

Signals and Systems Final

10:20a.m. ~ 12:20p.m., June 19, Fri., 2009

- Closed book, but open 1 sheet (both sides, 2 pages) of personal notes of A4 size
 - Total score: 110
 - Total 4 pages in one B4 sheet
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1. **[16]** Consider that a continuous-time signal $x(t)$ with its Laplace transform given by

$$X(s) = \frac{(s+3)}{(s+1)(s+2)}$$

- (a) Find all possible inverse bilateral Laplace transforms and justify your answer.
- (b) Sketch the region of convergence (ROC) in each case.
- (c) Determine each time function as causal, anticausal, or two-sided and justify your answer.
- (d) Determine each time function as bound-input and bound-output (BIBO) stable or not BIBO stable and justify your answer.

2. **[16]** Consider that a linear time-invariant (LTI) system $h(t)$ has its transfer function $H(z)$ given by

$$H(z) = \frac{\sum_{i=0}^M b_i z^{-i}}{\sum_{i=0}^N a_i z^{-i}}$$

Let $H_i(z)$ be the inverse system of $H(z)$.

- (a) Determine the constraint on the orders M and N if the LTI system $H(z)$ is causal and justify your answer.
 - (b) Determine the constraint on the orders M and N if the inverse system is also causal and justify your answer.
 - (c) Determine the constraint on the poles of $H(z)$ if $H(z)$ is stable and justify your answer.
 - (d) Determine the constraint on the poles and zeros of $H(z)$ if $H_i(z)$ is also stable and justify your answer.
3. **[8]** Consider a causal LTI system specified by the following linear constant coefficient equation (LCCDE) with input $x(t)$ and output $y(t)$:

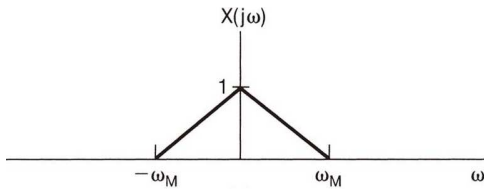
$$\frac{dy(t)}{dt} + 2y(t) = x(t)$$

- (a) Sketch the Bode plot of the frequency response of the LTI system.
- (b) Determine the group delay of the LTI system.

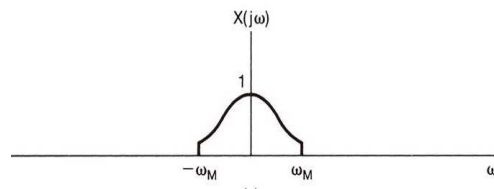
4. [8] Consider two message signals $x_1(t)$ and $x_2(t)$ with frequency spectra shown by Figures (a) and (b), respectively. We form the modulated signal $\Phi(t) = x_1(t) \cos \omega_c t + x_2(t) \sin \omega_c t$ for transmission, where $\omega_M \ll \omega_c$. In the receiver, we generate two signals $v_1(t)$ and $v_2(t)$ as

$$v_1(t) = \Phi(t) \cos \omega_c t \quad \text{and} \quad v_2(t) = \Phi(t) \sin \omega_c t,$$

respectively. Finally, $v_1(t)$ and $v_2(t)$ are filtered respectively by an ideal low-pass filter with cutoff frequency equal to $2\omega_M$ and unit amplitude to provide the output signals $y_1(t)$ and $y_2(t)$.



(a) $X_1(j\omega)$

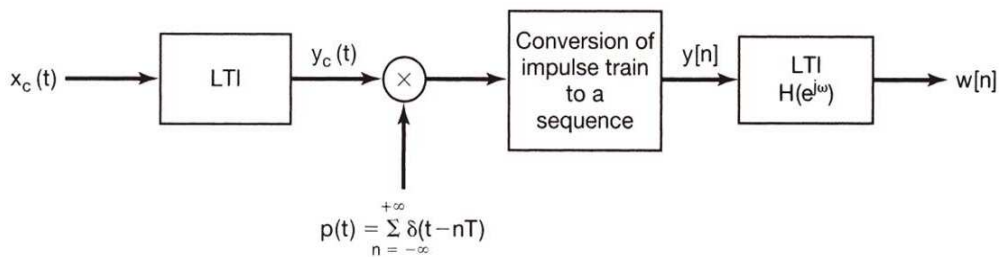


(b) $X_2(j\omega)$

- Determine $v_1(t)$ and sketch its frequency spectrum.
 - Determine $v_2(t)$ and sketch its frequency spectrum.
 - Determine $y_1(t)$ and sketch its frequency spectrum.
 - Determine $y_2(t)$ and sketch its frequency spectrum.
5. [12] Consider the system depicted below. Assume that the continuous-time LTI system is causal and specified by the LCCDE

