

Project : Digital Pulse Oximeter.

Principle:

$$1). \text{ Beer's Law: } I = I_0 \cdot e^{-\epsilon CL}$$

I_0 : light intensity entering sample.

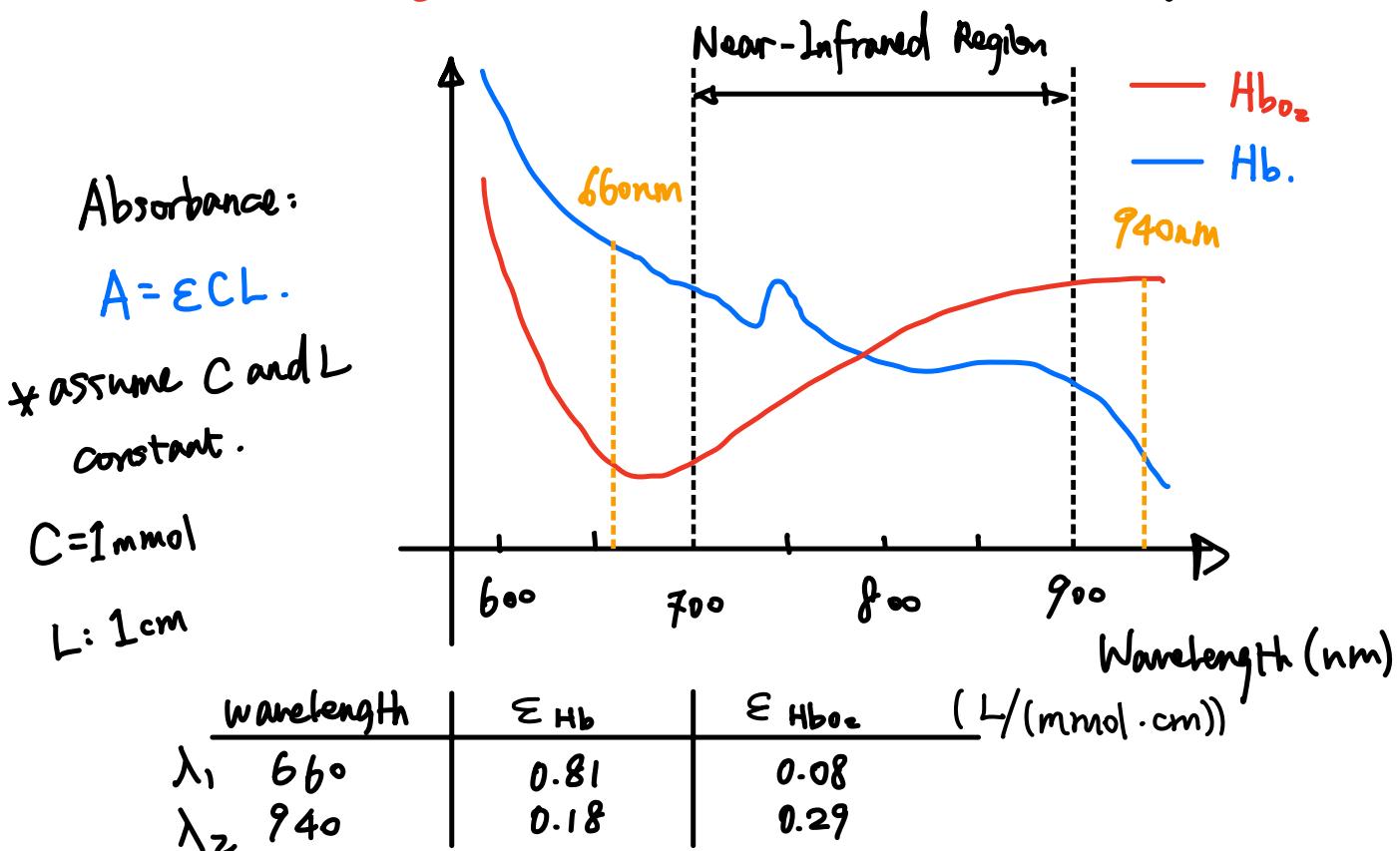
I : \sim leaving sample.

