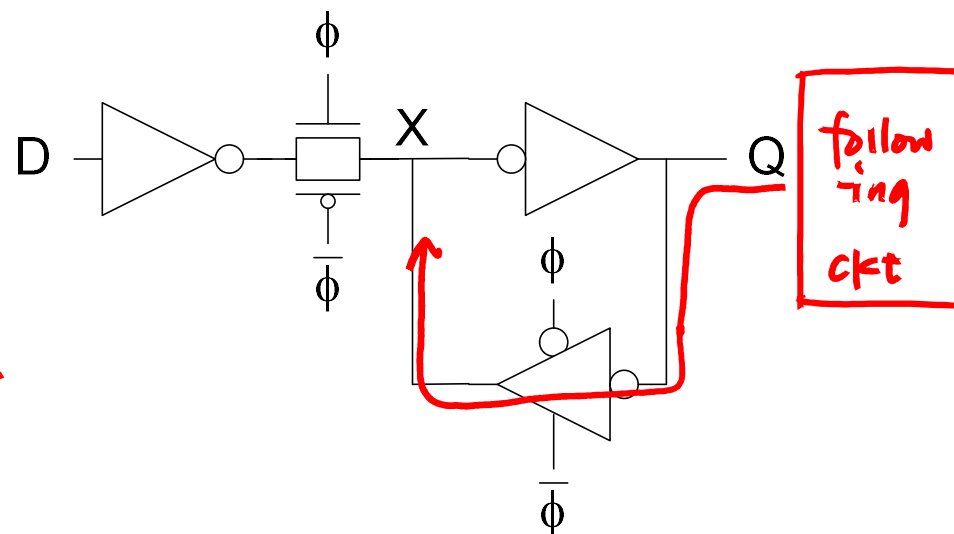
Latch Design (IV)

- Tristate feedback
- Pros:
 - Static
 - Static latches are now essential
- Cons:
 - Back-driving risk
 - Diffusion input
 - Noise can sneak through OFF transmission gate and destroy output



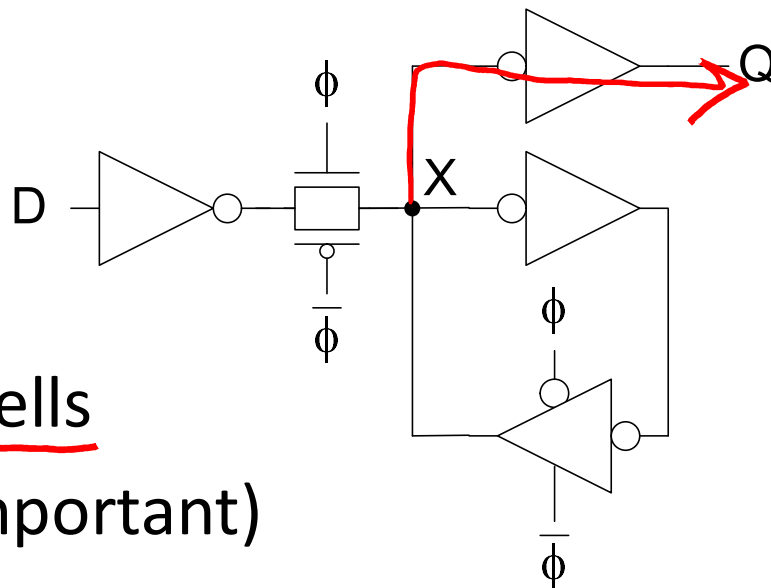
Latch Design (V)

- Buffered input with feedback
- Pros:
 - Fixes diffusion input
 - Non-inverting
- Cons:
 - Output noise back-driving



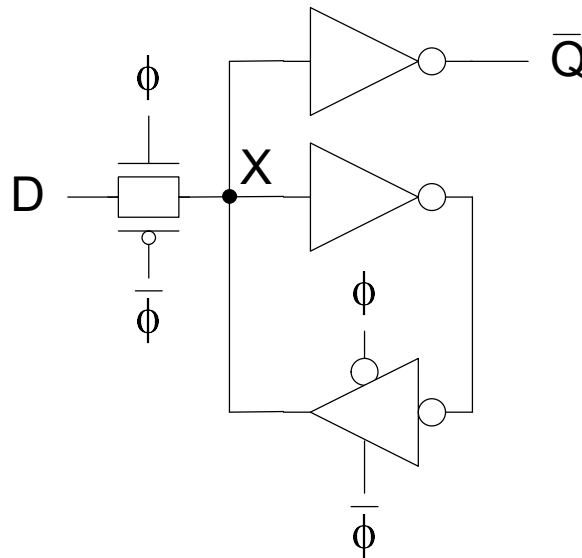
Latch Design (VI)

- Buffered output
 - No back-driving
- Widely used in standard cells
- Pros: very robust (most important)
- Cons:
 - Rather large
 - Rather slow (1.5 to 2 FO4 delays)
 - High clock loading



