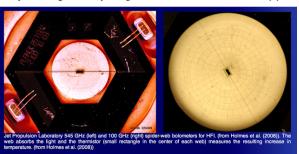
Exercise 05: Bolometer

- Gugliemo Grillo

Exercise on Bolometer



- A modern bolometer is miniature version of the calorimeter
- The calorimeter body behaves like a resistor with temperature dependent resistivity
- Thus the variations of the temperature T of the calorimeter can be measured by measuring the corresponding variation of its electrical resistance R(T)

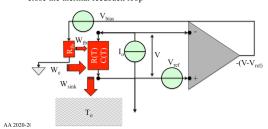


A possible feedback loop



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- · The picture shows a possible feedback loop.
 - the output $-(V-V_{ref})$ of the amplifier, is superimposed to a positive de voltage V_{bias} (the role of this will be clarified later). The sum of the two voltages makes a current $I_{fb} = (V_{bias} (V-V_{ref}))/R_{fb}$ flow through a resistor, which is perfectly thermally coupled to the bolometer.
 - The heat $W_{tb},$ generated by I_{fb} via Joule effect in the resistor, is used to close the thermal feedback loop



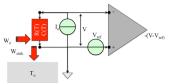
Numerical values

- $R_o = 1 \Omega$
- $T_o = 10 \ mK$
- $C_o = 150 \, pJ/K$
- $\kappa = 10 \, nW/K$
- $\rho = 2 k\Omega/K$
- $I_o = 1 \,\mu A$
- $R_{fb} = 100\Omega$
- $V_{bias} = 50 \,\mu V$

Model for a bolometer



- A constant bias current I_o, generated by a dc current generator, passing through the bolometer converts the change of resistance into a change of voltage V across the resistor. This is the final output of the instrument.
- The bolometer is in conductive thermal contact with a thermal sink at temperature T_0 , with a fairly temperature independent conductance κ . Thus $W_{sink} = \kappa(T-T_0)$
- As for all solids at low enough temperatures, which is where bolometers are used nowadays, the heat capacity depends on temperature as $C(T) = \frac{C_0}{T_0^3} T^3$, with C_0 a constant
- The voltage V across the bolometer is measured by an ideal, infinite input impedance differential amplifier with gain G=1.
- The way the amplifier is biased, produces an output $-(V-V_{ref})$, with $V_{ref}=I_oR(T_o)$ so that the amplifier output is zero when $T=T_o$



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Further notes

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- In order for the bolometer to work properly the $\delta T = T T_o$ must be so small that the all relevant equations might be linearized. The feedback loop is used to achieve this task.
- In the absence of external applied heath W_e, the bolometer will reach an equilibrium temperature T_e = T_o + ΔT. Again, for the magnetometer to work properly, ΔT but be small.
- V_{bias} is needed to obtain a double-sided feedback: as the Joule effect
 is quadratic in the voltage, you need to apply some heat W_{fbo} = V²_{bias}/
 R_{fb} also for zero feedback signal. Then

$$\delta W_{fb} = W_{fb} - W_{fbo} = \frac{(V_{bias} - (V - V_{ref}))^2}{R_{fb}} - \frac{V_{bias}^2}{R_{fb}}$$

can take any sign, and warm up or cool down the bolometer relative to T_e

• In the whole operating range of the experiment

$$R(T) = R_o + \rho (T - T_o)$$

S. Vitale

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Exercise

- Linearize the thermodynamic equations for δT and calculate the equilibrium value ΔT , that is the value for $W_e=0$
- · Calculate the voltage at amplifier output at equilibrium

$$V_e = V(T_o + \Delta T) - V_{ref}$$

- Calculate the equilibrium value for W_{fb} , W_{fbe}
- Consider W_e as a small input signal, and find the impulse and frequency responses $W_e \to \delta T' = \delta T \Delta T$ and $W_e \to \delta V = V V_{ref} V_e$
- What would those be if $V_{bias} = 0$, that is, open loop?
- Find the impulse and frequency response $W_e \rightarrow \delta W_{fb} = W_{fb} W_{fbe}$ and discuss how well $-\delta W_{fb}$ estimates W_e
- If $W_e = W_o Sin(2\pi ft)$ what is the maximum value that $|W_o|$ can take if one wants to keep $|\delta T'| \le 1 \, \mu K$.