$$\frac{dy_c(t)}{dt} + y_c(t) = x_c(t)$$

Now, we input a unit impulse to the system, i.e., $x_c(t) = \delta(t)$.



- Determine the output $y_c(t)$ of the continuous-time LTI system.
- Determine the frequency spectrum of $y[n]$.
- Determine the unit impulse response $h[n]$ of the discrete-time LTI system $H(e^{j\omega})$ such that $w[n] = \delta[n]$.

NOTE: There are problems in the back.

6. **[6]** For a real discrete-time sequence $x[n]$, prove that if $X(z)$, the z-transform of $x[n]$, has a pole (or zero) at $z = z_0$, then it must also have a pole (or zero) at $z = z_0^*$, where z_0^* denotes the complex conjugate of z_0 .
7. **[6]** Denote the unilateral z-transform of a discrete-time sequence $x[n]$ by $\mathcal{X}(z)$. Prove that the unilateral z-transform of $x[n+1]$ is $z\mathcal{X}(z) - zx[0]$.
8. **[10]** Prove the initial-value theorem of Laplace transform. That is, if a continuous-time signal $x(t) = 0$ for $t < 0$ and contains no impulse or higher order singularities at the origin, then

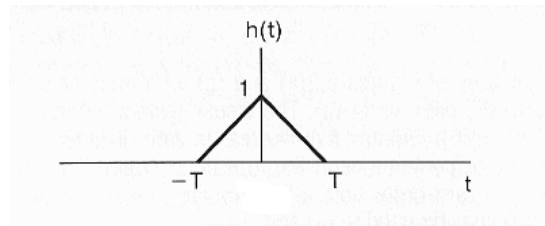
$$x(0^+) = \lim_{s \rightarrow \infty} sX(s).$$

(Hint: Make use of the Taylor series expansion of $x(t)$ at $t = 0^+$ and the Laplace transform pair

$$e^{-at} \left(\frac{t^n}{n!} \right) u(t) \xrightarrow{L} \frac{1}{(s+a)^{n+1}}, \quad \text{Re}\{s\} > -a.$$

9. **[12]** Consider the case of frequency modulation where the modulated signal is expressed as $y(t) = \cos(\omega_c t + m \sin \omega_m t) = \cos(\omega_c t) \cos(m \sin \omega_m t) - \sin(\omega_c t) \sin(m \sin \omega_m t)$ with ω_c being the frequency of the carrier, ω_m the frequency of the modulating signal, and m the modulation index.
- (a) When m is sufficiently small (say, much smaller than $\pi/2$), the modulated signal can be approximated by $y(t) = \cos(\omega_c t) - m(\sin \omega_m t)(\sin \omega_c t)$. Sketch the spectrum of the approximated signal.
- (b) When m is large, the approximation shown above no longer applies. But it can be noted that the terms $\cos(m \sin \omega_m t)$ and $\sin(m \sin \omega_m t)$ represent periodic signals. What are the fundamental periods of these two periodic signals? Justify your answer.
- (c) Explain how the spectrum of $y(t)$ should look like for the case where m is large.
10. **[8]** Suppose $H(e^{j\omega}) = |H(e^{j\omega})| e^{j\angle H(e^{j\omega})}$ is the frequency response of a discrete-time system. The response of the system to the input $x[n] = \sin(\omega_0 n + \theta_0)$ would be of this form $y[n] = |H(e^{j\omega_0})| x[n - n_0]$ if $\angle H(e^{j\omega_0})$ is related to ω_0 in a certain way. Determine this relation.

11. [8] Show that, in terms of reconstruction of a signal from its samples, the linear interpolation method is equivalent to impulse-train sampling followed by convolution with a triangular impulse response $h(t)$ shown in the following figure:



where T is the sampling period.