

$K \in K^d$ DECENTRALIZED
 $K \in K^m$

\Rightarrow IN ORDER TO DEFINE K WITH A GIVEN
 STRUCTURE WE CAN USE LMIs BASED DESIGN

2023.10.24

LMIs

$$\begin{aligned} P > 0 \\ A^T P + P A < 0 \end{aligned} \quad \left\{ \begin{array}{l} A \text{ IS HURWITZ} \end{array} \right.$$

$$\begin{cases} P > 0 \\ F^T P F - P < 0 \end{cases} \Rightarrow F \text{ IS SCHUR STABLE}$$

LMIs FOR STABILITY

CENTRALIZED STATE-FEEDBACK CONTROLLER DESIGN

$$\begin{aligned} \dot{x} = Ax + Bu & \xrightarrow{\text{LYAPUNOV: } A_u \text{ IS HURVITZ} \Leftrightarrow \exists P = P^T \text{ s.t. AND } K \text{ s.t.}} \\ u = Kx & \quad K \in \mathbb{R}^{n \times n} \\ \xrightarrow{\text{LINEAR IN } P \quad \text{NON LINEAR IN } P \text{ AND } K} & \quad \dot{x} = \frac{(A+BK)x}{A_C} \end{aligned}$$

REFORMULATE IT AS AN LMI

$$Y = P^{-1} \quad Y(A^T P + P A + K^T B^T P + P B K) Y < 0 \quad \begin{array}{c} \text{LINEAR} \\ \downarrow L^T \\ Y A^T + A Y + Y K^T B^T + B K Y < 0 \end{array} \Rightarrow \begin{array}{c} \text{LINEAR} \\ \downarrow L \\ Y A^T + A Y + L^T B^T + B L < 0 \end{array}$$

\rightarrow SOLUTION TO LMI PROBLEM $\rightarrow Y; L \Rightarrow P = Y^{-1} \Rightarrow K = LY^{-1} \Rightarrow A+BK$ IS HURVITZ

DISCRETE TIME

$$x_{k+1} = Fx_k + Gu_k \quad \xrightarrow{\text{LYAPUNOV: } F_u \text{ IS SCHUR STABLE IFF } \exists P = P^T \text{ s.t. AND } K \text{ s.t.}}$$

$$u_k = Kx_k \quad \xrightarrow{\text{LINEAR IN } P \quad \text{NON LINEAR IN } P \text{ AND } K} \quad \begin{cases} P > 0 \\ F_u P F_u^T - P < 0 \rightarrow (F+GK)^T P (F+GK) - P < 0 \end{cases}$$

$$\begin{array}{c} \text{LINEAR IN } P \\ \text{NL IN } P \text{ AND } K \end{array} \quad \xrightarrow{\text{NL IN } P, K} \quad \begin{array}{c} F^T P F - P < 0 \rightarrow F P F^T - P + F L^T G^T + G L F^T + G L P^{-1} L^T G^T < 0 \\ \xrightarrow{\text{HEAVILY NL}} G K P K^T C^T = G L P^{-1} P K^T G^T = G L P^{-1} L^T G^T \end{array}$$

$$-(F P F^T - P + F L^T G^T + G L F^T + G L P^{-1} L^T G^T) < 0 \quad \xrightarrow{\text{LINEAR}} P - F P F^T - F L^T G^T - G L F^T - G L P^{-1} L^T G^T > 0 \quad \textcircled{*}$$

SCHUR COMPLEMENT

$$\begin{cases} P > 0 \\ Q - S^{-1} P S > 0 \end{cases} \Leftrightarrow \begin{bmatrix} Q & S \\ S^{-1} & P \end{bmatrix} > 0$$

THANKS TO THE SCHUR COMPLEMENT $\textcircled{*}$ IS EQUIVALENT TO

$$\begin{bmatrix} P - F P F^T - F L^T G^T - G L F^T & G L \\ G L^T & P \end{bmatrix} > 0 \quad \Rightarrow \text{LMI!} \Rightarrow P, L \Rightarrow K = L P^{-1} \Rightarrow F+GK \text{ IS SCHUR STABLE!}$$

LMI FOR PERFORMANCE (HOW TO PLACE EIGENVALUES IN A REGION?)