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In [1]:

packages used

import scipy.constants
from scipy import signal
from scipy.integrate import quad

import numpy as np

import matplotlib.pyplot as plt

```
import seaborn as sns

from IPython.display import display, Math

# Given constans
R0 = 1 #Ohm
T0 = 10e-3 #K
C0 = 150e-12 # J/K
KAPPA = 10e-9 # W/K
RHO = 2e3 #Ohm/K
I0 = 1e-6 #A
Rfb = 100 #Ohm
Vbias = 50e-6 #V
```

Equilibrium Value

Let's start by writing the thermodynamical equation for the system both in terms of heat and voltages:

$$C_0rac{T^3}{T_0^3}rac{d\delta T}{dt}=W_e-W_{sink}+W_{fb}$$
 $C_orac{(T+\delta T)^3}{T_0}rac{d\delta T}{dt}=W_e-k\delta T+rac{[V_{bias}-(V-V_{ref})]^2}{R_{fb}}$

The second one, by substituting the relation $V(T) = I_0 \rho(T - T_0) + I_0 R(T_0)$ and neglecting terms of order higher than linear, becomes:

$$C_oig(3rac{\delta T}{T_0}+1ig)rac{d\delta T}{dt}+ig(k+2rac{V_{bias}}{R_{fb}}
ho I_oig)\delta T=W_e+rac{V_{bias}^2}{R_{fb}}$$

It's then possible to evaluate ΔT by imposing $W_e=0$ and noting that at equilibrium $\frac{d\delta T}{dt}=0$. Thus we have:

$$\Delta T = rac{V_{bias}^2}{R_{fb}}ig[k + 2rac{V_{bias}}{R_{fb}}
ho I_oig]^{-1}$$

```
In [2]: DeltaT = Vbias**2/Rfb / (KAPPA+2*Vbias*RHO*I0/Rfb)
    display(Math(r'T_e-T_0=\Delta T = {:.5e} K'.format(DeltaT)))
```

$$T_e - T_0 = \Delta T = 2.08333e - 03K$$

Amplifier's voltage output at equilibrium

The quantity desired is obtained through:

$$V_e = V(T_0 + \Delta T) - V_{ref} = \rho(T_0 + \Delta T)I_0 - \rho T_0 I_0 = \rho I_0 \Delta T$$

```
In [3]: Ve = RHO*I0*(DeltaT)
  display(Math(r'V_e = {:.5e} V'.format(Ve)))
```

Feedback heat at equilibrium

The heat generated through the feedback loop can be written as:

$$W_{fbe} = rac{[V_{bias} - (V - V_{ref})]^2}{R_{fb}} = rac{V_{bias}^2 - 2
ho\Delta T I_0 V_{bias} + (
ho I_0\Delta T)^2}{R_{fb}} \simeq rac{V_{bias}^2 - 2
ho\Delta T I_0 V_{bias}}{R_{fb}}$$

where we discarded the term depending on ΔT^2 .

$$W_{fbe} = 2.08333e - 11V$$

Impulse and frequency response

In order to evaluate both the impulse and the frequency response we assume the term $\frac{\delta T}{T_0} << 1$ and simplify the differential equation to:

$$C_o rac{d\delta T(t)}{dt} + ig(k + 2rac{V_{bias}}{R_{fb}}
ho I_oig)\delta T(t) = W_e(t) + rac{V_{bias}^2}{R_{fb}}$$