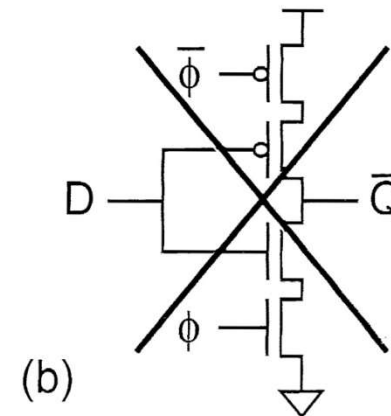
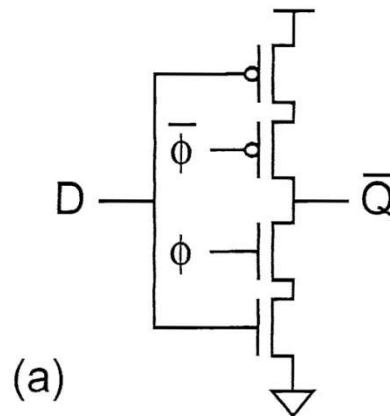
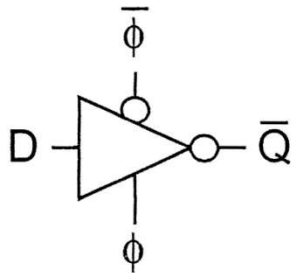
Latch Design (VII)

- Another alternative
 - Smaller and faster
 - Need to be careful with noise control