CONTINUOUS TIME $A+BK$, FOR FAST RESPONSE



$$\Rightarrow \operatorname{Re}(\operatorname{eig}(A+BK)) < -\alpha \Rightarrow \operatorname{Re}(\operatorname{eig}(A+BK)) + \alpha < 0$$

$$\operatorname{Re}(\operatorname{eig}(A+BK+\alpha I)) < 0$$

$\Rightarrow A+BK+\alpha I$ HURVITZ STABLE

\Rightarrow PLUG THIS IN THE LMI!

$$(A+BK+\alpha I)^T P + P(A+BK+\alpha I) < 0$$

$$A^T P + PA + K^T B^T P + PBK + 2\alpha P < 0$$

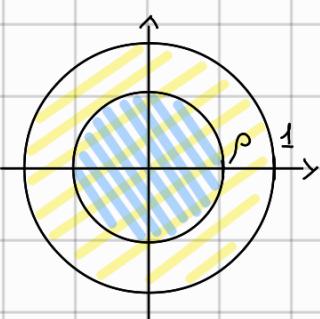
$$Y(A^T P + PA + K^T B^T P + PBK + 2\alpha P) Y < 0$$

$$YA^T + YA + YK^T B^T + BKY + 2\alpha Y < 0$$

$\downarrow L^T \quad \downarrow L$

$$YA^T + YA + L^T B^T + BL + 2\alpha Y < 0 \quad \text{LMI!}$$

DISCRETE TIME



FOR FASTER RESPONSE, $\rho < 1$

$$|\operatorname{eig}(A+BK)| < \rho \Rightarrow \frac{|\operatorname{eig}(A+BK)|}{\rho} < 1$$

$$\Rightarrow \left| \operatorname{eig}\left(\frac{A+BK}{\rho}\right) \right| < 1 \quad \frac{A+BK}{\rho} \text{ SCHUR STABLE}$$

$$\frac{(F+GK)}{\rho} P \frac{(F+GK)^T}{\rho} - P < 0 \quad (F+GK) P (F+GK)^T - \rho^2 P < 0$$

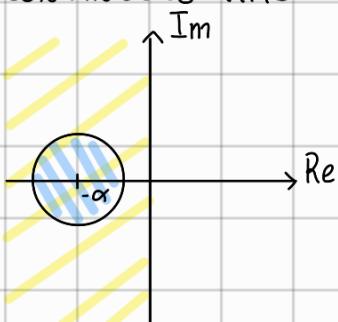
ENDING UP WITH

$$\begin{bmatrix} \rho^2 P - FPF^T - FL^T C^T - GLF^T & CL \\ LTG^T & P \end{bmatrix} > 0 \quad \text{LMI!}$$

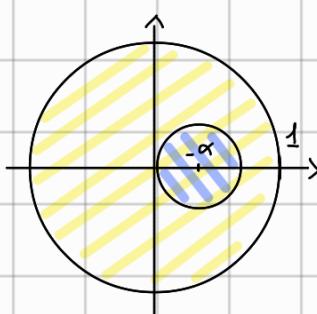
CONFINE EIGS IN A CIRCLE WITH CENTER IN $-\alpha$ AND RADIUS ρ (LIMIT OSCILLATIONS)

2023.10.26

CONTINUOUS TIME



DISCRETE TIME



e.g. $x_{k+1} = -0.5x_k$

$\Rightarrow x_{k+1}$ WILL CHANGE SIGN
EACH STEP

$$\rightarrow \frac{(A+BK+\alpha I)}{P} P \frac{(A+BK+\alpha I)^T}{P} - P < 0 \quad (A+BK+\alpha I)P(A+BK+\alpha I)^T - P^2 P < 0$$

$$(A+BK)P(A+BK)^T + \alpha P(A+BK)^T + \alpha (A+BK)P + \alpha^2 P - P^2 P < 0$$

$$(AP+BKP)P^{-1}(AP^T+BKP)^T - (AP+BL)P^{-1}(AP+BL)^T$$

$$\alpha PA^T + \alpha PK^T B + \alpha AP + \alpha BKP = \alpha PA^T + \alpha L^T B + \alpha AP + \alpha BL$$

$$\frac{(\rho^2 - \alpha^2)P - \alpha(PA^T + AP + L^T B + BL)}{Q} - \frac{(AP+BL)P^{-1}(AP+BL)^T}{S} > 0$$