Case 1:
$$W_e
ightarrow \delta T' = \delta T - \Delta T$$

If we perform the desired change of variable the equation becomes:

$$egin{split} C_o rac{d\delta T'(t)}{dt'} + ig(k+2rac{V_{bias}}{R_{fb}}
ho I_oig)\delta T'(t) + ig(k+2rac{V_{bias}}{R_{fb}}
ho I_oig)\Delta T &= W_e(t) + rac{V_{bias}^2}{R_{fb}} \ \ &\Rightarrow C_o rac{d\delta T'(t)}{dt} + ig(k+2rac{V_{bias}}{R_{fb}}
ho I_oig)\delta T'(t) &= W_e(t) \end{split}$$

where we make use of the relation (4). Performing a Fourier transform we get:

$$C_o i \omega \delta T'(\omega) + \left(k + 2rac{V_{bias}}{R_{fb}}
ho I_o
ight)\delta T'(\omega) = W_e(\omega)$$
 [Frequency response] $h(\omega) = rac{\delta T'(\omega)}{W_e(\omega)} = rac{1/C_0}{i\omega + rac{k}{C_0} + 2rac{V_{bias}}{R_{fb}C_0}}$ [Impulse response] $h(t) = rac{\delta T'(t)}{W_e(t)} = rac{1}{2\pi C_0}e^{-\left(rac{k}{C_0} + 2rac{V_{bias}}{R_{fb}C_0}
ight)t} heta(t)$

Case 2:
$$W_e
ightarrow \delta V = V - V_{ref} - V_e$$

We could substitute and redo all the calculation. Instead we note that:

$$\delta V(t) = I_0
ho(\delta T'(t) + \Delta T) - I_0
ho(T_0 + \Delta T - T_0) = I_0
ho \ \delta T'(t)$$
 $o \delta V(\omega) = I_0
ho \ \delta T'(\omega)$

We can then perform the substitution in the Fourier space and get:

$$[ext{Frequency response}] \; h(\omega) = rac{\delta V(\omega)}{W_e(\omega)} = rac{I_0
ho/C_0}{i\omega + rac{k}{C_0} + 2rac{V_{bias}}{R_{fb}C_0}}$$

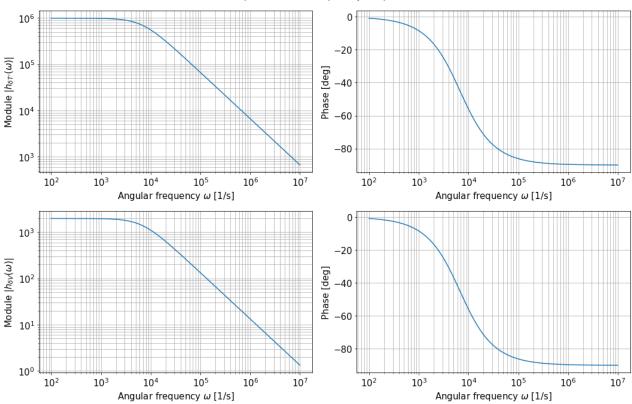
$$[\text{Impulse response}] \; h(t) = \frac{\delta V(t)}{W_e(t)} = \frac{I_0 \rho}{2\pi C_0} e^{-\left(\frac{k}{C_0} + 2\frac{V_{bias}}{R_{fb}C_0}\right)t} \theta(t)$$