ϵ : extinction coefficient

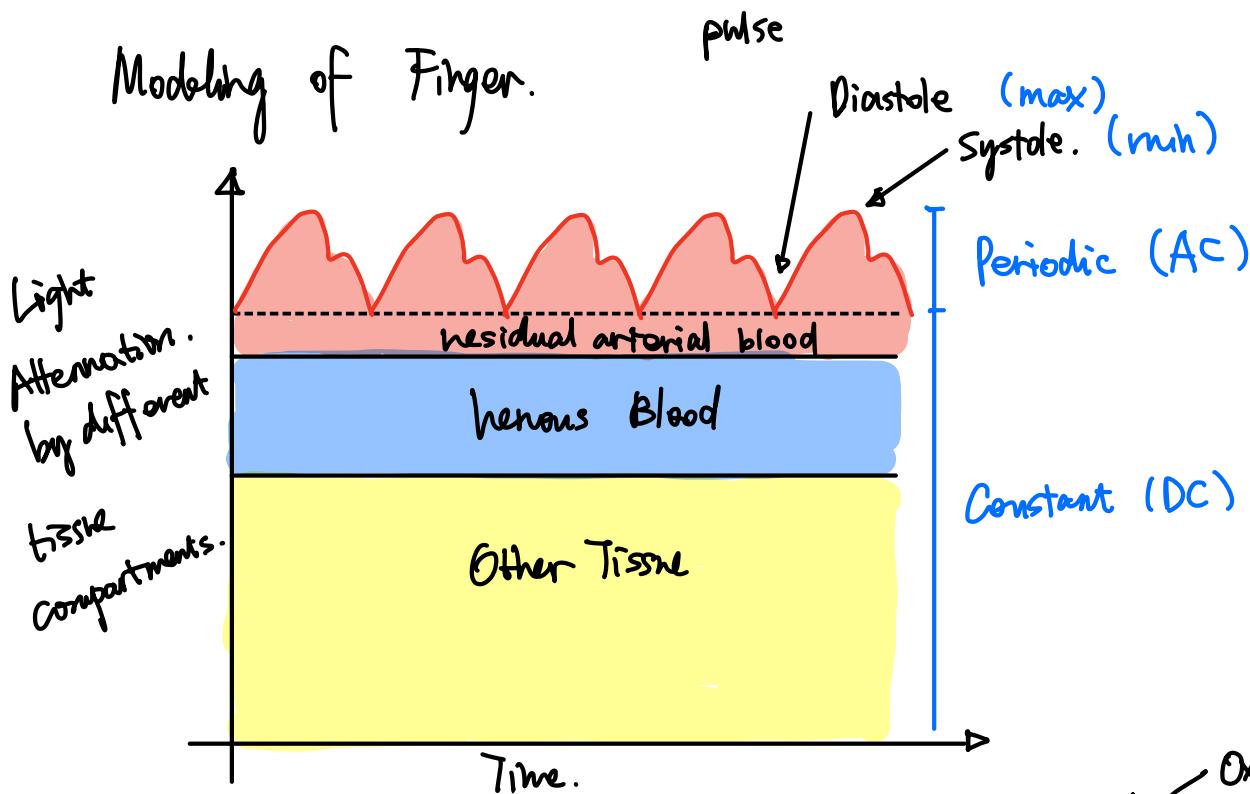
C : sample concentration

L : sample length (depth)

2) Absorbance of Hemoglobin (Hb) and Oxygenated Hemoglobin (HbO_2) at various wavelength.



Modeling of Finger.



at systole:

$$I_{min} = I_0 / 0 - \varepsilon_{DC} C_{DC} \cdot L_{DC} - (\varepsilon_{Hb} C_{Hb} + \varepsilon_{HbO_2} C_{HbO_2}) L_{max}$$

at diastole:

$$I_{max} = I_0 / 0 - \varepsilon_{DC} C_{DC} \cdot L_{DC} - (\varepsilon_{Hb} C_{Hb} + \varepsilon_{HbO_2} C_{HbO_2}) L_{min}$$

ratio:

$$\frac{I_{min}}{I_{max}} = \frac{I_0 / 0 - \varepsilon_{DC} C_{DC} \cdot L_{DC} - (\varepsilon_{Hb} C_{Hb} + \varepsilon_{HbO_2} C_{HbO_2}) L_{max}}{I_0 / 0 - \varepsilon_{DC} C_{DC} \cdot L_{DC} - (\varepsilon_{Hb} C_{Hb} + \varepsilon_{HbO_2} C_{HbO_2}) L_{min}}$$