S

S^T

USING THE SCHUR COMPLEMENT

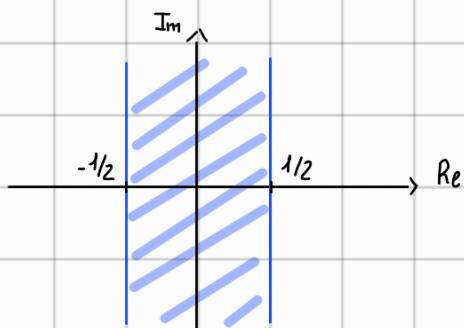
$$\begin{bmatrix} Q & S \\ S^T & P \end{bmatrix} > 0 \Rightarrow \begin{bmatrix} (\rho^2 - \alpha^2)P - \alpha(PA^T + AP + L^T B + BL) & (AP+BL) \\ (AP+BL)^T & P \end{bmatrix} > 0$$

CONFINE EIGS IN A REGION D

$$D := \left\{ z \in \mathbb{C} \text{ s.t. } f_D(z) < 0 \right\}, \quad f_D(z) = \underbrace{\Lambda}_{\substack{\text{REAL} \\ \text{SYMMETRIC} \\ \text{SQUARE} \\ \text{MATRIX}}} + \underbrace{\Theta z + \Theta^T z^*}_{\substack{\text{REAL} \\ \text{SQUARE} \\ \text{COMPLEX CONJUGATE}}}$$

EXAMPLE 1

$$\Lambda = \begin{bmatrix} -1 & 0 \\ 0 & -1 \end{bmatrix}, \quad \Theta = \begin{bmatrix} 1 & 0 \\ 0 & -1 \end{bmatrix}, \quad f_D(z) = \begin{bmatrix} -1 & 0 \\ 0 & -1 \end{bmatrix} + \begin{bmatrix} z & 0 \\ 0 & -z \end{bmatrix} + \begin{bmatrix} z^* & 0 \\ 0 & z^* \end{bmatrix} = \begin{bmatrix} -1+z+z^* & 0 \\ 0 & -1-(z+z^*) \end{bmatrix} < 0$$



$$= \begin{bmatrix} -1+2\operatorname{Re}(z) & 0 \\ 0 & -1-2\operatorname{Re}(z) \end{bmatrix} \Rightarrow \begin{cases} -1+2\operatorname{Re}(z) < 0 \Rightarrow \operatorname{Re}(z) < 1/2 \\ -1-2\operatorname{Re}(z) < 0 \Rightarrow \operatorname{Re}(z) > -1/2 \end{cases}$$

$$z+z^* = \operatorname{Re}(z) + j\operatorname{Im}(z) + \operatorname{Re}(z) - j\operatorname{Im}(z) = 2\operatorname{Re}(z)$$

EXAMPLE 2

$$\Lambda = \begin{bmatrix} -1 & 0 \\ 0 & -1 \end{bmatrix}, \quad \Theta = \begin{bmatrix} 0 & 1 \\ -1 & 0 \end{bmatrix}, \quad f_D(z) = \begin{bmatrix} -1 & 0 \\ 0 & -1 \end{bmatrix} + \begin{bmatrix} 0 & z \\ -z & 0 \end{bmatrix} + \begin{bmatrix} 0 & -z^* \\ z^* & 0 \end{bmatrix} = \begin{bmatrix} -1 & z-z^* \\ -(z-z^*) & -1 \end{bmatrix} < 0$$

$$= \begin{bmatrix} -1 & 2j\operatorname{Im}(z) \\ -2j\operatorname{Im}(z) & -1 \end{bmatrix} < 0 \quad \text{BOTH EIGS OF } f_D(z) < 0$$



$$\operatorname{tr}(f_D(z)) = \lambda_1 + \lambda_2 < 0 \quad \operatorname{tr}(f_D(z)) = -2 < 0$$

$$\det(f_D(z)) = \lambda_1 \lambda_2 > 0 \quad \det(f_D(z)) = 1 - 4j\operatorname{Im}(z) > 0$$

$$\Rightarrow -1/2 < \operatorname{Im}(z) < 1/2$$

COMBINING THE 2 $\left(\begin{matrix} 1 & 1 \\ -1 & -1 \end{matrix} \right)$ YOU OBTAIN THE INTERSECTION OF THE REGIONS.
OBVIOUSLY THE REGIONS CAN BE SHIFTED AND SCALED

