



$$\begin{aligned}
 2. (a) \quad G_{2AS}(z) &= (1-z^{-1}) \mathcal{Z} \left[ \frac{G(s)}{s} \right] \\
 &= (1-z^{-1}) \mathcal{Z} \left[ \frac{2}{s^2(s+6)^2} \right] = (1-z^{-1})
 \end{aligned}$$



$$3. (a) \quad \dot{x}(t) = Ax(t) + Bu(t)$$

$$\begin{aligned}\Phi(T) &= \mathcal{L}^{-1}\{[sI - A]^{-1}\} = \mathcal{L}^{-1}\left\{\begin{bmatrix} s & -1 \\ 1 & s \end{bmatrix}^{-1}\right\} = \mathcal{L}^{-1}\left\{\frac{\begin{bmatrix} s & 1 \\ -1 & s \end{bmatrix}}{s^2 + 1}\right\} \\ &= \mathcal{L}^{-1}\left\{\begin{bmatrix} \frac{s}{s^2+1} & \frac{1}{s^2+1} \\ \frac{-1}{s^2+1} & \frac{s}{s^2+1} \end{bmatrix}\right\} = \begin{bmatrix} \cos T & \sin T \\ -\sin T & \cos T \end{bmatrix}\end{aligned}$$

$$\begin{aligned}\Theta(T) &= \int_0^T \Phi(\tau) B d\tau = \int_0^T \begin{bmatrix} \cos \tau & \sin \tau \\ -\sin \tau & \cos \tau \end{bmatrix} \begin{bmatrix} 0 \\ 1 \end{bmatrix} d\tau \\ &= \int_0^T \begin{bmatrix} \sin \tau \\ \cos \tau \end{bmatrix} d\tau = \begin{bmatrix} -\cos \tau \\ \sin \tau \end{bmatrix}_0^T = \begin{bmatrix} -\cos T - 1 \\ \sin T \end{bmatrix}\end{aligned}$$

$$x(k+1) = \Phi(T)x(k) + \Theta(T)u(k)$$

$$= \begin{bmatrix} \cos T & \sin T \\ -\sin T & \cos T \end{bmatrix} x(k) + \begin{bmatrix} -\cos T - 1 \\ \sin T \end{bmatrix} u(k)$$

$$y(k) = [1 \quad 1] x(k)$$

$$(b) \quad W_0 = \begin{bmatrix} C \\ CA \end{bmatrix} = \begin{bmatrix} 1 & 1 \\ \cos T - \sin T & \sin T + \cos T \end{bmatrix}$$

$$|W_0| = \sin T + \cos T - \cos T + \sin T = 2\sin T = 0$$

$$\Rightarrow T = k\pi \quad (k \text{ is an integer})$$

$$(c) \quad x(k+1) = Ax(k) + Bu(k) = Ax(k) + B(r(k) - y(k))$$

$$y(k) = Cx(k) + du(k) = Cx(k) + dr(k) - dy(k)$$

$$\Rightarrow y(k) = \frac{1}{1+d} (Cx(k) + dr(k)) = \frac{C}{1+d} x(k) + \frac{d}{1+d} r(k)$$

$$u(k) = r(k) - y(k) = \frac{1}{1+d} r(k) - \frac{C}{1+d} x(k)$$

$$\Rightarrow x(k+1) = Ax(k) + \frac{B}{1+d} r(k) - \frac{BC}{1+d} x(k)$$

$$= \left[A - \frac{BC}{1+d}\right] x(k) + \frac{B}{1+d} r(k)$$

$$\therefore A_d = \left[A - \frac{BC}{1+d}\right], \quad B_d = \frac{B}{1+d}, \quad C_d = \frac{C}{1+d}, \quad D_d = \frac{d}{1+d}$$



$$4. (a) \quad x(k+1) = Ax(k) + Bu(k) = [A - BK]x(k) + Bkr$$

$$[A - BK] = \begin{bmatrix} 1 & 1 \\ 0 & 1 \end{bmatrix} - \begin{bmatrix} 0.125 \\ 0.25 \end{bmatrix} \begin{bmatrix} k_1 & k_2 \end{bmatrix} = \begin{bmatrix} 1 - 0.125k_1 & 1 - 0.125k_2 \\ -0.25k_1 & 1 - 0.25k_2 \end{bmatrix}$$

$$|zI - [A - BK]| = \begin{vmatrix} z - 1 + 0.125k_1 & -1 + 0.125k_2 \\ 0.25k_1 & z - 1 + 0.25k_2 \end{vmatrix}$$