Clocked CMOS: C²MOS

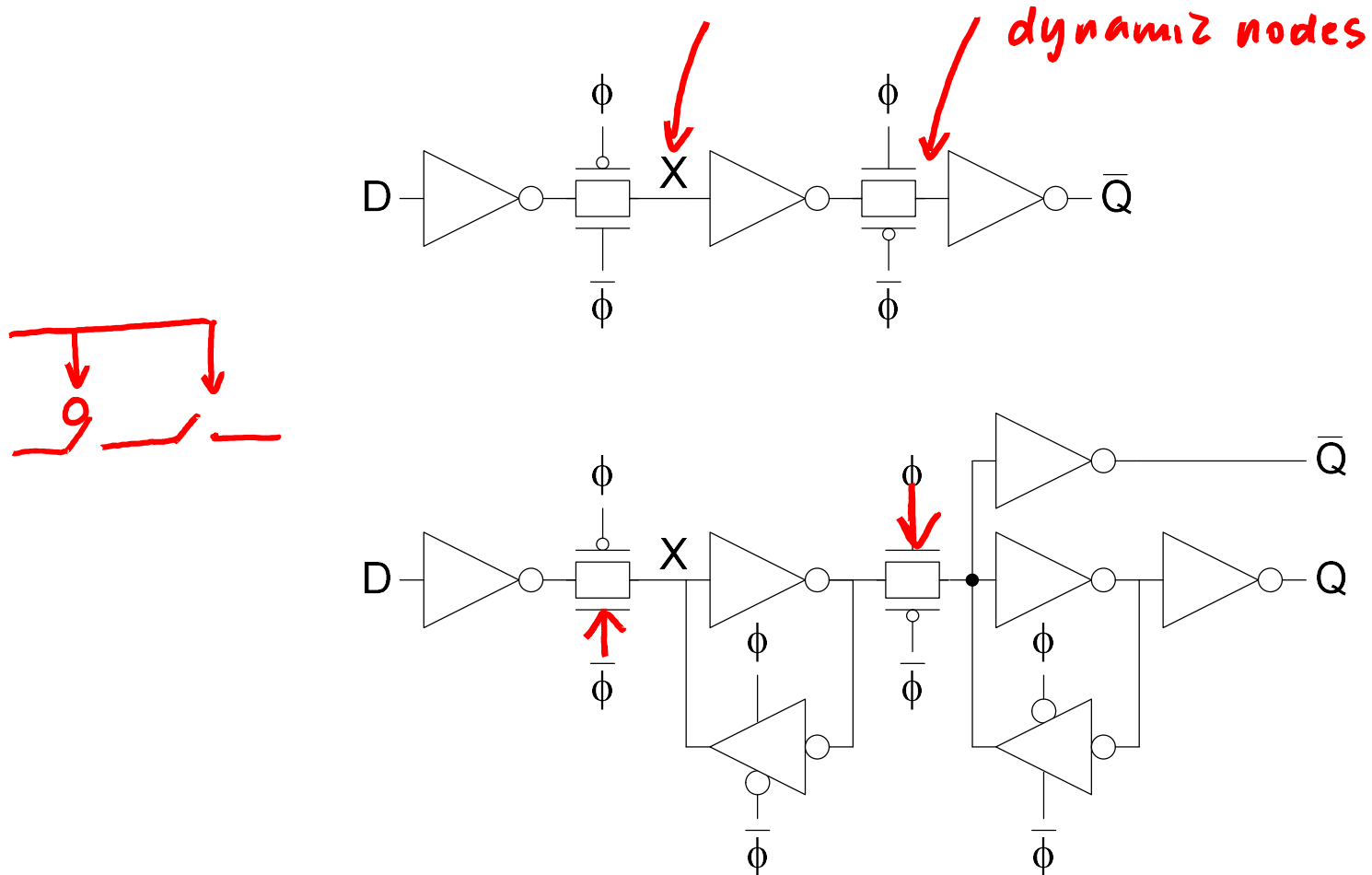
- C²MOS latch
 - Smaller
 - Slower



Bad design: toggling in D causes charge sharing noise when opaque

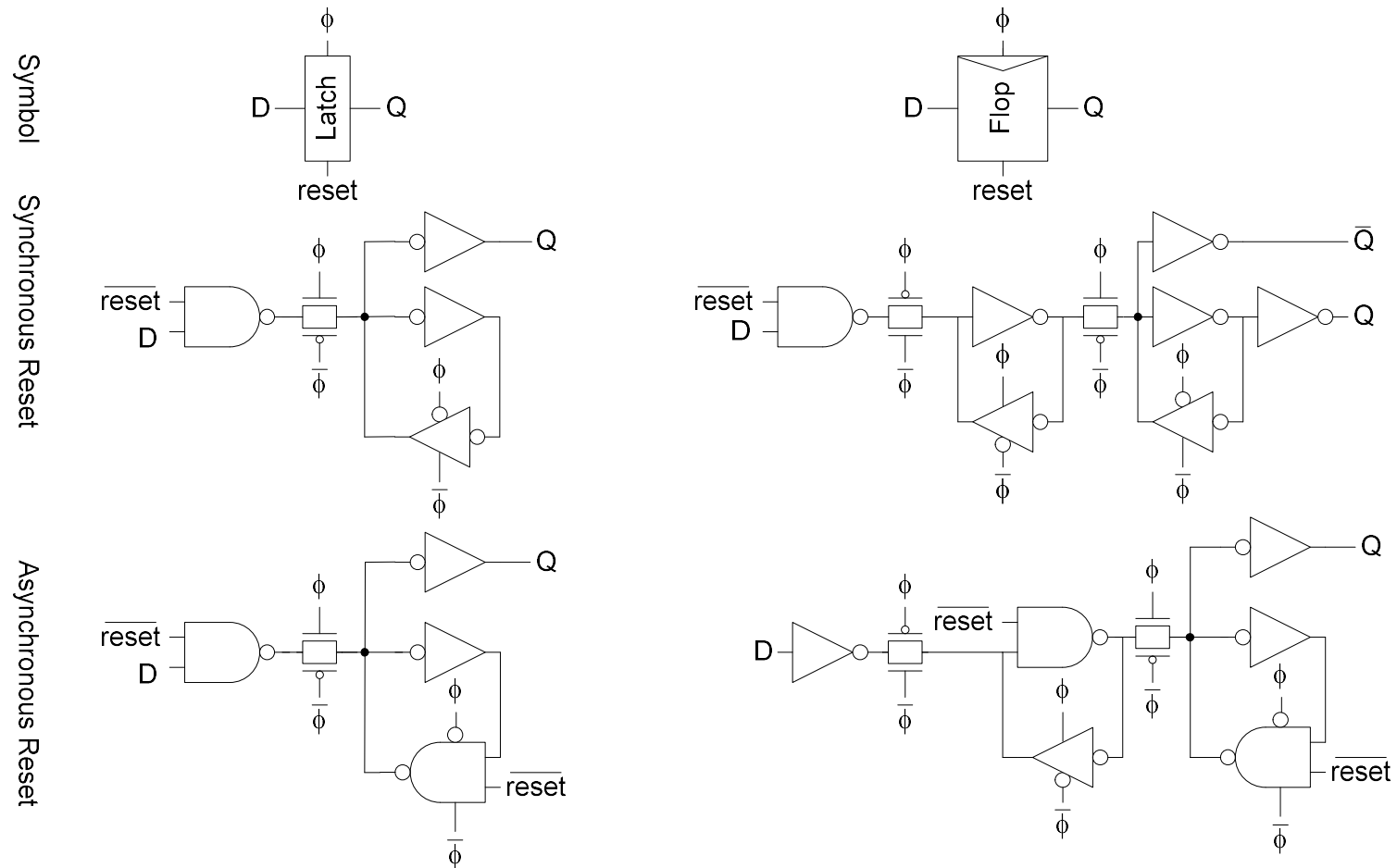
Flip-Flop Design

- Flip-flop is built as a pair of back-to-back latches



Reset

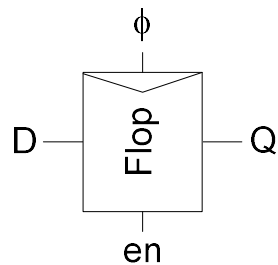
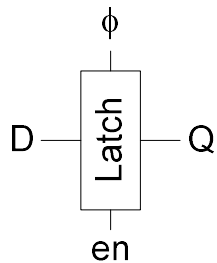
- Force output low when **reset** signal is asserted
- Synchronous vs. asynchronous



