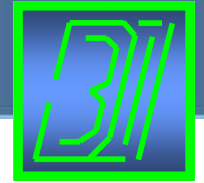




TRƯỜNG ĐẠI HỌC
BÁCH KHOA HÀ NỘI

Nguyễn Công Phương



Electric Circuit Theory

Magnetically Coupled Circuits

Contents

- I. Basic Elements Of Electrical Circuits
- II. Basic Laws
- III. Electrical Circuit Analysis
- IV. Circuit Theorems
- V. Active Circuits
- VI. Capacitor And Inductor
- VII. First Order Circuits
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- IX. Sinusoidal Steady State Analysis
- X. AC Power Analysis
- XI. Three-phase Circuits
- XII. Magnetically Coupled Circuits**
- XIII. Frequency Response
- XIV. The Laplace Transform
- XV. Two-port Networks

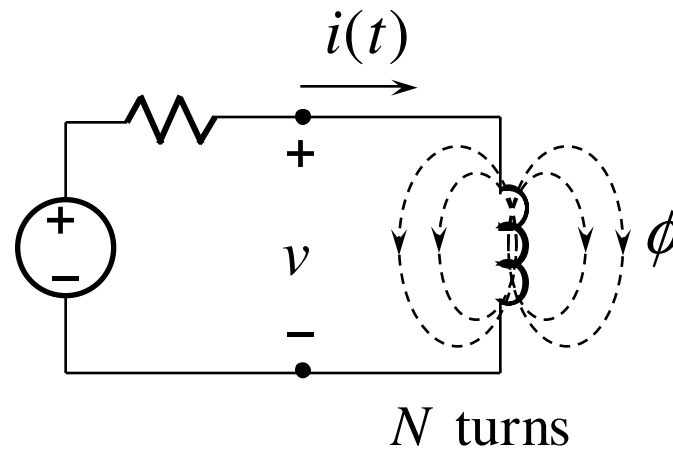


Magnetically Coupled Circuits

1. Mutual Inductance
2. Dot Convention
3. Analysis of Magnetically Coupled Circuits
4. Energy in a Coupled Circuit
5. Transformers



Mutual Inductance (1)

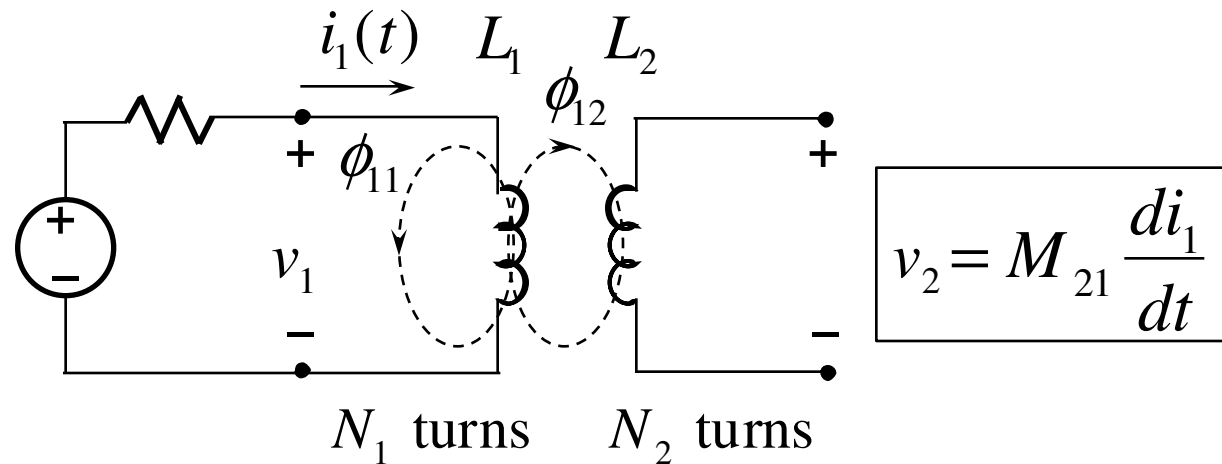


Faraday's law:
$$v = N \frac{d\phi}{dt} = N \frac{d\phi}{di} \frac{di}{dt} = L \frac{di}{dt}$$

$$L = N \frac{d\phi}{di}$$



Mutual Inductance (2)



$$\phi_1 = \phi_{11} + \phi_{12}$$

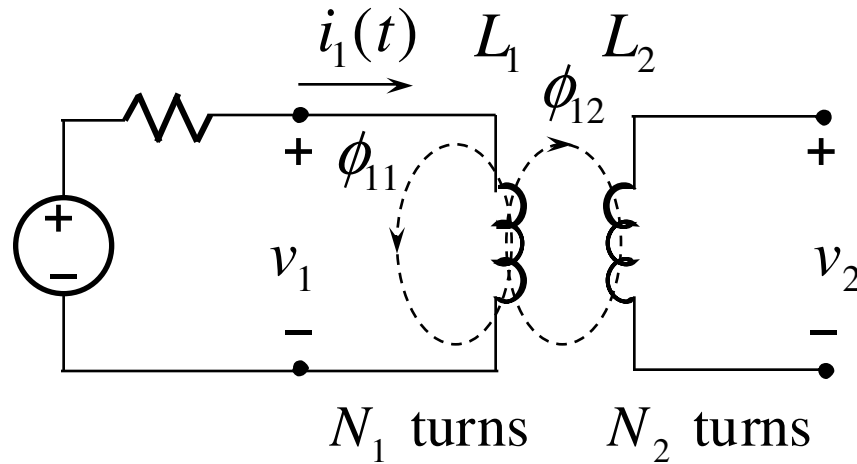
$$v_1 = N_1 \frac{d\phi_1}{dt}$$

$$= N_1 \frac{d\phi_1}{di_1} \frac{di_1}{dt} = L_1 \frac{di_1}{dt}$$

$$v_2 = N_2 \frac{d\phi_{12}}{dt}$$

$$= N_2 \frac{d\phi_{12}}{di_1} \frac{di_1}{dt} = M_{21} \frac{di_1}{dt}$$

Mutual Inductance (3)



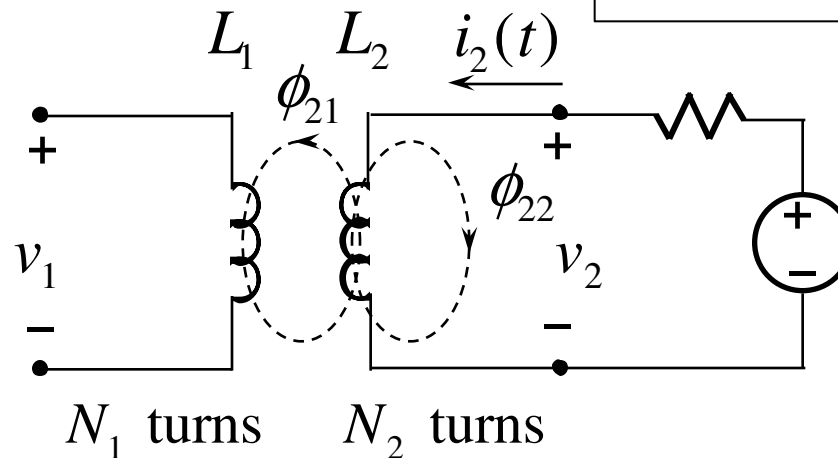
$$v_1 = L_1 \frac{di_1}{dt}$$

$$v_2 = M_{21} \frac{di_1}{dt}$$

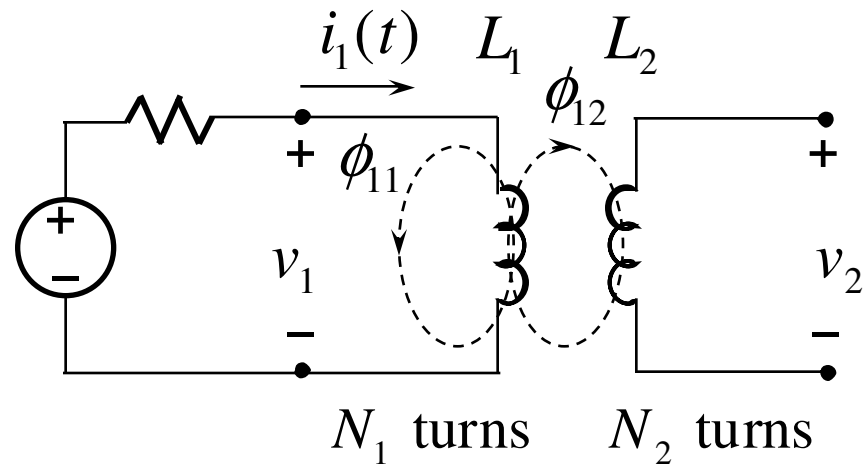
$$M_{12} = M_{21} = M = k\sqrt{L_1 L_2}$$

$$v_1 = M_{12} \frac{di_2}{dt}$$

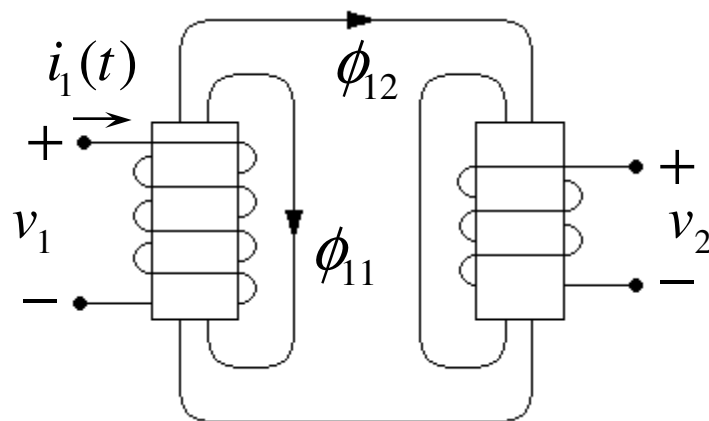
$$v_2 = L_2 \frac{di_2}{dt}$$



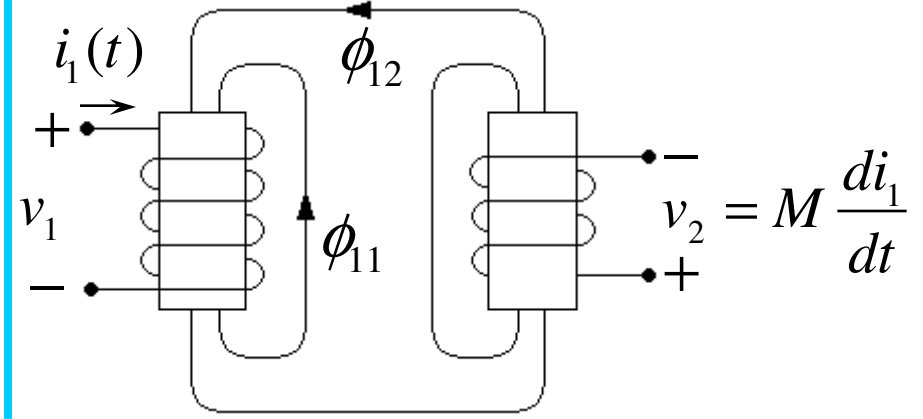
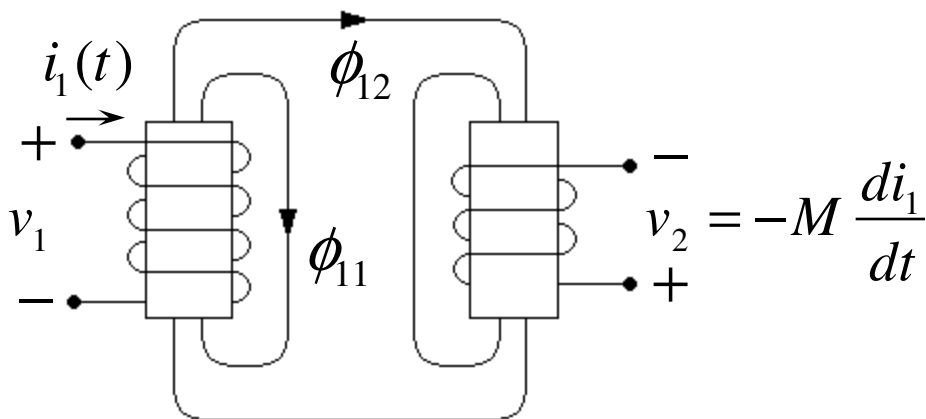
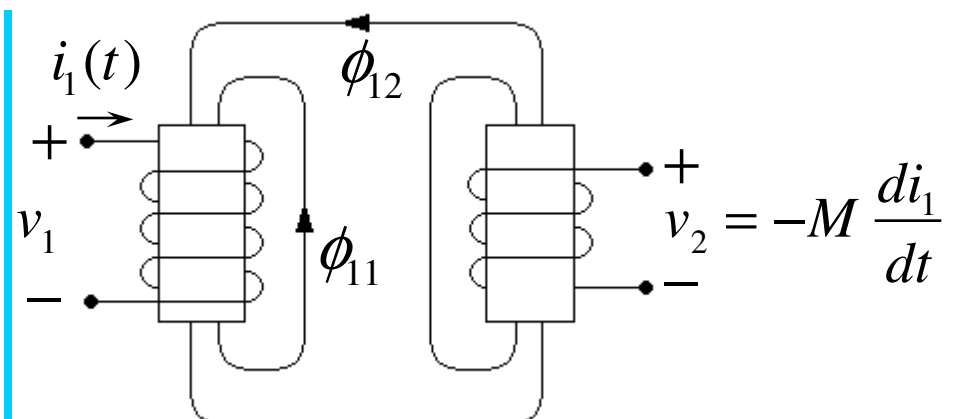
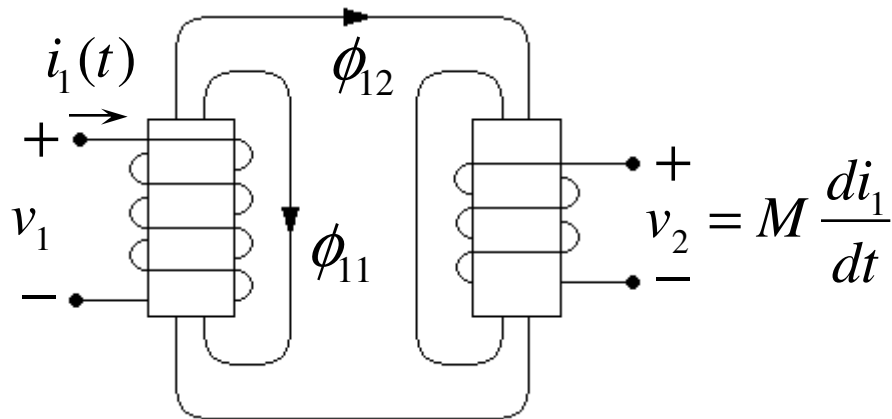
Mutual Inductance (4)



$$v_2 = M_{21} \frac{di_1}{dt}$$



Mutual Inductance (5)



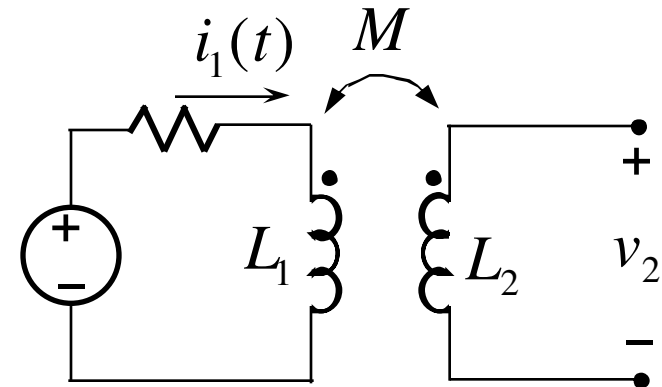
Magnetically Coupled Circuits

1. Mutual Inductance
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Dot Convention (1)

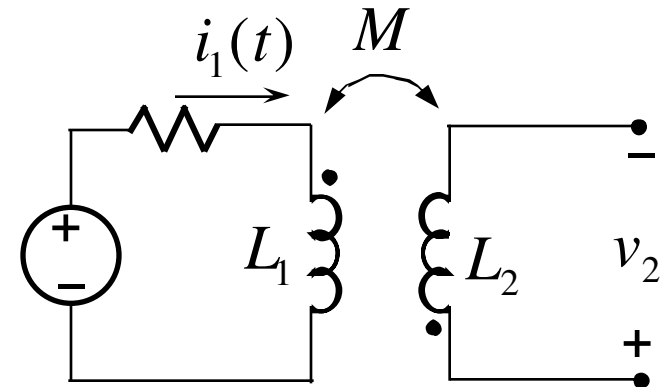
- If a current **enters** a dotted terminal of one coil, it induces a **positive** voltage at the dotted terminal of the second coil
- If a current **leaves** a dotted terminal of one coil, it induces a **negative** voltage at the dotted terminal of the second coil



$$v_2 = M \frac{di_1}{dt}$$

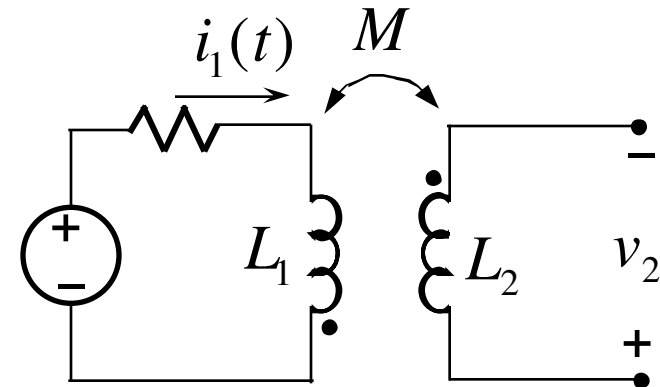
Dot Convention (2)

- If a current **enters** a dotted terminal of one coil, it induces a **positive** voltage at the dotted terminal of the second coil
- If a current **leaves** a dotted terminal of one coil, it induces a **negative** voltage at the dotted terminal of the second coil



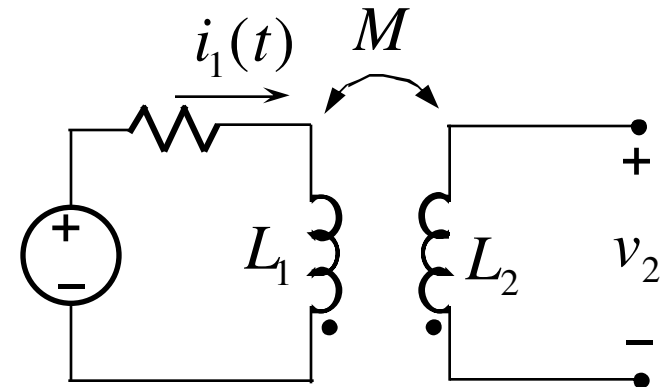
Dot Convention (3)

- If a current **enters** a dotted terminal of one coil, it induces a **positive** voltage at the dotted terminal of the second coil
- If a current **leaves** a dotted terminal of one coil, it induces a **negative** voltage at the dotted terminal of the second coil

