Digital Signal Processing

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Software Installation

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to	Abstract—This manual provides a simple in digital signal processing.	ntroduction
	1 Software Installation	

sudo apt-get update sudo apt-get install libffi-dev libsndfile1 python3 -scipy python3-numpy python3-matplotlib sudo pip install cffi pysoundfile

2 Digital Filter

2.1 Download the sound file from

Run the following commands

wget https://raw.githubusercontent.com/ gadepall/ EE1310/master/filter/codes/Sound_Noise.wav

2.2 You will find a spectrogram at https://academo.org/demos/spectrum-analyzer. Upload the sound file that you downloaded in Problem 2.1 in the spectrogram and play. Observe the spectrogram. What do you find? **Solution:** There are a lot of yellow lines between 440 Hz to 5.1 KHz. These represent the synthesizer key tones. Also, the key strokes are audible along with background noise.

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2.3 Write the python code for removal of out of band noise and execute the code.

Solution:

```
import soundfile as sf
from scipy import signal
#read.wav file
input signal,fs = sf.read('Sound Noise.wav'
#sampling frequency of input signal
sampl freq = fs
#order of the frequency
order = 4
#cutoff frequency 4khz.
cutoff freq = 4000.0
#digital frequency
Wn = 2*cutoff freq/sampl freq
# b and a are numerators and denominator
   polynomials respectively
b,a = signal.butter(order,Wn,'low')
#filter the input signal with butterworth filter
output signal = signal.filtfilt(b,a,input signal
\#output \quad signal = signal.lfilter(b,a,
   input signal)
#write the output signal into .wav file
sf.write('Sound With ReducedNoise.wav',
    output signal,fs)
```

2.4 The output of the python script Problem 2.3 is the audio file Sound With ReducedNoise.wav. Play the file in the spectrogram in Problem 2.2. What do you observe?

Solution: The key strokes as well as background noise is subdued in the audio. Also, the signal is blank for frequencies above 5.1 kHz.

3 DIFFERENCE EQUATION

3.1 Let

$$x(n) = \left\{ 1, 2, 3, 4, 2, 1 \right\} \tag{3.1}$$

Sketch x(n).

3.2 Let

$$y(n) + \frac{1}{2}y(n-1) = x(n) + x(n-2),$$

$$y(n) = 0, n < 0 \quad (3.2)$$

Sketch y(n).

Solution: The following code yields Fig. 3.2.

wget https://github.com/NareshBandaru13/ EE3900-A1/blob/main/codes/3-2.py

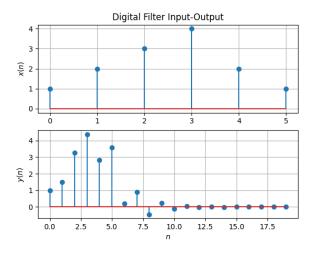


Fig. 3.2

3.3 Repeat the above exercise using a c code. **Solution:** The following code yields x data and y data

wget https://github.com/NareshBandaru13/ EE3900-A1/blob/main/codes/3-3.c

4 Z-TRANSFORM

4.1 The Z-transform of x(n) is defined as

$$X(z) = \mathbb{Z}\{x(n)\} = \sum_{n=-\infty}^{\infty} x(n)z^{-n}$$
 (4.1)

Show that

$$Z{x(n-1)} = z^{-1}X(z)$$
 (4.2)

and find

$$\mathcal{Z}\{x(n-k)\}\tag{4.3}$$

Solution: From (4.1),

$$Z\{x(n-k)\} = \sum_{n=-\infty}^{\infty} x(n-1)z^{-n}$$

$$= \sum_{n=-\infty}^{\infty} x(n)z^{-n-1} = z^{-1} \sum_{n=-\infty}^{\infty} x(n)z^{-n}$$
(4.4)
$$(4.5)$$

resulting in (4.2). Similarly, it can be shown that

$$\mathcal{Z}\{x(n-k)\} = z^{-k}X(z) \tag{4.6}$$

4.2 Obtain X(z) for x(n) defined in problem 3.1 **Solution:**

$$X(z) = \ddagger \{x(n)\} = \sum_{n=-\infty}^{\infty} x(n)z^{-n}$$

$$X(z) = ...0 + x(0)z^{0} + x(1)z^{-1} + x(2)z^{-2} + x(3)z^{-3} + x(4)z^{-1}$$

= 1 + 2z⁻¹ + 3z⁻² + 4z⁻³ + 2z⁻⁴ + 1z⁻⁵

4.3 Find

$$H(z) = \frac{Y(z)}{X(z)} \tag{4.7}$$

from (3.2) assuming that the Z-transform is a linear operation.

