

CG3207/EE3207E Computer Architecture

Getting Started – Vivado® Design Suite, Nexys™4 FPGA Development Board & VHDL/Verilog HDL

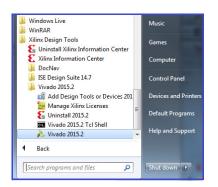
Shahzor Ahmad August 2016

Abstract

This manual provides an illustration of the tools you will use in CG3207/EE3207E through examples of a simple combinational and a simple sequential circuit. It will familiarize you with industry leader Xilinx®'s Vivado® Design Suite¹ — a comprehensive integrated development environment for FPGA design flow, Digilent Inc.'s Nexys™ 4 Development Board featuring an FPGA from Xilinx®'s state-of-the-art Artix®-7 family, and VHDL/Verilog HDL.

¹ This manual assumes Xilinx Vivado WebPACK 2015.2 installed on a Windows 7 PC, but the process should be similar on Windows 8, or later Vivado versions.

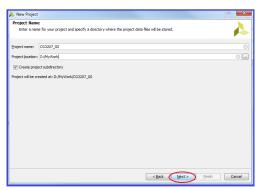
1. Creating a New Vivado Project



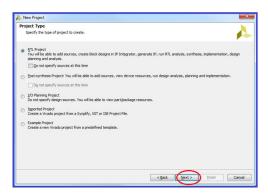
1.1. You can run Xilinx Vivado 2015.2 either by double clicking the icon on Desktop, or going to Start \rightarrow All Programs \rightarrow Xilinx Design Tools \rightarrow Vivado 2015.2 \rightarrow Vivado 2015.2.



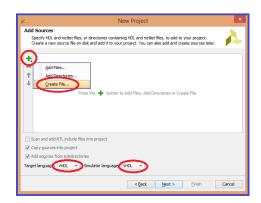
1.2. Having launched Vivado, click on **Create New Project**. The **Create a New Vivado Project**wizard will pop up. Click **Next**.

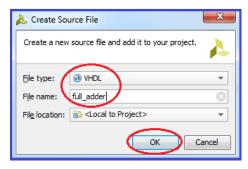


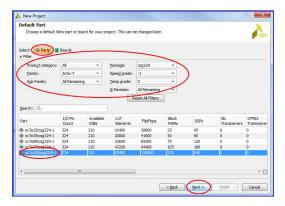
1.3. Enter a **Project name**, e.g., "CG3207_GS". Specify a **Project location**, e.g., "D:/MyWork". Click **Next**.

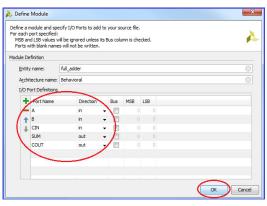


1.4. Ensure **Project Type** is set to **RTL Project**. Click **Next**.









1.5. In the **Add Sources** window, ensure **Target** and **Simulator language** are set to **VHDL** or **Verilog** (depending on your preference).

Click on the green plus button, and then on **Create File...** from the resulting pop-up menu to create a new VHDL source file.

1.6. A **Create Source File** window pops up. Ensure File type is set to your preferred HDL (either **VHDL** or **Verilog**). Enter "full_adder" as **File name**. Click **OK**, which adds a new file to our **Add Sources** window.

Click Next.

On the **Add Existing IP (optional)** window, click **Next**.

On the **Add Constraints (optional)** window, click **Next**. We shall add constraints later on.

1.7. In the Default Part window, we must specify the FPGA chip in use. The Nexys-4 board contains the Xc7a100T chip from the Artix-7 family. We filter by Parts. Choose Artix-7 for Family, csg324 for the Package, -1 for Speed grade and C for Temp grade². This filtering considerably narrows down the options, from which xc7a100tcsg324-1 should be selected.

Click **Next**. A summary is shown. Click **Finish**.

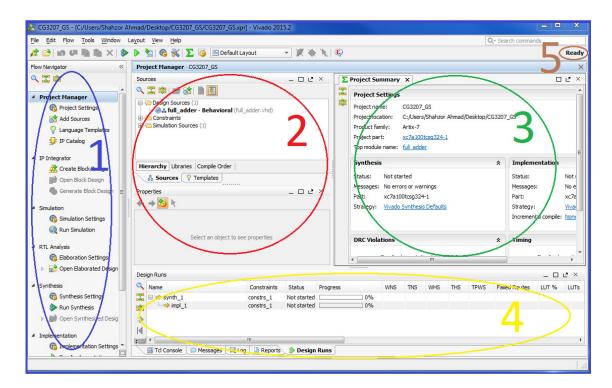
1.8. The wizard then asks us to define a VHDL Entity / Verilog Module to reside in the source file we just created. Leave Entity / Module name same as the source file (full_adder), and the Architecture name as Behavioral. Provide the input ports A, B and CIN³ as shown, and the output ports SUM and COUT.

Click Ok.