The plots for the impulse and frequency response are:

```
In [5]:
         def h T(omega):
             return 1/(1j*omega*C0 + KAPPA + 2*Vbias/Rfb)
         def h V(omega):
             return I0*RHO/(1j*omega*C0 + KAPPA + 2*Vbias/Rfb)
         w = np.linspace(int(1e2), int(1e7), int(1e6))
         points_h_T = h_T(w)
         mag = np.abs(points h T)
         phase = np.arctan(np.imag(points_h_T) / np.real(points_h_T))*180/np.pi
         # Bode plot
         plt.rcParams.update({'font.size': 15})
         fig, axis = plt.subplots(2,2, figsize=(15, 10))
         fig.suptitle('Bode plot for the Frequency responses')
         axis[0][0].plot(w, mag)
         axis[0][0].set xscale('log')
         axis[0][0].set_yscale('log')
         axis[0][0].set xlabel("Angular frequency $\omega$ [1/s]")
         axis[0][0].set_ylabel("Module |$h_{\delta T'}(\omega$)|")
         axis[0][0].grid(True, which="both")
         axis[0][1].plot(w, phase)
         axis[0][1].set_xscale('log')
         axis[0][1].set xlabel(r"Angular frequency $\omega$ [1/s]")
         axis[0][1].set_ylabel(r"Phase [deg]")
         axis[0][1].grid(True, which="both")
         points h V = h V(w)
         mag = np.abs(points_h_V)
         phase = np.arctan(np.imag(points h V) / np.real(points h V))*180/np.pi
         axis[1][0].plot(w, mag)
         axis[1][0].set xscale('log')
         axis[1][0].set_yscale('log')
         axis[1][0].set_xlabel(r"Angular frequency $\omega$ [1/s]")
         axis[1][0].set_ylabel(r"Module |$h_{\delta V}(\omega$)|")
         axis[1][0].grid(True, which="both")
         axis[1][1].plot(w, phase)
         axis[1][1].set_xscale('log')
         axis[1][1].set_xlabel(r"Angular frequency $\omega$ [1/s]")
         axis[1][1].set ylabel(r"Phase [deg]")
         axis[1][1].grid(True, which="both")
```

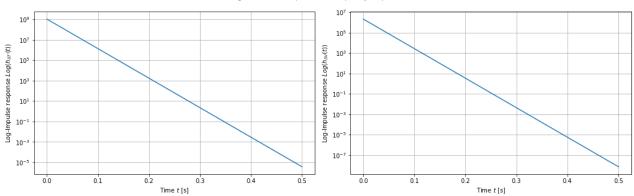
```
plt.tight_layout()
plt.show()
```

Bode plot for the Frequency responses



```
def h T(t):
In [6]:
             return np.exp(-(KAPPA/C0 + 2*Vbias/Rfb)*t)/(2*np.pi*C0)
         def h_V(t):
             return I0*RHO/(2*np.pi*C0)*np.exp(-(KAPPA/C0 + 2*Vbias/Rfb)*t)
         t = np.linspace(0, 0.5, int(1e6))
         # Bode plot
         plt.rcParams.update({'font.size': 10})
         fig, axis = plt.subplots(1, 2, figsize=(15, 5))
         fig.suptitle('Log-Plot for the impulse and frequency responses')
         axis[0].plot(t, h_T(t=t))
         axis[0].set_yscale('log')
         axis[0].set_xlabel("Time $t$ [s]")
         axis[0].set_ylabel("Log-Impulse response $Log(h_{\delta T'}(t))$")
         axis[0].grid(True, which="both")
         axis[1].plot(t, h_V(t=t))
         axis[1].set_yscale('log')
         axis[1].set_xlabel("Time $t$ [s]")
         axis[1].set_ylabel("Log-Impulse response $Log(h_{\delta V}(t))$")
         axis[1].grid(True, which="both")
         plt.tight_layout()
         plt.show()
```

Log-Plot for the impulse and frequency responses



Null V_{bias}

If $V_{bias} = 0$ the answer follow from the previous answers:

Case 1:
$$W_e
ightarrow \delta T' = \delta T - \Delta T$$

$$h(\omega) = rac{\delta T'(\omega)}{W_e(\omega)} = rac{1/C_0}{i\omega + rac{k}{C_0}}$$

$$h(t)=rac{\delta T'(t)}{W_e(t)}=rac{1}{2\pi C_0}e^{-rac{k}{C_0}t} heta(t)$$

Case 2:
$$W_e
ightarrow \delta V = V - V_{ref} - V_e$$

$$h(\omega) = rac{\delta V(\omega)}{W_e(\omega)} = rac{I_0
ho/C_0}{i\omega + rac{k}{C_0}}$$

