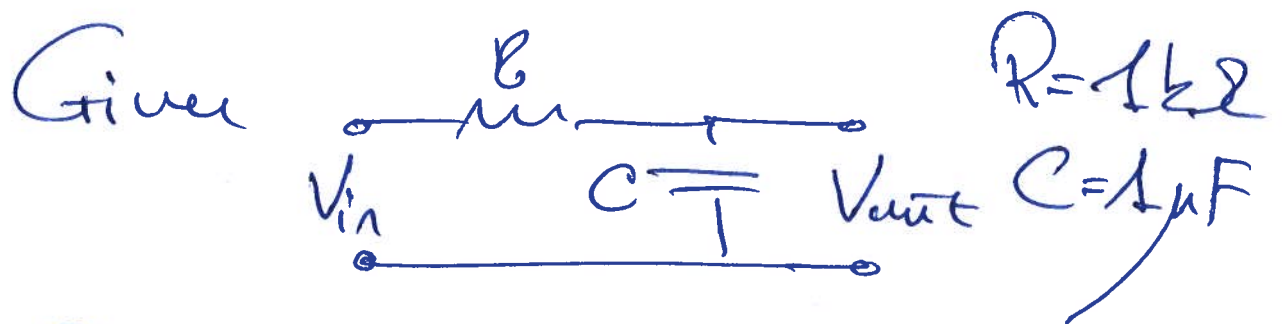


# Signal Processing, 6<sup>th</sup> lecture (1)

## Suggested Solutions

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1) Discrete-time impulse response

$$\begin{cases} h(t) = \frac{1}{RC} \cdot e^{-t/RC} \\ h[n] = T \cdot h(t) \Big|_{t=nT} = \frac{T}{RC} \cdot e^{-\frac{nT}{RC}} \\ T = \frac{1}{8000} = 125 \cdot 10^{-6} \text{ s.} \end{cases}$$



$$h[n] = 0.125 \cdot e^{-n \cdot 0.125}$$

---

Calculate  $h[n]$  for  $n=0,1,2,3,4$  ; ②

$$h[0] = 0.1250$$

$$h[1] = 0.1103$$

$$h[2] = 0.0974$$

$$h[3] = 0.0859$$

$$h[4] = 0.0758$$

2) Convolution Sum

$$y[n] = h[n] * x[n]$$

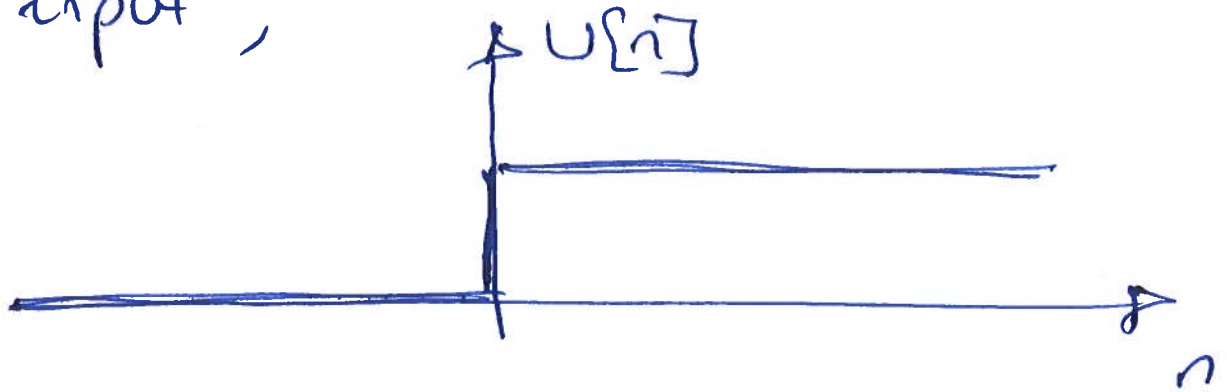
$$\Downarrow y[n] = \sum_{m=-\infty}^{\infty} h[m] \cdot x[n-m]$$

$$\Downarrow y[n] = \sum_{m=-\infty}^{\infty} 0.125 \cdot e^{-m \cdot 0.125} \cdot x[n-m]$$

$$y[n] = 0.125 \sum_{m=0}^{\infty} e^{-m \cdot 0.125} \cdot x[n-m]$$

For Causal LTI system

Step response is the output ③  
when  $x[n] = u[n]$  is applied to the  
input ;



Write a Matlab-program to do  
the calculation.