² These part details for the Artix-7 chip can be seen from the Nexys 4 schematics/reference manual available from Digilent:

⁽www.digilentinc.com/Products/Detail.cfm?Prod=NEXYS4)

³ Recall the full adder as opposed to a half adder takes a carry-in as an input too.

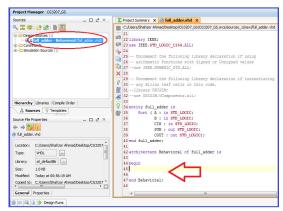


- 1.9. After creating a new project, one is presented with the Vivado IDE viewing environment as shown. The main parts of the GUI (graphical user interface) are highlighted⁴.
- 1 The Flow Navigator provides access to all the tools for a complete FPGA design flow HDL text entry, simulation, synthesis, implementation, bit-stream creation and FPGA programming (configuration).
- 2 The Data Windows Area displays design sources and data. Currently the Sources Window is seen.
- 3 The Workspace displays the Text Editor, Schematic Window, Synthesis reports such as Resource Utilization, etc. Currently **Project Summary** is seen.
- 4 The Results Windows display messages and log files as Vivado performs simulation, synthesis, and implementation, etc.
- 5 The Project Status Bar displays current status of the active design.

⁴ Xilinx Vivado Design Suite User Guide UG893 – Using the Vivado IDE (http://www.xilinx.com/support/documentation/sw_manuals/xilinx2015_2/ug893-vivado-ide.pdf)

2. Design Entry Using Vivado's Text Editor

2.1. VHDL



```
Port ( A : in STD LOGIC:
              B : in STD_LOGIC;
37
             CIN : in STD LOGIC:
              SUM : out STD LOGIC;
39
             COUT : out STD LOGIC):
40 end full adder:
42 architecture Behavioral of full_adder i
44 begi
46
          variable p, g: STD LOGIC ;
       begin
              p := a xor b :
48
              g := a and b ;
50
               sum <= p xor cin ;
51
              cout <= g or (p and cin)
52
```

2.1.1. Double click on the design source **full_adder.vhd** to open and display the file in the Vivado Editor that will be docked in the Workspace.

A VHDL **entity** definition for a full adder circuit is seen. The **architecture** is currently empty. Enter the **combinational process** between the begin and end statements of the architecture as shown in the adjacent figure. VHDL is case in-sensitive, unlike Verilog.

A process resumes whenever a signal in the sensitivity list changes and then suspends again till the next resume. It is often convenient to break a complex function into intermediate steps. We have thus employed internal variables p and g (which are invisible to the world outside the process), before computing the outputs sum and cout. Variable assignments are performed using blocking assignments (:=), which are instantaneous. That is, variables p and g immediately get their new values. Hence the following two signal assignments, sum and cout (performed using non-blocking assignment, <=), use the updated variable values. Unlike variables, signal assignments in a process take effect when the process suspends⁵. Also, variable assignments are evaluated sequentially, while signal assignments run in parallel to each other.

An oft-employed alternate architecture definition for a full adder is shown in adjacent figure. Observe p and g in this scenario are declared as signals, and their new values will take effect (together with sum and cout) only when the process suspends. Hence, sum and cout essentially use the old values of p and g (from the previous run of the process). The process eventually attains the correct answer, but needs to be evaluated twice compared to the process in the previous figure. Both architectures synthesize to the

same hardware, however. Note if one forgets to include p and g in the sensitivity list, simulation results for this architecture are incorrect (output signals remain undefined in the first cycle)! Vivado still synthesizes to the correct hardware, but that means a mismatch between simulation results and what the hardware actually does. Hence, it is recommended to use variable and not signal assignments for internal nets in combinational processes.

⁵ Signal assignment (http://www.vhdl.renerta.com/mobile/source/vhd00063.htm)

Since the full adder is really simple combinational logic, we may forego the process statement altogether, and use concurrent signal assignments as shown in the adjacent figure. A concurrent assignment is activated whenever there is a change in any signal on the R.H.S. It is independent of

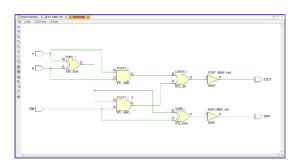
```
42 architecture Behavioral of full_adder is
43
44 begin
45
46 sum <= (a xor b) xor cin;
47 cout <= (a and b) or ((a xor b) and cin);
48
49 end Behavioral;
```

other statements in a given architecture, and is performed concurrently to other active statements. This architecture, similar to the first one, produces a simulation that is in agreement with the synthesised hardware.

2.1.2. Save your HDL entry by clicking File → Save File, or pressing Ctrl+S. If any syntax errors appear in your code (automatically checked by Vivado every time you perform a save), they are indicated in the Messages console in the Results Windows area at the bottom.



