1.

Aim:

To generate common discrete-time signals.

Apparatus required:

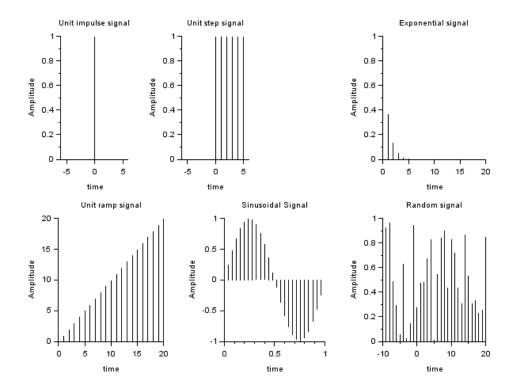
Personal computer, SCILAB software.

```
PROGRAM:
UNIT IMPULSE SIGNAL
clc;
clear all;
close;
N=5; //SET LIMIT
t1=-5:5;
x1=[zeros(1,N),ones(1,1),zeros(1,N)];
subplot(2,4,1);
plot2d3(t1,x1)
xlabel('time');
ylabel('Amplitude');
title('Unit impulse signal');
UNIT STEP SIGNAL
t2=-5:5;
x2=[zeros(1,N),ones(1,N+1)];
 subplot(2,4,2);
 plot2d3(t2,x2)
 xlabel('time');
 ylabel('Amplitude');
title('Unit step signal');
```

```
EXPONENTIAL SIGNAL
t3 =0:1:20;
x3 = exp(-t3);
     subplot(2,3,3);
     plot2d3(t3,x3);
     xlabel('time');
     ylabel('Amplitude');
     title('Exponential signal');
UNIT RAMP SIGNAL
     t4=0:20;
     x4=t4;
     subplot(2,3,4);
     plot2d3(t4,x4);
     xlabel('time');
     ylabel('Amplitude');
     title('Unit rampsignal');
SINUSOIDAL SIGNAL
     t5=0:0.04:1;
     x5=sin(2*%pi*t5);
     subplot(2,3,5);
     plot2d3(t5,x5);
     title('Sinusoidal
                           Signal')
     xlabel('time');
     ylabel('Amplitude');
RANDOM SIGNAL
     t6 = -10:1:20;
     x6=rand(1,31);
```

```
subplot(2,3,6);
plot2d3(t6,x6);
xlabel('time');
ylabel('Amplitude');
title('Random signal');
```

Output:



2. //Program to Compute the 8-point DFT of the Sequence x[n]=[1,1,1,1,1,1,0,0]