$$= 1 / 0 - \varepsilon_{DC} C_{DC} \cdot L_{DC} - (\varepsilon_{Hb} C_{Hb} + \varepsilon_{HbO_2} C_{HbO_2}) (L_{max} - L_{min}) + \varepsilon_{DC} C_{DC} \cdot L_{DC}$$

$$= 1 / 0 - (\varepsilon_{Hb} C_{Hb} + \varepsilon_{HbO_2} C_{HbO_2}) \Delta L$$

$$\Rightarrow \log \frac{I_{min}}{I_{max}} = - (\varepsilon_{Hb} C_{Hb} + \varepsilon_{HbO_2} C_{HbO_2}) \Delta L$$

$$A = (\varepsilon_{Hb} C_{Hb} + \varepsilon_{HbO_2} C_{HbO_2}) \Delta L = \log \frac{I_{max}}{I_{min}}$$

$$\Rightarrow C_{Hb} = \frac{A}{\Delta L} - \frac{\varepsilon_{HbO_2} C_{HbO_2}}{\varepsilon_{Hb}}, \quad C_{HbO_2} = \frac{A}{\Delta L} - \frac{\varepsilon_{Hb} C_{Hb}}{\varepsilon_{HbO_2}}$$

at λ_1 and λ_2 (660 nm and 940 nm to maximize diff)

$$A_{\lambda_1} = (\epsilon_{Hb, \lambda_1} C_{Hb} + \epsilon_{HbO_2, \lambda_1} C_{HbO_2}) \Delta L$$

$$A_{\lambda_2} = (\epsilon_{Hb, \lambda_2} C_{Hb} + \epsilon_{HbO_2, \lambda_2} C_{HbO_2}) \Delta L$$

$$\frac{A_{\lambda_1}}{A_{\lambda_2}} = \frac{\epsilon_{Hb, \lambda_1} C_{Hb} + \epsilon_{HbO_2, \lambda_1} C_{HbO_2}}{\epsilon_{Hb, \lambda_2} C_{Hb} + \epsilon_{HbO_2, \lambda_2} C_{HbO_2}}$$

$$(\epsilon_{Hb, \lambda_2} C_{Hb} + \epsilon_{HbO_2, \lambda_2} C_{HbO_2}) A_{\lambda_1} = (\epsilon_{Hb, \lambda_1} C_{Hb} + \epsilon_{HbO_2, \lambda_1} C_{HbO_2}) A_{\lambda_2}$$

$$C_{Hb} (\epsilon_{Hb, \lambda_2} A_{\lambda_1} - \epsilon_{Hb, \lambda_1} A_{\lambda_2}) = C_{HbO_2} (\epsilon_{HbO_2, \lambda_1} A_{\lambda_2} - \epsilon_{HbO_2, \lambda_2} A_{\lambda_1})$$

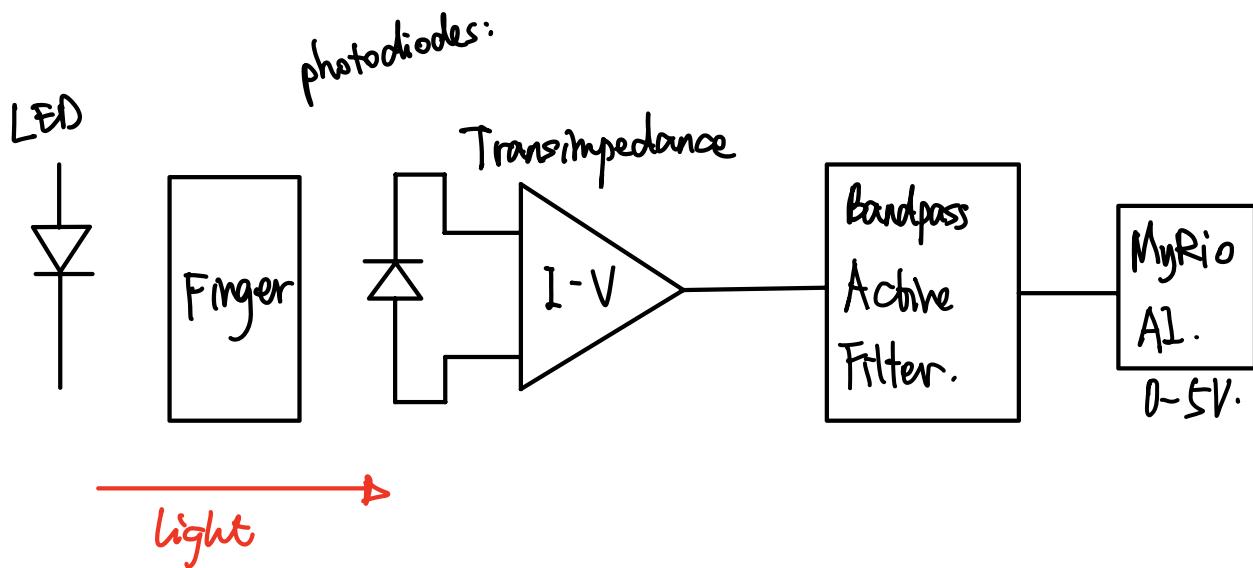
$$\Rightarrow C_{Hb} = C_{HbO_2} \frac{(\epsilon_{HbO_2, \lambda_1} A_{\lambda_2} - \epsilon_{HbO_2, \lambda_2} A_{\lambda_1})}{(\epsilon_{Hb, \lambda_2} A_{\lambda_1} - \epsilon_{Hb, \lambda_1} A_{\lambda_2})}$$