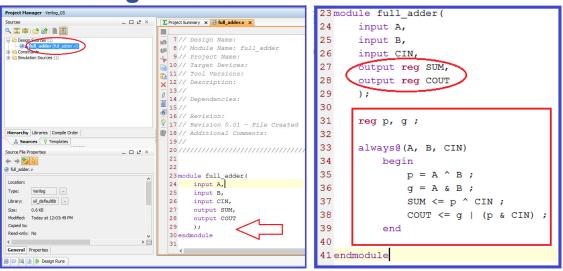
2.1.3. In the Flow Navigator pane, expand the RTL Analysis⁶ tools by clicking on the arrow on the left. Similarly, expand Open Elaborated Design. Click on Schematic. If a message Elaborate Design pops up, click OK. Vivado takes a few seconds to perform RTL Analysis and generate the schematic which is a representation of the pre-optimized design in terms of generic components – adders, multipliers, counters, gates, etc. independent of the target device or technology library.



2.1.4. The schematic displayed in the Workspace is that of a full adder circuit. Observe we had conveyed our design to Vivado through text that describes our intended hardware. RTL analysis can help debug and optimize our design. Select a logic instance in the schematic, right-click to select **Go to Source** or **Go to Definition**, to open the source file containing the module definition or instance of selected object respectively

⁶ HDL code for any reasonably complex design involving sequential circuits is written at a level of abstraction called Register Transfer Level (RTL) rather than actual gate-level representation.

2.2. Verilog



2.2.1. Double click on the design source **full_adder.v** to open and display the file in the Vivado Editor that will be docked in the Workspace. A Verilog module definition for a full adder circuit is seen. The body is currently empty. Modify the module output declarations and populate the body with combinational logic as seen in the adjacent figure. Note that Verilog is case-sensitive, unlike VHDL.

Verilog **always** blocks are evaluated every time the signals in the sensitivity list (or header) change, i.e., A, B, CIN in our example. In Verilog, all signals on the LHS of assignments in always blocks must be declared as **reg**⁷. This does not necessarily mean the signal is actually a register (registers are implied in hardware by assignments inside sequential always blocks). Since the outputs SUM and COUT are modified within our always block, they must be declared as reg.

It is often convenient to break a complex function into intermediate steps. We have thus employed internal signals p and g (which are invisible to the world outside the module), before computing the outputs SUM and COUT. Furthermore, we use **blocking assignments** (=) to instantaneously but sequentially update p and g. Hence the following two signal assignments, SUM and COUT use the updated values of p and g. Due to the use of **non-blocking assignment** (<=) the new values of SUM and COUT take effect concurrently when the always block suspends.

```
33 always@(A, B, CIN, p, g)
34 begin
35 p <= A ^ B;
36 g <= A & B;
37 SUM <= p ^ CIN;
38 COUT <= g | (p & CIN);
39 end
```

An alternate always block for a full adder design is shown in the adjacent figure. Observe the use of non-blocking assignments for p and g in this scenario. Their new values will take effect (together with SUM and COUT) concurrently when the always block suspends. Hence, SUM and COUT essentially use the old values of p and q (from the previous

run of the process). This code eventually attains the correct answer, but needs to be evaluated twice compared to the block in the previous figure. Both modules synthesize to the same hardware, however. Note if one forgets to include p and g in the sensitivity list, simulation results for this module are incorrect (output signals remain undefined in the first cycle)! Vivado still synthesizes to the correct hardware, but that means a mismatch between simulation results and what the hardware actually does.

⁷ "Structural Design with Verilog". David Harris. Available at (http://centaur.sch.bme.hu/~b4k1/labor1/honlaprol/verilog.pdf).

Since the full adder is really simple combinational logic, we may forego the always statement altogether, and use **continuous assignments** as shown in the adjacent figure. An **assign** statement updates the LHS any time the RHS changes, independent of other such continuous statements or always blocks in the module. Continuous assignments cannot appear inside an always block. Notice the use of wire to declare an internal signal prop. Technically, it is not necessary but good practice to declare single-bit wires. It is necessary to declare multi-bit busses. This module,

```
23 module full_adder(
24
      input A,
25
      input B,
26
27
28
      input CIN,
       output SUM,
      output COUT
29
30
       );
31
       wire prop ;
32
33
34
       assign prop = A ^ B ;
       assign SUM = prop ^ CIN ;
       assign COUT = (A & B) | (prop & CIN) ;
35
36
37 endmodule
```

similar to the first one, produces a simulation that is in agreement with the synthesised hardware.

2.2.2. Now follow steps 2.1.2 - 2.1.4 to save your design, check for any syntax errors and perform RTL analysis.

3. Testbench & Behavioral Simulation

Simulation is a major component in digital design flow. Stimuli are applied to the circuit, and the outputs are observed to verify the design behaves as intended. This ensures a system is tested before it is built.

A *testbench* is HDL code to test your design which is called *device under test* (DUT) or *unit under test* (UUT). The testbench contains statements to apply inputs to the DUT, which in our case is our full adder design.