$$= (z - 1 + \frac{1}{8}k_1)(z - 1 + \frac{1}{4}k_2) - \frac{1}{4}k_1(\frac{1}{8}k_2 - 1)$$

$$= z^2 - z + \frac{1}{4}k_2z - z + 1 - \frac{1}{4}k_2 + \frac{1}{8}k_1z - \frac{1}{8}k_1 + \frac{1}{32}k_1k_2 - \frac{1}{32}k_1k_2 + \frac{1}{4}k_1$$

$$= z^2 + (\frac{1}{8}k_1 + \frac{1}{4}k_2 - 2)z + (\frac{1}{8}k_1 - \frac{1}{4}k_2 + 1)$$

$$\alpha_c(z) = (z - z_1)(z - z_2) = z^2 - (z_1 + z_2)z + z_1z_2 = z^2 - z + 0.25$$

$$\Rightarrow \begin{cases} \frac{1}{8}k_1 + \frac{1}{4}k_2 - 2 = -1 \\ \frac{1}{8}k_1 - \frac{1}{4}k_2 + 1 = 0.25 \end{cases} \Rightarrow \begin{cases} \frac{1}{4}k_1 - 1 = -0.75, & k_1 = 1 \\ & k_2 = 3.5 \end{cases}$$

$$\Rightarrow k = [1 \quad 3.5], \quad y(k) = [1 \quad 1]x(k) = x_1(k) + x_2(k) = 1$$

$$\begin{bmatrix} x_1(k+1) \\ x_2(k+1) \end{bmatrix} = \begin{bmatrix} 0.875 & 0.5625 \\ -0.25 & 0.125 \end{bmatrix} \begin{bmatrix} x_1(k) \\ x_2(k) \end{bmatrix} + \begin{bmatrix} 0.125kr \\ 0.25kr \end{bmatrix}$$

$$\begin{cases} x_1(k+1) = 0.875x_1(k) + 0.5625x_2(k) + 0.125kr \\ x_2(k+1) = -0.25x_1(k) + 0.125x_2(k) + 0.25kr \\ x_1(k) + x_2(k) = 1 \end{cases}$$

$$kr = 0, \quad k = [4 \quad 6]$$

$$x(k+1) = [A - BK]x(k) = \begin{bmatrix} 1 & 1 \\ 0 & 1 \end{bmatrix} - \begin{bmatrix} 0.125 \\ 0.25 \end{bmatrix} \begin{bmatrix} 4 & 6 \end{bmatrix} = \begin{bmatrix} 1 & 1 \\ 0 & 1 \end{bmatrix} - \begin{bmatrix} 0.5 & 0.75 \\ 1 & 1.5 \end{bmatrix} = \begin{bmatrix} 0.5 & 0.25 \\ -1 & -0.5 \end{bmatrix}$$

$$|A - BK| = 0$$

$$(b) \quad S(2) = \begin{bmatrix} 0 & 0 \\ 0 & 0 \end{bmatrix}, \quad K(1) = \left( \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix} \begin{bmatrix} 0 & 0 \\ 0 & 0 \end{bmatrix} \begin{bmatrix} 0 \\ 1 \end{bmatrix} + 2 \right)^{-1} \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix} \begin{bmatrix} 0 & 0 \\ 0 & 0 \end{bmatrix} \begin{bmatrix} 0 \\ -2 \\ -1 \end{bmatrix} \\ = \begin{bmatrix} 0 & 0 \end{bmatrix},$$

$$S(1) = \left( \begin{bmatrix} 0 & 1 \\ -2 & -1 \end{bmatrix} - \begin{bmatrix} 0 \\ 1 \end{bmatrix} \begin{bmatrix} 1 & 0 \end{bmatrix} \right)^T \begin{bmatrix} 0 & 0 \\ 0 & 0 \end{bmatrix} \left( \begin{bmatrix} 0 & 1 \\ -2 & -1 \end{bmatrix} - \begin{bmatrix} 0 \\ 1 \end{bmatrix} \begin{bmatrix} 1 & 0 \end{bmatrix} \right) + 2 \begin{bmatrix} 0 \\ 0 \end{bmatrix} \begin{bmatrix} 1 & 0 \end{bmatrix} + \begin{bmatrix} 2 & 0 \\ 0 & 2 \end{bmatrix} \\ = Q = \begin{bmatrix} 2 & 0 \\ 0 & 2 \end{bmatrix},$$