$$SpO_2 = \frac{C_{HbO_2}}{C_{HbO_2} + C_{Hb}} = \frac{\cancel{C_{HbO_2}}}{\cancel{C_{HbO_2}} \left(1 + \frac{(\epsilon_{HbO_2, \lambda_1} A_{\lambda_2} - \epsilon_{HbO_2, \lambda_2} A_{\lambda_1})}{(\epsilon_{Hb, \lambda_2} A_{\lambda_1} - \epsilon_{Hb, \lambda_1} A_{\lambda_2})} \right)}$$

$$= \frac{(\epsilon_{Hb, \lambda_2} A_{\lambda_1} - \epsilon_{Hb, \lambda_1} A_{\lambda_2})}{(\epsilon_{Hb, \lambda_2} A_{\lambda_1} - \epsilon_{Hb, \lambda_1} A_{\lambda_2}) + (\epsilon_{HbO_2, \lambda_1} A_{\lambda_2} - \epsilon_{HbO_2, \lambda_2} A_{\lambda_1})}$$

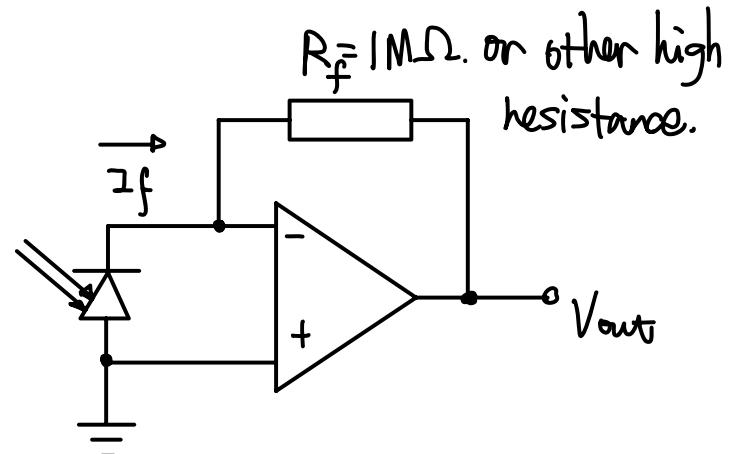
$$\text{or } = \underline{\hspace{10cm}}$$

Theoretical Schematic



- Power.

- Transimpedance Amplifier.

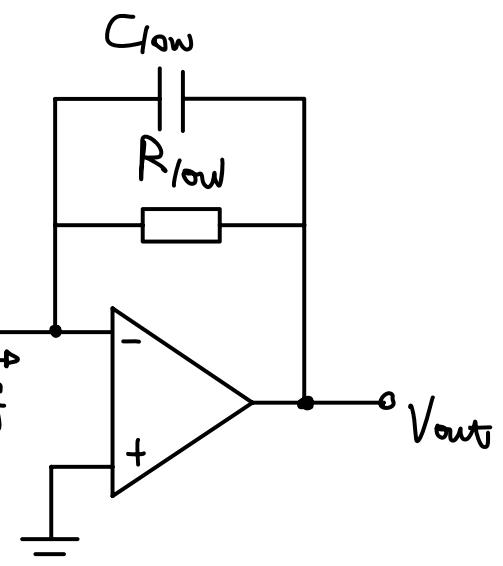


- Measure idle voltage / finger-inserted voltage.

