

**CT216 : End of Semester Project**

**Topic : Polar Codes**

**Analysis of Polar Codes: Channel Polarization  
and Code Rate**



**Group 3**

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# 1 Introduction

Polar Codes were introduced by Erdal Arikan in 2009. Polar Codes are a class of block codes that were the first ones proven to achieve the capacity of binary-input symmetric memoryless channels (BI-DMC) using a low-complexity encoding and decoding algorithm. Polar Codes utilize a method of channel polarization that enables them to achieve the capacity of a wide variety of binary-input symmetric memoryless channels. This channel polarization divides the channels into a set of reliable and unreliable channels. In this report/paper, we will analyze how the presence of a fraction of channels that remain unpolarized (reliable channels) affects the rate of convergence of the code rate to the channel capacity.

## 2 Background and Basic Concepts

### 2.1 Shannon's Channel Capacity Theorem

Shannon's Channel Capacity Theorem states that for any Discrete Memoryless channel  $W$ , there exists an associated non-negative constant  $I(W)$ , known as channel capacity such that

- For any  $\epsilon > 0$ , one can communicate over a channel  $W$  at asymptotic rate  $R = I(W) - \epsilon$  with vanishing probability of miscommunication.
- For any rate  $R > I(W)$ , it is impossible to communicate at rate  $R$  without the non-negligible probability of miscommunication *i.e.*, the probability of error at decoding approaches 1 as the number of channel uses increases.
- For rate  $R = I(W) - \epsilon$  where  $\epsilon$  is referred to as gap to capacity, we need to minimise it as much as we can for required error - correcting codes.

### 2.2 Performance of Linear Block Codes

Shannon's Theorem shows that random linear block codes perform well such that in order to get rate  $R = I(W) - \epsilon$ , we can take block length  $N = \frac{1}{\epsilon^2}$  with probability of decoding error  $e^{-\epsilon^2 N}$ .

But for random-codes there are disadvantages like :

- Non-Constructive Algorithms
- Exponential Time Decoding

So there is a need of explicit error-correcting codes with efficient encoding and decoding algorithms to communicate even at rates approaching capacity.

### 3 Channel Capacity and Binary Polar Codes

For a binary symmetric channel (BSC) with crossover probability  $p$ , the channel capacity  $C$  is given by the formula:

$$C = 1 - H(p)$$

where  $H(p)$  is the binary entropy function defined as:

$$H(p) = -p \log_2(p) - (1 - p) \log_2(1 - p)$$

Polar codes employ a transformation that polarizes the synthesized bit channels over iterations, causing them to become either very good (reliable) or very bad (unreliable) at transmitting information.

Also, Polar codes have property of achieving capacity at  $N \rightarrow \infty$ , with :

- poly(N) construction of message
- $O(N \log N)$  Encoding and Decoding Time Complexity

### 4 Enhanced Polar Transformation and Channel Behavior

#### 4.1 Transformation and Channel Definition

Given the polar transformation  $U \cdot G_N = X$ , where  $U$  are the original bits and  $X$  are the transformed bits. The bits  $U_i$  are transformed into  $X_i$  such that  $U_i$  depends on  $U_1, \dots, U_{i-1}$  in specific ways due to the recursive structure of  $G_N$ .

#### 4.2 Channel Passing and Receiving

After modulation (e.g., BPSK) and transmission through an AWGN channel, we receive a vector  $Y^N$ . Each bit  $U_i$  influences the formation of the entire received vector  $Y^N$ , but each virtual channel  $W_i$  associated with  $U_i$  is defined as:

$$W_i : U_i \rightarrow (Y^N, U^{i-1})$$

where  $U^{i-1}$  represents the sequence  $U_1, \dots, U_{i-1}$ .

### 5 Entropy Analysis

#### 5.1 Entropy of $U_i$ Given Past $U$ Values and $Y^N$

$$H(U_i | U_{i-1}, Y^N)$$

This is not generally equal for each  $U_i$  due to the different ways each bit is polarized. Polarization aims to make some channels (bits) very reliable (low entropy) and others very unreliable (high entropy).