3.1. Expand the **Project Manager** in the Flow Navigator. Click **Add Sources**. In the Add Sources window that pops up, select the **Add or create simulation sources** radio button, and click **Next**. Similar to **Step 1.6** in Section 1, create a VHDL / Verilog file called "test_full_adder" by specifying the name and clicking **OK**. Click **Finish**. Unlike **Step 1.8**, however, do not specify any I/O ports for the module (a testbench is not intended to be synthesized into hardware and does not have I/O ports). Click **OK** on the **Define Module** window. Click **Yes** on the **Define Module** message that pops up.

3.1. VHDL

```
34 entity test_full_adder is
36 end test_full_adder;
38 architecture Behavioral of test full adder is
      -- DECLARE DEVICE UNDER TEST (DUT)
      component full_adder is
         Port ( A : in STD_LOGIC;
                B : in STD LOGIC;
43
                CIN : in STD LOGIC;
                SUM : out STD_LOGIC;
46
                COUT : out STD_LOGIC);
     end component:
       -- DECLARE INTERNAL SIGNALS
50
     signal A, B, CIN, SUM, COUT : STD_LOGIC ;
51
52 begin
       -- INSTANTIATE DUT
     dut: full_adder port map( A, B, CIN, SUM, COUT ) ;
55
56
57
      -- STIMULI
     process
59
60
             A <= '0'; B <= '0'; CIN <= '0'; wait for 10ns;
61
             A <= '0'; B <= '0'; CIN <= '1'; wait for 10ns;
             A <= '0'; B <= '1'; CIN <= '0'; wait for 10ns;
              A <= '0'; B <= '1'; CIN <= '1'; wait for 10ns;
64
              A <= '1': B <= '0': CIN <= '0': wait for 10ns :
             A <= '1'; B <= '0'; CIN <= '1'; wait for 10ns;
65
              A <= '1'; B <= '1'; CIN <= '0'; wait for 10ns;
66
              A <= '1'; B <= '1'; CIN <= '1'; wait for 10ns;
67
69
70
       end process :
72 end Behavioral;
Sources
                                                   \Box \Box \times
 🔍 🔀 🖨 📄 🔠
 □·· Design Sources (1)
     i Simulation Sources (1)
    in test_full_adder - Behavioral (test_full_adder.vhd) (1)
            ···· ··· ··· ··· ··· ··· dut - full_adder - Behavioral (full_adder.vhd)
```

3.1.1. Similar to Step 2.1.1, open up the simulation source **test_full_adder.vhd** in the Vivado Editor. This is our **testbench**. The entity declaration is already present, with no port definitions. Code the architecture as shown in the adjacent figure.

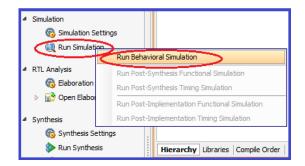
The declarative part of our testbench's architecture contains a component declaration for the DUT defined earlier (the entity full_adder), as well as internal nets (used to handle I/O for the DUT).

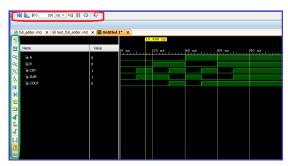
The first statement after begin instantiates the DUT. Observe from **Simulation Sources** in the **Data Windows** area that this causes full_adder to become a leaf node in the hierarchy for test full adder.

Then follows a process which provides input patterns to our DUT. The wait statements suspend the process for the specified interval between each stimulus⁸. The last wait suspends the process indefinitely, else the process would keep repeating the pattern.

(http://www.vhdl.renerta.com/mobile/source/vhd00 081.htm)

⁸ Wait statement





3.1.2. Expand the **Simulation** tools in the **Flow Navigator**, click **Run Simulation** → **Run Behavioral Simulation**.

The Vivado simulator pops open a few seconds later, having run our testbench.

3.1.3. Play around with the tools in the red and green boxes, use the mouse wheel to zoom in/out, and move the yellow bar around (observing the Value column as you do).

Initially, you may not see the waveform as in the figure as the time resolution is very small, and the simulation by default has run for a larger period of time.

Select View → Zoom Fit or use the Zoom Fit icon in the green box.

Verify from this simulation that our full adder design behaves as intended.

3.1.4. By default, the Vivado simulator adds signals from the testbench to the **Waveform** window. In more complex designs (such as CG3207/EE3207E Labs), it would be desirable to **add and monitor signals** from the instantiated DUT or even sub-entities/modules at various levels of the design hierarchy. To add and monitor signals from a desired entity/module, it may be selected from the **Scopes** window. The signals to be monitored may be selected from the **Objects** window, dragged and dropped into the **Waveform** window⁹.

⁹ For more information on adding and monitoring signals from instantiated DUT and sub-modules, and working with the Waveform window, see UG937

⁽http://www.xilinx.com/support/documentation/sw_manu_als/xilinx2015_2/ug937-vivado-design-suite-simulation-tutorial.pdf).