Solution: Applying (4.6) in (3.2),

$$Y(z) + \frac{1}{2}z^{-1}Y(z) = X(z) + z^{-2}X(z)$$
 (4.8)

$$\implies \frac{Y(z)}{X(z)} = \frac{1 + z^{-2}}{1 + \frac{1}{2}z^{-1}} \tag{4.9}$$

4.4 Find the Z transform of

$$\delta(n) = \begin{cases} 1 & n = 0 \\ 0 & \text{otherwise} \end{cases}$$
 (4.10)

and show that the Z-transform of

$$u(n) = \begin{cases} 1 & n \ge 0 \\ 0 & \text{otherwise} \end{cases}$$
 (4.11)

is

$$U(z) = \frac{1}{1 - z^{-1}}, \quad |z| > 1 \tag{4.12}$$

Solution: It is easy to show that

$$\delta(n) \stackrel{\mathcal{Z}}{\rightleftharpoons} 1 \tag{4.13}$$

and from (4.11),

$$U(z) = \sum_{n=0}^{\infty} z^{-n}$$
 (4.14)

$$=\frac{1}{1-z^{-1}}, \quad |z| > 1 \tag{4.15}$$

using the fomula for the sum of an infinite geometric progression.

4.5 Show that

$$a^{n}u(n) \stackrel{\mathcal{Z}}{\rightleftharpoons} \frac{1}{1 - az^{-1}} \quad |z| > |a| \tag{4.16}$$

Solution: $X(z) = \sum_{n=-\infty}^{\infty} x(n)z^{-n}$

$$X(z) = \sum_{n=-\infty}^{-1} 0 + \sum_{n=0}^{\infty} a^{n} (1) z^{-n}$$
$$= \sum_{n=0}^{\infty} (az^{-1})^{n}$$
$$= \frac{1}{1 - az^{-1}}$$

Also the series is convergent for |z| > |a|

4.6 Let

$$H(e^{j\omega}) = H(z = e^{j\omega}).$$
 (4.17)

Plot $|H(e^{j\omega})|$. Comment. $H(e^{j\omega})$ is known as the Discret Time Fourier Transform (DTFT) of x(n).

Solution: The following code plots Fig. 4.6.

wget https://github.com/NareshBandaru13/ EE3900-A1/blob/main/codes/4-6.py

run the command

$$H(e^{j\omega}) = \frac{1 + e^{-2j\omega}}{1 + \frac{1}{2}e^{-j\omega}}$$
(4.18)

$$\Rightarrow |H(e^{j\omega})| = \frac{\left|1 + \cos 2\omega - j\sin 2\omega\right|}{\left|1 + \frac{1}{2}\cos \omega - \frac{1}{2}\sin \omega\right|}$$

$$= \sqrt{\frac{(1 + \cos 2\omega)^2 + (\sin 2\omega)^2}{(4.19)^2}}$$

$$= \sqrt{\frac{(1+\cos 2\omega)^2 + (\sin 2\omega)^2}{(1+\frac{1}{2}\cos \omega)^2 + (\frac{1}{2}\sin \omega)^2}}$$
(4.20)

$$=\sqrt{\frac{2+2\cos 2\omega}{\frac{5}{4}+\cos \omega}}\tag{4.21}$$

$$= \sqrt{\frac{2(2\cos^2\omega)4}{5 + 4\cos\omega}}$$

$$= \frac{4|\cos\omega|}{\sqrt{5 + 4\cos\omega}}$$
(4.22)

$$=\frac{4\left|\cos\omega\right|}{\sqrt{5+4\cos\omega}}\tag{4.23}$$

The period of the numerator is π and the de-

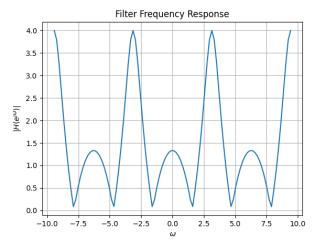


Fig. 4.6: $|H(e^{j\omega})|$

nominator is 2π taking the L.C.M of numerator and the denominator we get period as 2π