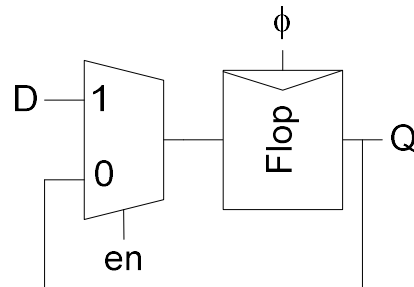
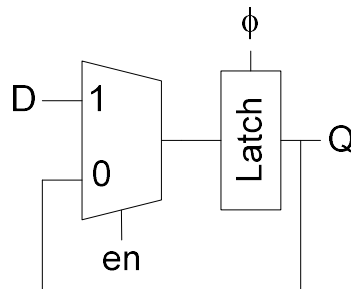
Enable

- Ignore clock when $en = 0$
 - MUX: increase D-to-Q delay
 - Clock gating: increase set-up time and skew

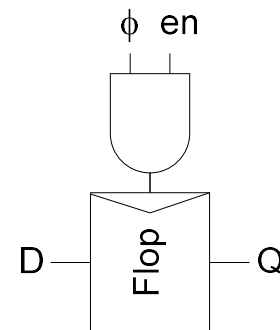
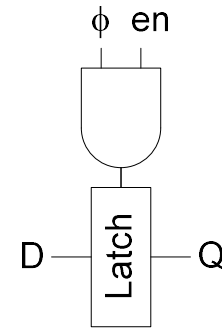
Symbol



Multiplexer Design

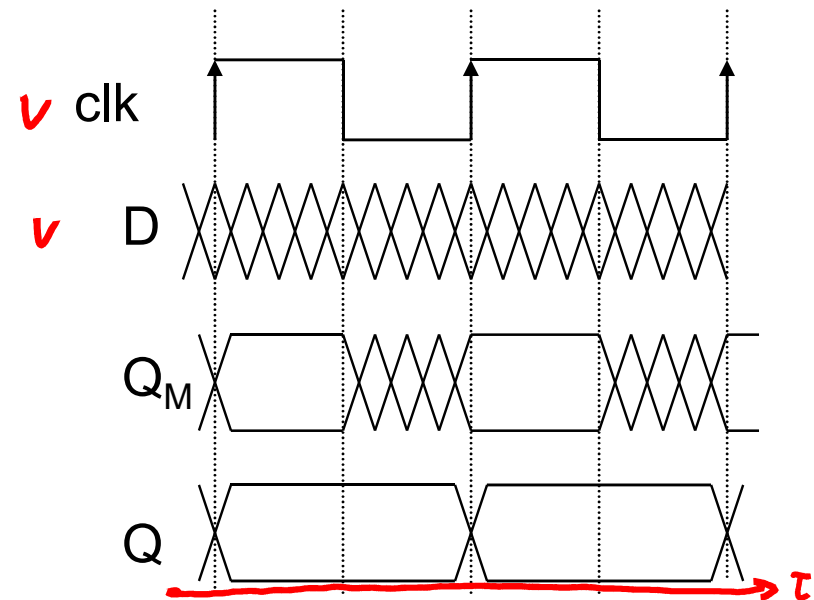
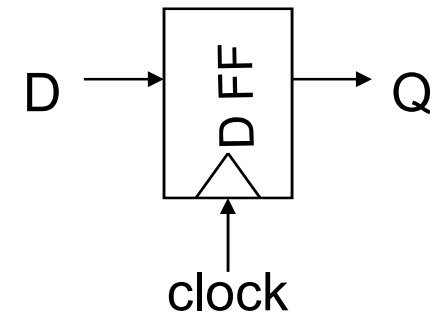
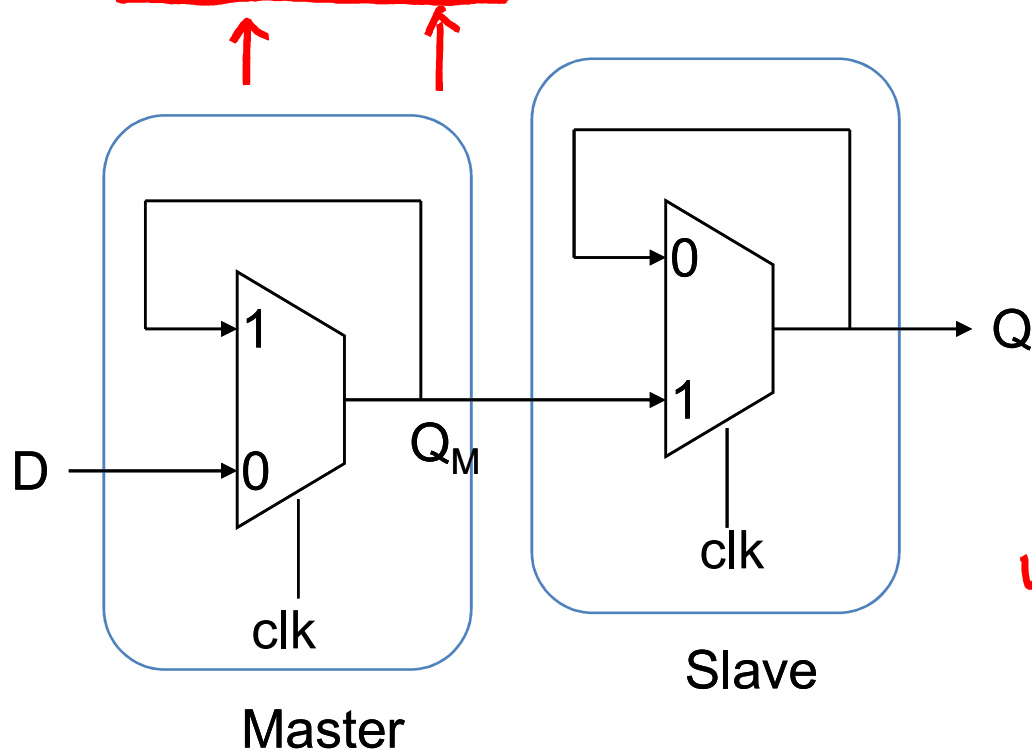