EXAMPLE 3 (SECTOR)

$$\Lambda = 0$$

$$\Theta = \begin{bmatrix} \sin \alpha & \cos \alpha \\ -\cos \alpha & \sin \alpha \end{bmatrix} \quad f_D(z) = \begin{bmatrix} S_\alpha z & C_\alpha z \\ -C_\alpha z & S_\alpha z \end{bmatrix} + \begin{bmatrix} S_\alpha z^* & C_\alpha z^* \\ C_\alpha z^* & S_\alpha z^* \end{bmatrix} = \begin{bmatrix} S_\alpha(z+z^*) & C_\alpha(z-z^*) \\ -C_\alpha(z-z^*) & S_\alpha(z+z^*) \end{bmatrix} =$$

$$= 2 \begin{bmatrix} S_\alpha \operatorname{Re}(z) & f_\alpha \operatorname{Im}(z) \\ -f_\alpha \operatorname{Im}(z) & S_\alpha \operatorname{Re}(z) \end{bmatrix}$$

THIS IS GUARANTEED BY $\bullet \operatorname{tr}(f_D(z)) = 4 \sin \alpha \operatorname{Re}(z) < 0$

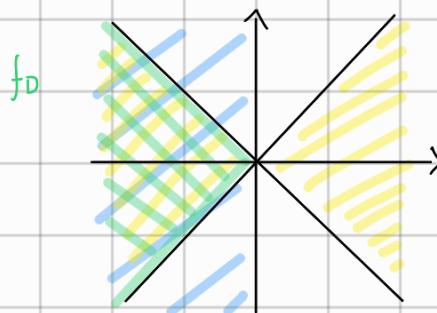
$$\bullet \det(f_D(z)) = \sin^2(\alpha) \operatorname{Re}^2(z) - \cos^2(\alpha) \operatorname{Im}^2(z) > 0$$

$$\alpha \in (0, \pi/2)$$

$$\Rightarrow \sin(\alpha) \operatorname{Re}(z) > 0 \Leftrightarrow \operatorname{Re}(z) < 0$$

$$\Rightarrow \sin^2(\alpha) \operatorname{Re}^2(z) - \cos^2(\alpha) \operatorname{Im}^2(z) > 0$$

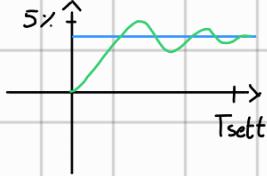
$$\left| \frac{\operatorname{Im}(z)}{\operatorname{Re}(z)} \right| < \left| \frac{\sin \alpha}{\cos \alpha} \right| \Rightarrow q \in (\cos(\alpha), 1)$$



e.g. REQUIREMENTS

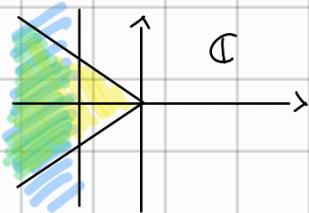
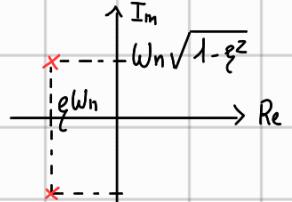
$$T_{\text{sett}} < 10$$

$$S \% < \bar{S}$$



$$T_{\text{sett}} = \frac{5}{|\operatorname{Re}(\lambda)|} \quad S \% = 100 e^{-\frac{\xi \pi}{\sqrt{1-\xi^2}}}$$

$$\frac{\xi}{\sqrt{1-\xi^2}} = \frac{\operatorname{Re}(\lambda)}{\operatorname{Im}(\lambda)}$$



THEOREM

$$\lambda_i \in D, \forall i = 1, \dots, n \text{ IF } \exists Y = Y^\top \text{ s.t. } M_D < 0$$

THE BLOCKS OF M_D ARE $m_{ij} = \lambda_i \delta_{ij} I + \theta_{ij} (A+BK)Y + \theta_{ji} Y(A+BK)^\top$

EXAMPLE 3

$$\Lambda = 0$$

$$\Theta = \begin{bmatrix} \sin \alpha & \cos \alpha \\ -\cos \alpha & \sin \alpha \end{bmatrix} \quad M_D = \begin{bmatrix} \sin \alpha (A+BK)Y + \sin \alpha Y(A+BK)^\top & \cos \alpha (A+BK)Y - \cos \alpha Y(A+BK) \\ -\cos \alpha (A+BK)Y + \cos \alpha Y(A+BK)^\top & \sin \alpha (A+BK)Y + \sin \alpha Y(A+BK)^\top \end{bmatrix} < 0$$

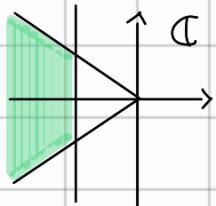
$$\rightarrow (A+BK)Y = AY + BKY = AY + BL \Rightarrow \text{ALWAYS LMI!}$$