3.2. Verilog

```
23 module test full adder (
25
26
27
28
      // DECLARE INPUT SIGNALS
      reg A ;
29
      reg B ;
30
      reg CIN ;
31
32
      // DECLARE OUTPUT SIGNALS
33
      wire SUM ;
34
      wire COUT ;
35
36
      // INSTANTIATE DUT
37
      full_adder dut(A, B, CIN, SUM, COUT) ;
38
39
      // STIMULI
40
      initial
             A = 0; B = 0; CIN = 0; #10;
42
              A = 0; B = 0; CIN = 1; #10;
              A = 0; B = 1; CIN = 0; #10;
44
              A = 0; B = 1; CIN = 1; #10;
46
              A = 1 ; B = 0 ; CIN = 0 ; #10 ;
47
              A = 1 ; B = 0 ; CIN = 1 ; #10 ;
48
              A = 1 ; B = 1 ; CIN = 0 ; #10 ;
49
              A = 1 ; B = 1 ; CIN = 1 ; #10 ;
50
51
52 endmodule
```

```
Sources

Design Sources (1)

Constraints

Simulation Sources (1)

Simulation Sources (1)
```

3.2.1. Similar to Step 2.2.1, open up the simulation source **test_full_adder.v** in the Vivado Editor. This is our **testbench**. The module declaration is already present, with no port definitions. Code the body as shown in the adjacent figure.

The declarative part of our testbench module contains an instantiation of the DUT defined earlier (the module full_adder), as well as internal nets (used to handle I/O for the DUT).

Upon instantiating the DUT, observe from **Simulation Sources** in the **Data Windows** area that this causes full_adder to become a leaf node in the hierarchy for test_full_adder.

Then follows an initial block which provides input patterns to our DUT. Similar to the case of an always block, signals appearing on the LHS of assignments in an initial block must be declared as reg. Note initial statements should only be used in testbenches for simulation, not in modules intended to be synthesized into actual hardware.

The timescale directive at the top of our test_full_adder.v file:

'timescale 1ns / 1ps
Specifies a time unit in this file to be 1ns, and
the simulation to have a precision of 1ps. The #
statements suspend the initial block for the
specified interval between each stimulus.

3.2.2. Now follow steps 3.1.2 - 3.1.4 to simulate your design and analyse the resulting waveforms.

4. Design Constraints

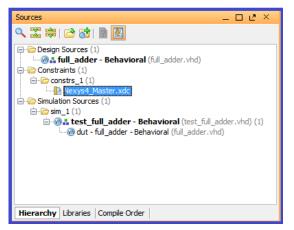
Design constraints such as timing and physical I/O pin mapping must be defined before synthesizing a design for a target FPGA. We shall not specify any timing constraints for now as our circuit is combinational; no sequential elements exist in our design (though hardly the case in practice).

4.1. Constraints are specified via a .xdc file (Xilinx Design Constraints). Rather than writing constraints from scratch, let us use Nexys 4 Master XDC provided by Digilent here https://reference.digilentinc.com/media/reference/programmable-logic/nexys-4/nexys4 master xdc.zip]. Download and unzip the file in some directory.



4.2. Click on **Project Manager** in the **Flow Navigator**. Click on **Add Sources**. Select the **Add or create constraints** radio button. Click **Next**.

On the Add or Create Constraints window, use the green plus button to Add Files..., and add the Nexys4_Master.xdc you just downloaded from the location you unzipped it into. Let the option Copy constraints files into project remain checked, so that a fresh copy is placed in your Vivado project folder.



4.3. The file is added to the **Constrains** folder in the **Sources** window. Open it up for editing in the Workspace.

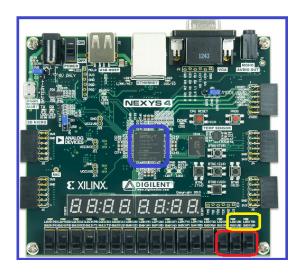
Currently all code is commented out in the file. Depending on our design, we simply need to uncomment the lines corresponding to the pins we use and rename the ports if necessary.



4.4. Un-comment the lines as in the adjacent figure. Rename the ports to \mathbb{A} , \mathbb{B} and \mathbb{CIN} as shown. This essentially maps inputs \mathbb{A} , \mathbb{B} and \mathbb{CIN} of our full_adder entity to SW0, SW1 and SW3 respectively on the Nexys 4 board.



4.5. Similarly, map the outputs SUM and COUT to LED0 and LED1 respectively as shown.

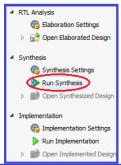


4.6. The said switches and LEDs are indicated by red and yellow boxes in the adjacent figure, respectively. The Artix-7 FPGA is also indicated in blue.

WARNING: the chips on the board are electrostatic sensitive (and costly!). Avoid touching. Handle the board by the edges.

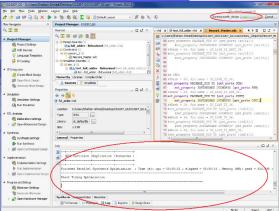
5. Synthesis

Logic synthesis transforms RTL HDL code into an optimized set of logic gates to reduce the amount of hardware and efficiently perform the intended function.