Clock Gating Design



Sequencing Elements Characterization

- Master-slave-based edge-triggered flip-flop



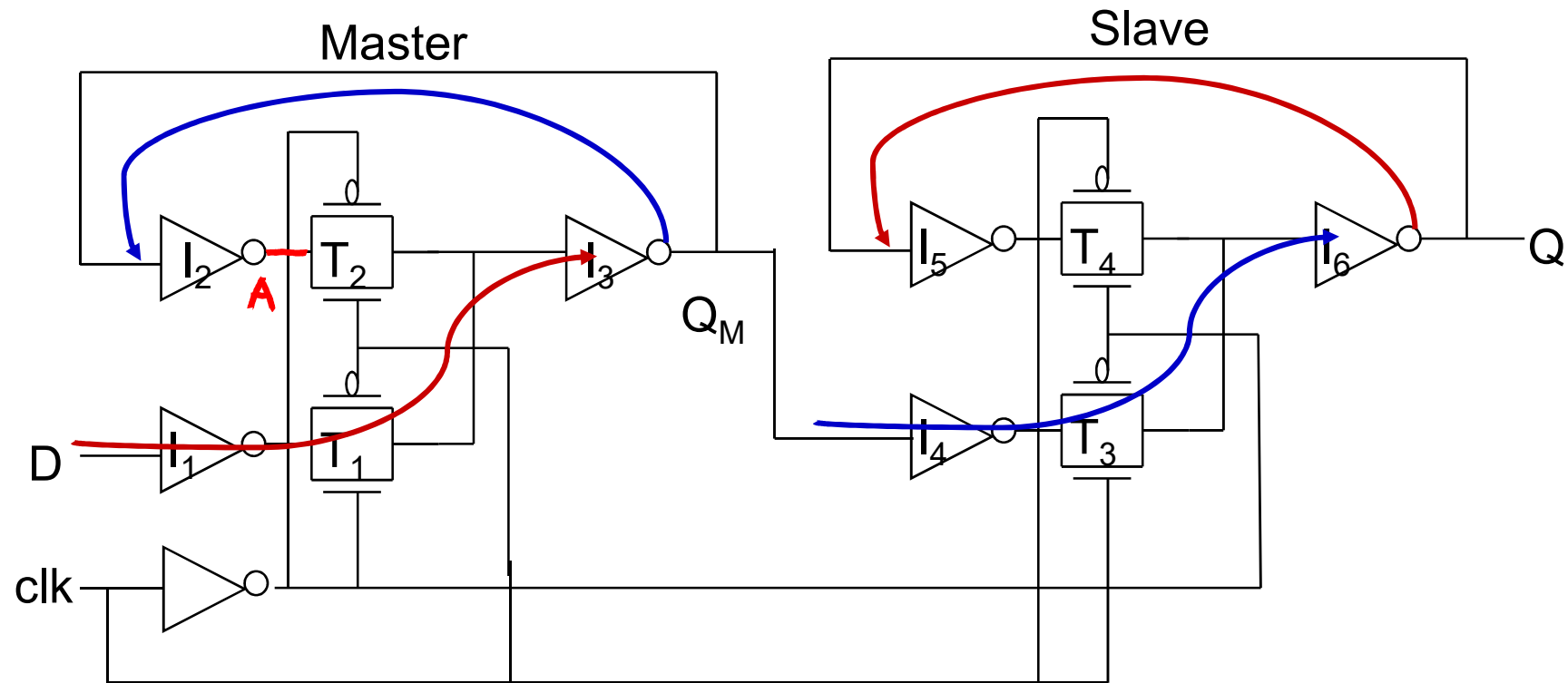
clk = 0 transparent

hold

clk = 0 \rightarrow 1 hold

transparent

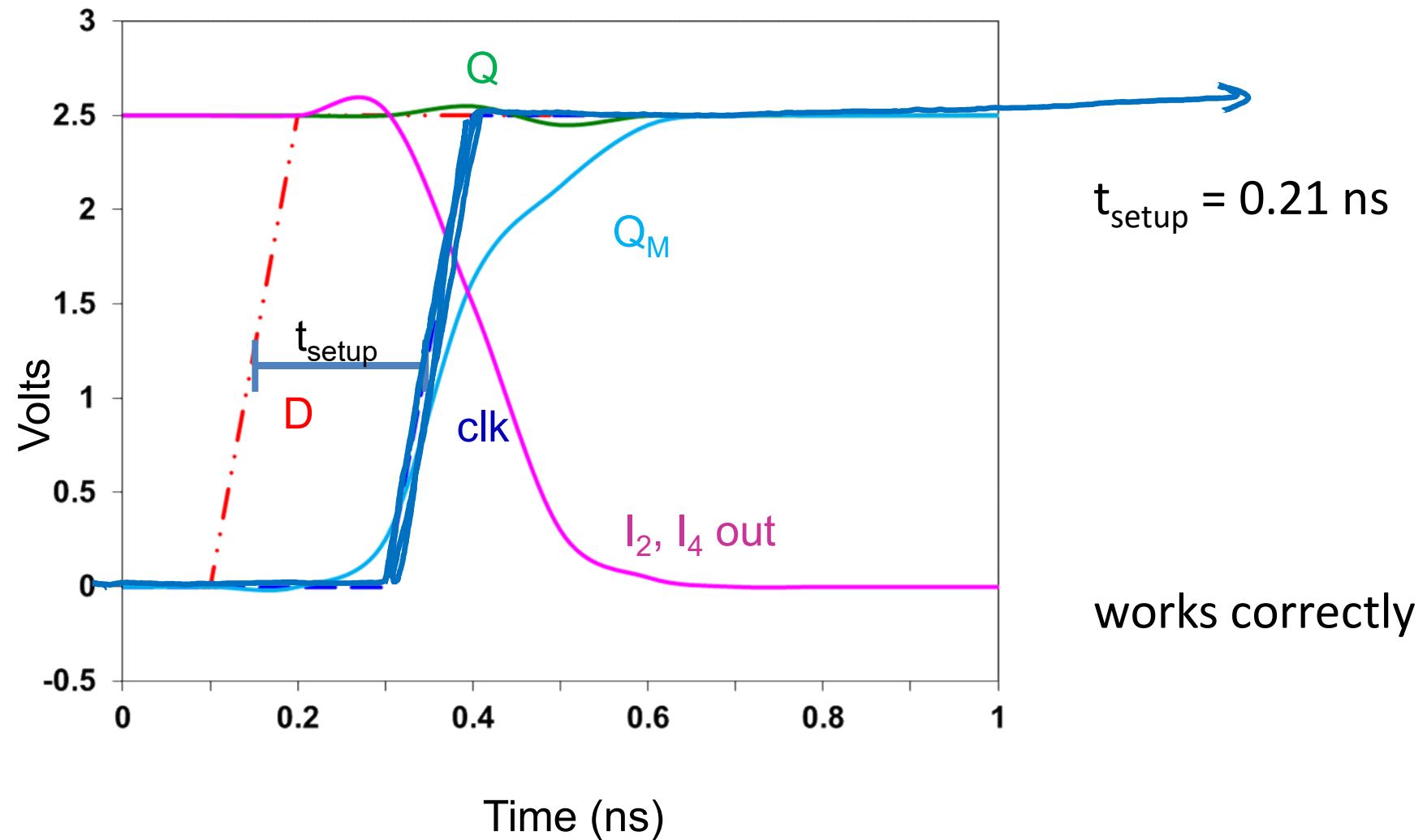
Implementation



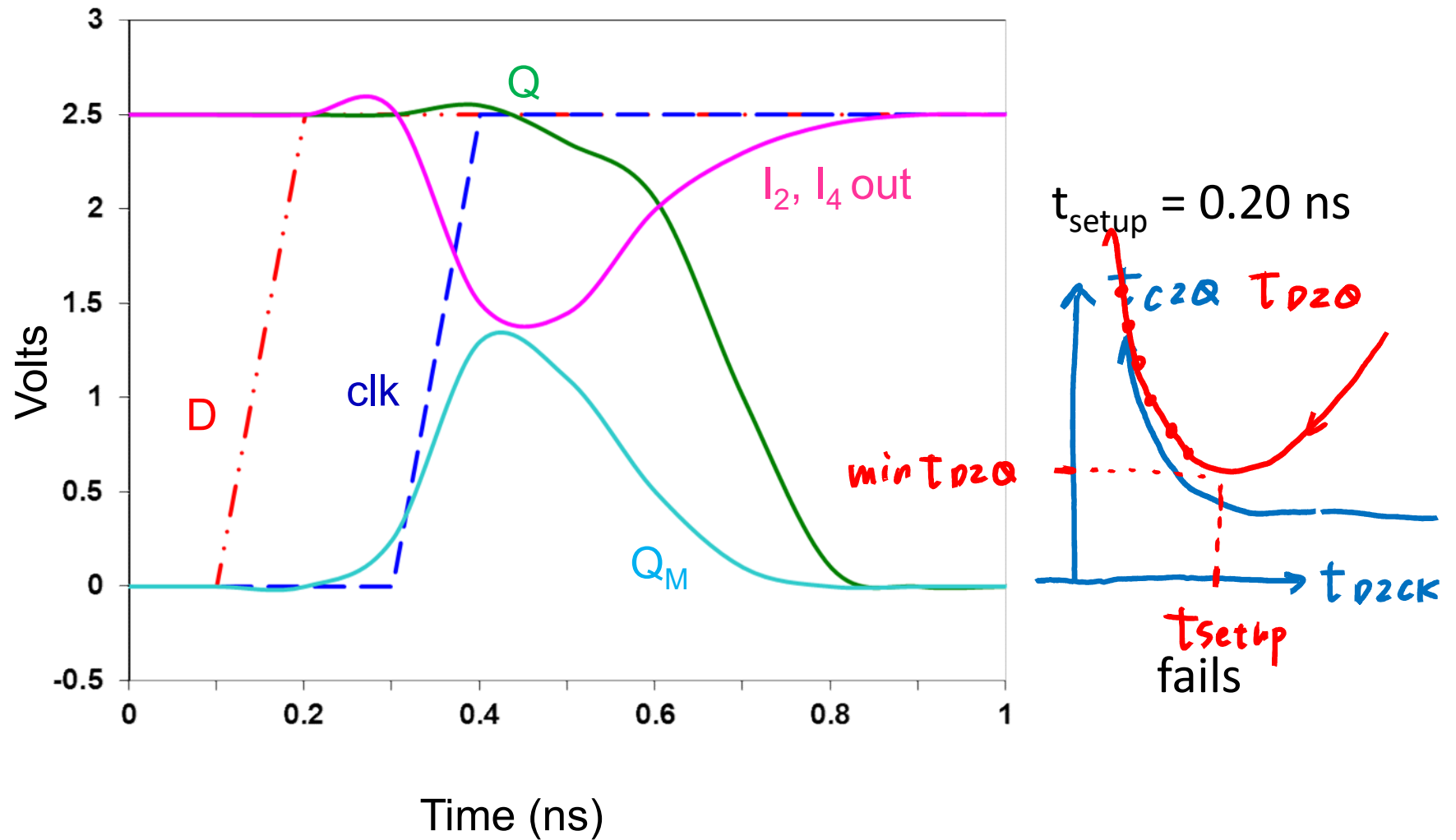
Timing Properties

- Assume
 - Propagation delays are t_{pd_inv} and t_{pd_tx}
 - Contamination delay is 0
 - Inverter delay to clk is 0
- **Setup time:** time before rising edge of clk that D must be valid
$$t_{su} = 3 * t_{pd_inv} + t_{pd_tx}$$
- **Propagation delay:** time for D to reach Q
$$t_{c2q} = t_{pd_inv} + t_{pd_tx}$$
- **Hold time:** time D must be stable after rising edge of clk
$$t_{hold} = \text{zero}$$
 - Slow clock can cause both latches transparent and increase hold time

Setup Time Simulation (I)