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```
% Beregning af Step-respons vha. foldnings-summen
%  $y[n] = \sum_{m=0}^{\infty} 0.125e^{(-0.125m)} x[n-m]$ 

% Steprespons'en beregnes ved hjælp af to nestede loops
% Den ydre løkke opdaterer sample nummeret mens den indre
% løkke beregner selve produkt-summen.
% Der beregnes 50 samples.

clear;

for n=0:49, % n er sample-nummeret
    sum = 0; % sum benyttes til beregning af produktsummen for sample n
            % og sættes følgelig lig 0 initialt

    % Den indre løkke beregner selve produktsummen -- baseres på 100 led
    for m=0:99,

        % Bestem værdien af impulsresponsen til tidspunktet,  $h$ 
        h = 0.125*exp(-0.125*m);

        % Bestem værdien af step-funktionen u[n-m]
        if (n-m) >= 0
            u = 1;
        else
            u = 0;
        end;

        sum = sum + (h * u);
    end;

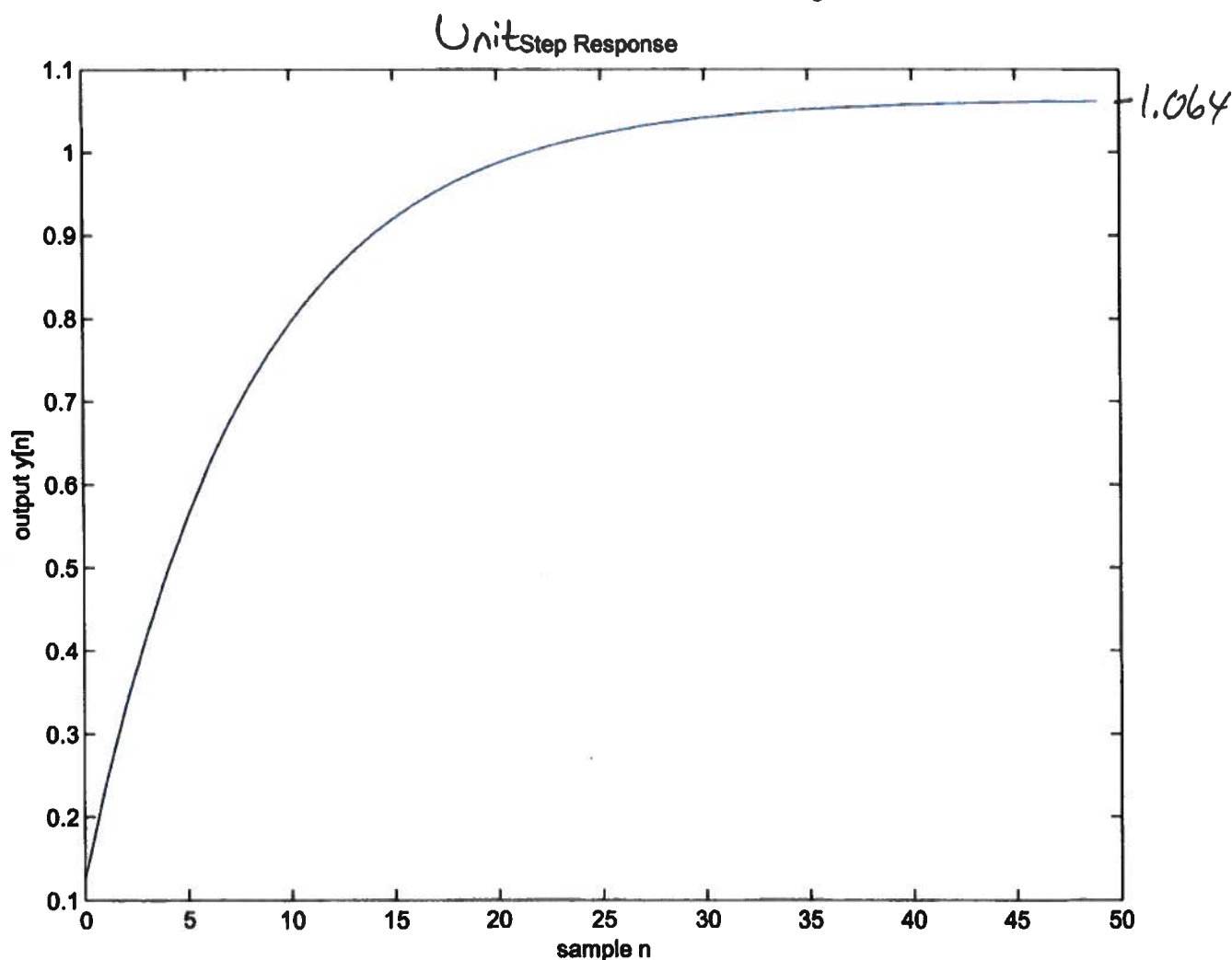
    % Opdater output-sekvensen
    y(n+1) = sum; % Bemærk at MatLab ikke kan indeksere 0
end;

