Part I. Paper Assignment

1. This problem examines conversions between various filter specifications. Given the absolute specifications $\delta_s=0.0001$ and $\omega_p=0.3\pi$, $\omega_s=0.5\pi$, determine the relative specifications A_s and ω_c , $\Delta\omega$.

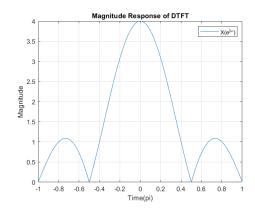
$$A_S=-20\log_{10}\delta_s=80~dB,~\omega_c=\frac{\omega_p+\omega_s}{_2}=0.4\pi,~\Delta\omega=\omega_s-\omega_P=0.2\pi$$

- 2. The Hann window function can be written as $w[n] = [0.5 0.5 \cos(2\pi n/M)]w_R[n]$ where $w_R[n]$ is the rectangular window of length M + 1.
 - (a) Express the DTFT of w[n] in terms of the DTFT of $w_R[n]$.
 - (b) Explain why the Hann window has the wider mainlobe but lower sidelobes than the rectangular window of the same length.

$$\begin{split} \text{(a)} \ & x[n] \cos(\omega_c n) \leftrightarrow \frac{1}{2} X \Big(e^{j(\omega + \omega_c)} \Big) + \frac{1}{2} X \Big(e^{j(\omega - \omega_c)} \Big) \\ \\ & \Rightarrow W \Big(e^{j\omega} \Big) = \frac{1}{2} W_R \Big(e^{j\omega} \Big) - \frac{1}{4} \Bigg(W_R \Big(e^{j\left(\omega + \frac{2\pi}{M}\right)} \Big) + W_R \Big(e^{j\left(\omega - \frac{2\pi}{M}\right)} \Big) \Bigg)$$

- (b) Hann window 用原本一半的 Rectangular window 再減掉 1/4 往左右平移 $e^{j2\pi/M}$ 的 Rectangular 訊號,也因此 Hann window 會有比較寬的 mainlobe 和比較小的 sidelobes。
- 3. Consider a FIR filter with impulse response h[n] = u[n] u[n 4].
 - (a) Determine and sketch the magnitude response $|H(e^{j\omega})|$.
 - (b) Determine and sketch the amplitude response $A(e^{j\omega})$. Compare this sketch with that in (a) and comment on the difference.
 - (c) Determine and sketch the phase response $\angle H(e^{j\omega})$.
 - (d) Determine and sketch the angle response $\Psi(e^{j\omega})$. Compare this sketch with that in (c) and comment on the difference.

(a)
$$H(e^{j\omega}) = \frac{1 - e^{-j4\omega}}{1 - e^{j\omega}} \Rightarrow |H(e^{j\omega})| = \left|\frac{1 - \cos 4\omega + j \sin 4\omega}{1 - \cos \omega - i \sin \omega}\right|$$

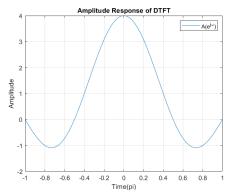


(b) Odd order M; even tap L = M + 1, symmetric \rightarrow Type II

$$H(e^{j\omega}) = h[0] + h[1]e^{-j\omega} + h[2]e^{-j2\omega} + h[3]e^{-j3\omega} = \left(2h[1]\cos\left(\frac{\omega}{2}\right) + \frac{1}{2}\right)e^{-j\omega} + h[2]e^{-j\omega} + h[3]e^{-j\omega} = \left(2h[1]\cos\left(\frac{\omega}{2}\right) + \frac{1}{2}\right)e^{-j\omega} + h[2]e^{-j\omega} + h[3]e^{-j\omega} = \left(2h[1]\cos\left(\frac{\omega}{2}\right) + \frac{1}{2}\right)e^{-j\omega} + h[2]e^{-j\omega} + h[2]e^{-j\omega} + h[3]e^{-j\omega} = \left(2h[1]\cos\left(\frac{\omega}{2}\right) + \frac{1}{2}\right)e^{-j\omega} + h[2]e^{-j\omega} + h[2]e^{-j\omega} + h[3]e^{-j\omega} = \left(2h[1]\cos\left(\frac{\omega}{2}\right) + \frac{1}{2}\right)e^{-j\omega} + h[2]e^{-j\omega} + h[2]e^{-j\omega}$$

$$2h[0]\cos\left(\tfrac{3\omega}{2}\right)\!\big)\,e^{-j\left(\tfrac{3}{2}\right)\omega}=A\big(e^{j\omega}\big)e^{-j\left(\tfrac{3}{2}\right)\omega}$$