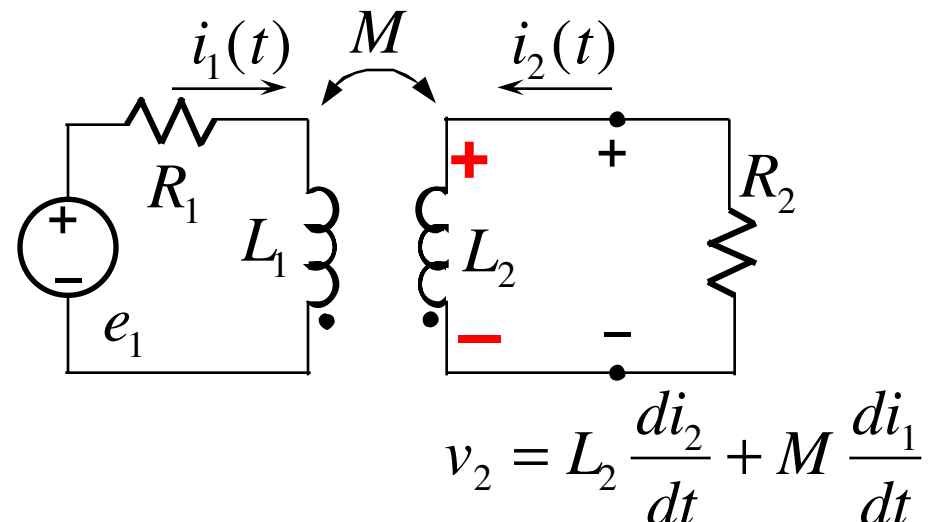
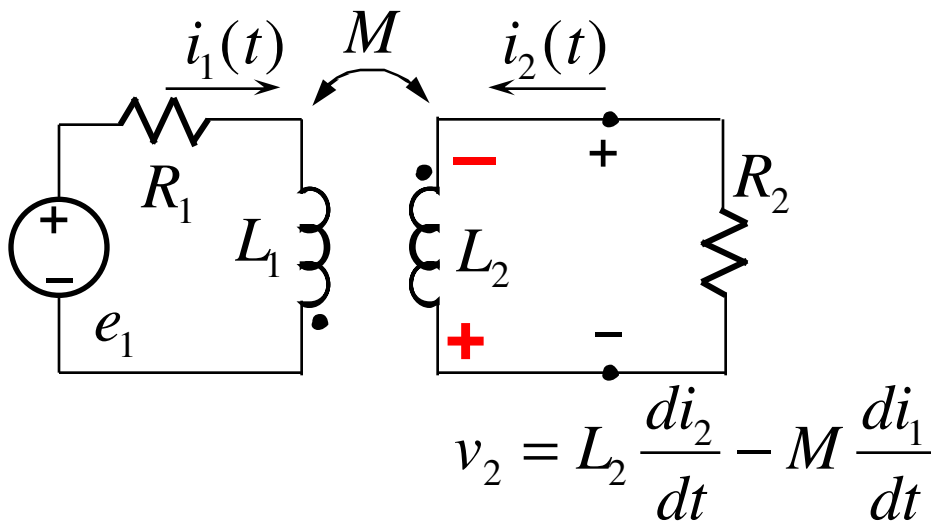
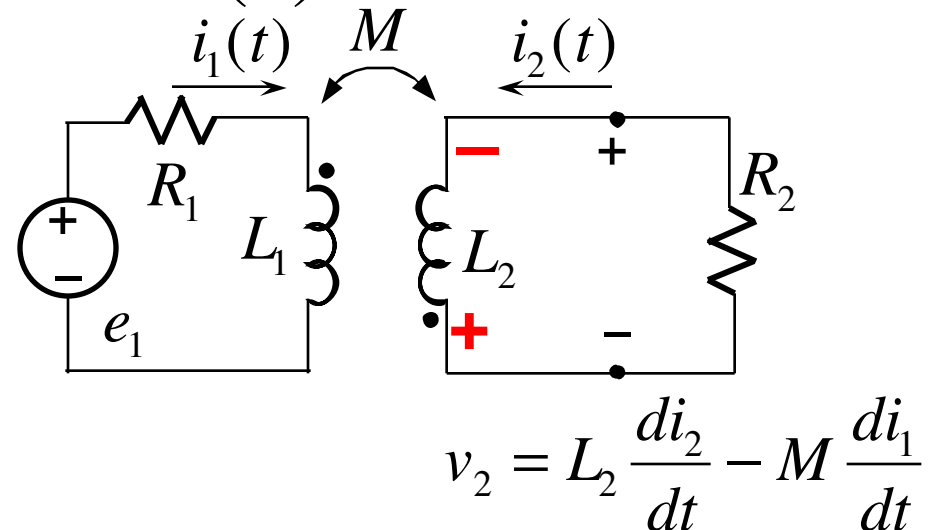
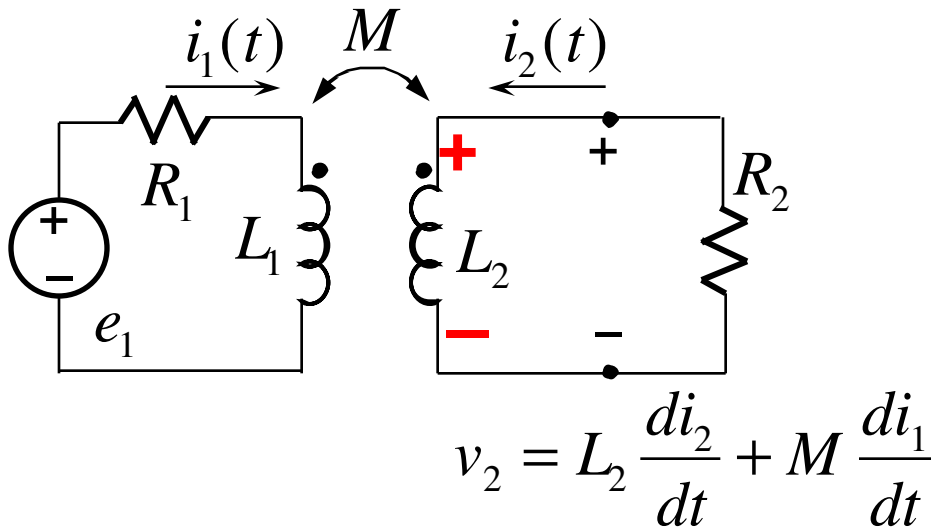


Dot Convention (4)

- If a current **enters** a dotted terminal of one coil, it induces a **positive** voltage at the dotted terminal of the second coil
- If a current **leaves** a dotted terminal of one coil, it induces a **negative** voltage at the dotted terminal of the second coil



Dot Convention (5)

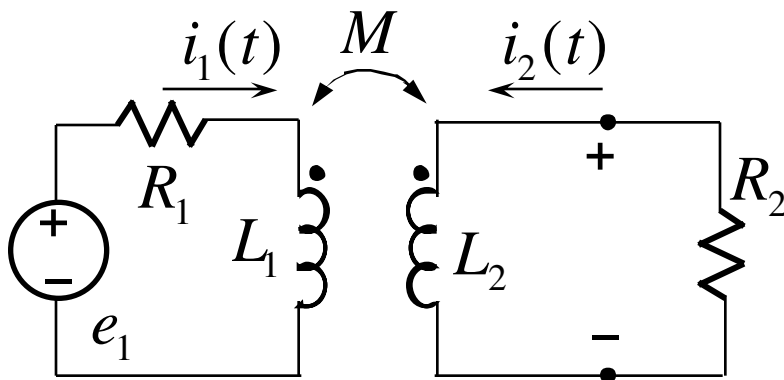


Magnetically Coupled Circuits

1. Mutual Inductance
2. Dot Convention
- 3. Analysis of Magnetically Coupled Circuits**
 - a) MCC in Phasor Domain**
 - b) Branch Current Method**
 - c) Mesh Current Method**
 - d) Equivalent Subcircuits**
4. Energy in a Coupled Circuit
5. Transformers

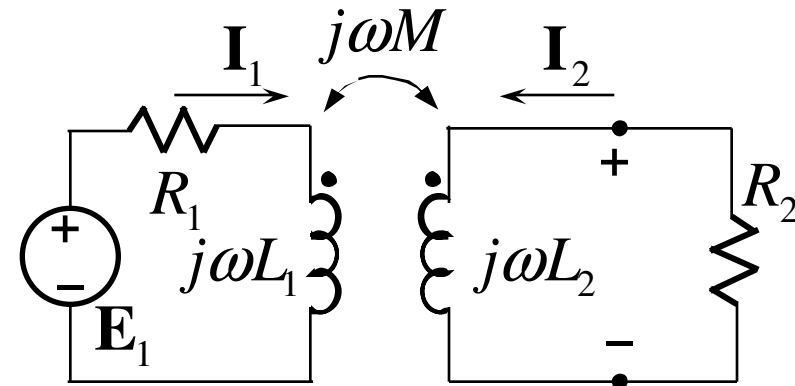


MCC in Phasor Domain (1)



$$v_1 = L_1 \frac{di_1}{dt} + M \frac{di_2}{dt}$$

$$v_2 = L_2 \frac{di_2}{dt} + M \frac{di_1}{dt}$$



$$\rightarrow \mathbf{V}_1 = j\omega L_1 \mathbf{I}_1 + j\omega M \mathbf{I}_2$$

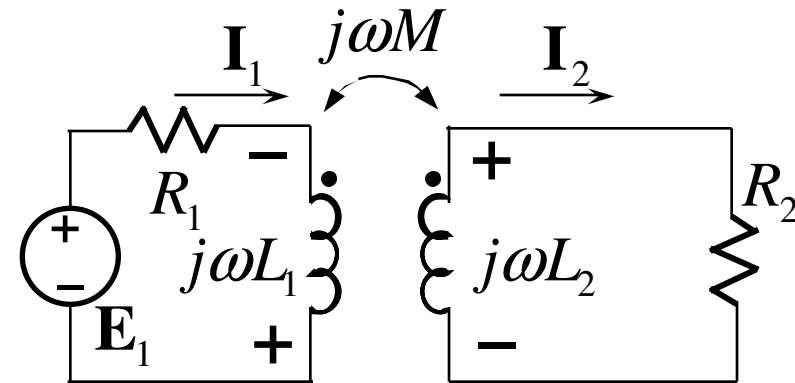
$$\rightarrow \mathbf{V}_2 = j\omega L_2 \mathbf{I}_2 + j\omega M \mathbf{I}_1$$

$$\boxed{\mathbf{Z}_M = j\omega M}$$

MCC in Phasor Domain (2)

Ex. 1

$\mathbf{E}_1 = 100 \angle 0^\circ \text{ V}; \omega = 100 \text{ rad/s};$
 $L_1 = 0.2 \text{ H}; L_2 = 0.3 \text{ H}; M = 0.1 \text{ H};$
 $R_1 = 30 \Omega; R_2 = 40 \Omega; \text{ Find currents?}$



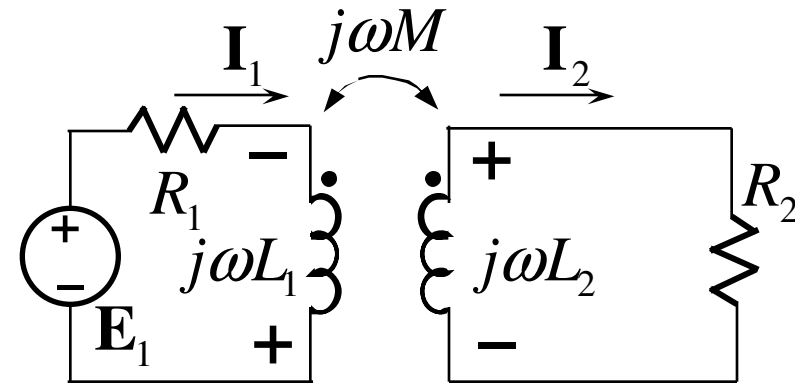
$$\left. \begin{aligned} \mathbf{V}_{M1} &= j\omega M \mathbf{I}_2 \\ \mathbf{V}_{1L} &= j\omega L_1 \mathbf{I}_1 \\ \mathbf{V}_{R1} + \mathbf{V}_{1L} - \mathbf{V}_{1M} &= \mathbf{E}_1 \end{aligned} \right\} \rightarrow R_1 \mathbf{I}_1 + j\omega L_1 \mathbf{I}_1 - j\omega M \mathbf{I}_2 = \mathbf{E}_1$$

$$\left. \begin{aligned} \mathbf{V}_{M2} &= j\omega M \mathbf{I}_1 \\ \mathbf{V}_{2L} &= j\omega L_2 \mathbf{I}_2 \\ \mathbf{V}_{R2} + \mathbf{V}_{2L} - \mathbf{V}_{2M} &= 0 \end{aligned} \right\} \rightarrow R_2 \mathbf{I}_2 + j\omega L_2 \mathbf{I}_2 - j\omega M \mathbf{I}_1 = 0$$

MCC in Phasor Domain (3)

Ex. 1

$\mathbf{E}_1 = 100 \angle 0^\circ \text{ V}; \omega = 100 \text{ rad/s};$
 $L_1 = 0.2 \text{ H}; L_2 = 0.3 \text{ H}; M = 0.1 \text{ H};$
 $R_1 = 30 \Omega; R_2 = 40 \Omega; \text{ Find currents?}$



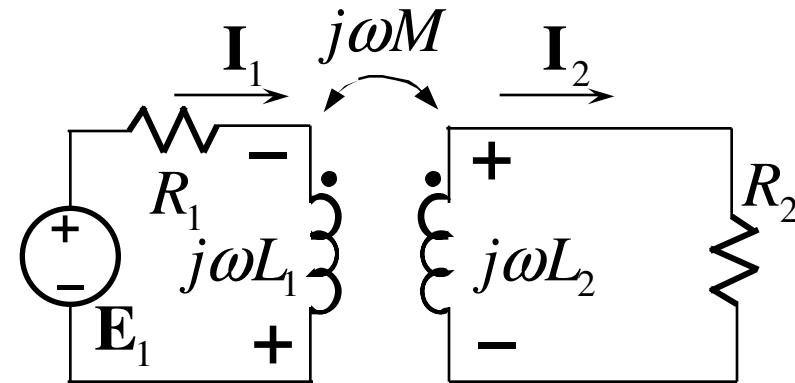
$$\left. \begin{aligned} R_1 \mathbf{I}_1 + j\omega L_1 \mathbf{I}_1 - j\omega M \mathbf{I}_2 &= \mathbf{E}_1 \\ R_2 \mathbf{I}_2 + j\omega L_2 \mathbf{I}_2 - j\omega M \mathbf{I}_1 &= 0 \end{aligned} \right\}$$

$$\rightarrow \begin{cases} 30\mathbf{I}_1 + j100 \times 0.2\mathbf{I}_1 - j100 \times 0.1\mathbf{I}_2 = 100 \angle 0^\circ \\ 40\mathbf{I}_2 + j100 \times 0.3\mathbf{I}_2 - j100 \times 0.1\mathbf{I}_1 = 0 \end{cases} \rightarrow \begin{cases} \mathbf{I}_1 = 2.34 - j1.39 \text{ A} \\ \mathbf{I}_2 = 0.50 + j0.21 \text{ A} \end{cases}$$

MCC in Phasor Domain (4)

Ex. 1

$\mathbf{E}_1 = 100 \angle 0^\circ \text{ V}; \omega = 100 \text{ rad/s};$
 $L_1 = 0.2 \text{ H}; L_2 = 0.3 \text{ H}; M = 0.1 \text{ H};$
 $R_1 = 30 \Omega; R_2 = 40 \Omega;$ Find currents?



$$\mathbf{V}_{M1} = j\omega M \mathbf{I}_2; \mathbf{V}_{M2} = j\omega M \mathbf{I}_1$$

$$\begin{cases} \mathbf{V}_{R1} + \mathbf{V}_{1L} - \mathbf{V}_{1M} = \mathbf{E}_1 \\ \mathbf{V}_{R2} + \mathbf{V}_{2L} - \mathbf{V}_{2M} = 0 \end{cases}$$

$$\rightarrow \begin{cases} R_1 \mathbf{I}_1 + j\omega L_1 \mathbf{I}_1 - j\omega M \mathbf{I}_2 = \mathbf{E}_1 \\ R_2 \mathbf{I}_2 + j\omega L_2 \mathbf{I}_2 - j\omega M \mathbf{I}_1 = 0 \end{cases}$$

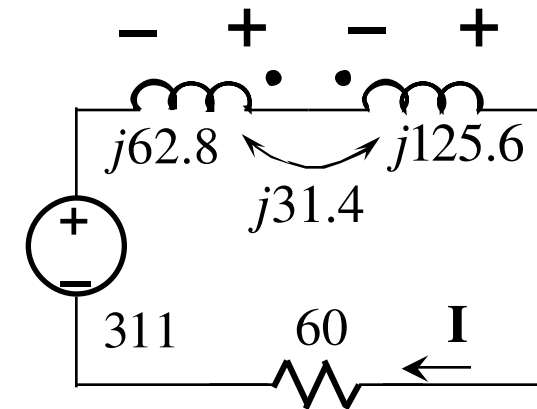
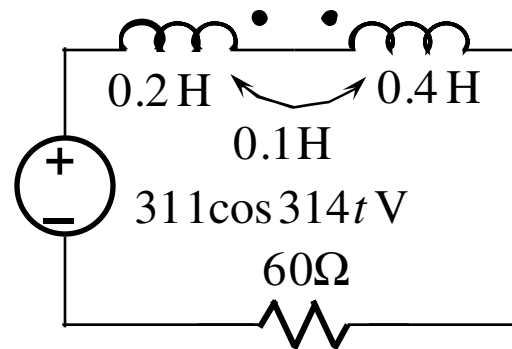
$$\rightarrow \begin{cases} \mathbf{I}_1 = 2.34 - j1.39 \text{ A} \\ \mathbf{I}_2 = 0.50 + j0.21 \text{ A} \end{cases}$$

1. Write voltages of mutual inductance
2. Assign signs at dotted terminals (using dot convention)
3. Write KVL equations
4. Write the set of equations & solve for it

MCC in Phasor Domain (5)

Ex. 2

Find current?



$$j62.8\mathbf{I} - j31.4\mathbf{I} + j125.6\mathbf{I} - j31.4\mathbf{I} + 60\mathbf{I} = 311$$

$$\rightarrow \mathbf{I} = 2.23 \angle -64.5^\circ \text{ A}$$

$$\rightarrow i = 2.23 \cos(314t - 64.5^\circ) \text{ A}$$

1. Write voltages of mutual inductance
2. Assign signs at dotted terminals (using dot convention)
3. Write KVL equations
4. Write the set of equations & solve for it

MCC in Phasor Domain (6)

Ex. 3

Find current?