4.7 Express h(n) in terms of $H(e^{2w})$.

Solution:

$$\int_{-\pi}^{\pi} H(e^{j\omega})e^{j\omega n}d\omega \tag{4.24}$$

$$= \int_{-\pi}^{\pi} \sum_{k=-\infty}^{\infty} h(k) e^{-j\omega k} e^{j\omega n} d\omega \qquad (4.25)$$

$$=\sum_{k=-\infty}^{\infty}h(k)\int_{-\pi}^{\pi}e^{j\omega(n-k)}d\omega$$
 (4.26)

Now,

$$\int_{-\pi}^{\pi} e^{j\omega(n-k)} d\omega = \begin{cases} \int_{-\pi}^{\pi} d\omega & n-k=0\\ \frac{\exp(j\omega(n-k))}{j(n-k)} \Big|_{-\pi}^{\pi} & n-k \neq 0 \end{cases}$$

$$(4.27)$$

$$= \begin{cases} 2\pi & n-k=0\\ 0 & n-k \neq 0 \end{cases}$$
 (4.28)

$$=2\pi\delta(n-k)\tag{4.29}$$

Thus,

$$\int_{-\pi}^{\pi} H(e^{j\omega})e^{j\omega n}\omega = 2\pi \sum_{k=-\infty}^{\infty} h(k)\delta(n-k) \quad (4.30)$$

$$= 2\pi h(n) * \delta(n) \tag{4.31}$$

$$=2\pi h(n) \tag{4.32}$$

Therefore, h(n) is given by the inverse DTFT (IDTFT) of $H(e^{j\omega})$

$$h(n) = \frac{1}{2\pi} \int_{-\pi}^{\pi} H(e^{j\omega}) e^{j\omega n} \omega \qquad (4.33)$$

5 Impulse Response

5.1 Using long division, find

$$h(n), \quad n < 5 \tag{5.1}$$

for H(z) in (4.9).

Solution: From (4.9), we can write

$$H(z) = \frac{1 + z^{-2}}{1 + \frac{z^{-1}}{2}}$$
 (5.2)

$$\begin{array}{r}
1+z^{-1}/2| \begin{array}{rrr} 2z^{-1} & -4 \\
1 & +z^{-2} \\
2z^{-1} & +z^{-2} \\
\hline
 & 1 & -2z^{-1} \\
-4 & -2z^{-1} \\
\hline
 & 5
\end{array}$$

So we can replace (4.9) as,

$$\frac{1+z^{-2}}{1+\frac{z^{-1}}{2}} = 2z^{-1} - 4 + \frac{5}{1+z^{-1}/2}$$
 (5.3)

Now we can expand the second term of above expression as an infinite geometric series,

$$\frac{5}{1+z^{-1}/2} = 5\left(1 + \left(\frac{-1}{2z}\right) + \left(\frac{-1}{2z}\right)^2 + \dots\right) (5.4)$$

where we assume $\left|\frac{1}{2z}\right|$ < 1. So (5.3) will become.

$$= 2z^{-1} - 4 + 5 + \frac{-5}{2}z^{-1} + \frac{5}{4}z^{-2} + \frac{-5}{8}z^{-3} + \frac{5}{16}z^{-4} + \dots$$

$$= 1z^{0} + \frac{-1}{2}z^{-1} + \frac{5}{4}z^{-2} + \frac{-5}{8}z^{-3} + \frac{5}{16}z^{-4} + \dots$$

Now to get h(n) for n < 5 we will compare (5.6) with the below equation,

$$H(z) = \sum_{n = -\infty}^{n = \infty} h(n)z^{-n}$$
 (5.7)

h(n) will be the coefficient of z^{-n} . Using this, from (5.6) we can write,

$$h(0) = 1 (5.8)$$

$$h(1) = \frac{-1}{2} \tag{5.9}$$

$$h(2) = \frac{5}{4} \tag{5.10}$$

$$h(3) = \frac{-5}{8} \tag{5.11}$$

$$h(4) = \frac{5}{16} \tag{5.12}$$

And for n < 0 h(n) = 0.

For n > 5, we can get h(n) from the geometric series.

$$h(n) = 5\left(\frac{-1}{2}\right)^n \tag{5.13}$$

5.2 Find an expression for h(n) using H(z), given that

$$h(n) \stackrel{\mathcal{Z}}{\rightleftharpoons} H(z) \tag{5.14}$$

and there is a one to one relationship between h(n) and H(z). h(n) is known as the *impulse response* of the system defined by (3.2).