$$h(t)=rac{\delta V(t)}{W_e(t)}=rac{I_0
ho}{2\pi C_0}e^{-rac{k}{C_0}t} heta(t)$$

```
In [7]: def h_T(t):
    return np.exp(-(KAPPA/C0)*t)/(2*np.pi*C0)

def h_V(t):
    return I0*RHO/(2*np.pi*C0)*np.exp(-(KAPPA/C0)*t)

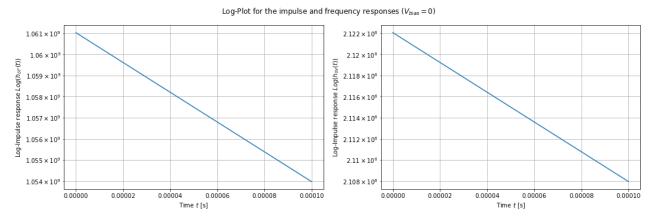
t = np.linspace(0, 1e-4, int(1e6))

# Bode plot
plt.rcParams.update({'font.size': 10})
fig, axis = plt.subplots(1, 2, figsize=(15, 5))
fig.suptitle('Log-Plot for the impulse and frequency responses ($V_{bias}=0$)')

axis[0].plot(t, h_T(t=t))
axis[0].set_yscale('log')
axis[0].set_xlabel("Time $t$ [s]")
axis[0].set_ylabel("Log-Impulse response $Log(h_{delta T'}(t))$")
axis[0].grid(True, which="both")
```

```
axis[1].plot(t, h_V(t=t))
axis[1].set_yscale('log')
axis[1].set_xlabel("Time $t$ [s]")
axis[1].set_ylabel("Log-Impulse response $Log(h_{\delta V}(t))$")
axis[1].grid(True, which="both")

plt.tight_layout()
plt.show()
```



Case: $W_e
ightarrow \delta W_{fb} = W_{fb} - W_{fbe}$

To start, let's write the relation between δT and δW_{fb} :

$$\delta W_{fb} = W_{fb} - W_{fbe} = rac{[V_{bias} - (V - V_{ref})]^2}{R_{fb}} - rac{V_{bias}^2 - 2
ho\Delta T I_o V_{bias} + (
ho I_0\Delta T)}{R_{fb}} = rac{-2V_{bias}
ho I_0(\delta)}{R_{fb}}$$

$$\delta W_{fb} = rac{-2V_{bias}
ho I_0}{R_{fb}}\delta T + rac{2V_{bias}
ho I_0}{R_{fb}}\Delta T + \mathcal{O}[(\delta T^2 - \Delta T^2)]$$

$$o \delta T = -rac{R_{fb}}{2
ho V_{bias}I_0}\delta W_{fb} + \Delta T$$

Where in (22) the $\mathcal{O}\big[(\delta T^2-\Delta T^2)\big]$ term was neglected. This is done because $\mathcal{O}\big[(\delta T^2-\Delta T^2)\big]=\mathcal{O}\big[(T_0^2-T_e^2+2T\;\Delta T)\big]$ is negligible. An esteem for the error at equilibrium is provided before the plot.

The differential equation then becomes:

$$-rac{C_0R_{fb}}{2V_{bias}
ho I_0}rac{d\ \delta W_{fb}(t)}{dt}-ig(k+2rac{V_{bias}}{R_{fb}
ho I_0}ig)rac{R_{fb}}{2V_{bias
ho I_0}}\ \delta W_{fb}=W_e(t)+ig[rac{V_{bias}^2}{R_{fb}}-ig(k+2rac{V_{bias}}{R_{fb}
ho I_0}ig)\ \Delta T_e^2$$

Where the term between square brackets on the right is zero due to (4). The equation can be rewritten as:

$$rac{d \ \delta W_{fb}(t)}{dt} + ig(rac{k}{C_0} + 2rac{V_{bias}
ho I_0}{C_0 R_{fb}}ig) \ \delta W_{fb}(t) = -rac{2V_{bias}
ho I_0}{C_0 R_{fb}}W_e(t)$$

Defining
$$au=rac{2V_{bias}
ho I_0}{C_0R_{fb}}$$
 and $\delta W_{fb}
ightarrow -\delta W_{fb}$:

$$rac{d \; \delta W_{fb}(t)}{dt} + \left(rac{k}{C_0} + au
ight) \; \delta W_{fb}(t) = au W_e(t)$$

$$i\omega \; \delta W_{fb}(\omega) + \left(rac{k}{C_0} + au
ight) \; \delta W_{fb}(\omega) = au W_e(\omega)$$