$$K(0) = \left( \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix} \begin{bmatrix} 2 & 0 \\ 0 & 2 \end{bmatrix} \begin{bmatrix} 0 \\ 1 \end{bmatrix} + 2 \right)^{-1} \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix} \begin{bmatrix} 2 & 0 \\ 0 & 2 \end{bmatrix} \begin{bmatrix} 0 \\ -2 \\ -1 \end{bmatrix} \\ = \frac{1}{4} \begin{bmatrix} -4 & -2 \end{bmatrix} = \begin{bmatrix} -1 & -\frac{1}{2} \end{bmatrix},$$

$$S(0) = \left( \begin{bmatrix} 0 & 1 \\ -2 & -1 \end{bmatrix} - \begin{bmatrix} 0 \\ 1 \end{bmatrix} \begin{bmatrix} -1 & -\frac{1}{2} \end{bmatrix} \right)^T \begin{bmatrix} 2 & 0 \\ 0 & 2 \end{bmatrix} \left( \begin{bmatrix} 0 & 1 \\ -2 & -1 \end{bmatrix} - \begin{bmatrix} 0 \\ 1 \end{bmatrix} \begin{bmatrix} -1 & -\frac{1}{2} \end{bmatrix} \right) \\ + 2 \begin{bmatrix} -1 \\ -\frac{1}{2} \end{bmatrix} \begin{bmatrix} -1 & -\frac{1}{2} \end{bmatrix} + \begin{bmatrix} 2 & 0 \\ 0 & 2 \end{bmatrix}$$

$$= \begin{bmatrix} 0 & 1 \\ -1 & -\frac{1}{2} \end{bmatrix}^T \begin{bmatrix} 2 & 0 \\ 0 & 2 \end{bmatrix} \begin{bmatrix} 0 & 1 \\ -1 & -\frac{1}{2} \end{bmatrix} + \begin{bmatrix} 2 & 1 \\ 1 & \frac{1}{2} \end{bmatrix} + \begin{bmatrix} 2 & 0 \\ 0 & 2 \end{bmatrix}$$

$$= \begin{bmatrix} 0 & -1 \\ 1 & -\frac{1}{2} \end{bmatrix} \begin{bmatrix} 2 & 0 \\ 0 & 2 \end{bmatrix} \begin{bmatrix} 0 & 1 \\ -1 & -\frac{1}{2} \end{bmatrix} + \begin{bmatrix} 4 & 1 \\ 1 & \frac{5}{2} \end{bmatrix}$$

$$= \begin{bmatrix} 0 & -2 \\ 2 & -1 \end{bmatrix} \begin{bmatrix} 0 & 1 \\ -1 & -\frac{1}{2} \end{bmatrix} + \begin{bmatrix} 4 & 1 \\ 1 & \frac{5}{2} \end{bmatrix} = \begin{bmatrix} 2 & 1 \\ 1 & \frac{5}{2} \end{bmatrix} + \begin{bmatrix} 4 & 1 \\ 1 & \frac{5}{2} \end{bmatrix} = \begin{bmatrix} 6 & 2 \\ 2 & 5 \end{bmatrix}$$

$$u^*(0) = -K(0)x(0) = -\begin{bmatrix} -1 & -\frac{1}{2} \end{bmatrix} \begin{bmatrix} 1 \\ 1 \end{bmatrix} = \frac{3}{2}$$

$$x^*(1) = Ax(0) + Bu^*(0) = \begin{bmatrix} 0 & 1 \\ -2 & -1 \end{bmatrix} \begin{bmatrix} 1 \\ 1 \end{bmatrix} + \begin{bmatrix} 0 \\ 1 \end{bmatrix} \frac{3}{2} = \begin{bmatrix} 1 \\ -\frac{3}{2} \end{bmatrix}$$

$$u^*(1) = -K(1)x^*(1) = -\begin{bmatrix} 0 & 0 \end{bmatrix} \begin{bmatrix} 1 \\ -\frac{3}{2} \end{bmatrix} = 0$$



$$5. (a) \quad H(s) = \frac{50}{s(1+0.01s)(1+0.003s)} \quad , \quad \omega_c = 150 \text{ rad/s} \quad ,$$

$$H_{\text{sam}}(s) = \frac{1 - e^{-Ts}}{s} H(s) \quad , \quad T = 2 \text{ ms} = 2 \times 10^{-3} \text{ s} \quad , \quad s = j\omega_c = j150$$