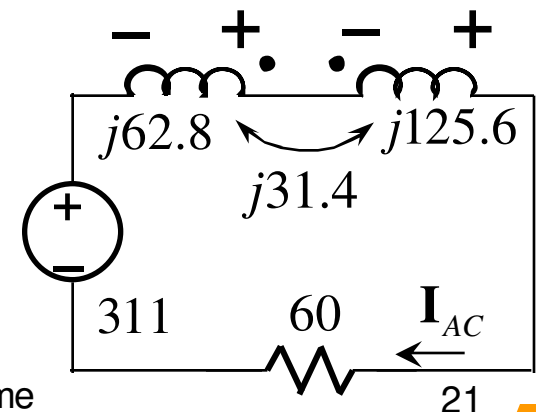
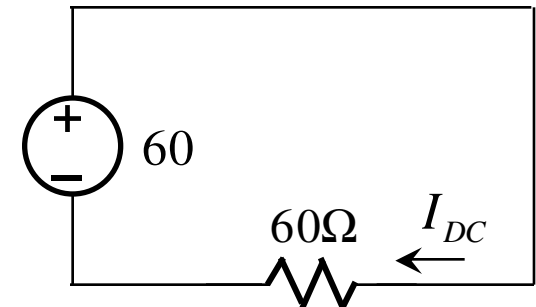
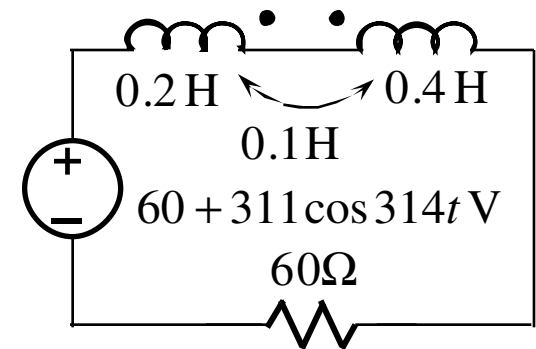
$$I_{DC} = \frac{60}{60} = 1 \text{ A}$$

$$(j62.8 - j31.4 + j125.6 - j31.4 + 60) \mathbf{I}_{AC} = 311$$

$$\rightarrow \mathbf{I}_{AC} = 2.23 \angle -64.5^\circ \text{ A}$$

$$\rightarrow i_{AC} = 2.23 \cos(314t - 64.5^\circ) \text{ A}$$

$$\rightarrow i = I_{DC} + i_{AC} = \boxed{1 + 2.23 \cos(314t - 64.5^\circ) \text{ A}}$$



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Branch Current Method (1)

Ex. 1

$$V_{1M} = Z_M I_2$$

$$V_{2M} = Z_M I_1$$

$$I_1 + I_2 - I_3 = 0$$

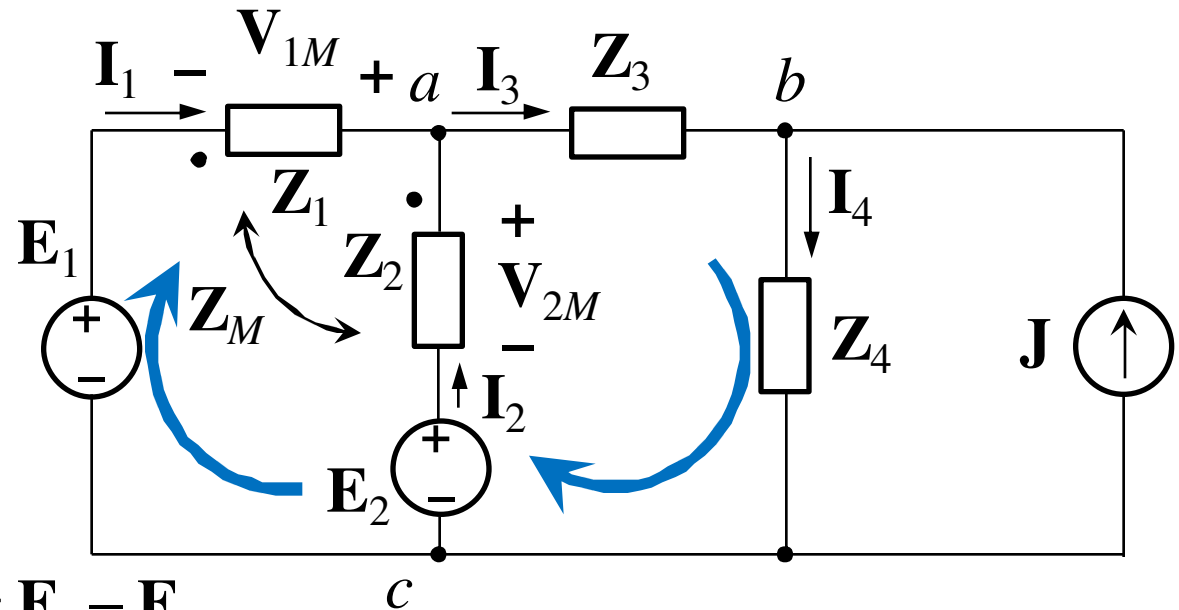
$$I_3 - I_4 + J = 0$$

$$V_{Z1} - V_{1M} - V_{Z2} + V_{2M} = E_1 - E_2$$

$$\begin{aligned} \rightarrow Z_1 I_1 - Z_M I_2 - Z_2 I_2 + Z_M I_1 &= \\ &= E_1 - E_2 \end{aligned}$$

$$V_{Z2} - V_{2M} + V_3 + V_4 = E_2$$

$$\rightarrow Z_2 I_2 - Z_M I_1 + Z_3 I_3 + Z_4 I_4 = E_2$$



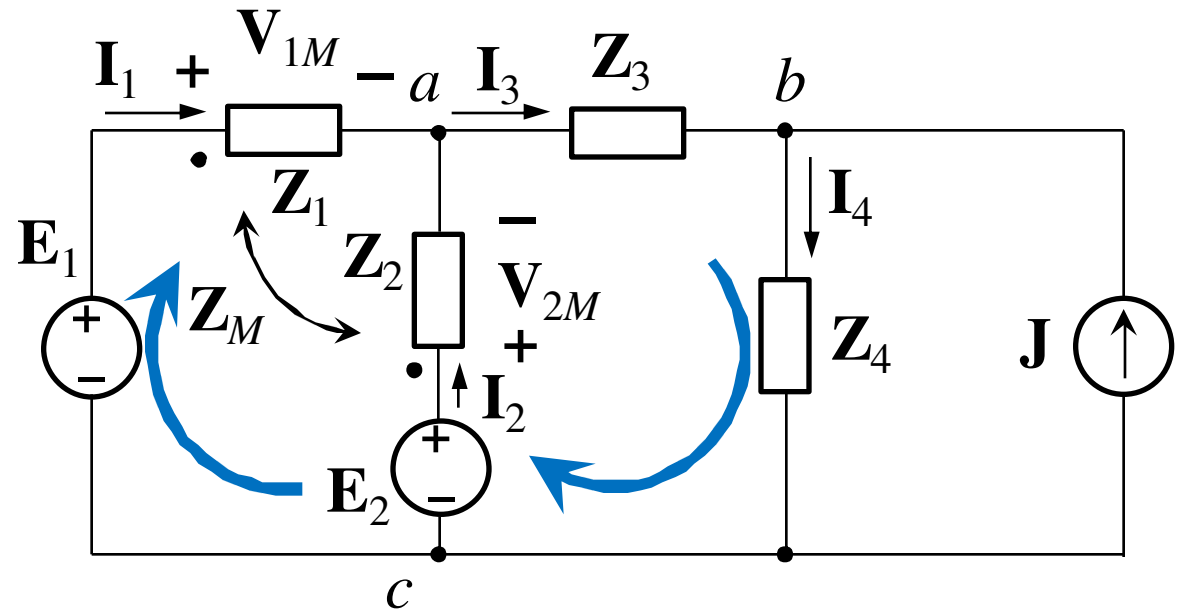
1. Write voltages of mutual inductance
2. Assign signs at dotted terminals (using dot convention)
3. Write KVL equations
4. Write the set of equations & solve for it

Branch Current Method (2)

Ex. 2

$$V_{1M} = Z_M I_2$$

$$V_{2M} = Z_M I_1$$



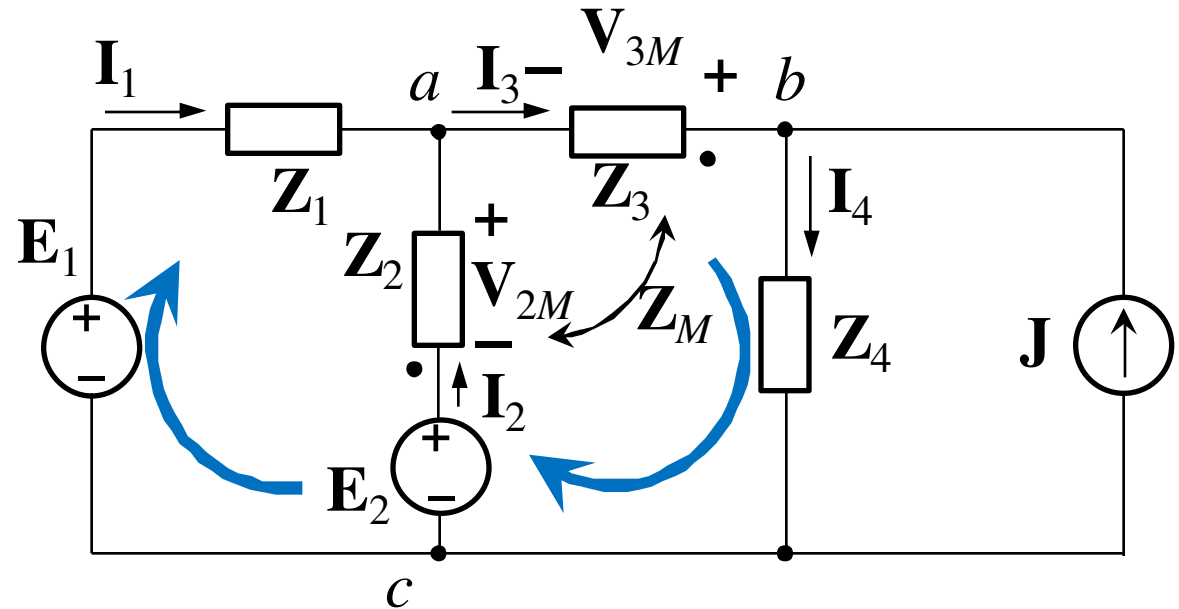
$$\begin{cases} I_1 + I_2 - I_3 = 0 \\ I_3 - I_4 + J = 0 \\ V_{Z1} + V_{1M} - V_{Z2} - V_{2M} = E_1 - E_2 \rightarrow Z_1 I_1 + Z_M I_2 - Z_2 I_2 - Z_M I_1 = E_1 - E_2 \\ V_{Z2} + V_{2M} + V_3 + V_4 = E_2 \rightarrow Z_2 I_2 + Z_M I_1 + Z_3 I_3 + Z_4 I_4 = E_2 \end{cases}$$

Branch Current Method (3)

Ex. 3

$$V_{2M} = Z_M I_3$$

$$V_{3M} = Z_M I_2$$



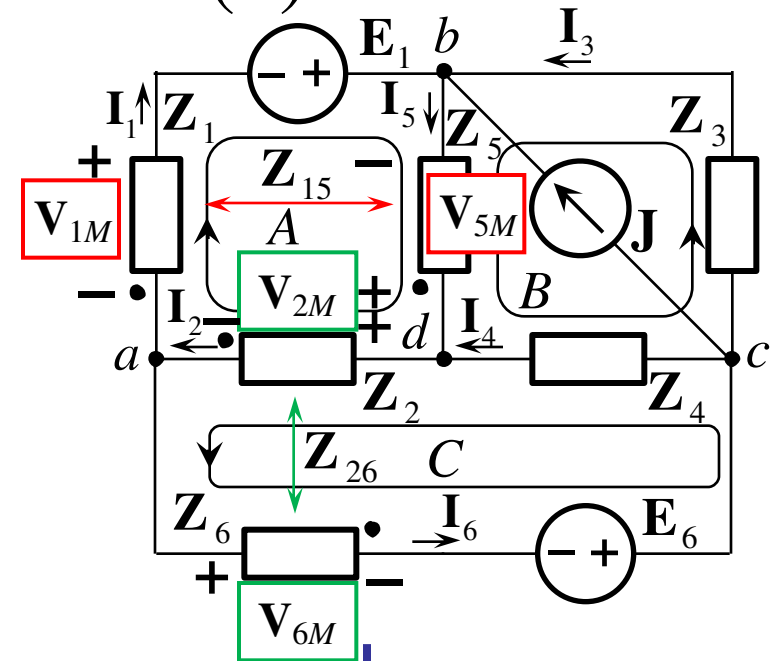
$$\begin{cases} I_1 + I_2 - I_3 = 0 \\ I_3 - I_4 + J = 0 \\ Z_1 I_1 - Z_2 I_2 + Z_M I_3 = E_1 - E_2 \\ Z_2 I_2 - Z_M I_3 + Z_3 I_3 - Z_M I_2 + Z_4 I_4 = E_2 \end{cases}$$

Branch Current Method (4)

Ex. 4

$$n_{KCL} = 4 - 1 = 3$$

$$n_{KVL} = 6 - 4 + 1 = 3$$



$$a : -I_1 + I_2 - I_6 = 0$$

$$b : I_1 + I_3 - I_5 + J = 0$$

$$c : -I_3 - I_4 + I_6 - J = 0$$

$$A : Z_1 I_1 - Z_{15} I_5 + Z_5 I_5 - Z_{15} I_1 + Z_2 I_2 + Z_{26} I_6 = E_1$$

$$B : Z_3 I_3 + Z_5 I_5 - Z_{15} I_1 - Z_4 I_4 = 0$$

$$C : Z_2 I_2 + Z_{26} I_6 + Z_6 I_6 + Z_{26} I_2 + Z_4 I_4 = E_6$$

$$V_{1M} = Z_{15} I_5$$

$$V_{5M} = Z_{15} I_1$$

$$V_{2M} = Z_{26} I_6$$

$$V_{6M} = Z_{26} I_2$$

Branch Current Method (5)

Ex. 5

$$b: I_2 + I_3 + I_6 = 0$$

$$c: I_4 - I_3 + I_5 = 0$$

$$d: -I_1 - I_4 - J = 0$$

$$A: Z_1 I_1 + Z_{12} I_2 - Z_{15} I_5 + Z_5 I_5 - Z_{15} I_1 - Z_4 I_4 + Z_{34} I_3 = E_1$$

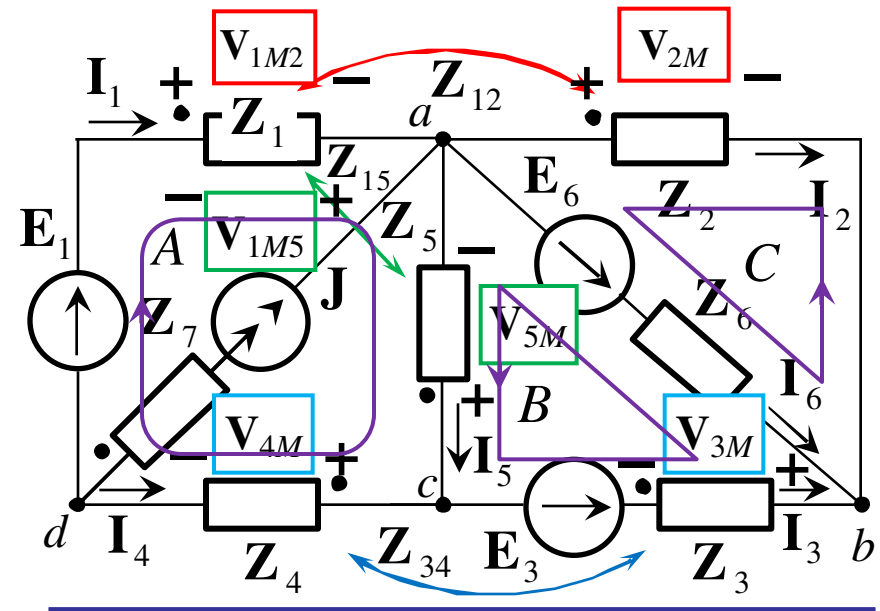
$$B: Z_5 I_5 - Z_{15} I_1 + Z_3 I_3 - Z_{34} I_4 - Z_6 I_6 = E_3 - E_6$$

$$C: Z_6 I_6 - Z_2 I_2 - Z_{12} I_1 = E_6$$

$$V_{1M2} = Z_{12} I_2; \quad V_{1M5} = Z_{15} I_5$$

$$V_{2M} = Z_{12} I_1; \quad V_{5M} = Z_{15} I_1$$

$$V_{3M} = Z_{34} I_4; \quad V_{4M} = Z_{34} I_3$$



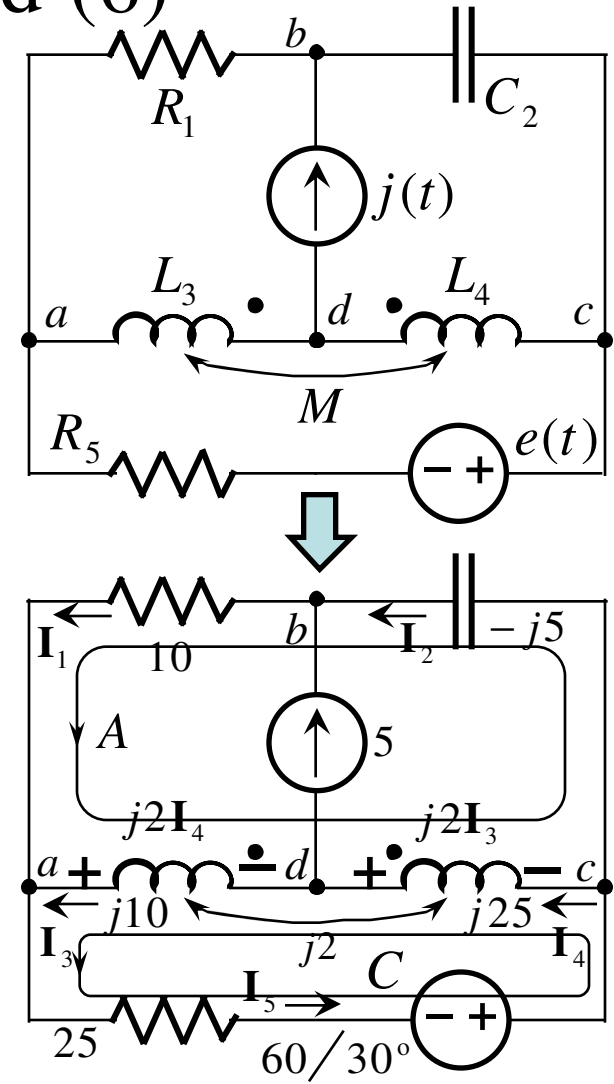
Ex. 6

Branch Current Method (6)

$R_1 = 10 \, \Omega$, $R_5 = 25 \, \Omega$, $L_3 = 0.2 \, \text{H}$, $L_4 = 0.5 \, \text{H}$, $C_2 = 4 \, \text{mF}$,
 $M = 0.04 \, \text{H}$, $j(t) = 5\sin(50t) \, \text{A}$, $e(t) = 60\sin(50t + 30^\circ) \, \text{V}$.

$$\begin{cases} a : \mathbf{I}_1 + \mathbf{I}_3 - \mathbf{I}_5 = 0 \\ b : -\mathbf{I}_1 + \mathbf{I}_2 + 5 = 0 \\ c : -\mathbf{I}_2 - \mathbf{I}_4 + \mathbf{I}_5 = 0 \\ A : 10\mathbf{I}_1 - j10\mathbf{I}_3 + j2\mathbf{I}_4 - j25\mathbf{I}_4 + j2\mathbf{I}_3 - j5\mathbf{I}_2 = 0 \\ C : 25\mathbf{I}_5 + j25\mathbf{I}_4 - j2\mathbf{I}_3 + j10\mathbf{I}_3 - 2j_2\mathbf{I}_4 = 60\angle 30^\circ \end{cases}$$

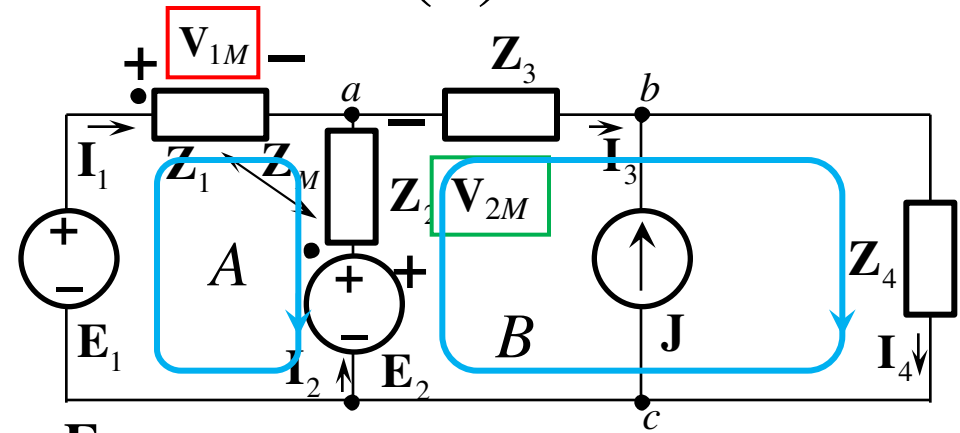
$$\rightarrow \begin{cases} \mathbf{I}_1 = 3.71\angle 25.17^\circ \, \text{A} \\ \mathbf{I}_2 = 2.28\angle 136.18^\circ \, \text{A} \\ \mathbf{I}_3 = 3.23\angle -155.51^\circ \, \text{A} \\ \mathbf{I}_4 = 2.46\angle -32.93^\circ \, \text{A} \\ \mathbf{I}_5 = 0.48\angle 29.75^\circ \, \text{A} \end{cases} \rightarrow \begin{cases} i_1 = 3.71\sin(50t + 25.17^\circ) \, \text{A} \\ i_2 = 2.28\sin(50t + 136.18^\circ) \, \text{A} \\ i_3 = 3.23\sin(50t - 155.51^\circ) \, \text{A} \\ i_4 = 2.46\sin(50t - 32.93^\circ) \, \text{A} \\ i_5 = 0.48\sin(50t + 29.75^\circ) \, \text{A} \end{cases}$$



Ex. 7

Branch Current Method (7)

$Z_1 = 10 + j15\Omega$; $Z_2 = 20 + j10\Omega$; $Z_M = j2\Omega$;
 $Z_3 = -j20\Omega$; $Z_4 = 25\Omega$; $E_1 = 100\text{ V}$;
 $E_2 = 150\angle 30^\circ\text{ V}$; $J = 5\angle 45^\circ\text{ A}$.