COMPUTATIONAL COMPLEXITY DEPENDS ON:

- # CONSTRAINTS (CAN'T BE REDUCED)
- # FREE VARIABLE $\rightarrow Y \in \mathbb{R}^{n \times n}$ SYMMETRIC $\rightarrow 1+2+3+\dots+n = \frac{n(n+1)}{2}$ FREE VARIABLES
 $L \in \mathbb{R}^{m \times n}$ $\rightarrow m \times n$ FREE VARIABLES

LMI OPTIMIZATION PROBLEMS

① MINIMIZE CONTROL EFFORT



$$U = Kx$$

$$\text{minimize } \|K\|$$

$$\hookrightarrow \text{minimize } \alpha_K$$

$$\text{SUBJECT TO } \alpha_K \geq \|K\| \Rightarrow$$

FREE VARIABLES ARE $Y(P)$ AND L

$$\|K\| = \|LY^{-1}\| \leq \|L\| \cdot \|Y^{-1}\|$$

$$\text{minimize } \alpha_L + \alpha_Y \quad (1)$$

$$\text{SUBJECT TO } \alpha_L \geq \|L\|, \alpha_Y \geq \|Y\| \quad (2)$$

LET'S REWRITE (1) AND (2) AS LMIs

$$(1) \alpha_L \geq \|L\| \quad \alpha_L^2 I \geq L^T L \quad \text{P} \quad I - L^T I L \geq 0 \quad \begin{bmatrix} \gamma_L I & L^T \\ L & I \end{bmatrix} > 0$$

$\gamma_L = \alpha_L^2$ SCHUR COMPL

$$(2) \alpha_Y I - I Y^{-1} I \quad \text{SCHUR COMPL} \rightarrow \begin{bmatrix} \alpha_Y I & I \\ I & Y^{-1} \end{bmatrix} > 0$$

LMIs ALLOW TO DEFINE A STRUCTURE FOR THE FREE MATRICES

Y --- DEC $K \in K^d$ BLOCK-DIAGONAL HOW TO FIX A STRUCTURE TO $K = LY^{-1}$ BY FIXING A DIST $K \in K^m$ STRUCTURED \rightarrow STRUCTURE TO Y AND L ?

e.g. $K = \begin{bmatrix} k_1 & 0 \\ 0 & k_2 \end{bmatrix}$ IMPOSING $L = \begin{bmatrix} l_1 & 0 \\ 0 & l_2 \end{bmatrix}$ ISN'T ENOUGH, BECAUSE IF $Y = \begin{bmatrix} y_{11} & y_{12} \\ y_{21} & y_{22} \end{bmatrix} \Rightarrow K = LY^{-1} = \begin{bmatrix} l_1 y_{11} & l_1 y_{12} \\ l_2 y_{21} & l_2 y_{22} \end{bmatrix}$

BUT IF $Y = \begin{bmatrix} y_{11} & 0 \\ 0 & y_{22} \end{bmatrix} \Rightarrow K = LY^{-1} = \begin{bmatrix} l_1 y_{11} & 0 \\ 0 & l_2 y_{22} \end{bmatrix}$ DIAGONAL NOT DIAGONAL

THE RULE IS:

- L MUST HAVE THE SAME BLOCK STRUCTURE AS K
- Y MUST BE BLOCK-DIAGONAL!

IMPLICATIONS

- REDUCTION OF # OF FREE VARIABLES $Y = \begin{bmatrix} Y_1 & 0 & \cdots & 0 \\ 0 & Y_2 & \ddots & \vdots \\ 0 & 0 & \cdots & Y_n \end{bmatrix}$ $\frac{n(n+1)}{2}$ FREE VARIABLES
→ FOR EACH BLOCK $\frac{n(n+1)}{2}$ FREE VARIABLES

SAME HAPPENS FOR L WHEN STRUCTURED ACCORDINGLY TO THE CASE.

- WE LOSE THEORETICAL GUARANTEES

IN THE LYAPUNOV THEOREM $P = P^T > 0 \rightarrow$ IN DISTRIBUTED AND DECENTRALIZED DESIGN WE FIX A STRUCTURE TO P → IF A SOLUTION $\exists \Rightarrow$ THE CORRESPONDING CONTROLLER IS STABILIZING.
IF A CONTROLLER $\exists \Rightarrow$ WE CAN FIND IT WITH LMIs

H₂-NORM MINIMIZATION CONTROL

2023.11.02

CONT. TIME $\dot{x} = Ax + Bu + B_w w$ NOISE/DISTURBANCE
 $z = Cx + Du$ PERFORMANCE OUTPUT

$u = Kx$ s.t. THE H₂-NORM OF $G_{zw}(s)$ (THE TR. FUN. BETWEEN W AND Z) IS MINIMIZED.