Compare to myRio AI. port.

$$V_{out} = - I_f R_f .$$

- Bandpass Active Filter



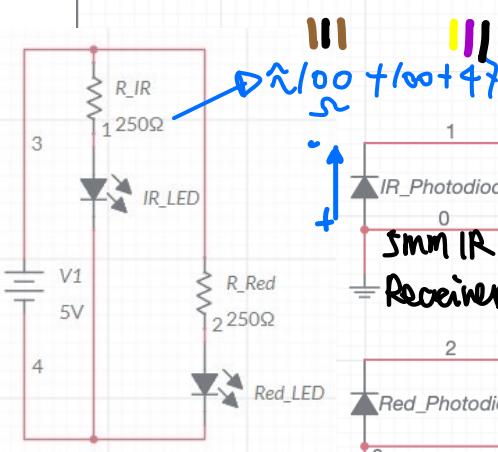
$$\frac{V_{out}}{V_{in}} = \frac{R_{low}}{R_{high}} , f_c = \frac{1}{2\pi R C}$$

Wiring:

* color code: V_{CC} GND Signal (Red) Signal (IR)

$\pm 5 \pm 12$

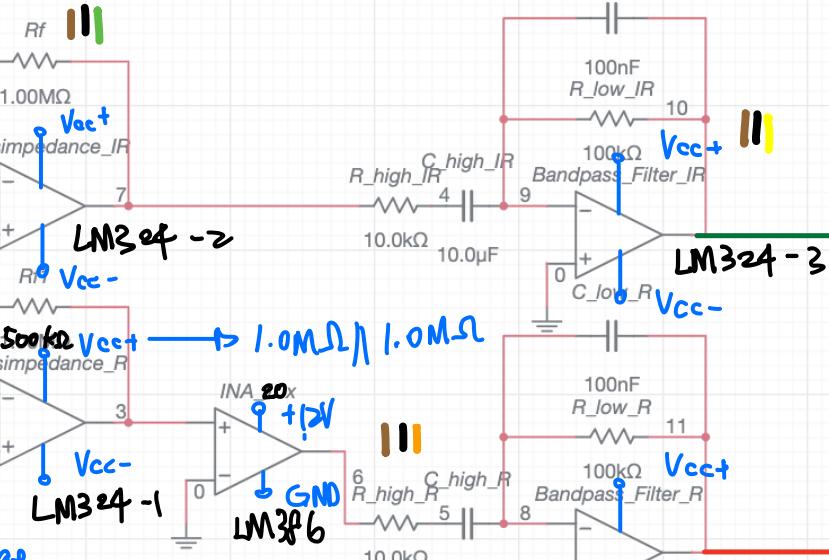
$V_{CC+} : 12V$
 $V_{CC-} : -12V$



* common mode voltage $\frac{1}{2} \text{sum}(IN+, IN-)$

LM324 Supply Voltage: $\pm(3, 30)V$

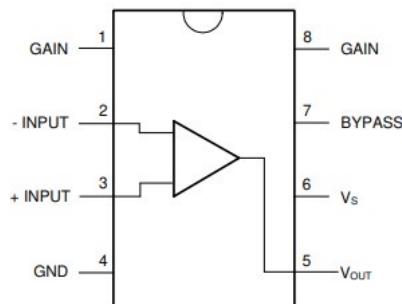
C_{low_IR}



* 2 DC Supply voltage

- ① $+5V$ right rail +
- ② $+12V$ left rail +
- ③ $-12V$ right rail -
- ④ GND left rail -

LM386



LM324

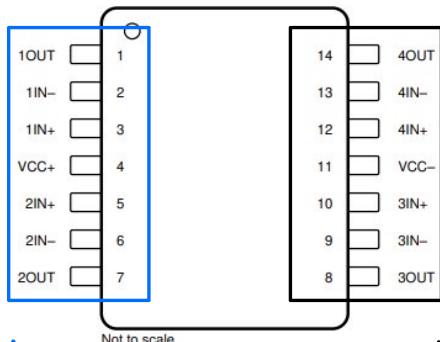
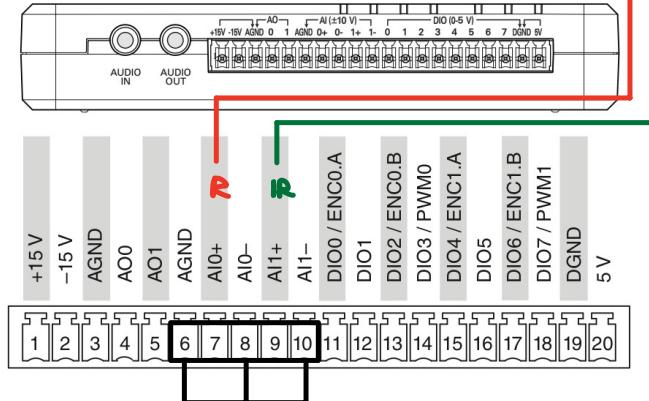