5.1. Expand **Synthesis** tools in the **Flow Navigator**, and click **Run Synthesis**.

Synthesis is usually the most time-consuming part of the FPGA design flow. In our simple example, however, it should only take a few moments.

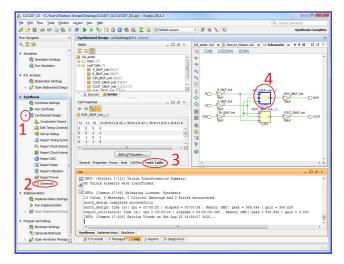


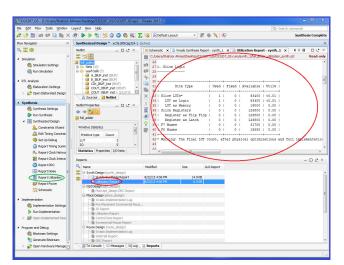
5.2. While Vivado performs synthesis, the **Project Status Bar** at the top right of the IDE provides visual indication.

A log appears in the **Log** tab in the **Results Windows**, while any warnings or errors are displayed in the **Messages** tab.



5.3. Upon completion, a **Synthesis Completed** messages pops up, asking to proceed to perform implementation. Click **Cancel**, as we'd like to first analyse the synthesized output.





5.4. Expand **Open Synthesized Design** under **Synthesis** in the **Flow Navigator**. Click **Schematic**. This opens up the synthesized netlist in the Sources window describing the hardware. A schematic appears in the Workspace.

Select the LUT3¹⁰ labeled **SUM_OBUF_inst_i_1** in the schematic and then click on **Truth Table** in the Data Windows area. You will observe from the Truth Table that it indeed encodes the behaviour for SUM output. Similarly the CARRY output is encoded as another Truth Table **COUT_OBUF_inst_i_1**.

5.5. Resource utilization may be opened in the Workspace for analysis by clicking **Utilization Report** (red) in the **Reports** tab at the bottom of the IDE.

We observe only one slice¹¹ LUT is used as logic. No flip-flops are used as this is a purely combinational circuit.

Utilization may also be observed by clicking on **Report Utilization** in the **Flow Navigator**. Similarly, Report Timing Summary, Report Noise and Report Power may be used.

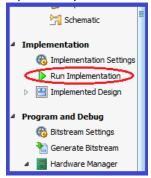
Post-implementation reports may be similarly analysed, and are more accurate as they are generated after placement and routing.

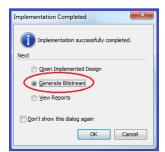
¹⁰ Field-Programmable Gate Arrays (FPGAs) don't really contain gates! Instead they contain lots of CLBs (configurable logic blocks), consisting of multi-input LUTs (look-up tables) to implement combinational logic, and flip-flops to implement registers. It is easy to appreciate from this example how a LUT has been configured to behave as a full adder.

¹¹ XC7A100T contains 15850 slices. Each slice contains 04 LUTs, 8 flip-flops. Two slices form a CLB. Other resources on the FPGA include dedicated DSP slices, Block RAM, Clock Management & PLL for frequency synthesis, ADCs and I/O.

6. Implementation, Bitstream Generation & FPGA Configuration

Implementation involves optimal placement and routing (PAR), where the netlist elements are mapped to the FPGA's physical resources and interconnected together. The I/O and timing constraints specified by the user are also respected.









6.1. Click on **Run Implementation** under **Implementation** in the **Flow Navigator**. If a Synthesis is Out-of-date message pops up, prompting to re-run synthesis, click **Yes**.

While Vivado performs synthesis and/or implementation, the **Project Status Bar** at the top right of the IDE provides visual indication.

A log appears in the **Log** tab in the **Results Windows**, while any warnings or errors are displayed in the **Messages** tab.

6.2. Once implementation is finished, Vivado asks whether to proceed to generate a bitstream (which would then be used to configure the FPGA with our design). Select **Generate Bitstream** radio button, and click **OK**.

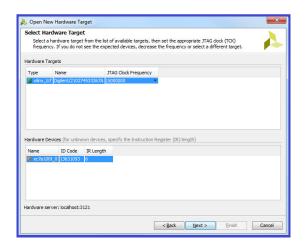
6.3. Once done, select **Open Hardware Manager** and click **OK**.

In the Flow Navigator, under Program and Debug → Hardware Manager → click Open Target → Open New Target.

Since Hardware Manager was opened just now, same may also be done from the pane just above the Sources Window / Text Editor.

6.4. Click **Next**. Ensure **Local server** is selected. At this point plug in your FPGA board into the USB port of your PC, and turn on the power switch (SW16)¹². Click **Next**.

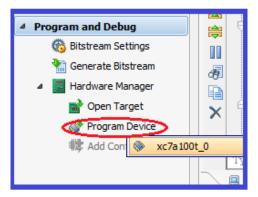
¹² Every time you power it up, the FPGA is auto configured to behave as a demo circuit by an on-board flash device. Some of the cool stuff: pressing BTNU triggers a recording for 5s. For the next 5s, you can hear it if you plug your earphones into the audio jack! Also try hooking up the VGA output to a screen, and connecting a mouse into the USB host port.