E.G. IF WE DEFINE $z = \begin{bmatrix} \sqrt{Q}x \\ \sqrt{R}u \end{bmatrix} = \begin{bmatrix} \sqrt{Q} \\ 0 \end{bmatrix}x + \begin{bmatrix} 0 \\ \sqrt{R} \end{bmatrix}u \Rightarrow$ WE DERIVE LQ CONTROLLER

IN CLOSED LOOP

THEOREM

$$\begin{cases} \dot{x} = (A+BK)x + B_w w \\ z = (C+DK)x \end{cases} \quad \|G_{zw}(s)\|_2^2 = \inf_y \left\{ \text{trace} \left[(C+DKy)Y(C+DK)^T \right] \right. \\ \left. \text{WHERE } (A+BK)Y + Y(A+BK)^T + B_w B_w^T < 0 \right\}$$

NOT VITAL IF NOT WELL KNOWN, JUST $B_w - I$

LET'S MAKE THIS AN LMI

$$\min_K \text{trace} \left[(C+DK)Y(C+DK)^T \right] \quad \text{WHERE } (A+BK)Y + Y(A+BK)^T + B_w B_w^T < 0$$

THIS IS ALSO STABILIZING, BECAUSE IF $(A+BK)Y + Y(A+BK)^T + B_w B_w^T < 0$ THEN $(A+BK)Y + Y(A+BK)^T < 0$

LET'S DEFINE A NEW MATRIX

LYAPUNOV THEOREM

$$\begin{aligned} \min_{Y,L,S} \text{trace}(S) \quad &\text{SUBJECT TO } (A+BK)Y + Y(A+BK)^T + B_w B_w^T < 0 \\ &S \geq (C+DK)Y(C+DK)^T \Rightarrow S \text{ IS AN UPPER BOUND} \\ &\Rightarrow S - (C+DK)Y(C+DK)^T \geq 0 \longrightarrow \underbrace{(CY+DKY)}_L Y^{-1} \underbrace{(CY+DKY)^T}_L < 0 \end{aligned} \quad \text{BOTH LMI UNSUITABLE}$$

⇒ SCHUR COMPONENT

$$\begin{bmatrix} S & CY+DKY \\ (CY+DKY)^T & Y \end{bmatrix} \geq 0 \quad \text{AND} \quad (A+BK)Y + Y(A+BK)^T + B_w B_w^T < 0 \Rightarrow AY + BL + YA^T + L^T B^T + B_w^T B_w < 0$$

DISCR. TIME

$$\begin{cases} X_{k+1} = F X_k + G U_k + B_w W_k \\ Z = C X_k + D U_k \end{cases}$$

$U_k = K X_k$ s.t. THE $\| \cdot \|_2$ -NORM OF $G_{zw}(z)$ (THE TR. FUN. BETWEEN w AND z) IS MINIMIZED.

THEOREM

$$\|G_{zw}(s)\|_2^2 = \inf_y \left\{ \text{trace} \left[(C + DK)Y(C + DK)^T \right] \text{ WHERE } (F + GK)Y(F + GK)^T - Y + B_w B_w^T < 0 \right\}$$

$$\min_K \text{tr} \left[(C + DK)Y(C + DK)^T \right] \text{ WHERE } (F + GK)Y(F + GK)^T - Y + B_w B_w^T < 0$$

FOLLOWING THE SAME STEPS

$$\min \text{trace}(S) \text{ SUBJECT TO SCHUR COMPONENT } \begin{bmatrix} S & CY + DL \\ (CY + DL)^T & Y \end{bmatrix} \geq 0$$

$$(F + GK)Y + Y(F + GK)^T - Y + B_w B_w^T < 0 \rightarrow \text{UNSUITABLE FOR AN LMI}$$

$$(F + GK)Y(F + GK)^T - Y + B_w B_w^T < 0 \rightarrow Y - B_w B_w - (F + GK)Y(F + GK)^T > 0$$

$$Y - B_w B_w - (FY + GKY)Y^{-1}(FY + GKY)^T > 0 \Rightarrow \text{SCHUR COMPLEMENT} \begin{bmatrix} Y - B_w B_w^T & (FY + GL) \\ (FY + GL)^T & Y \end{bmatrix} > 0$$