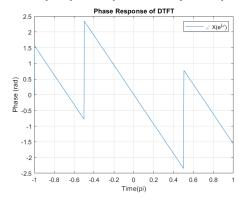
$$\Rightarrow A(e^{j\omega}) = \left(2\cos\left(\frac{\omega}{2}\right) + 2\cos\left(\frac{3\omega}{2}\right)\right)e^{-j\left(\frac{3}{2}\right)\omega}$$



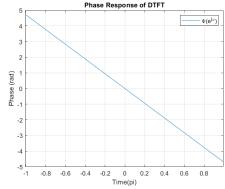
$$\Rightarrow \left|A\left(e^{j\omega}\right)\right| = H\left(e^{j\omega}\right)$$

(c)
$$H(e^{j\omega}) = \frac{1-e^{-j4\omega}}{1-e^{j\omega}} = |H(e^{j\omega})|e^{j\angle H(e^{j\omega})}$$

$$\Rightarrow \angle H\!\left(e^{j\omega}\right) = \angle\!\left(1-e^{-j4\omega}\right) - \angle\!\left(1-e^{j\omega}\right)$$



(d)
$$H(e^{j\omega}) = A(e^{j\omega})e^{-j\left(\frac{3}{2}\right)\omega} = A(e^{j\omega})e^{j\Psi(e^{j\omega})} \Rightarrow \Psi(e^{j\omega}) = -\left(\frac{3}{2}\right)\omega$$



$$\Rightarrow \Psi(e^{j\omega}) = unwrap \angle H(e^{j\omega})$$

4. Consider the type-IV linear-phase FIR filter characterized by antisymmetric impulse response and odd-M.

(a) Show that the amplitude response $A(e^{j\omega})$ is given by (10.38) with coefficients d[k] given in (10.39).

$$H(e^{j\omega}) = \sum_{k=1}^{\frac{M+1}{2}} \left(d[k] \sin\left[\omega\left(k - \frac{1}{2}\right)\right] \right) j e^{-\frac{j\omega M}{2}} \approx jA(e^{j\omega}) e^{-\frac{j\omega M}{2}}$$
 (10.38)
$$d[k] = 2h\left[\frac{M+1}{2} - k\right]$$
 (10.39)

(b) Show that the amplitude response $A(e^{j\omega})$ can be further expressed as (10.40) with coefficients $\hat{d}[k]$ given in (10.41).

$$A(e^{j\omega}) = \sin\left(\frac{\omega}{2}\right) \sum_{k=0}^{(M-1)/2} \tilde{d}[k] \cos(\omega k)$$
 (10.40)

$$d[k] = \begin{cases} \frac{1}{2}(2\tilde{d}[0] - \tilde{d}[1]), & k = 1\\ \frac{1}{2}(\tilde{d}[k-1] - \tilde{d}[k]), & 2 \le k \le (M-1)/2 \\ \frac{1}{2}\tilde{d}[(M-1)/2], & k = (M+1)/2 \end{cases}$$
(10.41)