Solution: The H(z) can be written as,

$$H(z) = \frac{1}{1 + \frac{z^{-1}}{2}} + \frac{z^{-2}}{1 + \frac{z^{-1}}{2}}$$
 (5.15)

From (4.16) we can write it as,

$$h(n) = \left(\frac{-1}{2}\right)^n u(n) + \left(\frac{-1}{2}\right)^{n-2} u(n-2) \quad (5.16)$$

5.3 Sketch h(n). Is it bounded? Justify Theoritically.

Solution: Download the code for the plot 5.3 from the below link,

wget https://github.com/NareshBandaru13/ EE3900-A1/blob/main/codes/5-3.py

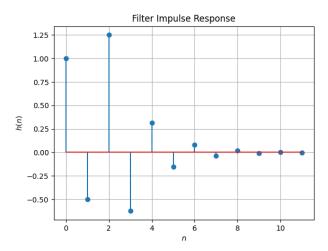


Fig. 5.3: h(n) as inverse of H(n)

From the plot it seems like h(n) is bounded and becomes smaller in magnitude as n increases. Using (5.16), we can get theoritical expression as,

$$h(n) = \begin{cases} 0 & , n < 0 \\ \left(\frac{-1}{2}\right)^n & , 0 \le n < 2 \\ 5\left(\frac{-1}{2}\right)^n & , n \ge 2 \end{cases}$$
 (5.17)

A sequence $\{x_n\}$ is said to be bounded if and only if there exist a positive real number M such that,

$$|x_n| \le M, \forall n \in \mathcal{N} \tag{5.18}$$

So to say h(n) is bounded we should able to find the M which satisfies (5.18). For n < 0,

$$|h(n)| \le 0 \tag{5.19}$$

For $0 \le n < 2$,

$$|h(n)| = \left|\frac{-1}{2}\right|^n \tag{5.20}$$

$$= \left(\frac{1}{2}\right)^n \le 1 \tag{5.21}$$

And for $n \geq 2$,

$$|h(n)| = \left| 5\left(\frac{-1}{2}\right) \right|^n$$
 (5.22)

$$= \left(\frac{5}{2}\right)^n \le \frac{5}{4} \tag{5.23}$$

From above three cases, we can get M as,

$$M = \max\left\{0, 1, \frac{5}{4}\right\} \tag{5.24}$$

$$=\frac{5}{4}$$
 (5.25)

Therefore, h(n) is bounded using (5.18) with $M = \frac{5}{4}$ i.e.,

$$|h(n)| \le \frac{5}{4} \forall n \in \mathcal{N} \tag{5.26}$$

5.4 Convergent? Justify using the ratio test.

Solution: We can say a given real sequence $\{x_n\}$ is convergent if

$$\lim_{n \to \infty} \left| \frac{x_{n+1}}{x_n} \right| < 1 \tag{5.27}$$

This is known as Ratio test. In this case the limit will become,

$$\lim_{n \to \infty} \left| \frac{h(n+1)}{h(n)} \right| = \lim_{n \to \infty} \left| \frac{5\left(\frac{-1}{2}\right)^{n+1}}{5\left(\frac{-1}{2}\right)^n} \right|$$
 (5.28)

$$= \lim_{n \to \infty} \left| \frac{-1}{2} \right|$$
 (5.29)
$$= \frac{1}{2}$$
 (5.30)

$$=\frac{1}{2}$$
 (5.30)

As $\frac{1}{2}$ < 1, from root test we can say that h(n)is convergent.

5.5 The system with h(n) is defined to be stable if

$$\sum_{n=-\infty}^{\infty} h(n) < \infty \tag{5.31}$$

Is the system defined by (3.2) stable for the impulse response in (5.14)?

Solution: From (5.16),

$$\sum_{n=-\infty}^{\infty} h(n) = \sum_{n=-\infty}^{\infty} \left(\left(\frac{-1}{2} \right)^n u(n) + \left(\frac{-1}{2} \right)^{n-2} u(n-2) \right)$$

(5.32)

$$=2\left(\frac{1}{1+\frac{1}{2}}\right) \tag{5.33}$$

$$=\frac{4}{3}$$
 (5.34)

: the system is stable.

5.6 Verify the above result using a python code.Solution: Download the python code from the below link

wget https://github.com/NareshBandaru13/ EE3900-A1/blob/main/codes/5-6.py

Then run the following command,

5.7 Compute and sketch h(n) using

$$h(n) + \frac{1}{2}h(n-1) = \delta(n) + \delta(n-2), \quad (5.35)$$

This is the definition of h(n).