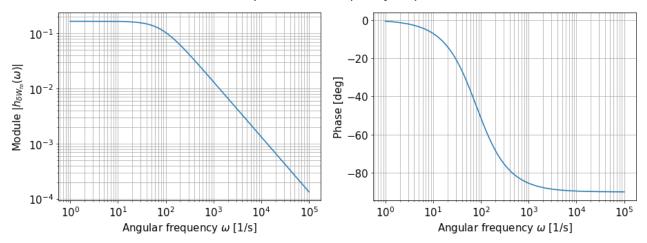
The frequency and impulse responses are then:

$$h_{\delta W_{fb}}(\omega) = rac{W_{fb}(\omega)}{W_e(\omega)} = rac{ au}{i\omega + (rac{k}{C_0} + au)}$$

$$h_{\delta W_{fb}}(t) = rac{W_{fb}(t)}{W_e(t)} = rac{ au}{2\pi} heta(t) e^{-(rac{k}{C_0} + au)t}$$

```
Te = T0 + DeltaT
In [33]:
          error = lambda T: (RHO**2 * IO**2 / Rfb)*(TO**2 - Te**2 + 2*T*DeltaT)
          display(Math(r'error({:.5e})\simeq{:.5e} W'.format(Te, error(Te))))
         error(1.20833e - 02) \simeq 1.73611e - 13W
          tau = 2*Vbias*RHO*I0 /(C0*Rfb)
 In [9]:
          def h W(omega):
              return -1*tau / (1j*omega+(KAPPA/C0 + tau) )
          w = np.linspace(int(1e0), int(1e5), int(1e6))
          points = h W(w)
          mag = np.abs(points)
          phase = np.arctan(np.imag(points) / np.real(points))*180/np.pi
          # Bode plot
          plt.rcParams.update({'font.size': 15})
          fig, axis = plt.subplots(1,2, figsize=(15, 5))
          fig.suptitle('Bode plot for the Frequency responses')
          axis[0].plot(w, mag)
          axis[0].set xscale('log')
          axis[0].set yscale('log')
          axis[0].set_xlabel("Angular frequency $\omega$ [1/s]")
          axis[0].set_ylabel("Module |$h_{\delta W_{fb}}(\omega$)|")
          axis[0].grid(True, which="both")
          axis[1].plot(w, phase)
          axis[1].set xscale('log')
          axis[1].set_xlabel(r"Angular frequency $\omega$ [1/s]")
          axis[1].set_ylabel(r"Phase [deg]")
          axis[1].grid(True, which="both")
```

Bode plot for the Frequency responses



Maximum value for $|W_0|$

In order to answer to the last problem, let's rewrite equation (8) with the given values:

$$C_o rac{d\delta T'(t)}{dt} + ig(k + 2rac{V_{bias}}{R_{fb}}
ho I_oig)\delta T'(t) = W_0 sin(2\pi ft)$$

Now we perform some simplification assuming that:

\begin{enumerate}

\item We are only interested in the behaviour near the maxima

\item The relation between $\d T'(t)\$ and $W_e(t)\$ is linear (this can be assumed because we ar \end{enumerate}

With these hypothesis the maxima of $W_e(t)$ correspond to the maxima of $\delta T'(t)$ and it's derivative with respect to the time is zero. The expression can then be simplified to:

$$ig(k+2rac{V_{bias}}{R_{fb}}
ho I_oig)\delta T'(t)=W_0 o\delta T'(t)=W_0ig(k+2rac{V_{bias}}{R_{fb}}
ho I_oig)^{-1}\leq 1\mu K$$

And the maximum value for W_0 is:

$$W_0 = \left(k + 2rac{V_{bias}}{R_{fb}}
ho I_o
ight)*1\mu K$$

 $|W_0| = 1.20000e - 08W$