(a)
$$H(e^{j\omega})=\sum_{n=0}^M h[n]e^{-j\omega n}$$
 , $h[n]=h[M-n]$

$$\Rightarrow H(e^{j\omega}) = h[0] + h[1]e^{-j\omega} + \dots + h[M-1]e^{-j(M-1)\omega} + h[M]e^{-jM\omega}
= \left(h[0]e^{j\frac{M}{2}\omega} + h[1]e^{j\left(\frac{M}{2}-1\right)\omega} + \dots + h[M-1]e^{-j\left(\frac{M}{2}-1\right)\omega} + h[M]e^{-j\frac{M}{2}\omega}\right)e^{-j\frac{M}{2}\omega}
= \left(2h[0]\sin\left(\frac{M}{2}\omega\right) + 2h[1]\sin\left(\left(\frac{M}{2}-1\right)\omega\right) + 2h[2]\sin\left(\left(\frac{M}{2}-2\right)\omega\right) + \dots + 2h\left[\frac{M-1}{2}\right]\sin\left(\left(\frac{M}{2}-\frac{M-1}{2}\right)\omega\right)\right)je^{-j\frac{M}{2}\omega}
= \left(d[k]\sin\left(\frac{M\omega}{2}\right) + d[k-1]\sin\left(\left(\frac{M}{2}-1\right)\omega\right) + \dots + d[1]\sin\left(\frac{\omega}{2}\right)\right)je^{-j\left(\frac{M}{2}\right)\omega}$$

$$\rightarrow H(e^{j\omega}) = \sum_{k=1}^{\frac{M+1}{2}} \left(d[k] \sin \left[\omega \left(k - \frac{1}{2} \right) \right] \right) j e^{-\frac{j\omega M}{2}} = j A(e^{j\omega}) e^{-\frac{j\omega M}{2}}$$

(b) $\sin(\alpha + \beta) = \sin \alpha \cos \beta + \cos \alpha \sin \beta \rightarrow 2 \sin \beta \cos \alpha = \sin(\alpha + \beta) - \sin(\alpha - \beta)$

$$\begin{split} &A(e^{j\omega}) = \sum_{k=1}^{\frac{M+1}{2}} \left(d[k] \sin\left[\omega\left(k - \frac{1}{2}\right)\right] \right) \\ &= d[1] \sin\left(\frac{\omega}{2}\right) + d[2] \sin\left(\frac{3\omega}{2}\right) + \dots + d[k-1] \sin\left(\left(\frac{M}{2} - 1\right)\omega\right) + d[k] \sin\left(\frac{M\omega}{2}\right) \\ &= d[1] \sin\left(\frac{\omega}{2}\right) + d[2] \left(2\sin\left(\frac{\omega}{2}\right)\cos(\omega) + \sin\left(\frac{\omega}{2}\right)\right) + d[3] \left(2\sin\left(\frac{\omega}{2}\right)\cos(2\omega) + \sin\left(\frac{3\omega}{2}\right)\right) + \dots + d[k-1] \left(2\sin\left(\frac{\omega}{2}\right)\cos\left(\frac{(M-3)\omega}{2}\right) + \sin\left(\frac{(M-4)\omega}{2}\right)\right) + \dots \end{split}$$

$$\begin{split} &d[k]\left(2\sin\left(\frac{\omega}{2}\right)\cos\left(\frac{(M-1)\omega}{2}\right)+\sin\left(\frac{(M-2)\omega}{2}\right)\right)\\ &=\sin\left(\frac{\omega}{2}\right)\left[\left(d[1]+\cdots+d\left[\frac{M+1}{2}\right]\right)+2\left(d[2]+\cdots+d\left[\frac{M+1}{2}\right]\right)\cos\omega+\\ &2\left(d[3]+\cdots+d\left[\frac{M+1}{2}\right]\right)\cos2\omega+\cdots+2d\left[\frac{M+1}{2}\right]\cos\left(\frac{M+1}{2}\omega\right)\right]\\ &=\sin\left(\frac{\omega}{2}\right)\sum_{k=0}^{(M-1)/2}\tilde{d}[k]\cos(\omega k)\\ & \Longrightarrow A(e^{j\omega})=\sin\left(\frac{\omega}{2}\right)\sum_{k=0}^{(M-1)/2}\tilde{d}[k]\cos(\omega k) \text{ with}\\ &d[k]=\begin{cases} \frac{1}{2}(2\tilde{d}[0]-\tilde{d}[1]),\ k=1\\ \frac{1}{2}(\tilde{d}[k-1]-\tilde{d}[k]),\ 2\leq k\leq (M-1)/2\\ \frac{1}{2}\tilde{d}[(M-1)/2],\ k=(M+1)/2 \end{cases} \end{split}$$