Solution: Download the code for the plot 5.7 from the below link,

wget https://github.com/NareshBandaru13/ EE3900-A1/blob/main/codes/5-7.py

Note that this is same as 5.3. For n < 0, h(n) = 0 and,

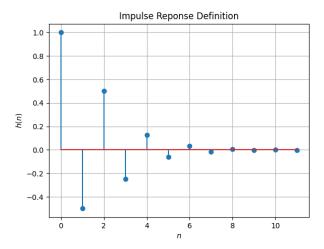


Fig. 5.7: From the definition of h(n)

$$h(0) = \delta(0) \tag{5.36}$$

$$= 1$$
 (5.37)

For n = 1.

$$h(1) + \frac{1}{2}h(0) = \delta(1) + \delta(-1)$$
 (5.38)

$$\implies h(1) = -\frac{1}{2}h(0) \tag{5.39}$$

$$= -\frac{1}{2} \tag{5.40}$$

n = 2,

$$h(2) + \frac{1}{2}h(1) = 0 + \delta(0)$$
 (5.41)

$$h(2) = 1 + \frac{1}{4} \tag{5.42}$$

$$=\frac{5}{4}\tag{5.43}$$

And for n > 2 RHS will be 0 so,

$$h(n) = -\frac{1}{2}h(n-1)$$
 (5.44)

Overall

$$h(n) = \begin{cases} 0 & , n < 0 \\ 1 & , n = 0 \\ -\frac{1}{2} & , n = 1 \\ \frac{5}{4} & , n = 2 \\ -\frac{1}{2}h(n-1) & , n > 2 \end{cases}$$
 (5.45)

5.8 Compute

$$y(n) = x(n) * h(n) = \sum_{k=-\infty}^{\infty} x(k)h(n-k)$$
 (5.46)

Comment. The operation in (5.46) is known as *convolution*.

Solution: Download the code for plot 5.8 from the below link

wget https://github.com/NareshBandaru13/ EE3900-A1/blob/main/codes/5-8.py

5.9 Express the above convolution using a Toeplitz matrix.

Solution: Download the python code from the below link for the plot 5.9,

wget https://github.com/NareshBandaru13/ EE3900-A1/blob/main/codes/5-9.py

Then run the following command,

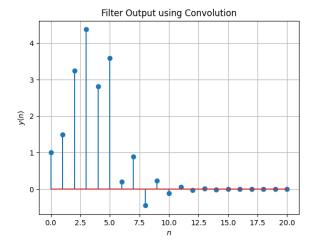


Fig. 5.8: y(n) using the convolution definition

python3 5-9.py

From (5.46), we express y(n) as

$$y(n) = \sum_{k=-\infty}^{\infty} x(k) h(n-k)$$
 (5.47)

To understand how we can use a Toeplitz matrix, we will see what we are doing in (5.46)

$$y(0) = x(0)h(0) (5.48)$$

$$y(1) = x(0)h(1) + x(1)h(0)$$
 (5.49)

$$y(2) = x(0)h(2) + x(1)h(1) + x(2)h(0)$$
(5.50)

.

The same thing can be written as,

$$y(0) = \begin{pmatrix} h(0) & 0 & 0 & \dots & 0 \end{pmatrix} \begin{pmatrix} x(0) \\ x(1) \\ x(2) \\ \vdots \\ x(5) \end{pmatrix}$$
 (5.51)

$$y(1) = \begin{pmatrix} h(1) & h(0) & 0 & 0 & . & . & .0 \end{pmatrix} \begin{pmatrix} x(0) \\ x(1) \\ x(2) \\ . \\ . \\ x(5) \end{pmatrix}$$
(5.52)

$$y(2) = \begin{pmatrix} h(2) & h(1) & h(0) & 0 & . & .0 \end{pmatrix} \begin{pmatrix} x(0) \\ x(1) \\ x(2) \\ . \\ x(5) \end{pmatrix}$$
(5.53)

.