% Plot stepresponsen %
x(1:50)=0:49;
plot(x,y)
title('Step Response');
xlabel('sample n');
ylabel('output y[n]');
```

PK

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Bemærk, at filteret i steady-state har et output, som numerisk er større end input ( $U[n]$ ).  
Se side 12.



3) Transfer function.

⑥

$$\left\{ \begin{aligned} H(z) &= \frac{Y(z)}{X(z)} = \frac{b}{1 - az^{-1}} \\ b &= \frac{T}{RC} \text{ and } a = e^{-T/RC} \end{aligned} \right.$$

⇓

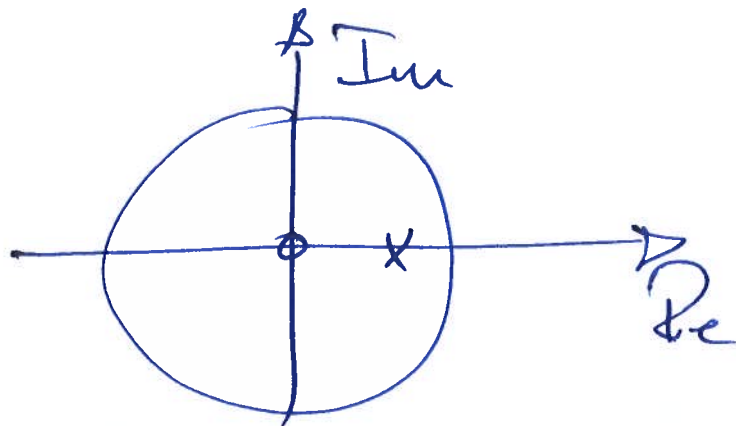
$$H(z) = \frac{0,125}{1 - 0,8825z^{-1}}$$

Poles and Zeros

$$H(z) = \frac{0,125}{1 - 0,8825z^{-1}} = \frac{0,125z}{z - 0,8825}$$

⇓

pole in  $z = 0,8825$  and zero in  $z = 0$



Pole is inside the unit circle

⇒ Stabil filter

(7)

4) Difference equation

$$H(z) = \frac{b}{1 - az^{-1}} = \frac{Y(z)}{X(z)}$$

⇒

$$Z^{-1}\{Y(z)\} = Z^{-1}\{aY(z)Z^{-1} + bX(z)\}$$

⇒

$$y[n] = ay[n-1] + bx[n]$$

Now, use this diff. equation to do the filtering of the noise contaminated signal.

% Anvend den styrende differens-ligning til beregning af et filtreret  
% output givet et inout-signal, som er overlejet med støj

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clear;