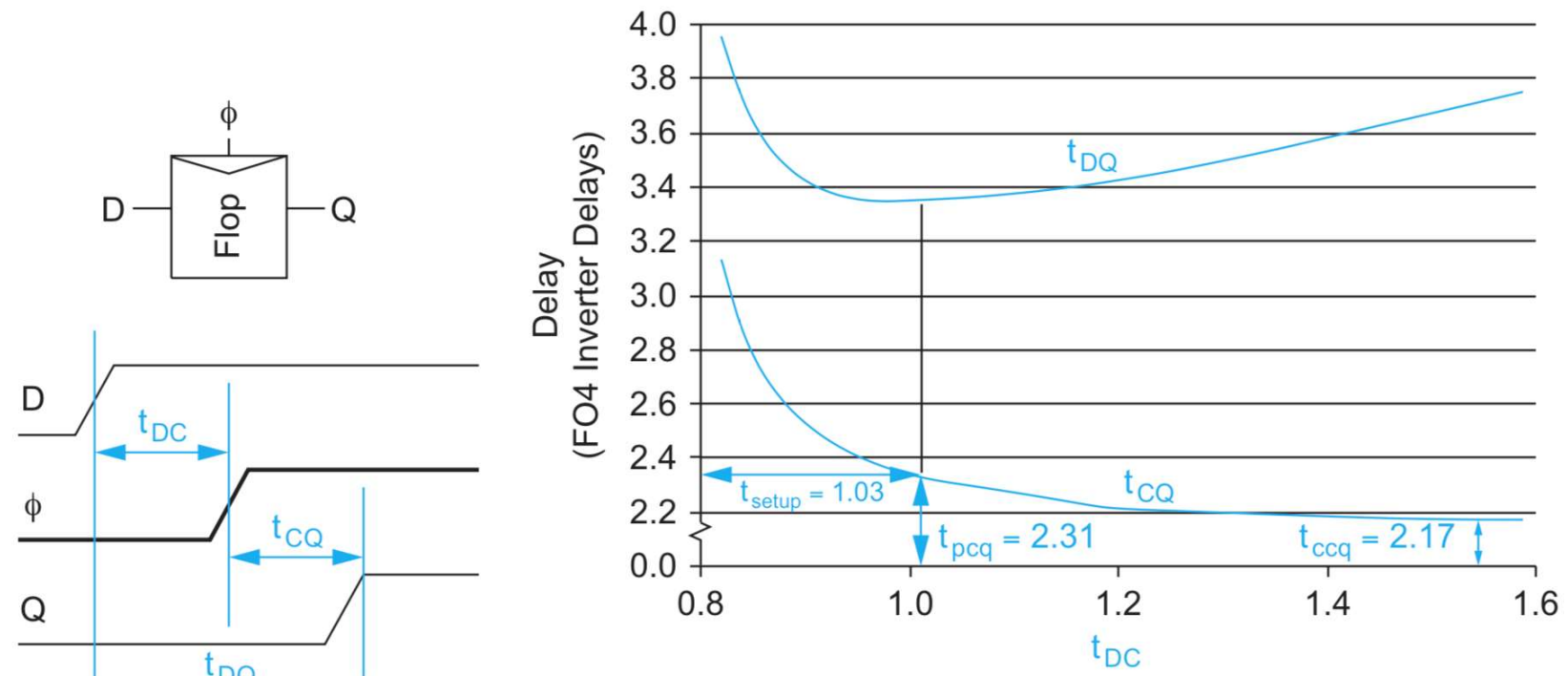


Setup Time Simulation (II)



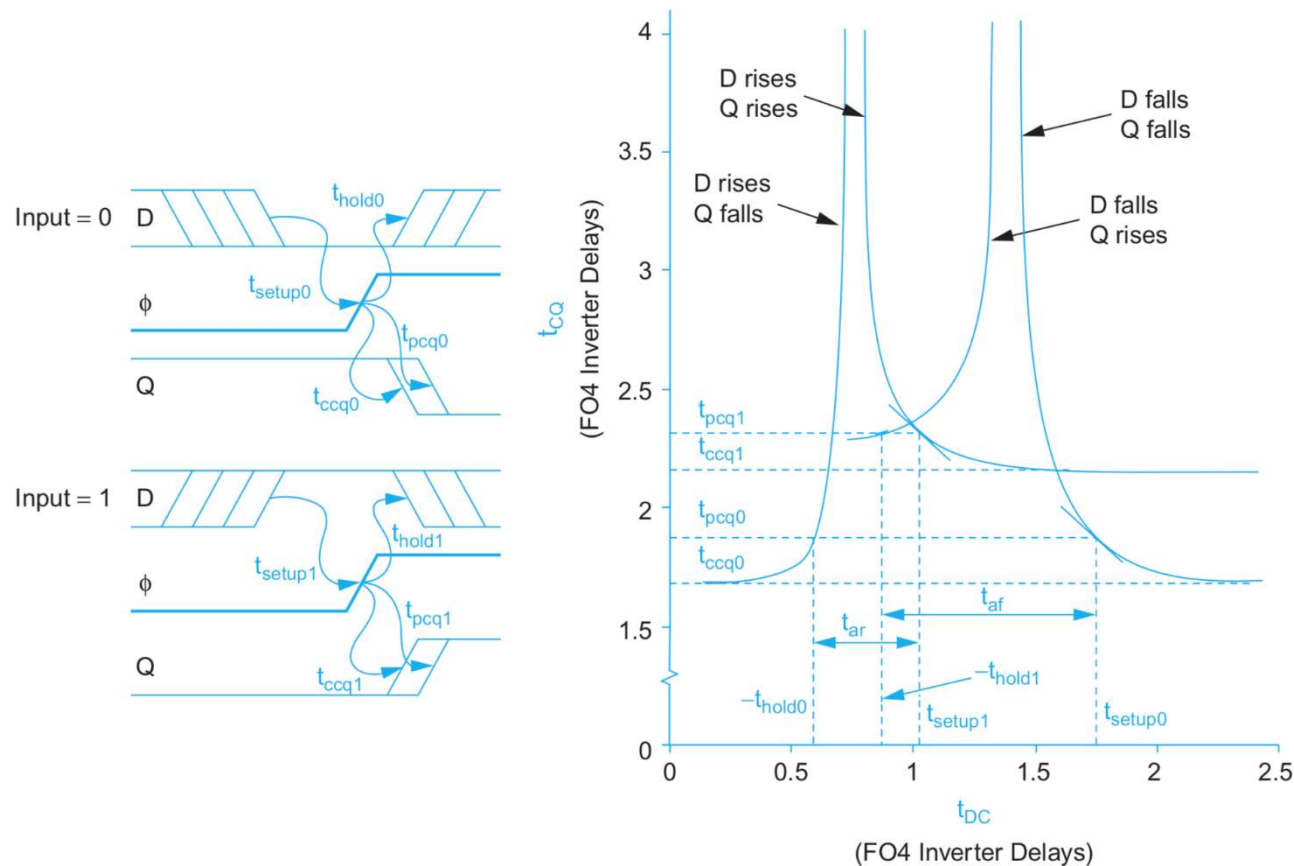
Sequencing Element Characterization (I)

- Flip flops
- **Setup time:** minimize t_{D2Q}



Sequencing Element Characterization (II)

- **Hold time:** min. time from clock to D for $t_{c2Q} < t_{pcq}$
- **Aperture width:** the width of timing window around clock edge during which data must not transition



Sequencing Element Characterization (III)

- Latches
- Setup time: t_{D2Q} 5% greater than min. value