clear; clc; close; x = [1,1,1,1,1,1,0,0];//DFT Computation X = fft(x,-1);//Display sequence X[k] in command window disp(X,"X[k]=");

```
RESULT
```

```
X[k]=
     column 1 to 5
  6. -0.7071068 - 1.7071068i 1. -i 0.7071068 + 0.2928932i 0
     column 6 to 8
  0.7071068 - 0.2928932i   1. + i   -0.7071068 + 1.7071068i
   3. //Program to Compute Circular Convolution y[n] where Y[k]=X1[k].X2[k]
      //x1[n]=[0,1,2,3,4]
       //x2[n]=[0,1,0,0,0]
       clear;
       clc;
       close;
       x1=[0,1,2,3,4];
       x2=[0,1,0,0,0];
       //DFT Computation
       X1 = fft(x1,-1);
       X2 = fft(x2,-1);
       Y=X1.*X2;
       //IDFT Computation
       y=round(fft(Y,1));
       //Display sequence y[n] in command window
       disp(y,"y[n]=");
       RESULT:
                   4. 0. 1. 2. 3.
         y[n]=
   4. //Program to Compute the DFT of given Sequence
       //x[n]=[1,-1,-1,1,1,1,-1] using DIT-FFT Algorithm.
       clear;
       clc;
       close;
       x = [1,-1,-1,-1,1,1,-1];
       //FFT Computation
       X = fft (x, -1);
       disp(X, 'X(z) = ');
       RESULT:
       X(z) =
            column 1 to 5
         0 - 1.4142136 + 3.4142136i 2. - 2.i 1.4142136 - 0.5857864i 4.
```

```
column 6 to 8
1.4142136 + 0.5857864i 2. + 2.i - 1.4142136 - 3.4142136i
```

```
5. //Program to find the DFT of a Sequence x[n]=[1,2,3,4,4,3,2,1]
   //using DIF Algorithm.
   clear;
   clc;
   close;
   x = [1,2,3,4,4,3,2,1];
   //FFT Computation
   X = fft (x, -1);
   disp(X, 'X(z) = ');
   RESULT:
   X(z) =
        column 1 to 5
     20. - 5.8284271 - 2.4142136i 0 - 0.1715729 - 0.4142136i 0
        column 6 to 8
    -0.1715729 + 0.4142136i 0 -5.8284271 + 2.4142136i
                        Unit II- INFINITE IMPULSE RESPONSE FILTERS
6. (i) //To Find out Bilinear Transformation of H(s)=2/((s+1)*(s+2))
   clear;
   clc;
   close;
   s=\%s;
   z=\%z;
   HS=2/((s+1)*(s+2));
   T=1;
   HZ=horner(HS,(2/T)*(z-1)/(z+1));
   disp(HZ, 'H(z) = ');
   RESULT:
   H(z) =
            2
     2 + 4z + 2z
      -----
           2
     -4z + 12z
       (ii) //To Find out Bilinear Transformation of H(s)=(s^2+4.525)/(s^2+0.692*s+0.504)
   clear;
```

```
clc;
close;
s=%s;
z=\%z;
HS=(s^2+4.525)/(s^2+0.692*s+0.504);
HZ=horner(HS,(2/T)*(z-1)/(z+1));
disp(HZ, 'H(z) = ');
RESULT:
H(z) =
  1.4478601 + 0.1783288z + 1.4478601z
      0.5298913 - 1.1875z + z
7. (i) //To Design the Filter using Impulse Invarient Method
clear;
clc;
close;
s=\%s;
T=0.2;
HS=10/(s^2+7*s+10);
elts=pfss(HS);
disp(elts,'Factorized HS = ');
//The poles comes out to be at -5 and -2
p1=-5;
p2=-2;
z=\%z;
HZ=T*((-3.33/(1-\%e^{(p_1*T)*z^{(-1)})}+(3.33/(1-\%e^{(p_2*T)*z^{(-1)})})
disp(HZ, 'HZ = ');
RESULT:
Factorized HS =
    (1)
  3.3333333
  -----
   2 + s
    (2)
 - 3.3333333
  -----
   5 + s
HZ =
       0.2014254z
```

```
0.2465970 - 1.0381995z + Z2
       HZ =
             0.2014254z
        0.2465970 - 1.0381995z + z2
(ii) //To Design the Filter using Impulse Invarient Method
      clear;
      clc;
      close;
      s=%s;
      T=1;
      HS=(2)/(s^2+3*s+2);
      elts=pfss(HS);
      disp(elts, 'Factorized HS = ');
      //The poles comes out to be at -2 and -1
      p1 = -2;
      p2=-1;
      z=\%z;
      HZ=(2/(1-\%e^{(p_2*T)*z^{(-1)})}-(2/(1-\%e^{(p_1*T)*z^{(-1)})})
      disp(HZ, 'HZ = ');
      RESULT:
      Factorized HS =
           (1)
          2
        1 + s
          (2)
        - 2
        2 + s
       HZ =
             0.4650883z
        0.0497871 - 0.5032147z + z2
       HZ =
             0.4650883z
        -----
```