$$V_{1M} = Z_M I_2$$

$$V_{2M} = Z_M I_1$$

$$\begin{cases}
 a: I_1 + I_2 - I_3 = 0 \\
 b: I_3 + J - I_4 = 0 \\
 A: Z_1 I_1 + Z_M I_2 - Z_2 I_2 - Z_M I_1 = E_1 - E_2 \\
 B: Z_2 I_2 + Z_M I_1 + Z_3 I_3 + Z_4 I_4 = E_2
 \end{cases}$$

$$\rightarrow \begin{cases}
 I_1 + I_2 - I_3 = 0 \\
 I_3 - I_4 = -5\angle 45^\circ \\
 (10 + j15 - j2)I_1 + [j2 - (20 + j10)]I_2 = 100 - 150\angle 30^\circ \\
 j2I_1 + (20 + j10)I_2 - j20I_3 + 25I_4 = 150\angle 30^\circ
 \end{cases}$$

$$\rightarrow \begin{cases}
 I_1 = -1,49 - j2,06\text{ A}; & I_2 = 2,40 + j0,79\text{ A} \\
 I_3 = 0,91 - j1,28\text{ A}; & I_4 = 4,44 + j2,26\text{ A}
 \end{cases}$$

Magnetically Coupled Circuits

1. Mutual Inductance
2. Dot Convention
- 3. Analysis of Magnetically Coupled Circuits**
 - a) MCC in Phasor Domain
 - b) Branch Current Method
 - c) Mesh Current Method**
 - d) Equivalent Subcircuits
4. Energy in a Coupled Circuit
5. Transformers

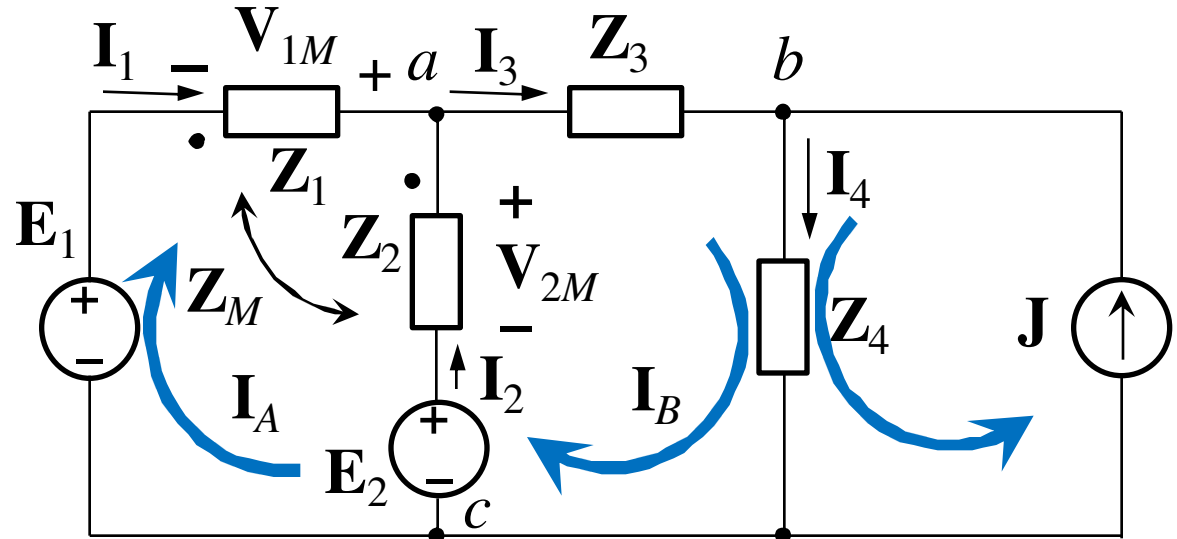


Mesh Current Method (1)

Ex. 1

$$V_{1M} = Z_M I_2$$

$$V_{2M} = Z_M I_1$$



$$\left\{ \begin{array}{l} Z_1 I_1 - Z_M I_2 - Z_2 I_2 + Z_M I_1 = E_1 - E_2 \\ Z_2 I_2 - Z_M I_1 + Z_3 I_3 + Z_4 I_4 = E_2 \\ I_1 = I_A, I_2 = I_B - I_A, I_3 = I_B, I_4 = I_B + J \end{array} \right\}$$

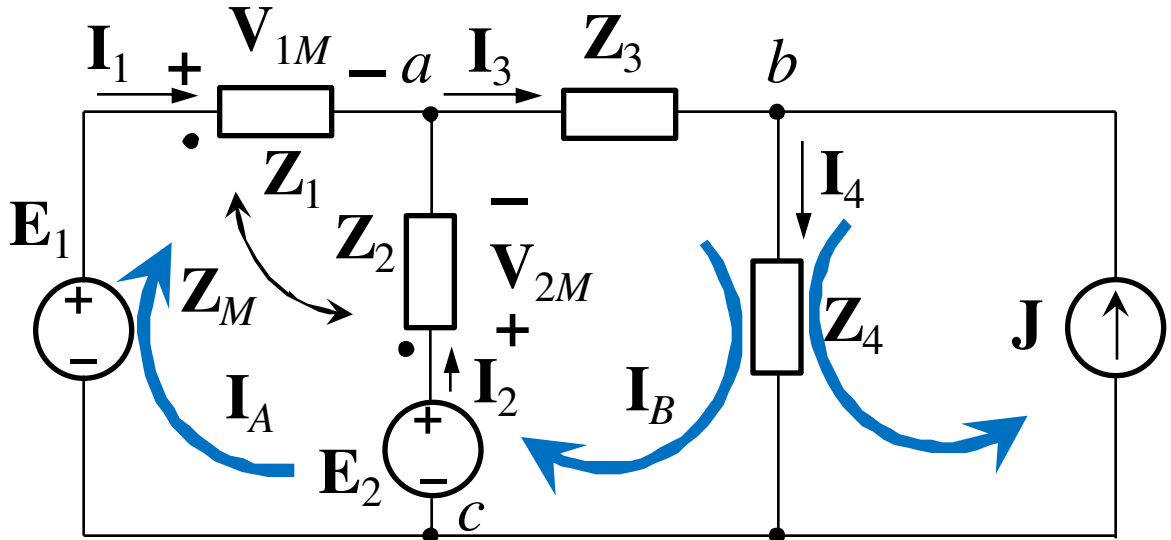
$$\rightarrow \left\{ \begin{array}{l} Z_1 I_A - Z_M (I_B - I_A) - Z_2 (I_B - I_A) + Z_M I_A = E_1 - E_2 \\ Z_2 (I_B - I_A) - Z_M I_A + Z_3 I_B + Z_4 (I_B + J) = E_2 \end{array} \right.$$

Mesh Current Method (2)

Ex. 2

$$V_{1M} = Z_M I_2$$

$$V_{2M} = Z_M I_1$$



$$\left\{ \begin{array}{l} Z_1 I_1 + Z_M I_2 - Z_2 I_2 - Z_M I_1 = E_1 - E_2 \\ Z_2 I_2 + Z_M I_1 + Z_3 I_3 + Z_4 I_4 = E_2 \end{array} \right\}$$

$$I_1 = I_A, I_2 = I_B - I_A, I_3 = I_B, I_4 = I_B + J$$

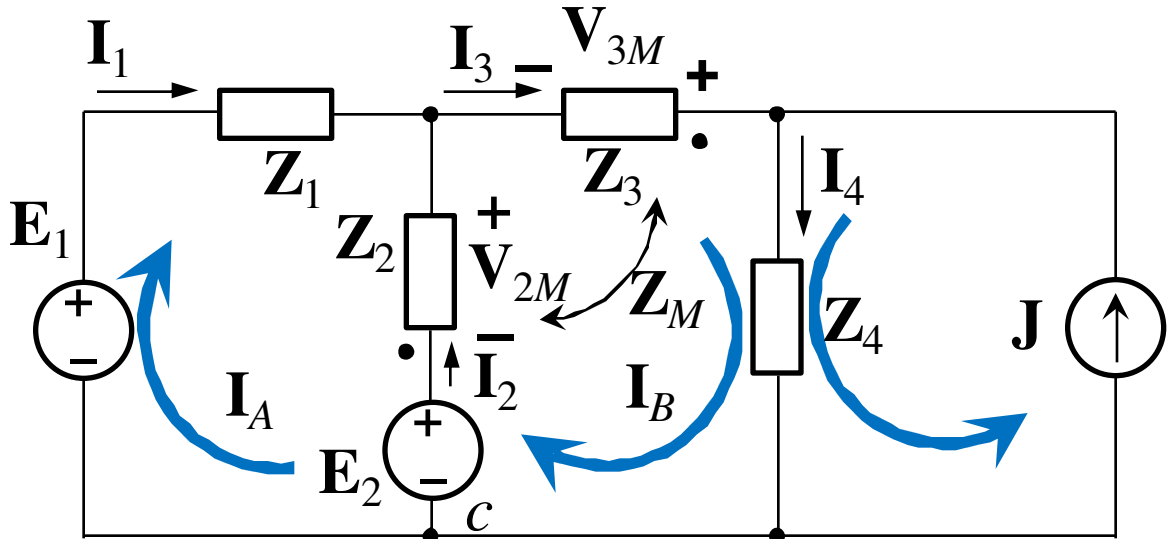
$$\rightarrow \left\{ \begin{array}{l} Z_1 I_A + Z_M (I_B - I_A) - Z_2 (I_B - I_A) + Z_M I_A = E_1 - E_2 \\ Z_2 (I_B - I_A) + Z_M I_A + Z_3 I_B + Z_4 (I_B + J) = E_2 \end{array} \right.$$

Mesh Current Method (3)

Ex. 3

$$V_{2M} = Z_M I_3$$

$$V_{3M} = Z_M I_2$$



$$\left\{ \begin{array}{l} Z_1 I_1 - Z_2 I_2 + Z_M I_3 = E_1 - E_2 \\ Z_2 I_2 - Z_M I_3 + Z_3 I_3 - Z_M I_2 + Z_4 I_4 = E_2 \\ I_1 = I_A, I_2 = I_B - I_A, I_3 = I_B, I_4 = I_B + J \end{array} \right\}$$

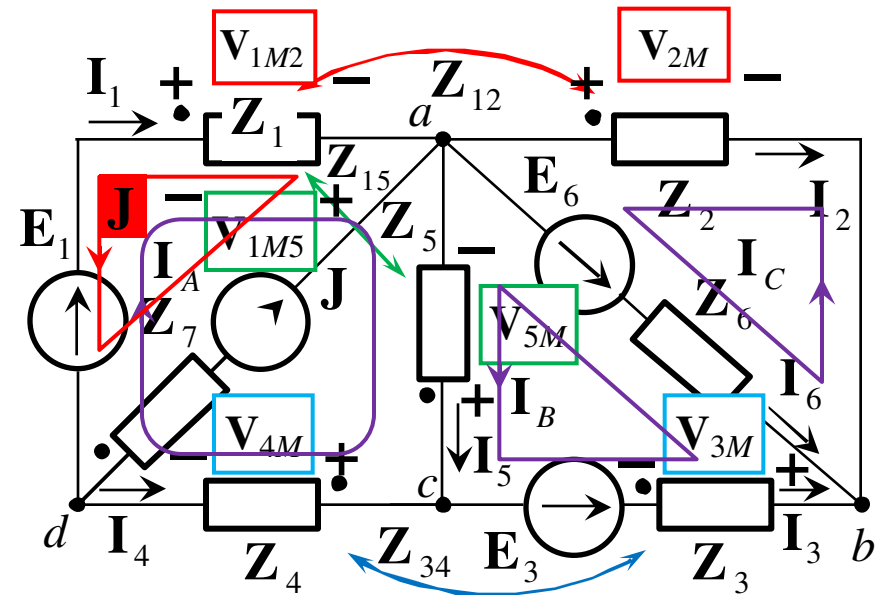
$$\rightarrow \left\{ \begin{array}{l} Z_1 I_A - Z_2 (I_B - I_A) + Z_M I_B = E_1 - E_2 \\ Z_2 (I_B - I_A) - Z_M I_B + Z_3 I_B - Z_M (I_B - I_A) + Z_4 (I_B + J) = E_2 \end{array} \right.$$

Mesh Current Method (5)

Ex. 5

$$\left\{ \begin{array}{l} A: Z_1 I_1 + Z_{12} I_2 - Z_{15} I_5 + Z_5 I_5 \\ \quad - Z_{15} I_1 - Z_4 I_4 + Z_{34} I_3 = E_1 \\ B: Z_5 I_5 - Z_{15} I_1 + Z_3 I_3 - Z_{34} I_4 \\ \quad - Z_6 I_6 = E_3 - E_6 \\ C: Z_6 I_6 - Z_2 I_2 - Z_{12} I_1 = E_6 \\ I_1 = I_A - J, \quad I_2 = -I_C, \quad I_3 = I_B \\ I_4 = -I_A, \quad I_5 = I_A + I_B, \quad I_6 = I_C - I_B \end{array} \right.$$

$$\rightarrow \left\{ \begin{array}{l} Z_1 (I_A - J) + Z_{12} (-I_C) - Z_{15} (I_A + I_B) + Z_5 (I_A + I_B) - Z_{15} (I_A - J) - Z_4 (-I_A) + Z_{34} I_B = E_1 \\ Z_5 (I_A + I_B) - Z_{15} (I_A - J) + Z_3 I_B - Z_{34} (-I_A) - Z_6 (I_C - I_B) = E_3 - E_6 \\ Z_6 (I_C - I_B) - Z_2 (-I_C) - Z_{12} (I_A - J) = E_6 \end{array} \right.$$



Ex. 6

Mesh Current Method (6)

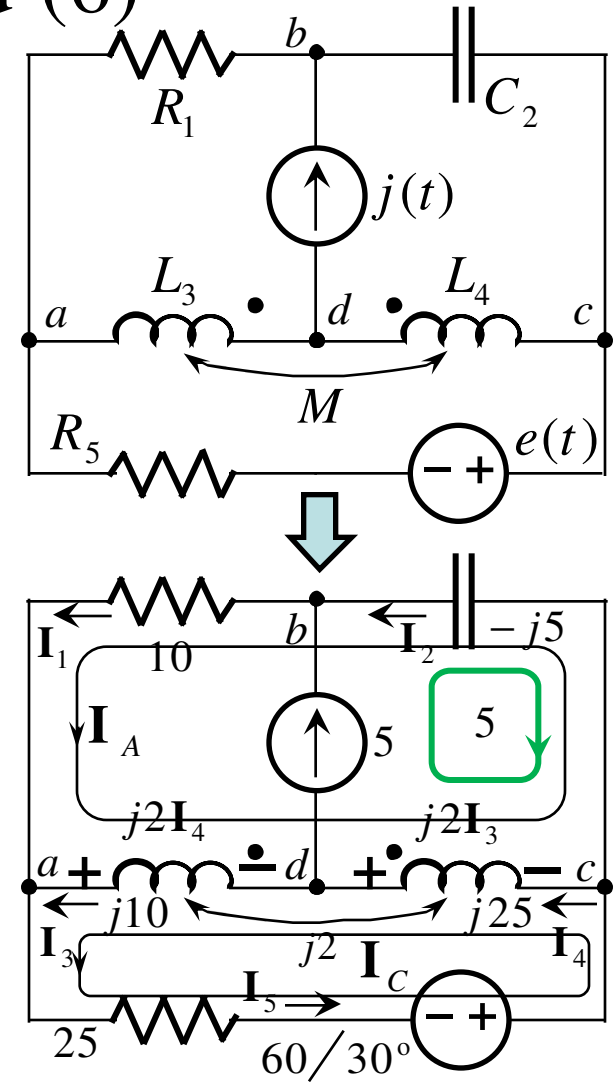
$R_1 = 10 \Omega$, $R_5 = 25 \Omega$, $L_3 = 0.2 \text{ H}$, $L_4 = 0.5 \text{ H}$, $C_2 = 4 \text{ mF}$,
 $M = 0.04 \text{ H}$, $j(t) = 5\sin(50t) \text{ A}$, $e(t) = 60\sin(50t + 30^\circ) \text{ V}$.

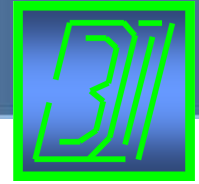
$$\begin{cases} A: 10\mathbf{I}_1 - j10\mathbf{I}_3 + j2\mathbf{I}_4 - j25\mathbf{I}_4 + j2\mathbf{I}_3 - j5\mathbf{I}_2 = 0 \\ C: 25\mathbf{I}_5 + j25\mathbf{I}_4 - j2\mathbf{I}_3 + j10\mathbf{I}_3 - 2j_2\mathbf{I}_4 = 60/30^\circ \\ \mathbf{I}_1 = \mathbf{I}_A, \mathbf{I}_2 = \mathbf{I}_A - 5, \mathbf{I}_3 = \mathbf{I}_C - \mathbf{I}_A, \mathbf{I}_4 = \mathbf{I}_C - \mathbf{I}_A + 5, \mathbf{I}_5 = \mathbf{I}_C \end{cases}$$

$$\rightarrow \begin{cases} (10 + j26)\mathbf{I}_A - j31\mathbf{I}_C = j90 \\ -j31\mathbf{I}_A + (25 + j31)\mathbf{I}_C = 51.96 - j85 \end{cases}$$

$$\rightarrow \begin{cases} \mathbf{I}_A = 3.36 + j1.58 \text{ A} \\ \mathbf{I}_C = 0.42 + j0.24 \text{ A} \end{cases}$$

$$\rightarrow \begin{cases} \mathbf{I}_1 = 3.71/25.17^\circ \text{ A} \\ \mathbf{I}_2 = 2.28/136.18^\circ \text{ A} \\ \mathbf{I}_3 = 3.23/-155.51^\circ \text{ A} \\ \mathbf{I}_4 = 2.46/-32.93^\circ \text{ A} \\ \mathbf{I}_5 = 0.48/29.75^\circ \text{ A} \end{cases} \rightarrow \begin{cases} i_1 = 3.71 \sin(50t + 25.17^\circ) \text{ A} \\ i_2 = 2.28 \sin(50t + 136.18^\circ) \text{ A} \\ i_3 = 3.23 \sin(50t - 155.51^\circ) \text{ A} \\ i_4 = 2.46 \sin(50t - 32.93^\circ) \text{ A} \\ i_5 = 0.48 \sin(50t + 29.75^\circ) \text{ A} \end{cases}$$



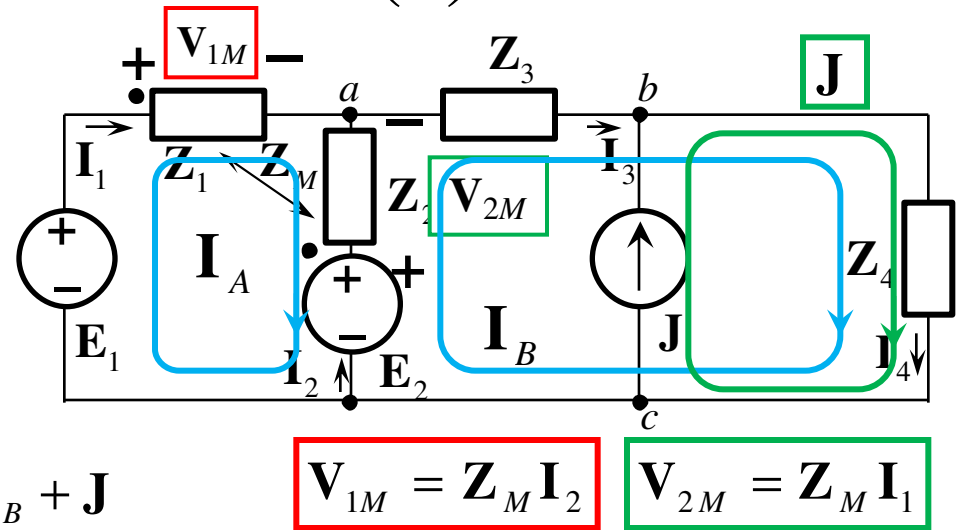


Ex. 7

Mesh Current Method (7)

$Z_1 = 10 + j15\Omega$; $Z_2 = 20 + j10\Omega$; $Z_M = j2\Omega$;
 $Z_3 = -j20\Omega$; $Z_4 = 25\Omega$; $E_1 = 100\text{ V}$;
 $E_2 = 150/\underline{30^\circ}\text{ V}$; $J = 5/\underline{45^\circ}\text{ A}$.

$$\begin{cases}
 A: Z_1 I_1 + Z_M I_2 - Z_2 I_2 - Z_M I_1 \\
 \quad = E_1 - E_2 \\
 B: Z_2 I_2 + Z_M I_1 + Z_3 I_3 + Z_4 I_4 = E_2 \\
 I_1 = I_A; I_2 = I_B - I_A; I_3 = I_B; I_4 = I_B + J
 \end{cases}$$



$$V_{1M} = Z_M I_2$$

$$V_{2M} = Z_M I_1$$

$$\rightarrow \begin{cases}
 (Z_1 + Z_2 - 2Z_M)I_A + (Z_M - Z_2)I_B = E_1 - E_2 \\
 (Z_M - Z_2)I_A + (Z_2 + Z_3 + Z_4)I_B = E_2 - Z_4 J
 \end{cases} \rightarrow \begin{cases}
 I_A = -1.49 - j2.06 \text{ A} \\
 I_B = 0.91 - j1.28 \text{ A}
 \end{cases}$$

$$\rightarrow \begin{cases}
 I_1 = I_A = -1.49 - j2.06 \text{ A} \\
 I_2 = I_B - I_A = 2.40 + j0.79 \text{ A} \\
 I_3 = I_B = 0.91 - j1.28 \text{ A} \\
 I_4 = I_B + J = 4.44 + j2.26 \text{ A}
 \end{cases}$$

Analysis of Magnetically Coupled Circuits

1. Write voltages of mutual inductance
2. Assign signs at dotted terminals (using dot convention)
3. Write KVL equations (branch current method or mesh current method)
4. Write the set of equations & solve for it



Magnetically Coupled Circuits

1. Mutual Inductance
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 - d) Equivalent Subcircuits**
4. Energy in a Coupled Circuit
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Equivalent Subcircuits (1)

Ex. 1

Find the Thevenin equivalent subcircuit?