6.5. If the board cannot be detected, power cycle the board, wait for Windows USB drivers to load up and try again.

If successfully detected, a hardware target (JTAG interface on the Nexys-4 board), and a hardware device (our Artix-7 chip) is shown.

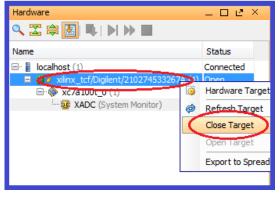
Select the xc7a100t_0 device. Click Next and Finish



6.6. Click **Program Device**. Then click the **xc7a100t_0** that pops up.



- 6.7. Click **Program** on the **Program Device** window that appears.
- 6.8. Once the FPGA is programmed with our design, verify the functionality by providing inputs with switches SW0, SW1 and SW2, and observing LEDs LD0 and LD1.



6.9. Before switching off your Nexys 4, disconnect it from Vivado by right-clicking on the the taget shown in **Hardware** window → **Close Target**.

Congratulations!
You just successfully completed your first
Artix-7 FPGA design flow ©

7. Design Flow for a Sample Sequential Circuit – VHDL

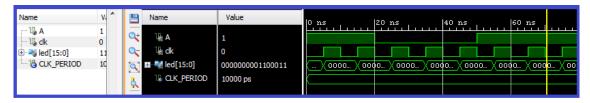
```
22 library IEEE:
23 use IEEE.STD LOGIC 1164.ALL;
24 use ieee.numeric_std.all;
26 entity shifter is
    Port ( clk : in STD_LOGIC;
            A : in STD LOGIC:
           led : out STD_LOGIC_VECTOR (15 downto 0));
32 architecture Behavioral of shifter is
   signal SHIFT_REG : STD LOGIC VECTOR (15 downto 0) := (others => '0') ;
            - 16-bit shift register
    signal CLK_1HZ : STD LOGIC ;
     signal counter_for_CLK_1HZ : STD_LOGIC_VECTOR (25 downto 0) := (others => '0') ;
          -- 26-bit register to count 50M at 10ns clock
38 begin
      --CLK 1HZ <= clk ;
40
      -- 15-bit SHIFT REGISTER
    process( CLK_1HZ )
          begin
              if CLK_1HZ'event and CLK_1HZ = '1' then
45
                  SHIFT REG(15 downto 1) <= SHIFT REG(14 downto 0) ;
46
                 SHIFT_REG(0) <= A ;
             end if :
48
         end process ;
49
50
      -- OUTPUT TO LEDS
51
     led <= SHIFT REG ;</pre>
      -- CLOCK DIVIDER
     process( clk )
          begin
              if clk'event and clk = '1' then
                 counter_for_CLK_1HZ <= std_logic_vector(unsigned(counter_for_CLK_1HZ) + 1) ;</pre>
                      -- increment counter
                 if unsigned (counter for CLK 1HZ) = 50000000 then
                     counter_for_CLK_1HZ <= (others => '0') ;
                      CLK 1HZ <= not CLK 1HZ ;
                 end if ;
65
              end if :
          end process ;
69 end Behavioral;
```

The HDL above implements a **16-bit shift register** via a clocked process. A crystal oscillator provides a 100MHz clock to the Artix-7 (connected to input port clk). The code above also implements a 26-bit counter to divide the clock and generate a 1Hz clock. This 1Hz signal is used to clock the shift register. Input port A – connected to SWO – is used to provide new input to bit 0 of the shift register every clock period. The status of the shift register is displayed on LEDs.

- 7.1. Create a project and design source as in Sec. 1. As in Sec. 2, edit the source file to look like the figure above.
- 7.2. Create a simulation source (see Sec. 3) to enter the testbench as in figure below. Observe a process using wait statements essentially generates the clk. Sample stimuli have been added, which can be used to verify the behaviour of the design. NOTE: for simulation, un-comment line#39 and comment out lines # 54 69 in the design source, while the opposite should be done for implementation (Why?). You should be able to obtain the simulation output as given in the figure and analyse it.

```
22 library IEEE;
23 use IEEE.STD_LOGIC_1164.ALL;
24
25 entity test_shifter is
26 -- Port ();
27 end test shifter;
28
29 architecture Behavioral of test shifter is
30
31 -- DECLARE DEVICE UNDER TEST (DUT)
32 component shifter is
33
     Port ( clk : in STD LOGIC;
            A : in STD LOGIC;
34
           led : out STD_LOGIC_VECTOR (15 downto 0));
35
36 end component;
38 -- DECLARE INTERNAL SIGNALS
39 signal A, clk : STD LOGIC ;
40 signal led : STD_LOGIC_VECTOR (15 downto 0) ;
41
42 -- CLOCK PERIOD DEF
43 constant CLK_PERIOD : time := 10 ns ;
44
45 begin
46
47
      -- TNSTANTTATE DUT
48
      dut: shifter port map( clk, A, led ) ;
50
      -- STIMULI
51
     process
52
         begin
             A <= '1'; wait for 2*CLK_PERIOD;
53
              A <= '0'; wait for 3*CLK_PERIOD;
             A <= '1';
56
             wait ;
57
      end process ;
58
59
      -- GENERATE CLOCK
60
      process
             clk <= '0';
62
             wait for CLK_PERIOD / 2;
63
             clk <= '1' :
64
65
              wait for CLK_PERIOD / 2 ;
66
          end process ;
67
68 end Behavioral;
```