0.0497871 - 0.5032147z + z2

```
8.
    Design a H.P.F. with given specifications
   clear;
   clc;
   close;
   ap=3;//db
   as=15;//db
   op=500;//rad/sec
   os=1000;//rad/sec
   N = log(sqrt((10^{(0.1*as)-1)}/(10^{(0.1*ap)-1)}))/log(os/op);
   disp(ceil(N),'Order of the filter, N =');
   s=\%s;
   HS=1/((s+1)*(s^2+s+1));//Transfer Function for N=3
   oc=1000//rad/sec
    Order of the filter, N =
        3.
    oc =
        1000.
    Normalized Transfer Function, H(s) =
                             s
        1.000D+09 + 2000000s + 2000s + s
9.
   (a) Design of a lowpass type 1 Chebyshev filter with 2dB ripple in the passband and -30 dB
   attenuation at the stopband edge. The sampling frequency is assumed to be 3000Hz and the cutoff
   frequencies at 37.5Hz and 75Hz respectively.
   >sf=3000;
   -->f1=37.5;
   -->f2=75;
    -->as=30;
   -->ap=2;
   -->om=[f1*(2*%pi)/sf,f2*(2*%pi)/sf];
    -->deltas=10.00**(-0.05*as);
   -->deltap=(1.0-10.0**(-0.05*ap));
   -->[cells,fact,zers,pols]=...
   -->eqiir('lp','ch1',om,deltap,deltas);
```

```
-->cells
        cells =
          ! -----
        1
        ! Nan +Nanz + z Nan +Nanz + z !
(b) Convert Analog LPF into [1]. High Pass [2]. Band Pass IIR Butterworth
Filter //Using Analog Filter Transformations //For the given cutoff
frequency Wc = 500 Hz
clear all;
omegap = 500;
omegas = 1000;
delta1 in dB = -3;
delta2 in dB = -40;
delta1 = 10^{(delta1 in dB/20)}
delta2 = 10^{(delta2 in dB/20)}
//Calculation of Filter Order
```

 $N = log10((1/(delta2^2))-1)/(2*log10(omegas/omegap))$

omega LPF = omegap; //Analog LPF Cutoff frequency omega HPF = omega LPF; //Analog HPF Cutoff frequency

[hs,pols,zers,gain] = analpf(N,'butt',[0,0],omega LPF)

omega2 = 600; //Upper Cutoff frequency omega1 = 300; //Lower Cutoff Frequency

BW = omega2 - omega1; //Bandwidth disp('Analog LPF Transfer function')

disp('Analog HPF Transfer function')

disp('Analog BPF Transfer function')

h HPF = horner(hs LPF,omega LPF*omega HPF/s)

h BPF = horner(hs LPF,omega LPF*(num/den)) //Plotting Low Pass Filter Frequency Response

clc; close;

N = ceil(N)

hs LPF = hs;

den = BW*s

s =poly(0,'s');

 $num = (s^2) + omega0$

N = 1;//

omegac = omegap;

//Poles and Gain Calculation [pols,gain]=zpbutt(N,omegac);

omega0 = (omega2*omega1);

hs LPF(2) = hs LPF(2)/500; hs LPF(3) = hs LPF(3)/500;

```
figure
fr=0:.1:1000;
hf=freq(hs LPF(2),hs LPF(3),%i*fr);
hm=abs(hf);
plot(fr,hm)
a=gca();
a.thickness = 3;
a.foreground = 1;
a.font style = 9;
xgrid(1)
xtitle('Magnitude Response of LPF Filter Cutoff frequency = 500Hz', 'Analog
Frequency--->','Magnitude');
//Plotting High Pass Filter Frequency Response
figure
fr=0:.1:1000;
hf HPF=freq(h HPF(2),h HPF(3),%i*fr);
hm HPF=abs(hf HPF);
plot(fr,hm_HPF)
a=gca();
a.thickness = 3;
a.foreground = 1;
a.font style = 9;
xgrid(1)
xtitle('Magnitude Response of HPF Filter Cutoff frequency = 500Hz', 'Analog
Frequency--->','Magnitude');
//Plotting Band Pass Filter Frequency Response
figure
fr=0:.1:1000;
hf_BPF=freq(h_BPF(2),h_BPF(3),%i*fr);
hm BPF=abs(hf BPF);
plot(fr,hm BPF)
a=gca();
a.thickness = 3;
a.foreground = 1;
a.font style = 9;
xgrid(1)
        xtitle('Magnitude Response of BPF Filter Upper Cutoff frequency =
        600Hz & Lower Cutoff frequency = 300Hz', 'Analog Frequency---
        >','Magnitude');
      10.
         Design a Chebyshev Filter with Given Specifications
        clear;
        clc:
        close;
        ap=2.5;//db
        as=30;//db
        op=20;//rad/sec
```

```
\label{eq:cosh} \begin{split} os=&50;//rad/sec\\ N=&acosh(sqrt((10^{(0.1*as)-1)/(10^{(0.1*ap)-1))})/acosh(os/op);\\ disp(ceil(N),'Order of the filter, N=');\\ Order of the filter, N=&\\ 3. \end{split}
```