load signal\_noise.dat % 1000 samples af støjbefængt signal %  
load signal.dat % 1000 samples af støjfrit signal %

y(1) = 0.125 \* signal\_noise(1);

for n=2:1000,  
y(n) = 0.8825 \* y(n-1) + 0.125 \* signal\_noise(n);  
end;

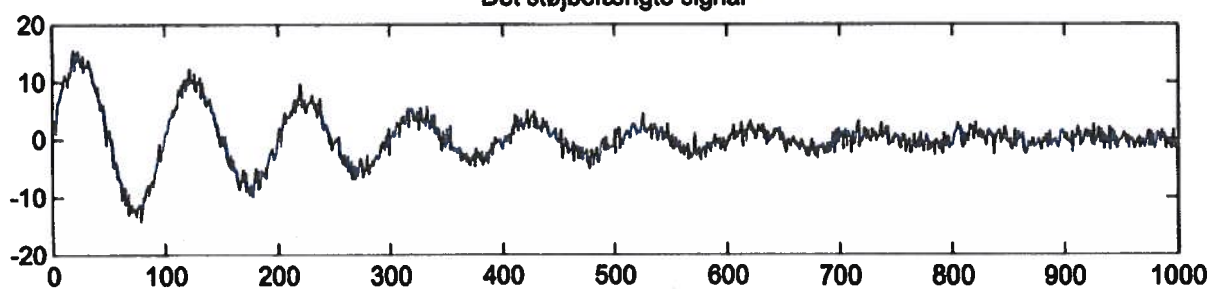
subplot(3,1,1);  
plot(signal\_noise);  
title('Det støjbefængte signal')  
subplot(3,1,2);  
plot(signal);  
title('Det ønskede signal')  
subplot(3,1,3);  
plot(y);  
title('Det filtrerede signal')



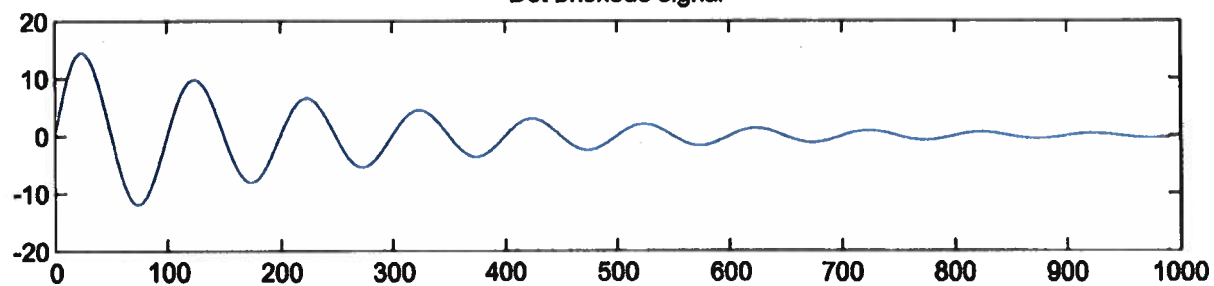
PC

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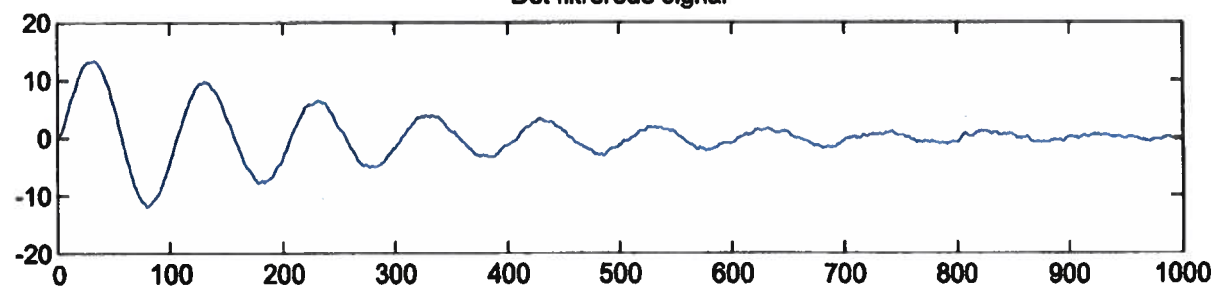
Det støjefængte signal



Det ønskede signal



Det filtrerede signal



5) Frequency response

$$H(e^{j\omega}) = H(z) \Big|_{z=e^{j\omega}}$$

⇓

$$H(e^{j\omega}) = \frac{0.125 \cdot e^{j\omega}}{e^{j\omega} - 0.8825}$$

Amplitude response

$$H(e^{j\omega}) = 0.125 \cdot \frac{\cos \omega + j \sin \omega}{\cos \omega - 0.8825 + j \sin \omega}$$

⇓

$$\begin{aligned} |H(e^{j\omega})| &= 0.125 \frac{|\cos \omega + j \sin \omega|}{|\cos \omega - 0.8825 + j \sin \omega|} \\ &= 0.125 \cdot \frac{1}{\sqrt{(\cos \omega - 0.8825)^2 + \sin^2 \omega}} \end{aligned}$$