#### 5.2 Polarization Effect

As a result of the polarization transform, for most indices  $i$ , as  $N$  (the total number of channels) becomes large, the entropy  $H(U_i | U_{i-1}, Y^N)$  tends towards 0 or 1. This is due to the nature of the transformation, which increasingly separates the transformed bits into those that are nearly deterministic and those that remain highly random.

## 6 Relating Entropy to Virtual Channels $W_i$

The entropy of each virtual channel  $W_i$ , which carries  $U_i$  over the noise and the past bits, is then defined as:

$$H(W_i) = H(U_i|U^{i-1}, Y^N)$$

This definition aligns with the behavior of  $H(U_i|U_{i-1}, Y^N)$ , which suggests that the polarization effect causes  $H(W_i)$  to polarize similarly:

- For most  $i$ ,  $H(W_i)$  approaches 0 or 1.
- But in practical condition this might not be the case always, so let's consider that  $\delta$  is the fraction of unpolarized channels *i.e.* entropy close to not either 0 or 1.

Let  $W$  be a binary-input symmetric channel. Then the polar transformation generates two channels:

- $W^+$  - This is typically the more reliable channel.
- $W^-$  - This is typically the less reliable channel.

The transformation seeks to satisfy the following inequality in terms of entropy:

$$H(W^+) < H(W) < H(W^-)$$

This inequality highlights the effectiveness of polarization where  $W^+$  becomes more reliable, and  $W^-$  less reliable because higher the entropy higher the randomness and higher the error in polarizing, thus allowing for the differentiation of channel capacities and their better utilization in coding.

## 7 Main Analysis

**Claim 1** *If a delta fraction  $\delta$  of channels remains unpolarized, the resulting polar code has a rate  $R$  that approaches the channel capacity  $C$  as the block length  $N$  tends to infinity and  $\delta$  tends to zero.*

By the theory of channel polarization, as  $N$  (the block length of the code) increases, the fraction of channels that polarize towards having a capacity close to either 0 or 1 increases. The rate  $R$  of the Polar Code can be defined as  $R = \frac{k}{N}$ , where  $k$  is the number of reliable channels used for transmitting information.

For large  $N$ , we have  $k \approx N(1 - H(p))$ , because in unpolarized channels, net information is non-uniformly divided between  $N$  different channels but after polarizing and leading  $N \rightarrow \infty$ , the net information is thus divided into only  $K$  good channels, where  $K$  is the total length of Information bits and other bad channels have their contribution in net Information  $\rightarrow 0$ , and thus  $R \approx 1 - H(p)$ , which aligns with the capacity  $C$ .

The presence of  $\delta$  fraction of unpolarized channels, having capacities not close to 0 or 1, implies that:

$$R = \frac{N(1 - \delta)(1 - H(p))}{N} = (1 - \delta)(1 - H(p))$$

because  $\delta$  proportion of channel won't participate in transmission of information and other good channels would only lead to contribute in transmission of information.

As  $N \rightarrow \infty$  and  $\delta \rightarrow 0$ ,  $R$  approaches  $1 - H(p)$  matching the capacity  $C$ . The fraction  $1 - \delta$  of channels that are highly reliable converges to 1, allowing the code rate to approach the channel capacity.

## 8 Conclusion

The analysis confirms that Polar Codes are capable of achieving channel capacity under the assumption of large block lengths and negligible fractions of unpolarized channels. This property is fundamental for the utilization of Polar Codes in practical communication systems where maximizing data transmission rate is crucial. Also, Polar Codes exemplify Matthew's effect. Matthew's effect in simple terms means the rich getting richer and the poor getting poorer. In the case of polar codes Matthew's effect refers to a phenomenon, where certain channels become highly reliable while others become less reliable due to polarization.

## References

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