- 7.3. Now revert back to the original source. Add the Nexys 4 master constraints file as in Sec. 4. Un-comment the lines for clock signal and input SWO (rename it to A) as in figure below. You will also need to un-comment constraint lines for all LEDs (no renaming required).
- 7.4. Synthesize and implement this design as in Secs. 5 6. Verify the functionality of the circuit in hardware.



```
6 ## Clock signal
7 ##Bank = 35, Pin name = IO_L12P_T1_MRCC_35, Sch name = CLK100MHZ
8 set_property PACKAGE_PIN E3 [get_ports clk]
9 set_property IOSTANDARD LVCMOS33 [get_ports clk]
10 #create_clock -add -name sys_clk_pin -period 10.00 -waveform {0 5} [get_ports clk]
11
12 ## Switches
13 ##Bank = 34, Pin name = IO_L21P_T3_DQS_34, Sch name = SWO
14 set_property PACKAGE_PIN U9 [get_ports A]
15 set_property IOSTANDARD LVCMOS33 [get_ports A]
```

8. Design Flow for a Sample Sequential Circuit – Verilog

```
23 module shifter (
24
      input clk,
25
      input A,
26
     output [15:0] led
27
28
29
      // 15-bit SHIFT REGISTER
30
     reg [15:0] SHIFT_REG = 0;
31
      always@(posedge clk_1Hz)
32
33
          SHIFT_REG[15:0] <= {SHIFT_REG[14:0], A};
34
             // shift all bits one position to the left, bring in A from the right
35
      end
36
37
      // OUTPUT TO LEDs
38
      assign led = SHIFT REG ;
39
     // CLOCK DIVIDER
40
41
     reg clk 1Hz = 0;
     reg [25:0] counter for clk 1Hz = 0;
43
          // 26-bit register to count 50M at 10ns clock
44
      always@(posedge clk)
45
      begin
          if( counter_for_clk_1Hz == 26'd50000000 )
46
47
48
                  counter_for_clk_1Hz <= 0 ;</pre>
49
                  clk_1Hz <= ~clk_1Hz ;
50
              end
51
          else
52
              counter_for_clk_1Hz <= counter_for_clk_1Hz + 1 ;</pre>
53
54 //
        assign clk_1Hz = clk;
55
56 endmodule
```

The HDL above implements a **16-bit shift register** using an <code>always</code> block sensitized to the positive clock edge. A crystal oscillator provides a 100MHz clock to the Artix-7 (connected to input port clk). The code above also implements a 26-bit counter to divide the clock and generate a 1Hz clock. This 1Hz signal is used to clock the shift register. Input port A – connected to SWO – is used to provide new input to bit 0 of the shift register every clock period. The status of the shift register is displayed on LEDs.

- 8.1. Create a project and design source as in Sec. 1. As in Sec. 2, edit the source file to look like the figure above.
- 8.2. Create a simulation source (see Sec. 3) to enter the testbench as in figure below. Sample stimuli have been added, which can be used to verify the behaviour of the design. NOTE: for simulation, uncomment line#54 and comment out lines #41-53 in the design source, while the opposite should be done for implementation (Why?). You should be able to obtain the simulation output as given in the figure on page 18 and analyse it.

```
23 module test_shifter(
24
25
26
27
     // DECLARE INPUT SIGNALS
28
    reg clk ;
29
    reg A ;
30
31
     // DECLARE OUTPUT SIGNALS
32
     wire [15:0] led;
33
34
     // INSTANTIATE DUT
35
     shifter dut(clk, A, led) ;
36
     // STIMULI
37
38
     initial
      begin
39
40
            clk = 0;
41
            A = 1'b1 ; #20 ;
            A = 1'b_0 ; #30 ;
             A = 1'b1;
44
         end
45
46
      // GENERATE CLOCK
47
      always
48
        begin
          #5 clk = ~clk ;
49
50
               // invert clk every 5 time units
51
          end
52
53 endmodule
```

- 8.3. Now revert back to the original source. Add the Nexys 4 master constraints file as in Sec. 4. Uncomment the lines for clock signal and input SWO (rename it to A) as in figure on page 18. You will also need to un-comment constraint lines for all LEDs (no renaming required).
- 8.4. Synthesize and implement this design as in Secs. 5 6. Verify the functionality of the circuit in hardware.