Figure 4. Primary/Secondary Signals on MSP Connector C



$$V_R = V_{A10+} - V_{A10-}$$

$$V_{IR} = V_{A11+} - V_{A11-}$$

Trans-Impedance

Active filter.

Resistor choice: (based on LM324; LM386 has preset gains)

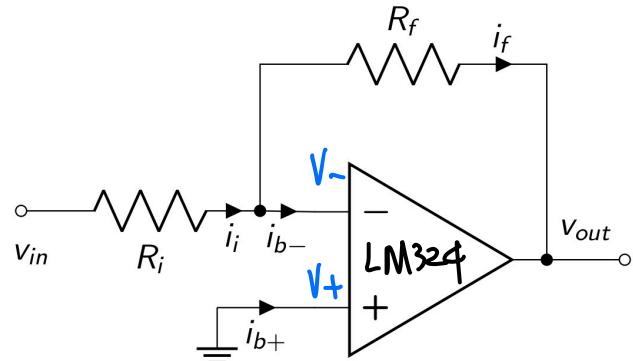
LM324:

- bias current (i_b)

$$\leq \pm 35 \text{ nA}$$

- input impedance (R_{in})

$$= 1 \text{ G}\Omega$$



To avoid signal clipping at A10, let ideal output

$$V_{out,i} = 5V \cdot 90\% = -4.5V$$

and the error tolerance 1%.

In worse case $V_{out} = (1 - 1\%) \times V_{out,i} = -4.4550V$

assume ideal non-inverting input & virtual short.

$$V_- = V_+ = 0V$$

$$i_f = \frac{V_- - V_{out}}{R_f} = \frac{4.4550V}{R_f}$$

$$V_{in} = -V_{out,i} \cdot \frac{R_i}{R_f} = 4.5V \cdot \frac{R_i}{R_f}$$

$$i_i = \frac{V_{in} - V_-}{R_i} = \frac{V_{in}}{R_i} = \frac{4.5V}{R_f}$$

According to KCL (at - input node): $i_i = i_{b-} + i_f$

$$i_{b-} = i_i - i_f \quad \text{and} \quad i_{b-} = 35 \text{ nA} = \frac{4.5V - 4.4550V}{R_f}$$

$$\text{so } R_{f,\text{upper}} = \frac{(4.5 - 4.4550)V}{35 \times 10^{-9} A} = 1.3 \times 10^6 \Omega = 1.3 M\Omega$$

- however, the input impedance $R_{in} = 1 G\Omega$

we want $R_f \lll R_{in}$

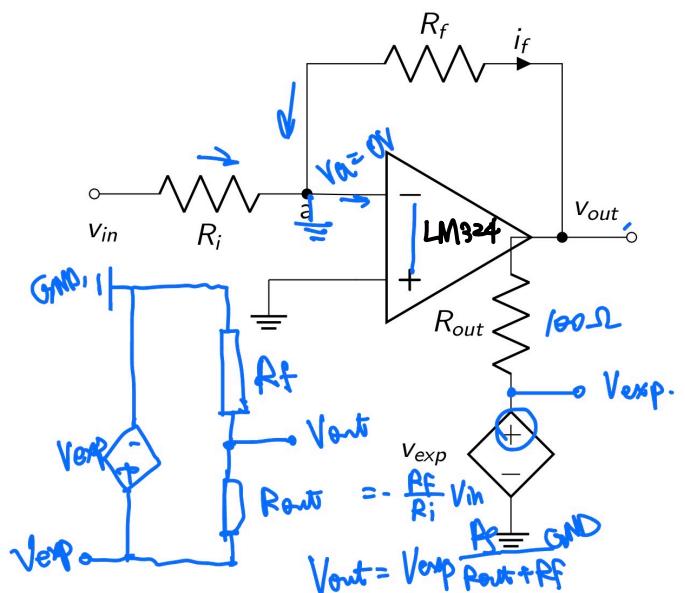
- * the desired upper bound of R_f is around $130 k\Omega$