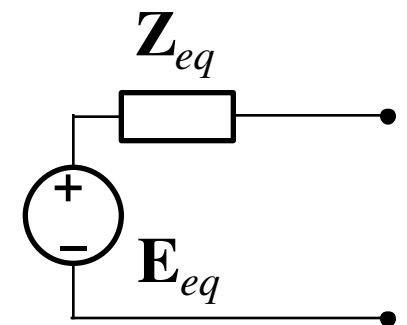
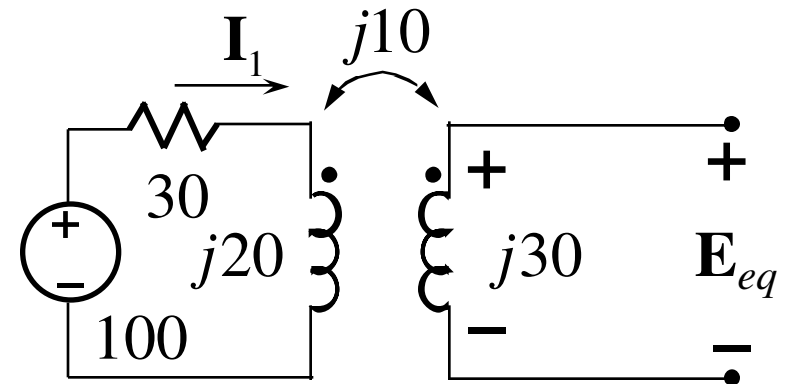
$$\mathbf{V}_{M2} = j10\mathbf{I}_1$$

$$\mathbf{E}_{eq} = \mathbf{V}_{M2} = j10\mathbf{I}_1$$

$$(30 + j20)\mathbf{I}_1 = 100$$

$$\rightarrow \mathbf{I}_1 = 2.31 - j1.54 \text{ A}$$

$$\rightarrow \mathbf{E}_{eq} = 15.38 + j23.08 \text{ V}$$



Equivalent Subcircuits (2)

Ex. 1

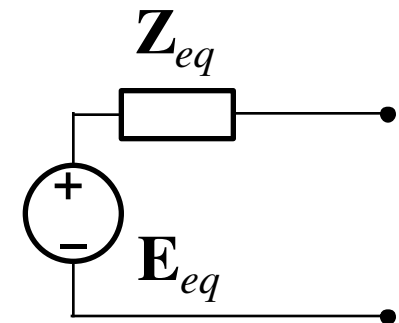
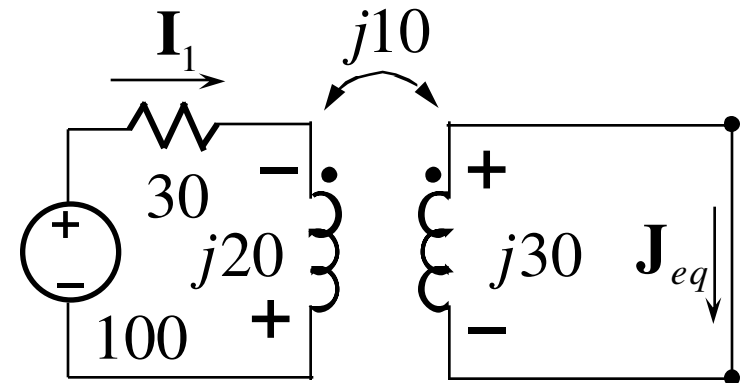
Find the Thevenin equivalent subcircuit?

$$\mathbf{Z}_{eq} = \frac{\mathbf{E}_{eq}}{\mathbf{J}_{eq}}$$

$$\begin{cases} (30 + j20)\mathbf{I}_1 - j10\mathbf{J}_{eq} = 100 \\ -j10\mathbf{I}_1 + j30\mathbf{J}_{eq} = 0 \end{cases}$$

$$\rightarrow \mathbf{J}_{eq} = 0.85 - j0.47 \text{ A}$$

$$\rightarrow \mathbf{Z}_{eq} = \frac{15.38 + j23.08}{0.85 - j0.47} = \boxed{2.31 + j28.46\Omega}$$



$$\boxed{\mathbf{E}_{eq} = 15.38 + j23.08 \text{ V}; \mathbf{Z}_{eq} = 2.31 + j28.46\Omega}$$

Equivalent Subcircuits (3)

Ex. 1

Find the Thevenin equivalent subcircuit?

$$\mathbf{Z}_{eq} = \frac{\mathbf{E}_{eq}}{\mathbf{J}_{eq}} = 2.31 + j28.46\Omega$$

Method 1

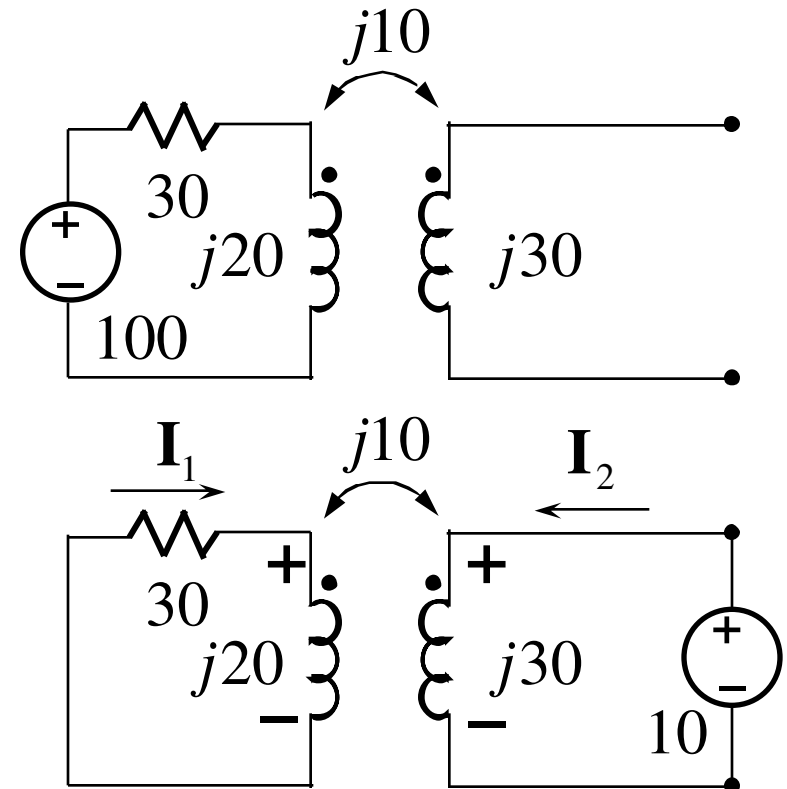
$$\mathbf{Z}_{eq} = \frac{10}{\mathbf{I}_2}$$

Method 2

$$\begin{cases} (30 + j20)\mathbf{I}_1 + j10\mathbf{I}_2 = 0 \\ j10\mathbf{I}_1 + j30\mathbf{I}_2 = 10 \end{cases}$$

$$\rightarrow \mathbf{I}_2 = 0.28 - j0.35 \text{ A}$$

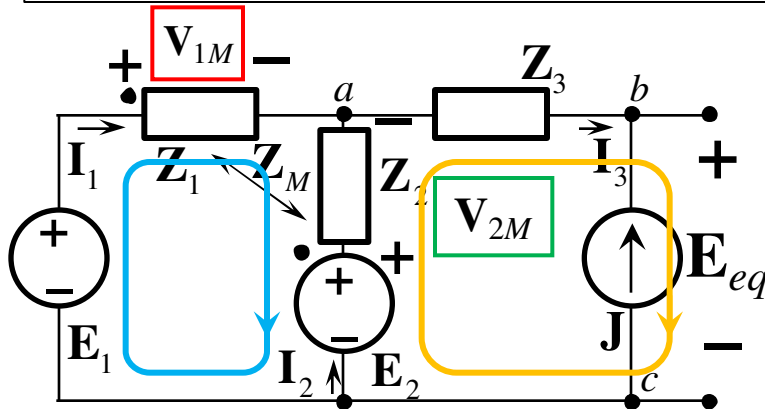
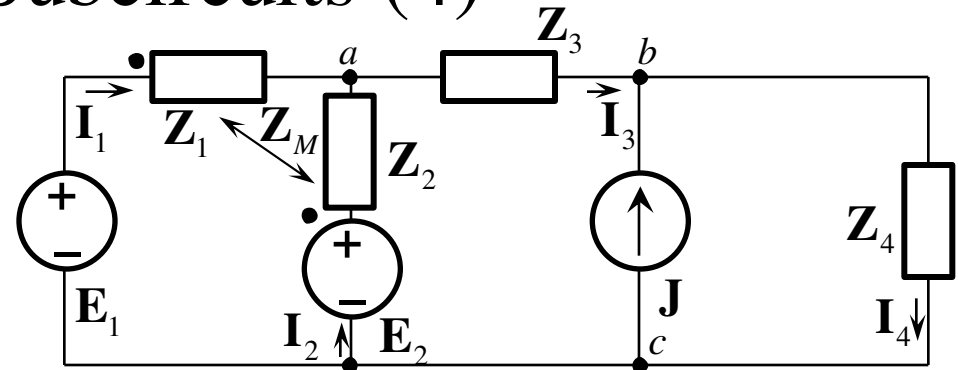
$$\rightarrow \mathbf{Z}_{eq} = \frac{10}{0.28 - j0.35} = \boxed{2.31 + j28.46\Omega}$$



Equivalent Subcircuits (4)

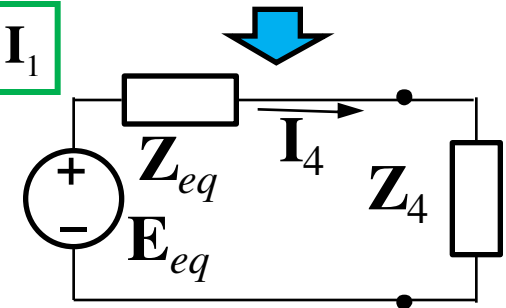
Ex. 2

$Z_1 = 10 + j15\Omega$; $Z_2 = 20 + j10\Omega$; $Z_M = j2\Omega$;
 $Z_3 = -j20\Omega$; $Z_4 = 25\Omega$; $E_1 = 100\text{ V}$;
 $E_2 = 150/30^\circ\text{ V}$; $J = 5/45^\circ\text{ A}$. Find I_4 ?



$$V_{1M} = Z_M I_2$$

$$V_{2M} = Z_M I_1$$

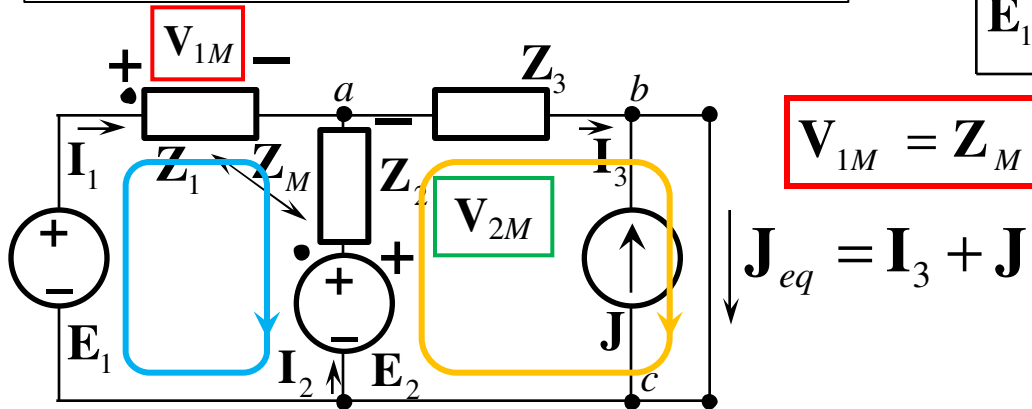
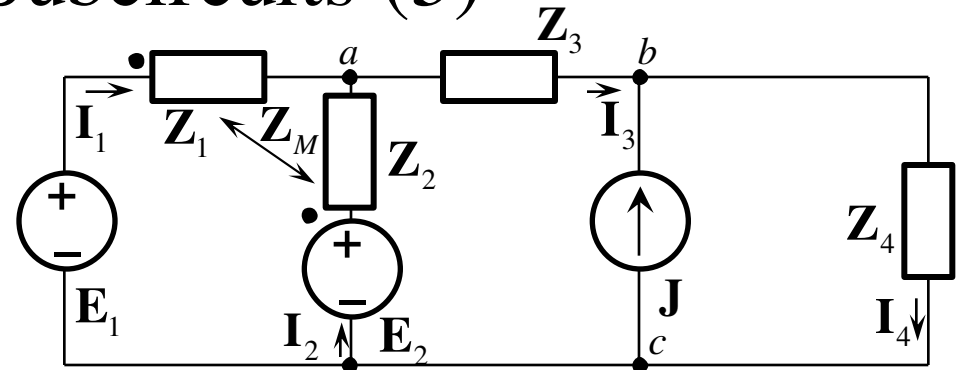


$$\left\{ \begin{array}{l} a : I_1 + I_2 = I_3 = -J \\ Z_1 I_1 + Z_M I_2 - Z_2 I_2 - Z_M I_1 = E_1 - E_2 \\ Z_2 I_2 + Z_M I_1 + Z_3 I_3 + E_{eq} = E_2 \end{array} \right. \rightarrow \begin{array}{l} I_1 = -4.34 - j2.76\text{ A}; I_2 = 0.81 - j0.78\text{ A} \\ E_{eq} = 171.19 + j20.42\text{ V} \end{array}$$

Equivalent Subcircuits (5)

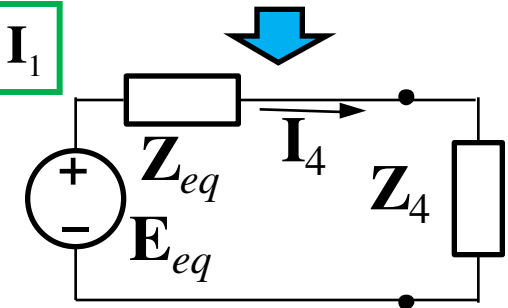
Ex. 2

$Z_1 = 10 + j15\Omega$; $Z_2 = 20 + j10\Omega$; $Z_M = j2\Omega$;
 $Z_3 = -j20\Omega$; $Z_4 = 25\Omega$; $E_1 = 100\text{ V}$;
 $E_2 = 150\angle 30^\circ\text{ V}$; $J = 5\angle 45^\circ\text{ A}$. Find I_4 ?



$$V_{1M} = Z_M I_2$$

$$V_{2M} = Z_M I_1$$



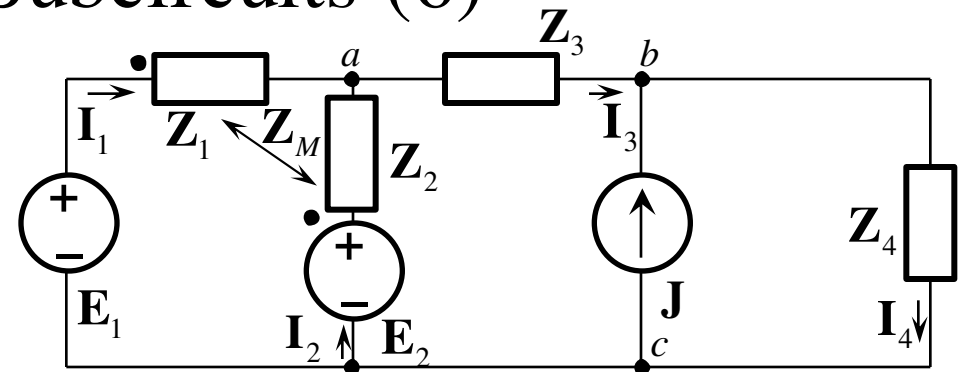
$$\begin{cases} I_1 + I_2 - I_3 = 0 \\ Z_1 I_1 + Z_M I_2 - Z_2 I_2 - Z_M I_1 = E_1 - E_2 \\ Z_2 I_2 + Z_M I_1 + Z_3 I_3 = E_2 \end{cases} \rightarrow J_{eq} = 5.25 + j11.09\text{ A}$$

$$\rightarrow I_3 = 1.71 + j7.56\text{ A}$$

Equivalent Subcircuits (6)

Ex. 2

$Z_1 = 10 + j15\Omega$; $Z_2 = 20 + j10\Omega$; $Z_M = j2\Omega$;
 $Z_3 = -j20\Omega$; $Z_4 = 25\Omega$; $E_1 = 100\text{ V}$;
 $E_2 = 150\angle 30^\circ\text{ V}$; $J = 5\angle 45^\circ\text{ A}$. Find I_4 ?