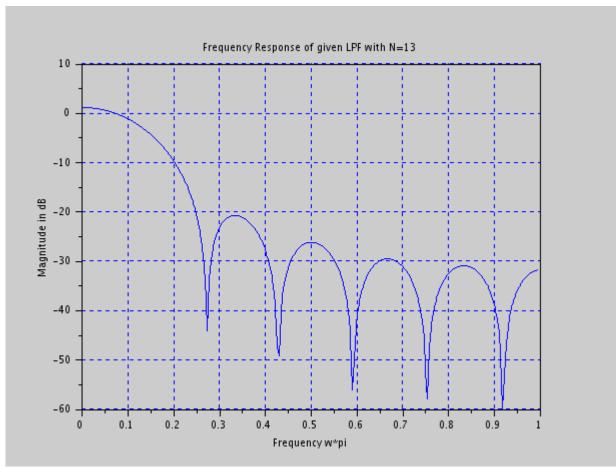
Unit III-FIR filter design

```
11.
   Plot Magnitude Responce of given L.P.F. with specifications:
   //N = 13, w = pi/6
   clear;
   clc;
   close;
   alpha=6;
   U=1;
   for n=0+U:1:12+U
   if n==7
   hd(n)=0.167;
   else
   hd(n)=(\sin(\%pi*(n-U-alpha)/6))/(\%pi*(n-U-alpha));
   end
   end
   [hzm, fr] = frmag (hd, 256);
   hzm_dB = 20* log10 (hzm)./ max (hzm);
   figure
   plot (2*fr, hzm_dB)
   a= gca ();
   xlabel ('Frequency w*pi');
   ylabel ('Magnitude in dB');
```

title ('Frequency Response of given LPF with N=13');

disp(hd,"Filter Coefficients,h(n)=");

xgrid (2)



Filter Coefficients,h(n)=

6.497D-18

0.0318310

0.0689161

0.1061033

0.1378322

0.1591549

0.167

0.1591549

0.1378322

0.1061033

0.0689161

0.0318310

6.497D-18

12. //Program to Plot Magnitude Responce of given L.P.F. Using Rectangular Window with specifications:

```
//N=7, fc=1000Hz, F=5000Hz
```

clear;

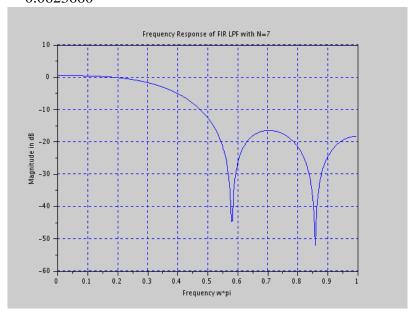
clc;

close;

N=7;

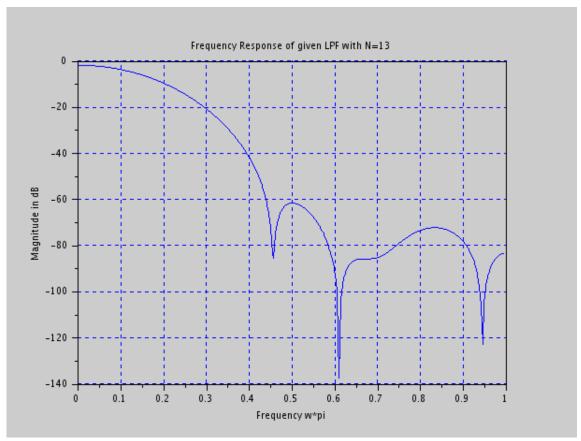
U=4;

```
h_Rect=window('re',N);
for n=-3+U:1:3+U
if n==4
hd(n)=0.4;
else
hd(n)=(\sin(2*\%pi*(n-U)/5))/(\%pi*(n-U));
h(n)=hd(n)*h_Rect(n);
end
[hzm, fr] = frmag (h, 256);
hzm_dB = 20* log10 (hzm)./ max (hzm);
figure
plot (2*fr, hzm_dB)
a= gca ();
xlabel ('Frequency w*pi');
ylabel ('Magnitude in dB');
title ('Frequency Response of FIR LPF with N=7');
xgrid (2)
disp(h,"Filter Coefficients,h(n)=");
RESULT:
Filter Coefficients,h(n)=
 - 0.0623660
  0.0935489
  0.3027307
  0.4
  0.3027307
  0.0935489
 - 0.0623660
```