- output impedance (R_{out})

$$= 300 \Omega$$

- output current (I_{out})

$$\lesssim -60 \text{ mA}$$



- * assume input side fully ideal, $V_a = 0V$

the output impedance R_{out} forms a voltage divider with R_f

if we would like 99% preserved $\frac{V_{out}}{V_{exp}}$

$$V_{out} = V_{exp} \cdot \frac{R_f}{R_f + R_{out}} \Rightarrow \frac{R_{f,\text{lower}}}{R_{f,\text{lower}} + 300 \Omega} = 0.99$$

$$\Rightarrow R_{f,\text{lower}} = 29.7 k\Omega$$

Thus for LM324: $30 k\Omega \leq R_f \leq 130 k\Omega$

Circuit Characterization.

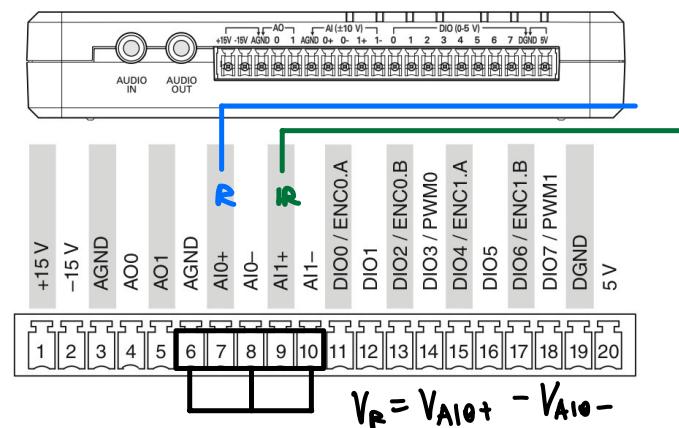
pin layout:

Fgen : offset sine signal. $\begin{cases} R: A/A01,4 \rightarrow t\text{INPUT}, 3 \\ I/R: A/A00,2 \rightarrow s\text{IN-}, 9 \end{cases}$

Circuit Output :

amplified & filtered signal $\begin{cases} R: f\text{OUT}, 14 \rightarrow C/A10^+, 3 \\ I/R: 3\text{OUT}, 8 \rightarrow C/A11^+, 5 \end{cases}$