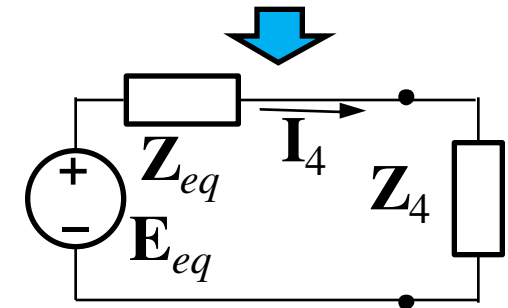


$$E_{eq} = 171.19 + j20.42\text{ V}$$

$$J_{eq} = 5.25 + j11.09\text{ A}$$

$$Z_{eq} = \frac{E_{eq}}{J_{eq}}$$

$$\rightarrow Z_{eq} = 7.47 - j11.90\Omega$$



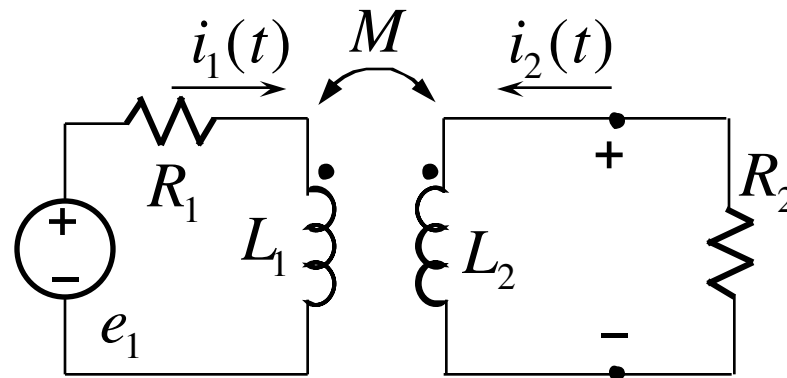
$$I_4 = \frac{E_{eq}}{Z_{eq} + Z_4} = \boxed{4.44 + j2.26\text{ A}}$$

Magnetically Coupled Circuits

1. Mutual Inductance
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Energy in a Coupled Circuit



$$w = \frac{1}{2} L_1 i_1^2 + \frac{1}{2} L_2 i_2^2 \pm M i_1 i_2$$

$$M = k \sqrt{L_1 L_2}$$

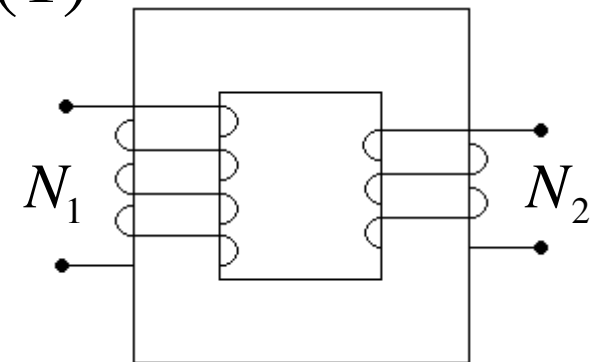
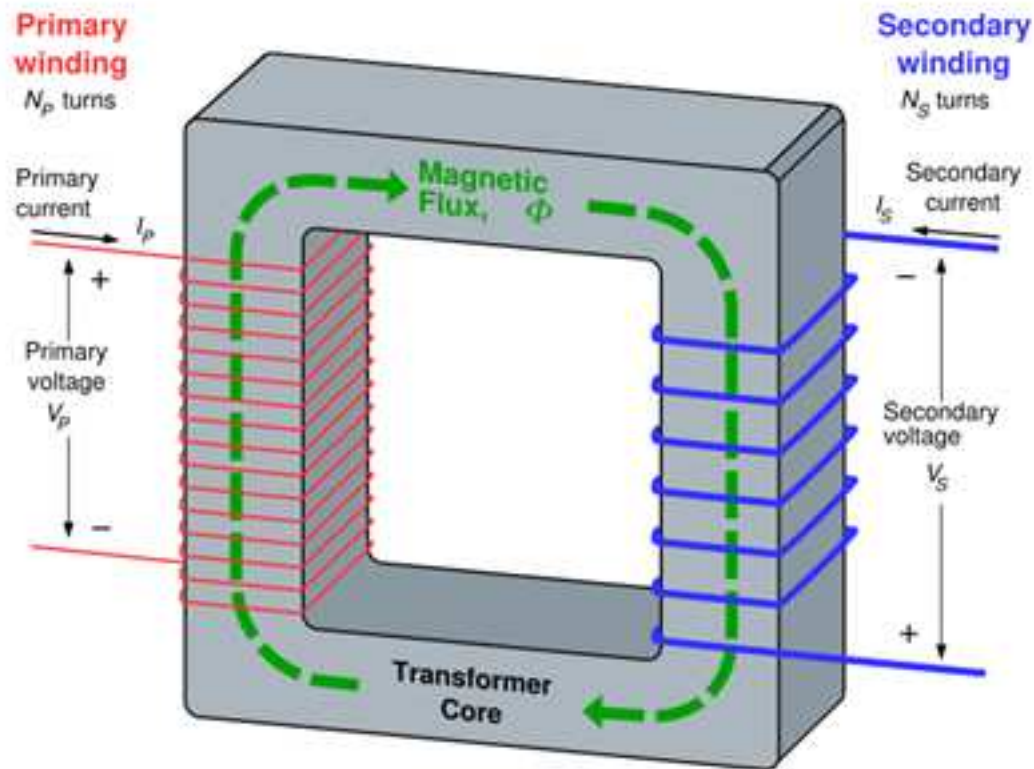
$$0 \leq k \leq 1$$

Magnetically Coupled Circuits

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 - a) Linear Transformers**
 - b) Ideal Transformers**
 - c) Ideal Autotransformers**
 - d) Three – Phase Transformers**



Linear Transformers (1)



http://www.splung.com/content/sid/3/page/applications_of_induction

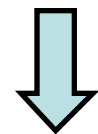
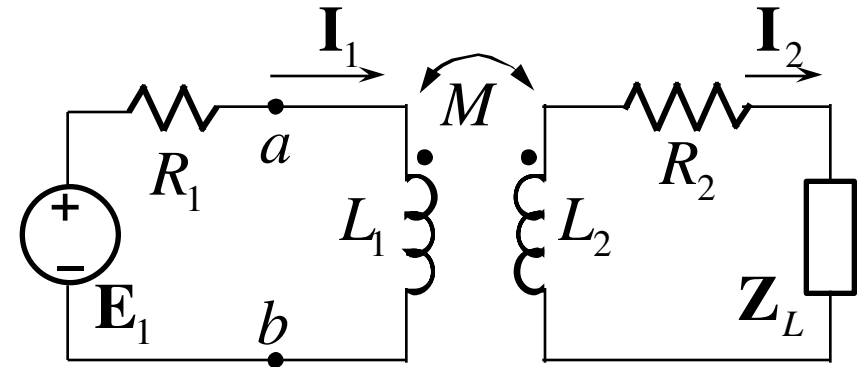
https://www.alibaba.com/product-detail/10KVA-Single-Phase-Transformer-to-VDE_60699387302.html

<https://sites.google.com/site/ncpdhbkhn/home>

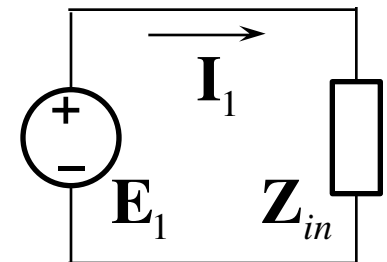
Linear Transformers (2)

$$\left\{ \begin{array}{l} (R_1 + j\omega L_1)\mathbf{I}_1 - j\omega M\mathbf{I}_2 = \mathbf{E}_1 \\ -j\omega M\mathbf{I}_1 + (R_2 + j\omega L_2 + \mathbf{Z}_L)\mathbf{I}_2 = 0 \end{array} \right.$$

$$\mathbf{Z}_{in} = \frac{\mathbf{E}_1}{\mathbf{I}_1}$$



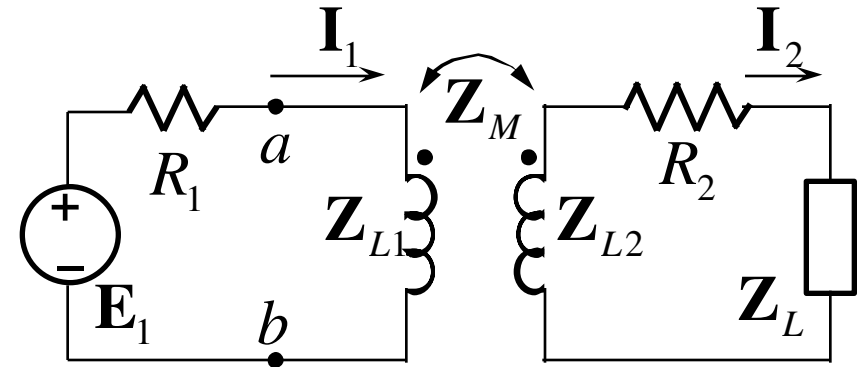
$$\rightarrow \mathbf{Z}_{in} = R_1 + j\omega L_1 + \frac{\omega^2 M^2}{R_2 + j\omega L_2 + \mathbf{Z}_L}$$



Ex.

Linear Transformers (3)

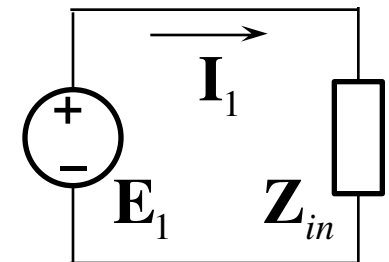
$E_1 = 100 \angle 30^\circ \text{ V}; R_1 = 60 \Omega; R_2 = 40 \Omega;$
 $Z_L = 80 + j10 \Omega; Z_{L1} = j20 \Omega; Z_{L2} = j40 \Omega;$
 $Z_M = j5 \Omega.$ Find I_1 ?



$$Z_{in} = R_1 + j\omega L_1 + \frac{\omega^2 M^2}{R_2 + j\omega L_2 + Z_L}$$

$$= 60 + j20 + \frac{5^2}{40 + j40 + 80 + j10} = 60.18 + j19.93 \Omega$$

$$I_1 = \frac{E_1}{Z_{in}} = \frac{100 \angle 30^\circ}{60.18 + j19.93} = 1.54 + j0.32 \Omega$$



Magnetically Coupled Circuits

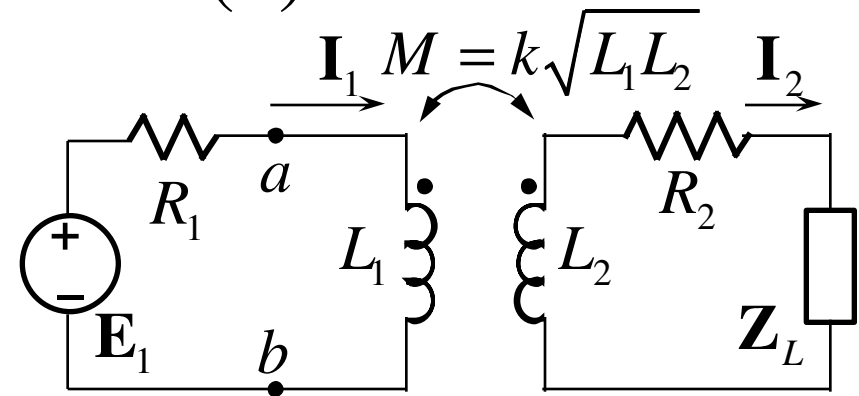
1. Mutual Inductance
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- b) Ideal Transformers**
- c) Ideal Autotransformers
- d) Three – Phase Transformers



Ideal Transformers (1)



A transformer is said to be ideal if:

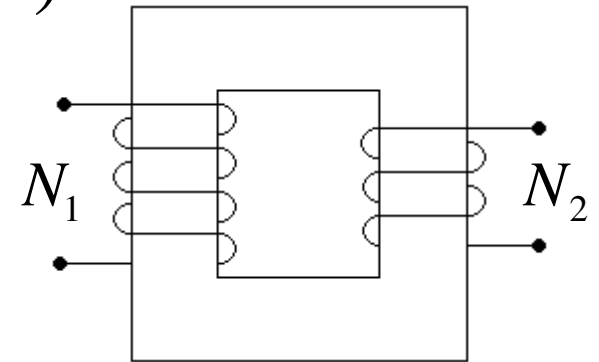
1. Coils have very large reactances ($L_1, L_2, M \rightarrow \infty$).
2. Coupling coefficient is equal to unity ($k = 1$).
3. Primary & secondary coils are lossless ($R_1 = R_2 = 0$).

Ideal Transformers (2)

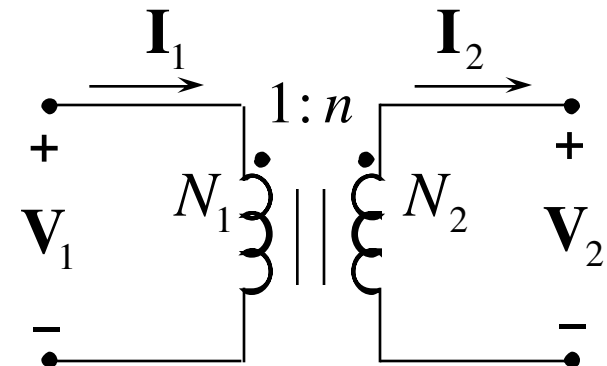
$$\left\{ \begin{array}{l} \mathbf{V}_1 = j\omega L_1 \mathbf{I}_1 - j\omega M \mathbf{I}_2 \\ \mathbf{V}_2 = -j\omega L_1 \mathbf{I}_2 + j\omega M \mathbf{I}_1 \end{array} \right\} \rightarrow \mathbf{V}_2 = \sqrt{\frac{L_2}{L_1}} \mathbf{V}_1 = n \mathbf{V}_1$$

If $k = 1 \rightarrow M = \sqrt{L_1 L_2}$

$$\rightarrow \frac{\mathbf{V}_2}{\mathbf{V}_1} = \sqrt{\frac{L_2}{L_1}} = n$$



$$\left\{ \begin{array}{l} v_1 = N_1 \frac{d\phi}{dt} \\ v_2 = N_2 \frac{d\phi}{dt} \end{array} \right\} \rightarrow \frac{v_2}{v_1} = \frac{N_2}{N_1} = n \rightarrow \boxed{\frac{\mathbf{V}_2}{\mathbf{V}_1} = \frac{N_2}{N_1} = n}$$



$$p_1 = p_2 \rightarrow v_1 i_1 = v_2 i_2 \rightarrow \frac{i_2}{i_1} = \frac{v_1}{v_2} \rightarrow \boxed{\frac{\mathbf{I}_2}{\mathbf{I}_1} = \frac{N_1}{N_2} = \frac{1}{n}}$$

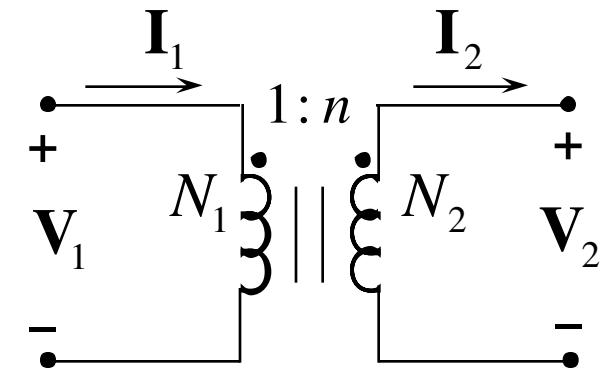
Ex. 1 Ideal Transformers (3)

Given an ideal step-down transformer rated at 22/0.4 kV, 1000 turns on the primary side. Find:

- a) The turn ratio?
- b) The number of turns on the secondary side?



Ideal Transformers (4)

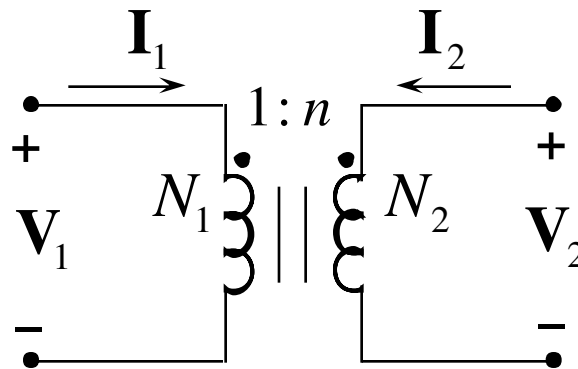


$$\left\{ \begin{array}{l} \mathbf{V}_1 = j\omega L_1 \mathbf{I}_1 - j\omega M \mathbf{I}_2 \\ \mathbf{V}_2 = -j\omega L_1 \mathbf{I}_2 + j\omega M \mathbf{I}_1 \end{array} \right\}$$

$$M = \sqrt{L_1 L_2}$$

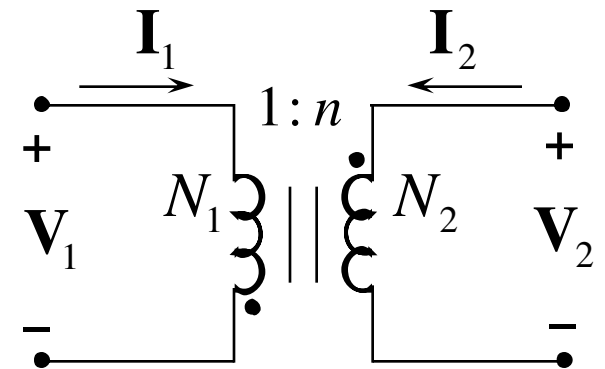
$$p_1 = p_2$$

$$\rightarrow \left\{ \begin{array}{l} \frac{\mathbf{V}_2}{\mathbf{V}_1} = \frac{N_2}{N_1} = n \\ \frac{\mathbf{I}_2}{\mathbf{I}_1} = \frac{N_1}{N_2} = \frac{1}{n} \end{array} \right.$$

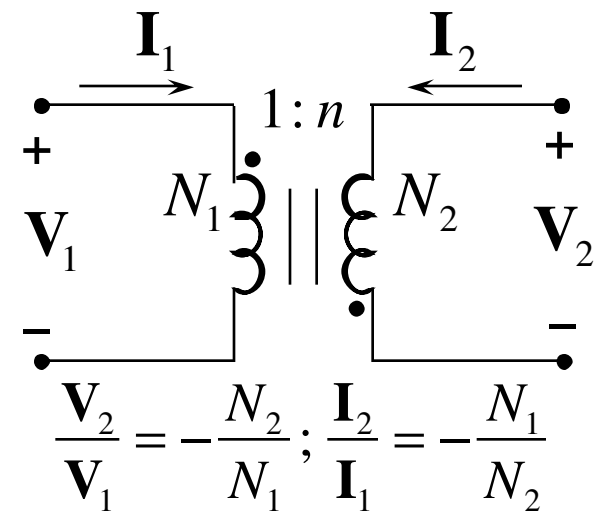


$$\left\{ \begin{array}{l} \mathbf{V}_1 = j\omega L_1 \mathbf{I}_1 + j\omega M \mathbf{I}_2 \\ \mathbf{V}_2 = j\omega L_1 \mathbf{I}_2 + j\omega M \mathbf{I}_1 \end{array} \right.$$

$$\rightarrow \left\{ \begin{array}{l} \frac{\mathbf{V}_2}{\mathbf{V}_1} = \frac{N_2}{N_1} = n \\ \frac{\mathbf{I}_2}{\mathbf{I}_1} = -\frac{N_1}{N_2} = -\frac{1}{n} \end{array} \right.$$

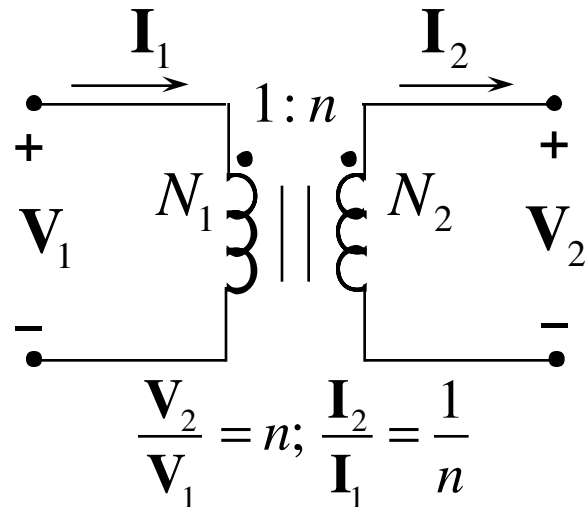


$$\frac{\mathbf{V}_2}{\mathbf{V}_1} = -\frac{N_2}{N_1}; \frac{\mathbf{I}_2}{\mathbf{I}_1} = \frac{N_1}{N_2}$$