**NANYANG TECHNOLOGICAL UNIVERSITY**  
**SEMESTER 1 EXAMINATION 2015-2016**  
**EE6203 – COMPUTER CONTROL SYSTEMS**

November/December 2015

Time Allowed: 3 hours

**INSTRUCTIONS**

1. This paper contains 5 questions and comprises 8 pages.
2. Answer all 5 questions.
3. All questions carry equal marks.
4. The Transform Table is included in Appendix A on pages 6 to 8.

$$C(z) = \frac{1}{(1+2z^{-1})(1-z^{-1})}$$

$$= \frac{A}{1+2z^{-1}} + \frac{B}{1-z^{-1}}$$

$$(2B-A)z^{-1} + (A+B) = 1$$

$$\begin{aligned} 2B-A &= 0 \\ A+B &= 1 \end{aligned} \quad \begin{aligned} B &= \frac{1}{3} \\ A &= \frac{2}{3} \end{aligned}$$

1. (a) Consider a first order system that has the following transfer function:

$$G(s) = \frac{C(s)}{E(s)} = \frac{s+1}{s+3}$$

$$G(z) = \frac{1}{1+2z^{-1}}$$

$$C(k) = \frac{2}{3}(-2)^k + \frac{1}{3}(1)^k$$

Given a sampling period  $T$  of 1 second, discretise the system using the forward difference method.

$$C(z) = G(z)E(z) = \frac{1}{(1+2z^{-1})(1-z^{-1})}$$

If the input signal  $e(kT)$  to the system is a unit-step function, determine the output response sequence  $c(kT)$  for  $k = 0, 1, 2, 3$ .

$$C(kT) = \frac{2}{3}(-2)^k + \frac{1}{3}(1)^k \quad (10 \text{ Marks})$$

- (b) Given the following digital controller

$$D(z) = \frac{z-0.5}{(z-1)(z+0.3)}$$

Show its implementation using the series programming approach.

(7 Marks)

- (c) Give your comment on part 1(a) if the backward difference approach is used.

(3 Marks)



2. (a) Given the following position control system for a robotic arm

$$G(s) = \frac{2}{s(s+6)^2}$$

find its discretized transfer function  $G_{ZAS}(z)$  if a zero-order-hold is used. Assume that the desired sampling period is 0.2 s.

(10 Marks)

- (b) For a finite settling time of  $k$  samples, we have the following deadbeat controller:

$$C(z) = \frac{1}{G_{ZAS}(z)} \left[ \frac{z^{-k}}{1-z^{-k}} \right]$$

Determine the deadbeat controller for the position control system in part 2(a) for  $k = 2$ .

(6 Marks)

- (c) Comment on how the time response of the controller in part 2(b) can be improved.

(4 Marks)

3. (a) A continuous-time system has a state-space representation given by

$$\begin{bmatrix} \frac{dx_1(t)}{dt} \\ \frac{dx_2(t)}{dt} \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ -1 & 0 \end{bmatrix} \begin{bmatrix} x_1(t) \\ x_2(t) \end{bmatrix} + \begin{bmatrix} 0 \\ 1 \end{bmatrix} u(t)$$

$$y(t) = \begin{bmatrix} 1 & 1 \end{bmatrix} \begin{bmatrix} x_1(t) \\ x_2(t) \end{bmatrix}$$

where  $x_1(t)$  and  $x_2(t)$  are the states,  $u(t)$  and  $y(t)$  are the input and output variables, respectively. Suppose that the system is sampled with a zero-order hold.

Obtain a discretised state-space model for the system in terms of the sampling period  $T$ .

(9 Marks)

- (b) Determine the values of the sampling period  $T$  that make the discretised system obtained in part 3(a) unobservable.

(4

Marks)



Note: Question No. 3 continues on Page 3

(c) A discrete-time system has a state-space representation given by

$$\begin{aligned}\mathbf{x}(k+1) &= \mathbf{A}\mathbf{x}(k) + \mathbf{B}u(k) \\ y(k) &= \mathbf{C}\mathbf{x}(k) + du(k)\end{aligned}$$

where  $\mathbf{x}(k)$ ,  $u(k)$  and  $y(k)$  are the states, input and output variables, respectively. If

$$u(k) = r(k) - y(k)$$

where  $r(k)$  is the reference input, obtain an overall closed-loop state-space representation of the following form:

$$\begin{aligned}\mathbf{x}(k+1) &= \mathbf{A}_{cl}\mathbf{x}(k) + \mathbf{B}_{cl}r(k) \\ y(k) &= \mathbf{C}_{cl}\mathbf{x}(k) + \mathbf{D}_{cl}r(k)\end{aligned}$$

State the assumptions that you have made, if any.