Figure 4. Primary/Secondary Signals on MSP Connector C

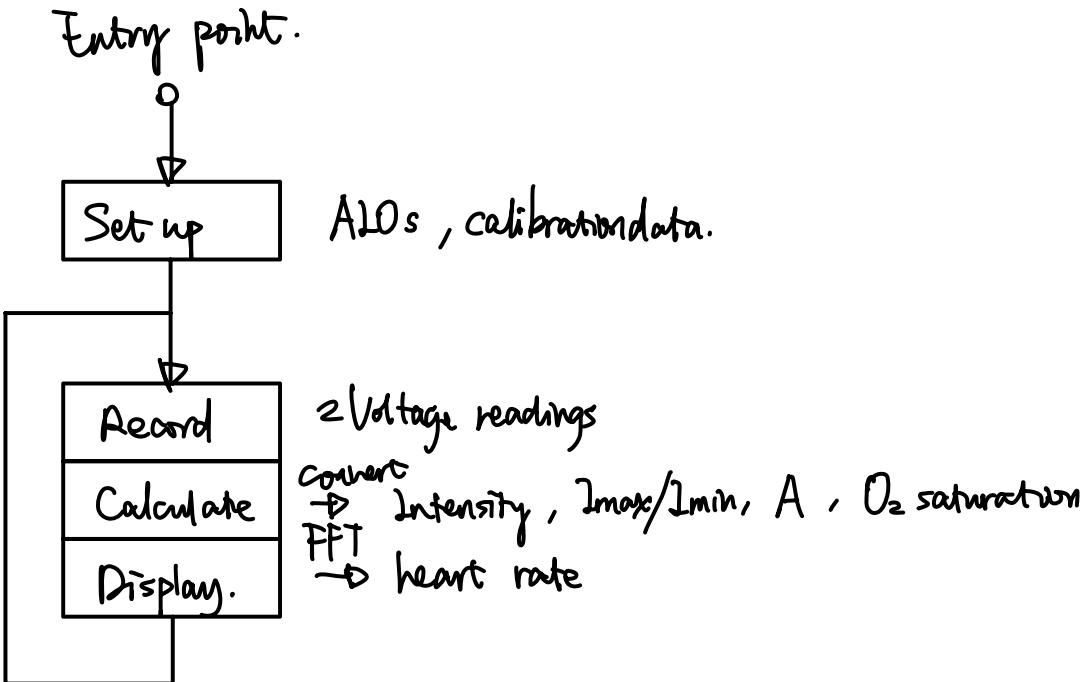


- Characterize : Graph vs. Input Frequency

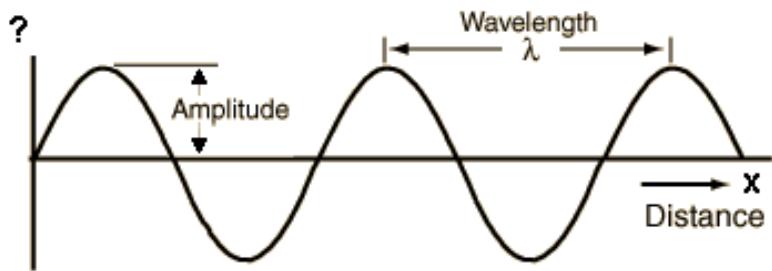
Labview Program Design:

① Super Loop vs. Producer consumer loop.

Super Loop:



② Test Program:



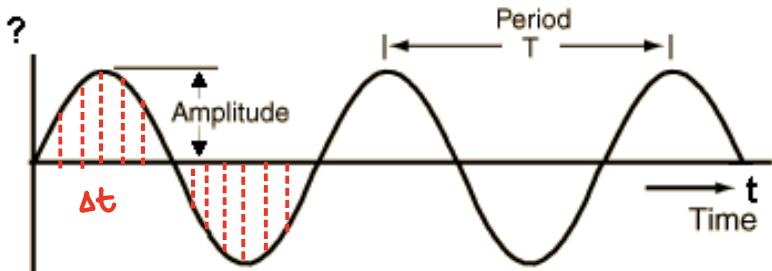
Sine Pattern

Inputs

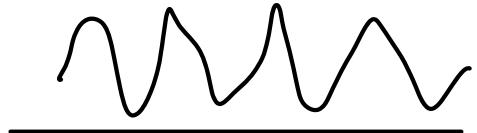
- # sample
- amplitude
- cycles

Output

- sine wave Y array



$$\text{Period } T, \text{ freq } f = \frac{1}{T}$$



Equations:

$$\text{Waveform duration} = \text{Period} \cdot \# \text{ of period.}$$

$$\Delta t \cdot \# \text{sample} = T \cdot \# \text{ cycle.}$$

$$\Delta t \cdot \# \text{sample/cycle} \cdot \cancel{\# \text{cycle}} = \frac{1}{f} \cdot \cancel{\# \text{cycle.}}$$

$$\Delta t = \frac{1}{f \cdot \# \text{sample/cycle}}$$

$$\text{if } f = 10 \text{ Hz} = 10 \text{ s}^{-1}, \# \text{sample/cycle} = 128/\text{cycle}.$$

$$\Delta t = \frac{1}{10 \text{ s}^{-1} \cdot 128} = \frac{1}{1280 \text{ s}^{-1}} = \frac{1}{1280} \text{ s.}$$

Use Scope & Fgen for NI myRIO.

③ Sub V1:

1. SpO₂ calculator.

Inputs: (I_{max}, I_{min}) at λ_{1,2}

$$A_{\lambda_i} = \log \frac{I_{\max, \lambda_i}}{I_{\min, \lambda_i}}, \quad \lambda_i, i \in 1, 2$$

wavelength	Σ_{Hb}	Σ_{HbO_2}	(L/(mmol · cm))
λ ₁ 660	0.81	0.08	
λ ₂ 940	0.18	0.29	

$$SpO_2 = \frac{(\Sigma_{Hb, \lambda_2} A_{\lambda_1} - \Sigma_{Hb, \lambda_1} A_{\lambda_2})}{(\Sigma_{Hb, \lambda_2} A_{\lambda_1} - \Sigma_{Hb, \lambda_1} A_{\lambda_2}) + (\Sigma_{HbO_2, \lambda_1} A_{\lambda_2} - \Sigma_{HbO_2, \lambda_2} A_{\lambda_1})}$$

Output: SpO₂

2. FFT- Frequency Analysis.