13. Plot Magnitude Responce of given L.P.F. with specifications:

```
//N=13,w=pi/6
//Using Hamming Window
clear;
clc;
close;
N=13;
alpha=6;
U=1;
h_hamm=window('hm',N);
for n=0+U:1:12+U
if n==7
hd(n)=0.167;
else
hd(n)=(\sin(\%pi*(n-U-alpha)/6))/(\%pi*(n-U-alpha));
end
h(n)=hd(n)*h_hamm(n);
end
[hzm, fr] = frmag (h, 256);
hzm_dB = 20* log10 (hzm)./ max (hzm);
figure
plot (2*fr, hzm_dB)
a= gca ();
xlabel ('Frequency w*pi');
ylabel ('Magnitude in dB');
title ('Frequency Response of given LPF with N=13');
xgrid (2)
disp(h,"Filter Coefficients,h(n)=");
disp(h,"Filter Coefficients,h(n)=");
```



```
Filter Coefficients,h(n)=
```

5.198D-19

0.0045082

0.0213640

0.0572958

0.1061308

0.1493465

0.167

0.1493465

0.1061308

0.0572958

0.0213640

0.0045082

5.198D-19

14.

```
//Program to Plot Magnitude Responce of ideal H.P.F.
```

//using Hanning Window

//wc1=0.25*pi

//N=11

clear;

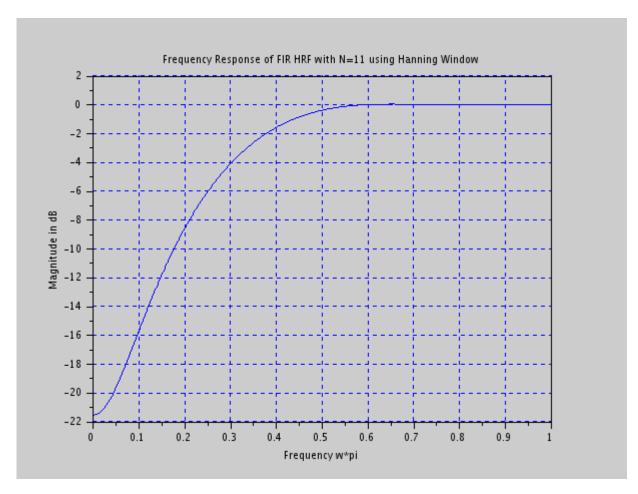
clc;

close;

N=11;

U=6;

```
h_hann=window('hn',N);
for n=-5+U:1:5+U
if n==6
hd(n)=0.75;
else
hd(n)=(sin(\%pi*(n-U))-sin(\%pi*(n-U)/4))/(\%pi*(n-U));
h(n)=h_hann(n)*hd(n);
end
[hzm, fr] = frmag (h, 256);
hzm_dB = 20* log10 (hzm)./ max (hzm);
figure
plot (2*fr, hzm_dB)
a= gca ();
xlabel ('Frequency w*pi');
ylabel ('Magnitude in dB');
title ('Frequency Response of FIR HRF with N=11 using Hanning Window');
xgrid (2);
```

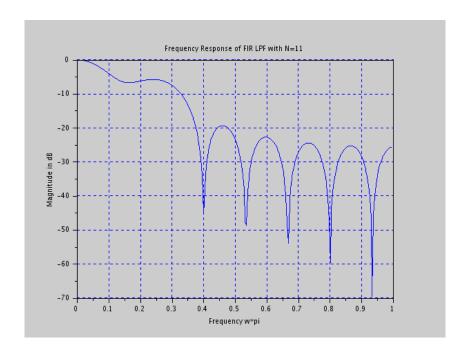


15. //Program to design Frequency Sampling Method FIR L.P.F. filter with following specifications:

//N=15, wc=pi/4

```
clear;
clc;
close;
N=15;
U=1;
for n=0+U:1:N-1+U
h(n)=(1+\cos(2*\% pi*(7-n)/N)+\cos(4*\% pi*(7-n)/N))/N;
end
[hzm, fr] = frmag(h, 256);
hzm_dB = 20* log10 (hzm)./ max (hzm);
figure;
plot (2*fr, hzm_dB);
a= gca ();
xlabel ('Frequency w*pi');
ylabel ('Magnitude in dB');
title ('Frequency Response of FIR LPF with N=11');
xgrid (2)
```

RESULT:



Unit IV- FINITE WORDLENGTH EFFECTS AND MULTIRATE SIGNAL PROCESSING

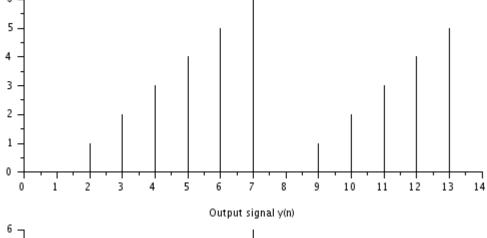
```
16. //To Compare the Variance of Output due to A/D Conversion process //y(n)=0.8y(n-1)+x(n) clear; clc ;
```

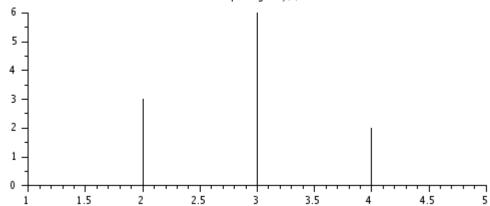
```
close;
n=8; //Bits
r=100; //Range
Q=2*r/(2^n); //Quantization Step Size
Ve=(Q^2)/12;
Vo=Ve*(1/(1-0.8^2));
disp(Q,'QUANTIZATION STEP SIZE =');
disp(Ve,'VARIANCE OF ERROR SIGNAL =');
disp(Vo,'VARIANCE OF OUTPUT =');
RESULT:
QUANTIZATION STEP SIZE =
  0.78125
VARIANCE OF ERROR SIGNAL =
  0.0508626
VARIANCE OF OUTPUT =
  0.1412851
17. Output noise power
   / Output noise power
   clc;clear;close;
   z=poly(0, 'z');
   H=0.5*z/(z-0.5);
   B=8;
   pn=2^{(-2*B)/12};
                        //Noise power
   X=H*horner(H,1/z)/z;
   r=roots(denom(X));
   rl=length(r);
   rc=coeff(denom(X))
   q1=[];q2=[];
   for n=1:rl
                     //Loop to separate poles inside the unit circle
     if (abs(r(n))<1) then
        q1=[q1 r(n)];
     else
        q2=[q2 r(n)];
     end
   end
   P=numer(X)/rc(length(rc));
   Q1=poly(q1,'z');
   Q2=poly(q2,'z');
   I=residu(P,Q1,Q2);
                         //Residue Calculation
                      //Output Noise power
   po=pn*I;
   disp(pn,'Input Noise power');
   disp(po,'Output Noise power');
```

```
- 0.5 1.25 - 0.5
   Input Noise power
     0.0000013
   Output Noise power
     0.0000004
18. (i) Deadband interval
    clc; clear;
   //y(n)=0.9y(n-1)+x(n)
   //Input x(n)=0
   n=-1;y=12;
                  //Initial Condition y(-1)=12
   flag=1;
   while n<8
     n=n+1;
     y=[y 0.9*y(n+1)];
     yr=round(y);
   end
   disp(n,'n=');
   disp(y, 'y(n)-exact');
   disp(yr,'y(n)-rounded');
   disp([-yr(n+2) yr(n+2)], 'Deadband interval')
   n=
     8.
    y(n)-exact
        column 1 to 8
     12. 10.8 9.72 8.748 7.8732 7.08588 6.377292 5.7395628
        column 9 to 10
     5.1656065 4.6490459
    y(n)-rounded
     12. 11. 10. 9. 8. 7. 6. 6. 5. 5.
    Deadband interval
    - 5. 5.
                            (or)
   (ii) Deadband interval
   clc; clear;
   //y(n)=0.9y(n-1)+x(n)
   a=0.9;
   l=ceil(0.5/(1-abs(a)));
   disp([-l 1],'Deadband interval ')
   Deadband interval
    - 6. 6.
```

rc =

```
19. Decimation
    clc;clear;close;
    x=[0:6 0:6];
    y=x(1:3:length(x));
    disp(x,'Input signal x(n)=');
    disp(y,'Output signal of decimation process by factor three y(n)');
    subplot(2,1,1);
    plot2d3(x);title('Input signal x(n)');
    subplot(2,1,2);
    plot2d3(y);title('Output signal y(n)');
```





Input signal x(n)=

column 1 to 13

0. 1. 2. 3. 4. 5. 6. 0. 1. 2. 3. 4. 5 column 14 6.