```
% Plot filterets amplituderespons
```

```
% Først genereres en frekvens-akse med 100 punkter i intervallet [0;PI]
```

```
for i=0:99,  
    omega(i+1) = (pi*i)/100;  
end;
```

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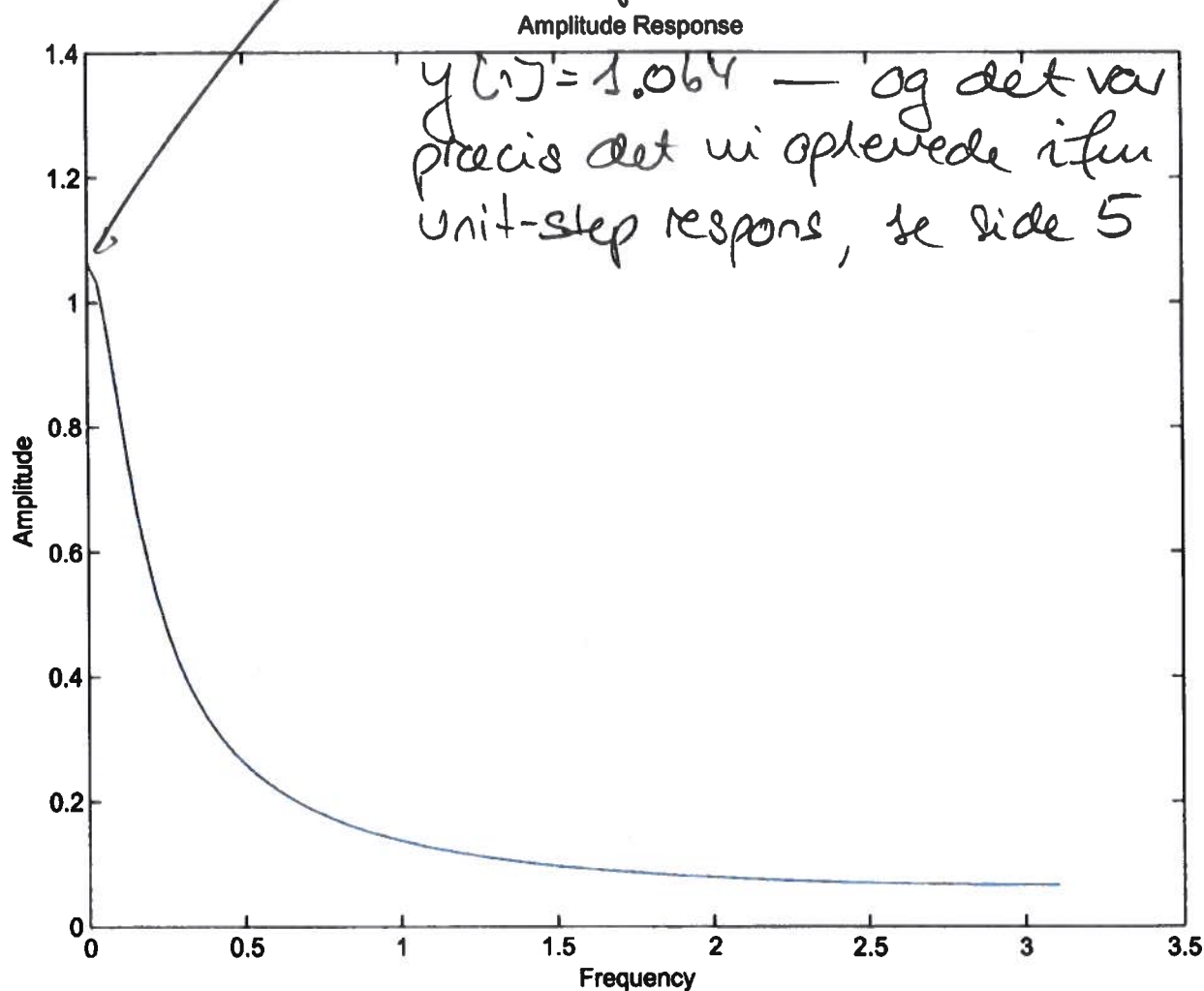
```
% Dernæst beregnes amplitude-værdierne for hvert enkelt frekvens-værdi
```