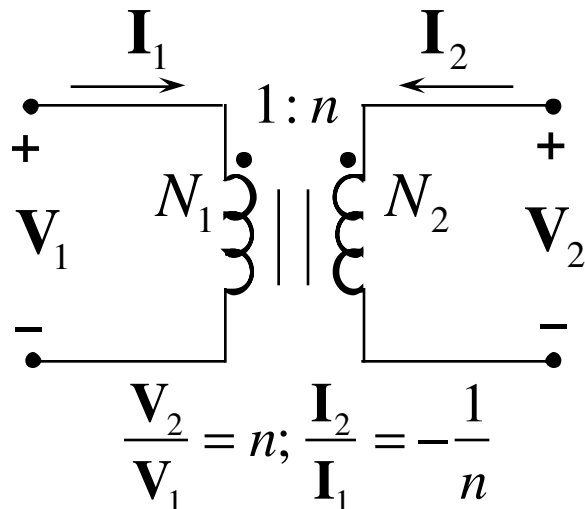
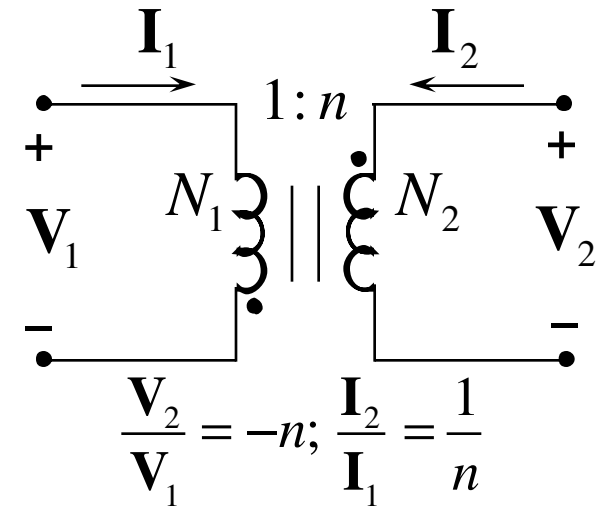


$$\frac{\mathbf{V}_2}{\mathbf{V}_1} = -\frac{N_2}{N_1}; \frac{\mathbf{I}_2}{\mathbf{I}_1} = -\frac{N_1}{N_2}$$

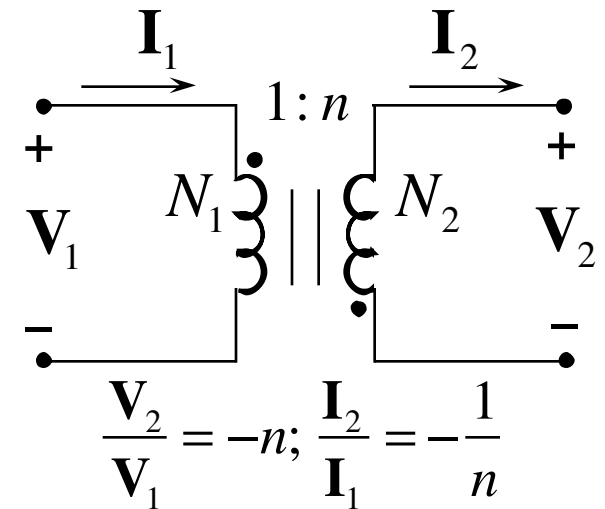
Ideal Transformers (5)



- If v_1 & v_2 are both positive or both negative at the dotted terminals, use $+n$.
Otherwise, use $-n$



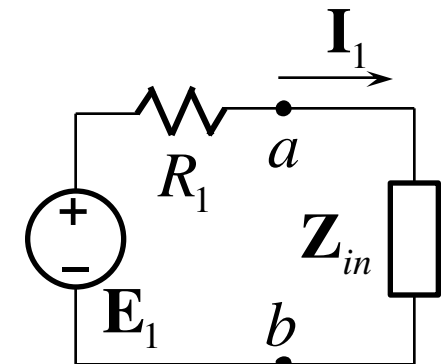
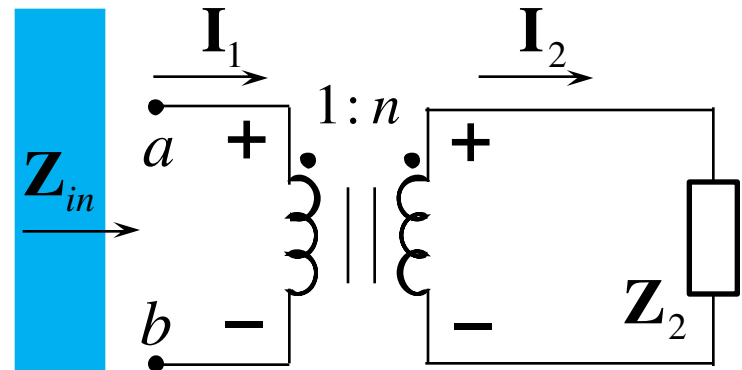
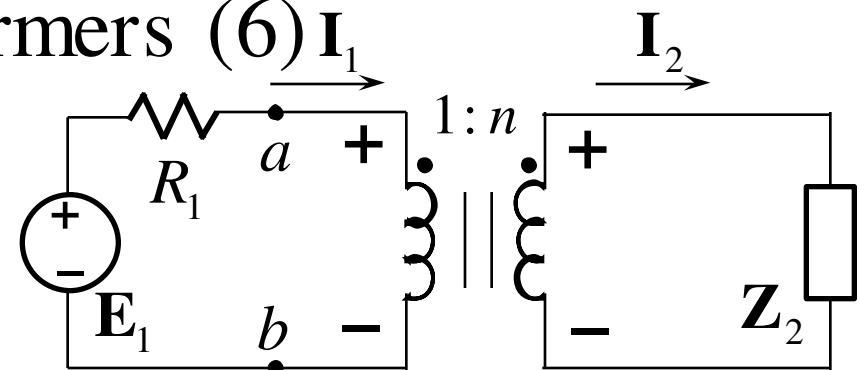
- If i_1 & i_2 both enter into or both leave the dotted terminals, use $-n$.
Otherwise, use $+n$



Ideal Transformers (6)

$$\left. \begin{aligned} \mathbf{Z}_{in} &= \frac{\mathbf{V}_{ab}}{\mathbf{I}_1} = \frac{\mathbf{V}_1}{\mathbf{I}_1} \\ \mathbf{V}_1 &= \mathbf{V}_2 / n \\ \mathbf{I}_1 &= n\mathbf{I}_2 \\ \mathbf{V}_2 &= \mathbf{Z}_2 \mathbf{I}_2 \end{aligned} \right\}$$

$$\rightarrow \mathbf{Z}_{in} = \frac{\mathbf{Z}_2}{n^2}$$



Ex. 2

Ideal Transformers (7)

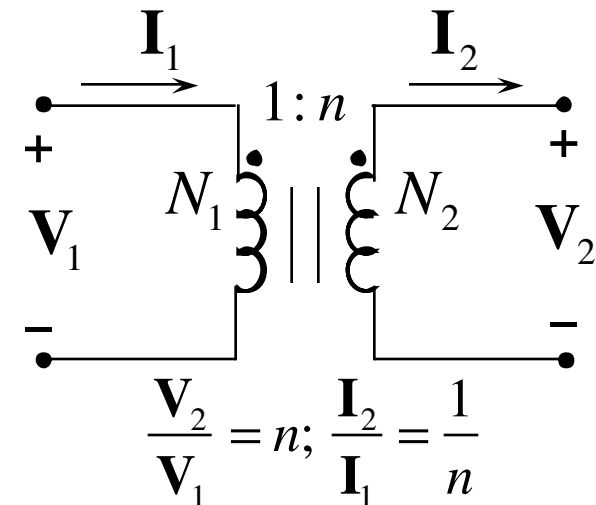
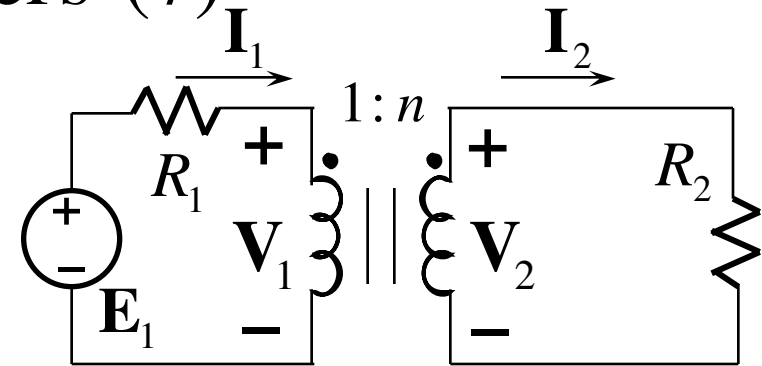
Given an ideal transformer, find currents if
 $E_1 = 100 \angle 0^\circ \text{ V}; n = 5; R_1 = 6\Omega; R_2 = 100\Omega$?

Method 1

$$\left. \begin{aligned} R_1 \mathbf{I}_1 + \mathbf{V}_1 &= \mathbf{E}_1 \\ -\mathbf{V}_2 + R_2 \mathbf{I}_2 &= 0 \end{aligned} \right\}$$

$$\rightarrow \begin{cases} 6\mathbf{I}_1 + \mathbf{V}_1 = 100 \angle 0^\circ \\ -5\mathbf{V}_1 + 100 \times \frac{\mathbf{I}_1}{5} = 0 \end{cases}$$

$$\rightarrow \mathbf{I}_1 = \boxed{10 \text{ A}} \rightarrow \mathbf{I}_2 = \frac{\mathbf{I}_1}{5} = \boxed{2 \text{ A}}$$



Ex. 2

Ideal Transformers (8)

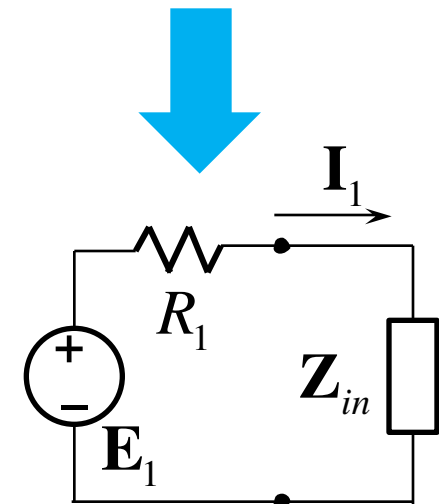
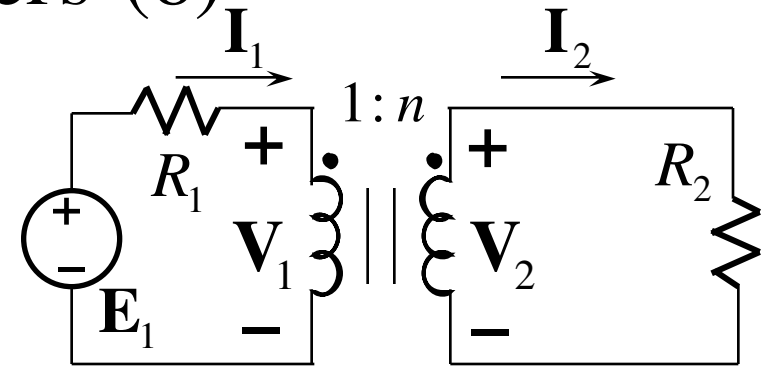
Given an ideal transformer, find currents if
 $E_1 = 100 \angle 0^\circ \text{ V}; n = 5; R_1 = 6\Omega; R_2 = 100\Omega$?

Method 2

$$Z_{in} = \frac{R_2}{n^2} = \frac{100}{25} = 4\Omega$$

$$I_1 = \frac{E_1}{R_1 + Z_{in}} = \frac{100}{6 + 4} = \boxed{10 \text{ A}}$$

$$\frac{I_1}{I_2} = n = 5 \rightarrow I_2 = \frac{I_1}{5} = \boxed{2 \text{ A}}$$



Ex. 3

Ideal Transformers (9)

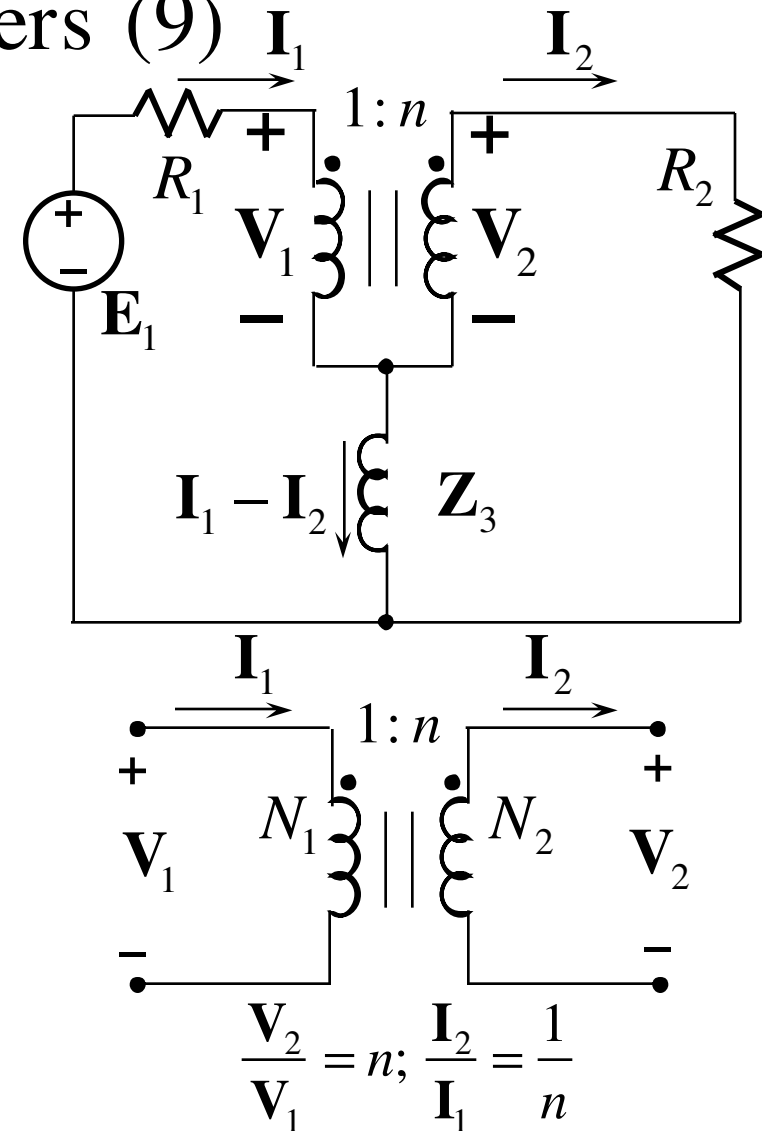
Given an ideal transformer, find currents if
 $\mathbf{E}_1 = 100 \angle 0^\circ \text{ V}; n = 5; R_1 = 6\Omega; R_2 = 100\Omega;$
 $\mathbf{Z}_3 = j20\Omega?$

$$\left. \begin{aligned} R_1 \mathbf{I}_1 + \mathbf{V}_1 + \mathbf{Z}_3 (\mathbf{I}_1 - \mathbf{I}_2) &= \mathbf{E}_1 \\ -\mathbf{V}_2 + R_2 \mathbf{I}_2 - \mathbf{Z}_3 (\mathbf{I}_1 - \mathbf{I}_2) &= 0 \end{aligned} \right\}$$

$$\rightarrow \begin{cases} 6\mathbf{I}_1 + \mathbf{V}_1 + j20 \left(\mathbf{I}_1 - \frac{\mathbf{I}_1}{5} \right) = 100 \angle 0^\circ \\ -5\mathbf{V}_1 + 100 \times \frac{\mathbf{I}_1}{5} - j20 \left(\mathbf{I}_1 - \frac{\mathbf{I}_1}{5} \right) = 0 \end{cases}$$

$$\rightarrow \mathbf{I}_1 = \boxed{3.79 - j4.85 \text{ A}}$$

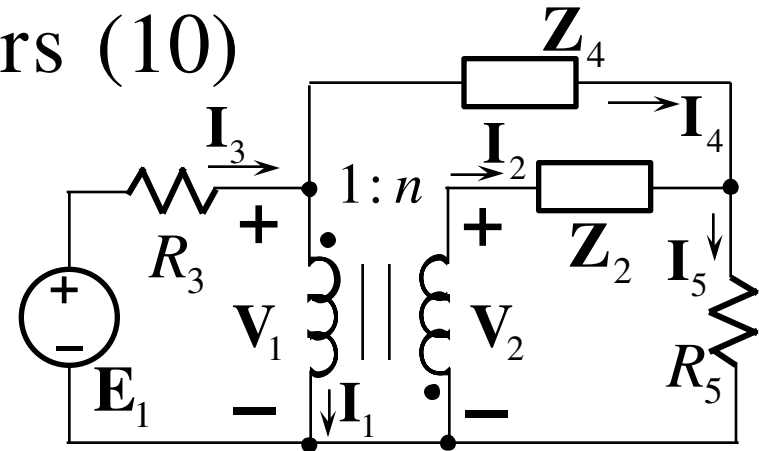
$$\rightarrow \mathbf{I}_2 = \frac{\mathbf{I}_1}{5} = \boxed{1.90 - j2.43 \text{ A}}$$



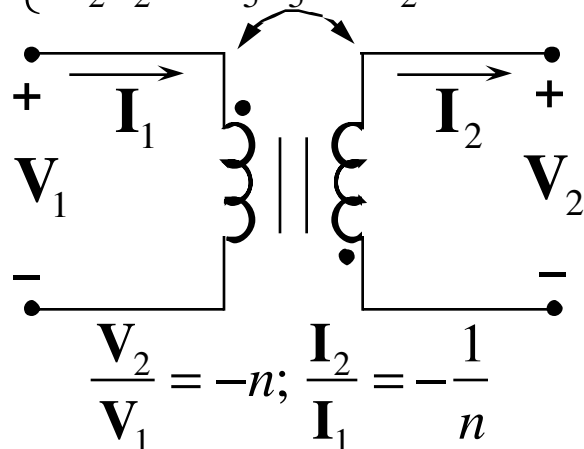
Ex. 4

Ideal Transformers (10)

$$\begin{aligned} \mathbf{E}_1 &= 100 \angle 0^\circ \text{ V}; n = 2; R_3 = 6\Omega; R_5 = 100\Omega; \\ \mathbf{Z}_2 &= j20\Omega; \mathbf{Z}_4 = 30 - j40\Omega. \end{aligned}$$



$$\begin{cases} \mathbf{I}_3 - \mathbf{I}_1 - \mathbf{I}_4 = 0 \\ \mathbf{I}_2 + \mathbf{I}_4 - \mathbf{I}_5 = 0 \\ R_3 \mathbf{I}_3 + \mathbf{V}_1 = \mathbf{E}_1 \\ \mathbf{Z}_4 \mathbf{I}_4 - \mathbf{Z}_2 \mathbf{I}_2 + \mathbf{V}_2 - \mathbf{V}_1 = 0 \\ \mathbf{Z}_2 \mathbf{I}_2 + R_5 \mathbf{I}_5 - \mathbf{V}_2 = 0 \end{cases}$$

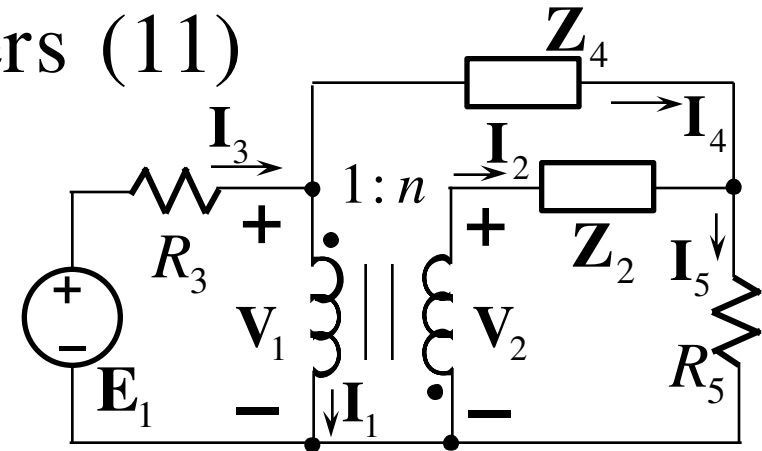


$$\rightarrow \begin{cases} \mathbf{I}_3 + 2\mathbf{I}_2 - \mathbf{I}_4 = 0 \\ \mathbf{I}_2 + \mathbf{I}_4 - \mathbf{I}_5 = 0 \\ R_3 \mathbf{I}_3 + \mathbf{V}_1 = \mathbf{E}_1 \\ \mathbf{Z}_4 \mathbf{I}_4 - \mathbf{Z}_2 \mathbf{I}_2 - 2\mathbf{V}_1 - \mathbf{V}_1 = 0 \\ \mathbf{Z}_2 \mathbf{I}_2 + R_5 \mathbf{I}_5 + 2\mathbf{V}_1 = 0 \end{cases}$$

