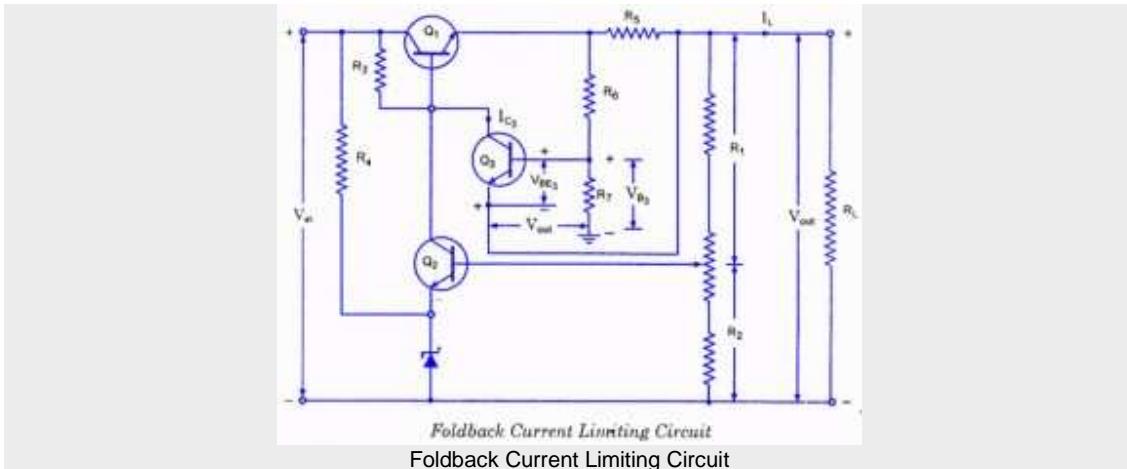


## Foldback Current Limiting

A problem with the simple current limiting circuit is that there is a large amount of power dissipation in the series pass transistor  $Q_1$  when the regulator remains short-circuited. This foldback current-limiting circuit is the solution of the above problem.



The circuit of a transistor series voltage regulator with foldback current limiting is illustrated in the figure. In this circuit, the base of transistor  $Q_3$  is biased by a voltage divider network consisting of resistors  $R_6$  and  $R_7$ . The load current  $I_L$  flows through resistor  $R_5$ , causing a voltage drop of  $I_L R_5$  across it. Thus a voltage of  $(I_L R_5 + V_{out})$  acts across the voltage divider ( $R_6 - R_7$ ) network. The voltage applied to the base of transistor  $Q_3$  is equal to the voltage drop across resistor  $R_7$  and is given as

$$V_{B3} = [R_7 / (R_6 + R_7)] (I_L R_5 + V_{out}) = K(I_L R_5 + V_{out}) \quad [1]$$

$$\text{where } K = R_7 / (R_6 + R_7) \quad [2]$$

The emitter of transistor  $Q_3$  is connected to the positive terminal of  $V_{out}$ . Applying Kirchhoff's voltage law to closed mesh of  $Q_3$  shown in the figure we have

$$V_{out} + V_{BE3} = V_{B3} \quad [3]$$

$$\text{or } V_{BE3} = V_{B3} - V_{out} = K(I_L R_5 + V_{out}) - V_{out} = K I_L R_5 - (1-K) V_{out} \quad [4]$$

The magnitude of base drive of transistor  $Q_1$  is controlled by  $V_{be3}$  of transistor  $Q_3$ . If the load resistance  $R_L$  decreases, the load current  $I_L$  will increase, causing voltage drop  $I_L R_5$  to increase. This causes  $V_{B3}$  to increase and therefore  $V_{be3}$  to increase. This makes transistor  $Q_3$  begin to conduct. The increased collector current  $I_{C3}$  of transistor  $Q_3$  flows through the resistor  $R_3$  thereby decreasing the base voltage of transistor  $Q_1$ . This results in reduction of the conduction level of

transistor Q<sub>1</sub>, limiting further increase in load current  $I_L$ . Using the base-emitter saturation voltage  $V_{BE3} = 0.7$  volts, the maximum current is

$$I_{Lmax} = [0.7 + (1-K) V_{out}] / KR_5 \quad [5]$$

#### Example for non-foldback current limiting

For  $R_5 = 1$  ohm and  $K = 1$ ,  $I_{Lmax} = 0.7$  amps (independent of  $V_{out}$ )

#### Example for foldback current limiting

For  $R_5 = 1$  ohm,  $V_{out} = 5$  volts, and  $K = 0.7$ ,  $I_{Lmax} = 3.1$  amps.

For  $R_5 = 1$  ohm,  $V_{out} = \text{short circuit}$ , and  $K = 0.7$ ,  $I_{Lshort} = 1.0$  amps.

Check the actual  $I_{Lmax}$  and  $I_{Lshort}$  by varying  $R_5$  and bench test it.

One final note. In Eq [4], for  $I_L=0$ ,  $V_{BE3} = -(1-K) V_{out}$ . Thus the base emitter junction is reverse biased when the load current is zero. The  $V_{EBO}$  rating of most transistors is very low, sometimes about 5 volts. A properly placed diode can protect this junction.

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Adapted from <http://www.circuitstoday.com/category/power-supplies/page/2>

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