Using Toeplitz matrix of h(n) we can simplify it as,

$$y(n) = \begin{pmatrix} h(0) & 0 & 0 & \dots & 0 \\ h(1) & h(0) & 0 & \dots & 0 \\ h(2) & h(1) & h(0) & \dots & 0 \\ & & & & \\ 0 & 0 & 0 & \dots & h(m-1) \end{pmatrix} \begin{pmatrix} x(0) \\ x(1) \\ x(2) \\ \vdots \\ x(5) \end{pmatrix}$$
(5.54)

Now from (??) we will take n

$$x(n) = \begin{pmatrix} 1\\2\\3\\4\\2\\1 \end{pmatrix}$$
 (5.55)

And from (5.17)

$$h(n) = \begin{pmatrix} 1 \\ -0.5 \\ 1.25 \\ \cdot \\ \cdot \\ \cdot \end{pmatrix}$$
 (5.56)

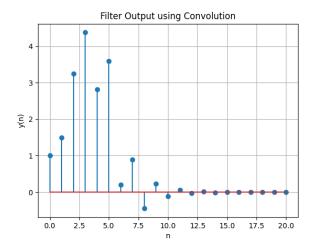


Fig. 5.9: Convolution of x(n) and h(n) using toeplitz matrix

Now using (5.54),

$$= \begin{pmatrix} 1\\1.5\\3.25\\ \cdot\\ \cdot\\ \cdot \end{pmatrix}$$
 (5.59)

5.10 Show that

$$y(n) = \sum_{k=-\infty}^{\infty} x(n-k)h(k)$$
 (5.60)

Solution: Substitute k := n - k in (5.46), we

will get

$$y(n) = \sum_{k=-\infty}^{\infty} x(k)h(n-k)$$
 (5.61)

$$=\sum_{n-k=-\infty}^{\infty}x(n-k)h(k)$$
 (5.62)

$$=\sum_{k=-\infty}^{\infty}x(n-k)h(k)$$
 (5.63)

(5.64)

6 DFT

6.1 Compute

$$X(k) \stackrel{\triangle}{=} \sum_{n=0}^{N-1} x(n)e^{-j2\pi kn/N}, \quad k = 0, 1, \dots, N-1$$
(6.1)

and H(k) using h(n).

Solution: Download the below python code for the plot ??,

wget https://github.com/Charanyash/EE3900— Digital_Signal_Processing/blob/master/ Sound%201/Codes/dft.py

And run the following command in the terminal.

python3 dft.py

6.2 Compute

$$Y(k) = X(k)H(k) \tag{6.2}$$

Solution: Download the below python code for the plot ??,

wget https://github.com/Charanyash/EE3900—Digital_Signal_Processing/blob/master/Sound%201/Codes/Y_K.py

Then run the following command in the terminal,

python3 Y K.py

6.3 Compute

$$y(n) = \frac{1}{N} \sum_{k=0}^{N-1} Y(k) \cdot e^{j2\pi kn/N}, \quad n = 0, 1, \dots, N-1$$
(6.3)

Solution: Download the below python code for the plot ??,

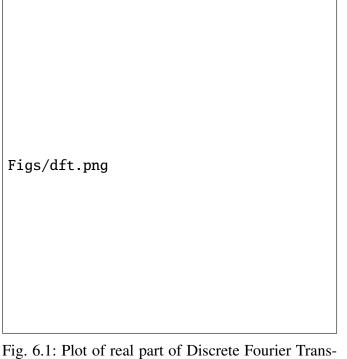


Fig. 6.1: Plot of real part of Discrete Fourier Transforms of x(n) and h(n)

Figs/Y_k.png

Fig. 6.2: Y(k) as the product of X(k) and H(k)

wget https://github.com/Charanyash/EE3900—Digital_Signal_Processing/blob/master/Sound%201/Codes/yndft_dif.py

Then run the following command,

python3 yndft dif.py

Figs/yndft_dif.png

Fig. 6.3: y(n) using IDFT and difference equation

6.4 Repeat the previous exercise by computing X(k), H(k) and y(n) through FFT and IFFT.Solution: Download the below python code for the plot ??,

wget https://github.com/Charanyash/EE3900—Digital_Signal_Processing/blob/master/Sound%201/Codes/yn_ifft.py

Then run the following command,

python3 yn ifft.py

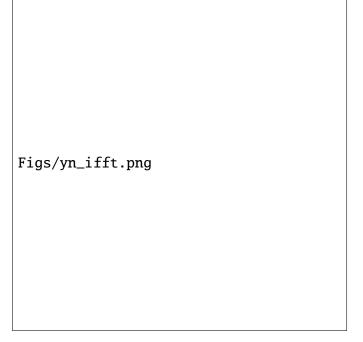


Fig. 6.4: The plot of y(n) using IFFT