Output signal of decimation process by factor three y(n)

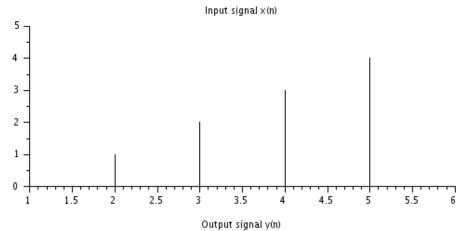
0. 3. 6. 2. 5.

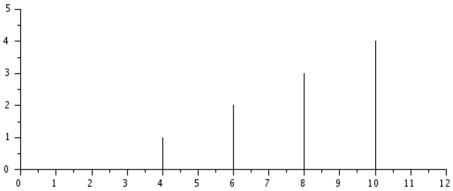
20. Interpolation

clc;clear;close;

x=0:5;

```
y=[];\\ for i=1:length(x)\\ y(1,2*i)=x(i);\\ end\\ disp(x, 'Input signal <math>x(n)=');\\ disp(y, 'Output signal of interpolation process with factor two y(n)');\\ subplot(2,1,1);\\ plot2d3(x);title('Input signal x(n)');\\ subplot(2,1,2);\\ plot2d3(y);title('Output signal y(n)');
```





Input signal x(n)=

0. 1. 2. 3. 4. 5.

Output signal of interpolation process with factor two y(n)

0. 0. 0. 1. 0. 2. 0. 3. 0. 4. 0. 5.

```
21. (i) //To Design an Analog Butterworth Filter
clear;
clc:
close;
op=0.2*\%pi;
os=0.4*%pi;
e1=0.9;
11=0.2;
epsilon=sqrt(1/(e1^2)-1);
lambda = sqrt(1/(11^2)-1);
N=log(lambda/epsilon)/log(os/op);
disp(ceil(N), Order of the filter, N = ');
s=\%s;
HS=1/((s^2+0.76537*s+1)*(s^2+1.8477*s+1));//Transfer Function for N=4
oc=op/epsilon^(1/ceil(N));
HS1=horner(HS,s/oc);
disp(HS1, 'Normalized Transfer Function, H(s) = ');
RESULT:
Order of the filter, N =
  4.
Normalized Transfer Function, H(s) =
  1 + 3.469403s + 6.0185667s2 + 6.1159251s3 + 3.1075263s4
9. (ii) //To Design an Analog Butterworth Filter
clear;
clc;
close;
ap=2;//db
as=10;//db
op=20;//rad/sec
os=30;//rad/sec
N = log(sqrt((10^{(0.1*as)-1)}/(10^{(0.1*ap)-1)}))/log(os/op);
disp(ceil(N), 'Order of the filter, N = ');
s=\%s;
HS=1/((s^2+0.76537*s+1)*(s^2+1.8477*s+1));//Transfer Function for N=4
oc=op/(10^{(0.1*ap)-1})^{(1/(2*ceil(N)))};
HS1=horner(HS,s/oc);
disp(HS1, 'Normalized Transfer Function, H(s) = ');
RESULT:
Order of the filter, N =
```

```
4. Normalized Transfer Function, H(s) =

1
------
1 + 0.1221815s + 0.0074644s2 + 0.0002671s2 + 0.0000048s4
```

Unit V- DIGITAL SIGNAL PROCESSORS

Extra lab experiments

1. SOUND PLAY COMMAND IN SCILAB

```
Reading a Speech Signal &
//Play the signal
clear;
clc;
[y,Fs,bits_y]=wavread("E:\4.wav");
sound(y,Fs,16)
```