```
amplitude = zeros(100,1);
```

```
for i=1:100,  
    naevner = sqrt((cos(omega(i)) - 0.8825)^2 + (sin(omega(i)))^2);  
    amplitude(i) = 0.125/naevner;  
end;
```

```
plot(omega,amplitude)  
xlabel('Frequency');  
ylabel('Amplitude');  
title('Amplitude Response');
```

Bemærk, at filteret har en DC-forstærkning som er lig 1.064. Det betyder, at hvis vi påtrykker  $x(t) = 1$  på input (altså en DC) så vil systemet svare med værdier



(13)

# 3dB Frequencies ;

The analog filter has  $H(s)$  ;

$$H(s) = \frac{1}{sRC + 1}$$

⇓

$$H(j\Omega) = \frac{1}{j\Omega RC + 1} \Big|_{s=j\Omega}$$

⇓

$$|H(j\Omega)| = \left| \frac{1}{1 + j\Omega RC} \right|$$

⇓

$$|H(j\Omega)| = \frac{1}{\sqrt{1 + (\Omega RC)^2}}$$

The 3dB freq. is the freq. where the amplitude has decreased 3dB as related to DC ( $\Omega=0$ ).

$$-3\text{dB} = 20 \log x \Rightarrow x = 0.707 \sim \frac{1}{\sqrt{2}}$$

$$|H(j\Omega)|_{-3dB} = \frac{1}{\sqrt{2}} = \frac{1}{\sqrt{1 + (\Omega RC)^2}}$$

(14)

$$\Downarrow \quad \Omega \cdot RC = 1$$

$$\Downarrow \quad \Omega = \frac{1}{RC}$$

$$\Downarrow \quad 2\pi f = \frac{1}{RC}$$

$$\Downarrow \quad \underline{\underline{f = 159 \text{ Hz}}}$$

Similarly, we can calculate the freq. for the discrete-time filter;

$$|H(e^{j\omega})|_{-3dB} = \frac{1}{\sqrt{2}} \cdot |H(e^{j0})|$$

$$= \frac{1.06383}{1.41421} = 0.7522$$

$$\text{Thus; } |H(e^{j\omega_{3dB}})| = 0.7522$$

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⇓

$$0.125$$

$$\frac{0.125}{\sqrt{(\cos \omega_{3dB} - 0.8825)^2 + \sin^2 \omega_{3dB}}} = 0.7522$$

⇓

(After several calculations)

$$\begin{cases} \omega_{3dB} = 0.125 \text{ rad} \\ \omega_{3dB} = \frac{2\pi f_{3dB}}{f_s} \end{cases}$$

⇓

$$0.125 = \frac{2\pi \cdot f_{3dB}}{8000}$$

⇓

$$f_{3dB} = \underline{\underline{159 \text{ Hz}}}$$

So, the 3dB freq. is identical for the analog and the digital filter.