Ex. 4

Ideal Transformers (11)

$$\begin{aligned} \mathbf{E}_1 &= 100 \angle 0^\circ \text{ V}; n = 2; R_3 = 6\Omega; R_5 = 100\Omega; \\ \mathbf{Z}_2 &= j20\Omega; \mathbf{Z}_4 = 30 - j40\Omega. \end{aligned}$$



$$\rightarrow \begin{cases} \mathbf{I}_3 + 2\mathbf{I}_2 - \mathbf{I}_4 = 0 \\ \mathbf{I}_2 + \mathbf{I}_4 - \mathbf{I}_5 = 0 \\ R_3\mathbf{I}_3 + \mathbf{V}_1 = \mathbf{E}_1 \\ \mathbf{Z}_4\mathbf{I}_4 - \mathbf{Z}_2\mathbf{I}_2 - 2\mathbf{V}_1 - \mathbf{V}_1 = 0 \\ \mathbf{Z}_2\mathbf{I}_2 + R_5\mathbf{I}_5 + 2\mathbf{V}_1 = 0 \end{cases}$$

$$\rightarrow \begin{cases} \mathbf{I}_3 + 2\mathbf{I}_2 - \mathbf{I}_4 = 0 \\ \mathbf{I}_2 + \mathbf{I}_4 - \mathbf{I}_5 = 0 \\ 6\mathbf{I}_3 + \mathbf{V}_1 = 100 \\ (30 - j40)\mathbf{I}_4 - j20\mathbf{I}_2 - 3\mathbf{V}_1 = 0 \\ j20\mathbf{I}_2 + 100\mathbf{I}_5 + 2\mathbf{V}_1 = 0 \end{cases}$$

$$\rightarrow \begin{cases} \mathbf{I}_2 = -3.63 - j0.012 \text{ A} \\ \mathbf{I}_3 = 10.09 + j0.87 \text{ A} \\ \mathbf{I}_4 = 2.83 + j0.84 \text{ A} \\ \mathbf{I}_5 = -0.79 + j0.83 \text{ A} \\ \mathbf{I}_1 = -2\mathbf{I}_2 = 7.25 + j0.025 \text{ A} \end{cases}$$

Magnetically Coupled Circuits

1. Mutual Inductance
2. Dot Convention
3. Analysis of Magnetically Coupled Circuits
4. Energy in a Coupled Circuit
- 5. Transformers**
 - a) Linear Transformers
 - b) Ideal Transformers
 - c) Ideal Autotransformers**
 - d) Three – Phase Transformers

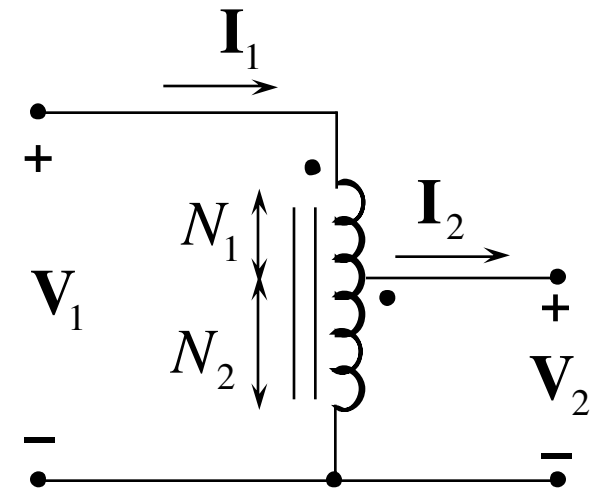


Ideal Autotransformers (1)

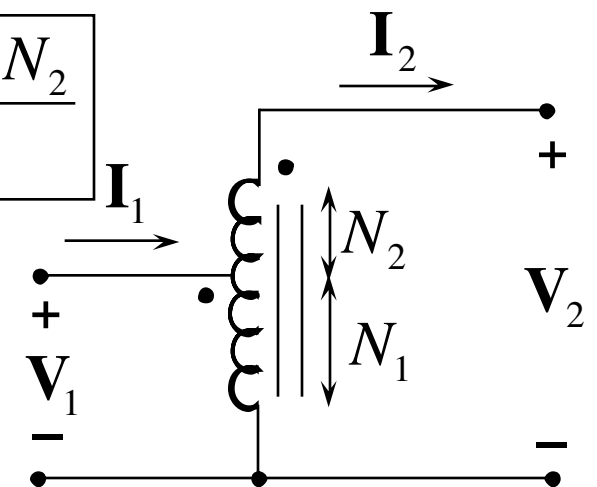
$$\left. \begin{aligned} v_1 &= (N_1 + N_2) \frac{d\phi}{dt} \\ v_2 &= N_2 \frac{d\phi}{dt} \end{aligned} \right\}$$

$$\frac{V_1}{V_2} = \frac{N_1 + N_2}{N_2}$$

$$p_1 = p_2 \rightarrow \frac{I_1}{I_2} = \frac{N_2}{N_1 + N_2}$$



$$\frac{V_1}{V_2} = \frac{N_1}{N_1 + N_2}; \quad \frac{I_1}{I_2} = \frac{N_1 + N_2}{N_1}$$



<https://proactivemarketplace.org/new-technipower-variatic-variable-autotransformer-model-w20.html>

<https://sites.google.com/site/ncpdhbkhn/home>

Ex. Ideal Autotransformers (2)

Given an ideal autotransformer, find currents if $\mathbf{E}_1 = 100/0^\circ$ V; $\mathbf{Z}_2 = 5 + j10\Omega$?

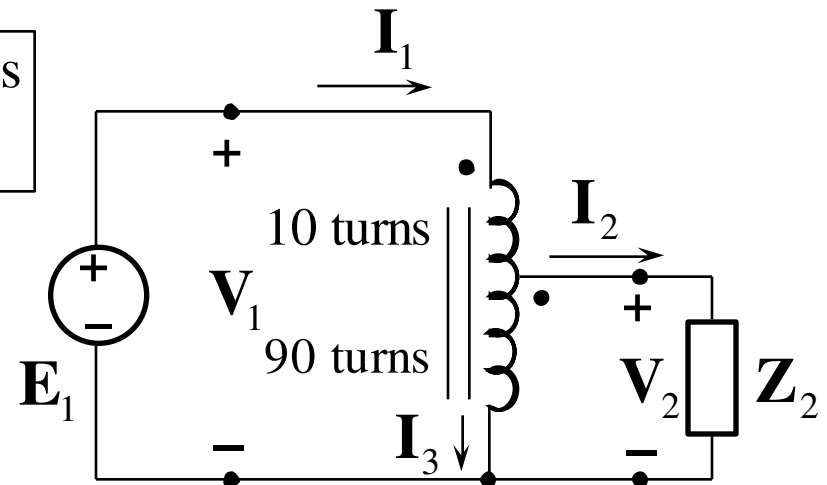
$$\frac{\mathbf{V}_1}{\mathbf{V}_2} = \frac{N_1 + N_2}{N_2} = \frac{10 + 90}{90} = 1.11$$

$$\rightarrow \mathbf{V}_2 = \frac{\mathbf{V}_1}{1.11} = \frac{100/0^\circ}{1.11} = 90/0^\circ \text{ V}$$

$$\rightarrow \mathbf{I}_2 = \frac{\mathbf{V}_2}{\mathbf{Z}_2} = \frac{90/0^\circ}{5 + j10} = \boxed{3.60 - j7.20 \text{ A}}$$

$$\frac{\mathbf{I}_1}{\mathbf{I}_2} = \frac{N_2}{N_1 + N_2} = \frac{90}{10 + 90} = 0.9 \rightarrow \mathbf{I}_1 = 0.9\mathbf{I}_2 = \boxed{3.24 - j6.48 \text{ A}}$$

$$\mathbf{I}_3 = \mathbf{I}_1 - \mathbf{I}_2 = \boxed{-0.36 + j0.72 \text{ A}}$$

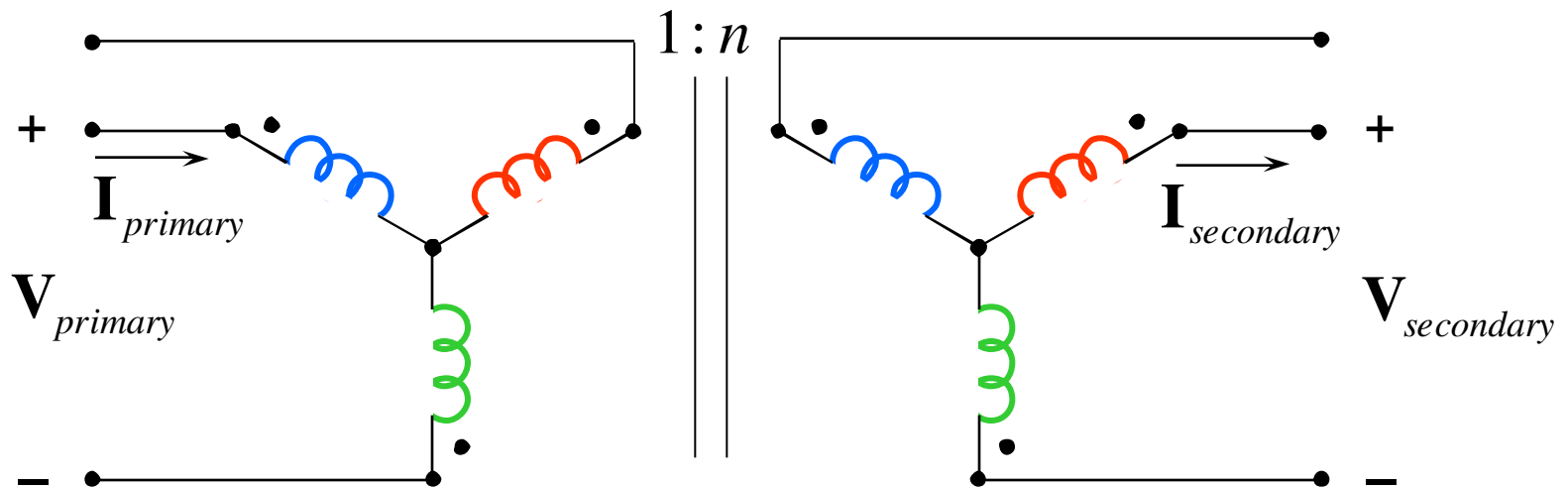


Magnetically Coupled Circuits

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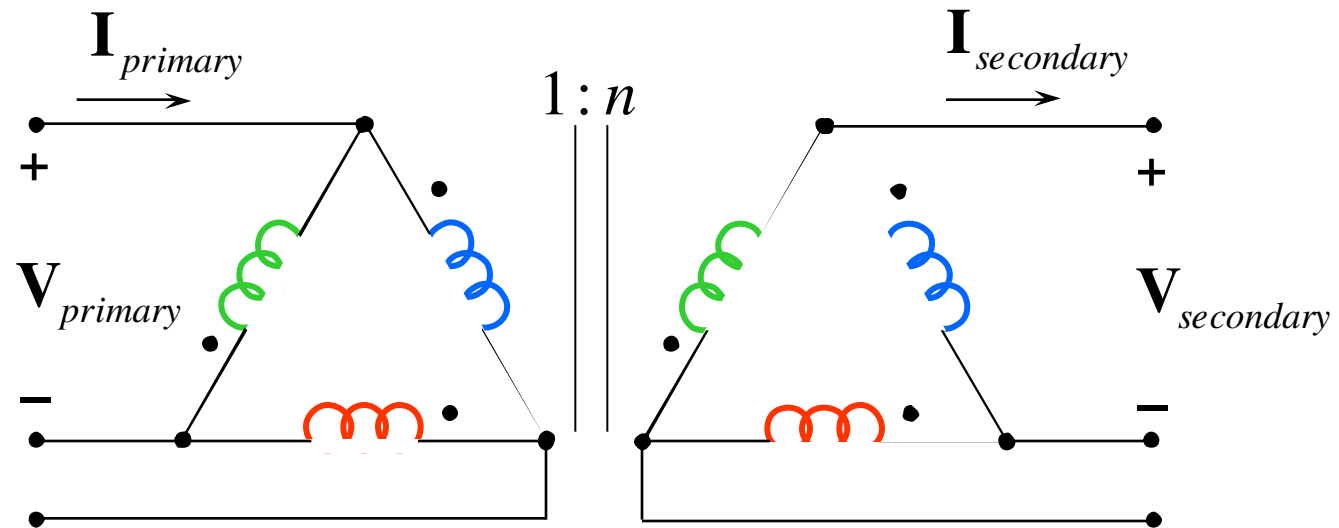
Three – Phase Transformers (1), Y–Y



$$V_{secondary} = nV_{primary}$$

$$I_{secondary} = \frac{I_{primary}}{n}$$

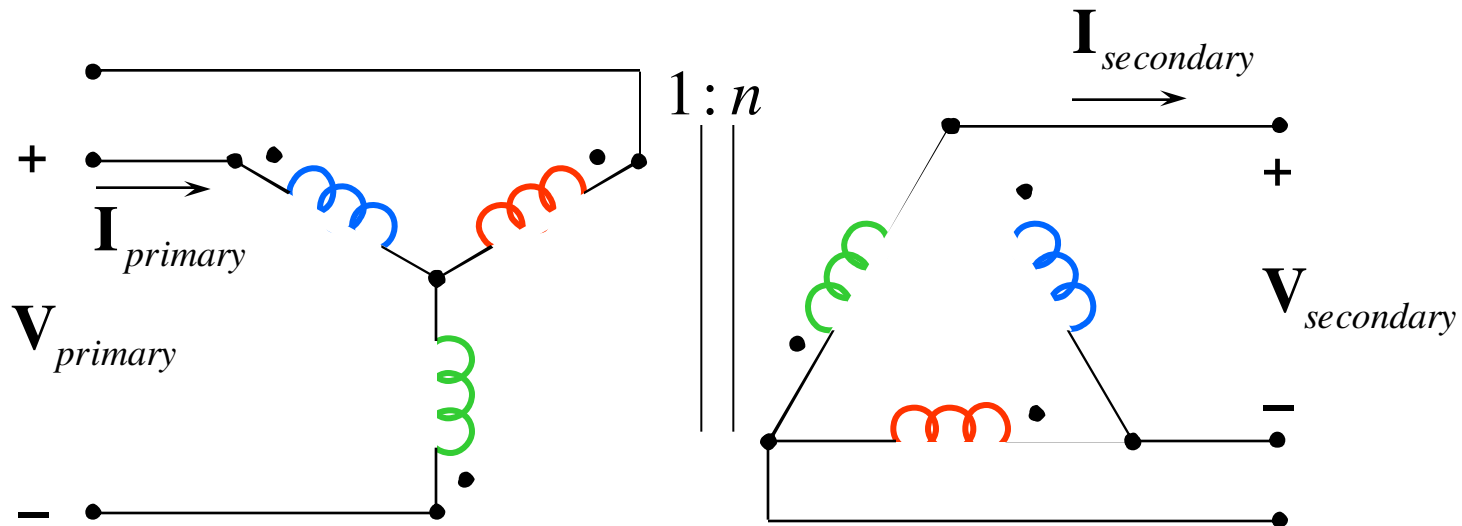
Three – Phase Transformers (2), Δ – Δ



$$V_{secondary} = n V_{primary}$$

$$I_{secondary} = \frac{I_{primary}}{n}$$

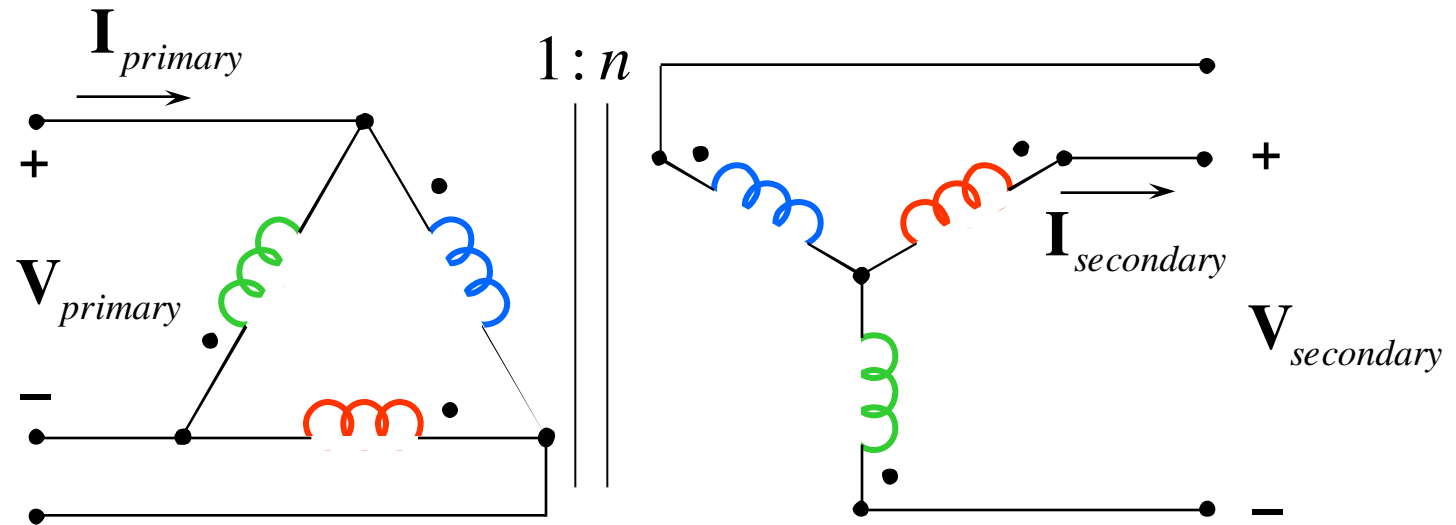
Three – Phase Transformers (3), Y– Δ



$$V_{secondary} = \frac{n V_{primary}}{\sqrt{3}}$$

$$I_{secondary} = \frac{\sqrt{3} I_{primary}}{n}$$

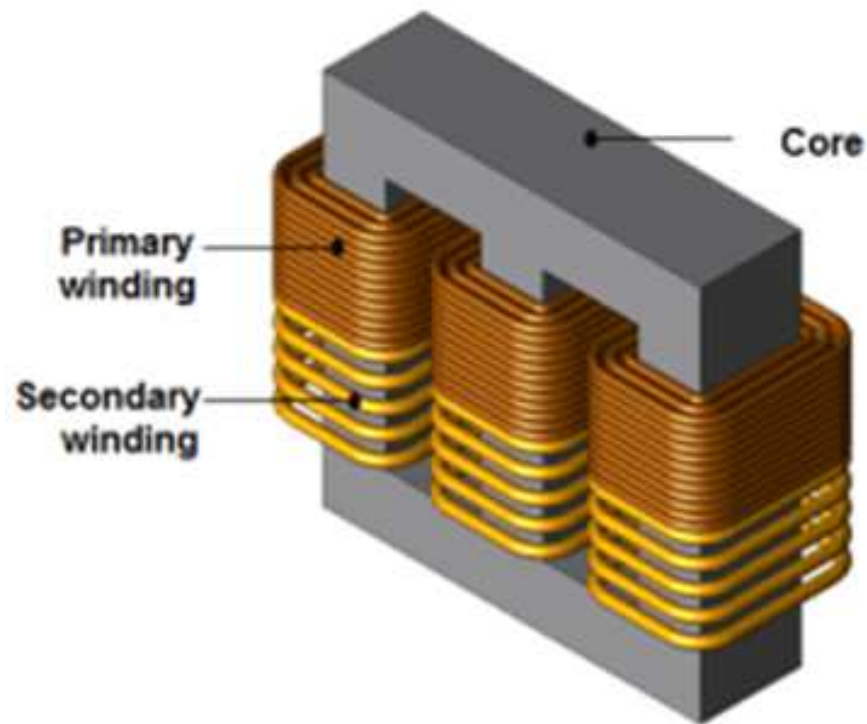
Three – Phase Transformers (4), Δ –Y



$$V_{secondary} = n\sqrt{3}V_{primary}$$

$$I_{secondary} = \frac{I_{primary}}{n\sqrt{3}}$$

Three – Phase Transformers (5)



https://www.jmag-international.com/catalog/132_threephasetransformer_loss/

<https://sites.google.com/site/ncpdhbkhn/home>