(7

Marks)

4. (a) A discrete-time system has a state-space representation given by

$$\begin{aligned}\begin{bmatrix} x_1(k+1) \\ x_2(k+1) \end{bmatrix} &= \mathbf{A} \begin{bmatrix} x_1(k) \\ x_2(k) \end{bmatrix} + \mathbf{B}u(k) \\ y(k) &= \mathbf{C} \begin{bmatrix} x_1(k) \\ x_2(k) \end{bmatrix}\end{aligned}$$

where

$$\mathbf{A} = \begin{bmatrix} 1 & 1 \\ 0 & 1 \end{bmatrix}, \quad \mathbf{B} = \begin{bmatrix} 0.125 \\ 0.25 \end{bmatrix}, \quad \mathbf{C} = [1 \quad 1]$$

$\begin{bmatrix} x_1(k) \\ x_2(k) \end{bmatrix}$ ,  $u(k)$  and  $y(k)$  are the states, input and output variables, respectively. A controller of the following form

$$u(k) = -\mathbf{K} \begin{bmatrix} x_1(k) \\ x_2(k) \end{bmatrix} + k_r r(k)$$

is to be implemented, where  $r(k)$  is the reference input.

Note: Question No. 4 continues on Page 4

- (i) Design  $k_r$  and  $\mathbf{K}$  such that the closed-loop poles are at  $z_{1,2} = 0.5 \pm j0.5$  and the final value of  $y(k)$  is unity for a unit step input  $r(k)$ .

(9

Marks)

- (ii) If  $k_r = 0$  and  $\mathbf{K} = [4 \ 6]$ , discuss the type of response that the closed-loop system will exhibit when subjected to a non-zero  $\mathbf{x}(0)$ .

(4

Marks)

- (b) A process is described by the following state-space representation

$$\mathbf{x}(k+1) = \mathbf{A}\mathbf{x}(k) + \mathbf{B}u(k)$$

where  $\mathbf{A} = \begin{bmatrix} 0 & 1 \\ -2 & -1 \end{bmatrix}$  and  $\mathbf{B} = \begin{bmatrix} 0 \\ 1 \end{bmatrix}$ . The performance index for the system is given by

$$J = \frac{1}{2} \mathbf{x}^T(N) \mathbf{S}(N) \mathbf{x}(N) + \frac{1}{2} \sum_{k=0}^{N-1} (\mathbf{x}^T(k) \mathbf{Q} \mathbf{x}(k) + r u^2(k))$$

and the design equations for  $k = N-1, \dots, 1, 0$  are

$$\mathbf{K}(k) = (\mathbf{B}^T \mathbf{S}(k+1) \mathbf{B} + r)^{-1} \mathbf{B}^T \mathbf{S}(k+1) \mathbf{A}$$

$$u^*(k) = -\mathbf{K}(k) \mathbf{x}(k)$$

$$\mathbf{S}(k) = (\mathbf{A} - \mathbf{B} \mathbf{K}(k))^T \mathbf{S}(k+1) (\mathbf{A} - \mathbf{B} \mathbf{K}(k)) + r \mathbf{K}(k)^T \mathbf{K}(k) + \mathbf{Q}$$

If  $\mathbf{Q} = \begin{bmatrix} 2 & 0 \\ 0 & 2 \end{bmatrix}$ ,  $\mathbf{S}(N) = \begin{bmatrix} 0 & 0 \\ 0 & 0 \end{bmatrix}$ ,  $r = 2$  and  $N = 2$ , find the optimal controls  $u^*(0)$  and  $u^*(1)$  such that  $J$  is minimised. Let the initial state

$$\mathbf{x}(0) = \begin{bmatrix} 1 \\ 1 \end{bmatrix}.$$

(7 Marks)



EE6203

5. (a) A conveyor belt system uses a DC motor for direct drive for its position control and has the following loop transfer function:

$$H(s) = \frac{50}{s(1 + 0.01s)(1 + 0.003s)}$$

What is the system phase margin at  $\omega_c = 150$  rad/s if a ZOH is used and the sampling period is 2 ms?

(10 Marks)

- (b) Given a digital controller that has the following transfer function

$$G(z) = \frac{Y(z)}{X(z)} = \frac{z}{z+0.9}$$

If the input is a unit impulse at  $k = 0$ , what is the output of this controller for the time steps  $k = 0, 1, 2, \dots, 6$  assuming that there is a rounding error of one decimal significant digit. Repeat for the case if it is a truncation error of one decimal significant digit.

(8 Marks)

- (c) Comment on the results that you obtained in part 5(b).

(2 Marks)