2. This function is used to design an equalizer which will be useful to compensate the aperture effect produced in the flat top sampling.

```
function [E]=EqualizerFor_ApertureEff(Ts)
```

```
//Equalizer to Compensate for aperture effect
T_Ts = 0.01:0.01:Ts;
E(1) =1;
for i = 2:length(T_Ts)
    E(i) = ((%pi/2)*T_Ts(i))/(sin((%pi/2)*T_Ts(i)));
end
a =gca();
a.data_bounds = [0,0.8;0.8,1.2];
plot2d(T_Ts,E,5)
xlabel('Duty cycle T/Ts')
ylabel('1/sinc(0.5(T/Ts))')
title('Normalized equalization (to compensate for aperture effect) plotted versus T/Ts')
endfunction
```

3. This function sholud be used along with function adapt_filt

```
function[y]= pow_1(x,N)
xold = 0.0;
```

```
for n =1:N
     sumx = xold+(x(n)^2);
     xold = sumx;
end
     y = sumx/N;
endfunction
```

4. This program is used to perform speech noise cancellation using LMS adaptive filter in scilab

```
//Caption: Speech Noise Cancellation using LMS Adaptive Filter
clc;
//Reading a speech signal
[x,Fs,bits]=wavread("E:\4.wav");
order = 40; // Adaptive filter order
x = x';
N = length(x);
t = 1:N;
//Plot the speech signal
figure(1)
subplot(2,1,1)
plot(t,x)
title('Noise free Speech Signal')
//Generation of noise signal
noise = 0.1*rand(1, length(x));
//Adding noise with speech signal
for i = 1:length(noise)
    primary(i)= x(i)+noise(i);
end
//Plot the noisy speech signal
subplot(2,1,2)
plot(t,primary)
title('primary = speech+noise (input 1)')
//Reference noise generation
for i = 1:length(noise)
   ref(i)= noise(i)+0.025*rand(10);
end
//Plot the reference noise
figure(2)
```

```
subplot(2,1,1)
plot(t,ref)
title('reference noise (input 2)')
//Adaptive filter coefficients initialized to zeros
w = zeros(order,1);
Average_Power = pow_1(x,N)
mu = 1/(10*order*Average_Power); //Adaptive filter step size
//Speech noise cancellation
for k = 1:110
for i =1:N-order-1
       buffer = ref(i:i+order-1); //current order points of reference
       desired(i) = primary(i)-buffer'*w; // dot product the reference & filter
       w = w+(buffer.*mu*desired(i)); //update filter coefficients
   end
end
//Plot the Adaptive Filter output
subplot(2,1,2)
plot([1:length(desired)],desired)
title('Denoised Speech Signal at Adaptive Filter Output')
//Calculation of Mean Squarred Error between the original speech signal and
//Adaptive filter output
for i =1:N-order-1
   err(i) = x(i)-desired(i);
square_error(i)= err(i)*err(i);
end
MSE = (sum(square_error))/(N-order-1);
MSE_dB = 20*log10(MSE);
//Playing the original speech signal
sound(x,Fs,16)
//Delay between playing sound signals
for i = 1:1000
j = 1;
end
//Playing Noisy Speech Signal
sound(primary,Fs,16)
//Delay between playing sound signals
for i = 1:1000
j = 1;
```

5. This program is used to get speech or voice signal informations such as sampling rate, bit deoth and tine duration of speech signal etc

```
//Caption: Reading a Speech Signal &
//[1]. Displaying its sampling rate
//[2]. Number of bits used per speech sample
//[3]. Total Time duration of the speech signal in seconds
clear;
clc;
[y,Fs,bits]=wavread("E:\4.wav");
a = gca();
plot2d(y);
a.x location = 'origin';
title('Speech signal with Sampling Rate = 8 KHz, No. of Samples = 8360')
disp(Fs, 'Sampling Rate in Hz Fs = ');
disp(bits,'Number of bits used per speech sample b =');
N = length(y);
T = N/Fs;
disp(N,'Total Number of Samples N =')
disp(T,'Duration of speech signal in seconds T=')
//Result
//Sampling Rate in Hz Fs =
// 8000.
//Number of bits used per speech sample b =
// 16.
//Total Number of Samples N =
// 8360.
//Duration of speech signal in seconds T=
// 1.045
```