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PROF. G. MASERA

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Berchialla Luca	236032
Laurasi Gjergji	238259
Mattei Andrea	233755
Lombardo Domenico Maria	233959

Introduction to the assigned problem

This relation deals with the design of a simple digital filter. From the given functional specifications, a final digital circuit has been implemented in VHDL including memories, control unit and data-path.

The design process will be analyzed using a bottom-up approach, starting from the individual components and their interconnections to the final custom designed FSM.

Overall specs description

As already mentioned, the final purpose of this activity is to implement a digital filter following the relation:

$$Y(n) = -0.5X(n) - 2X(n-1) + 4X(n-2) + 0.25X(n-3)$$

where $X(n)$ are the input stream data and $Y(n)$ the corresponding filtered data generated by the top equation.

The circuit starts by means of a *START* signal which enable a loading process of the input data into a 1 kByte memory. Then, the circuit automatically filters the data stream exploiting the equation already provided and loads the output filtered data into a second memory. Finally, the circuit reports end of process asserting an *HIGH* logical value to the *DONE* signal and awaits until the next *START*.

Every used component will now be discussed as single blocks, and individually debugged:

Memories

As already discussed, the final implementation will save the input data in a 1kByte memory while memorizing the output data into an output memory.

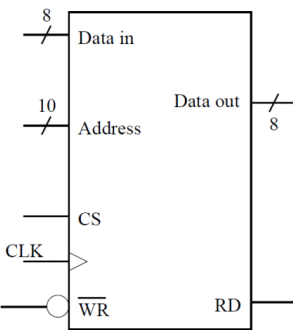


Figure 2 - Memory pinout

Figure 1: 1kByte memory

The memories store 1024 samples represented as 8 bit wide 2’s complement values. The writing operation is synchronous with the positive edge of the clock, while the reading is asynchronous.

The samples must be stored in order, from address 0 to 1023.

Once the desired address is selected and the RD signal asserted, the output data are shown in the *dataout* parallel output.

The figure below shows the testbench for the register file.

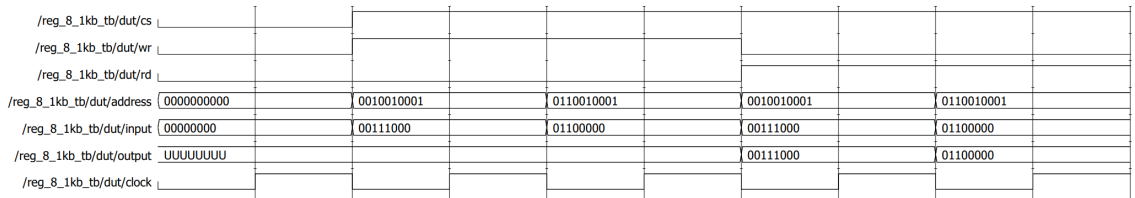


Figure 2: Memory testbench

Shift registers and registers

To be able to implement the already stated main equation, the circuit needs to perform 2s multiple multiplication and/or division.

The final implementation exploits the left and right shifting operations.

A 11 bit shift register has been implemented with parallel load and parallel output.

The *LOAD* signal must be asserted to load the *parallel input* data into the register, in a synchronous way with the positive edge of the clock.

The *SL* and *SR* commands permit the left or right shift of the internal memorized data.

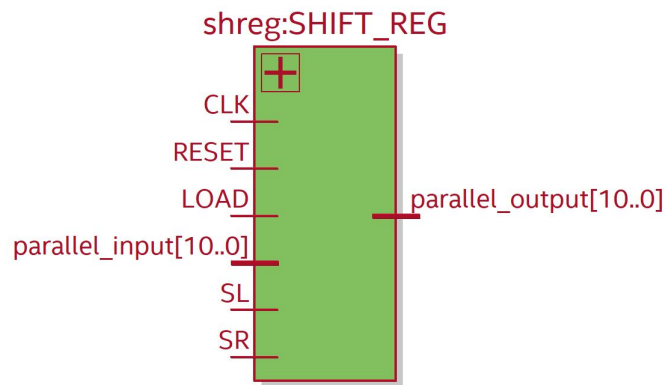


Figure 3: Shift register

The testbench for the 8-bit version is shown below:

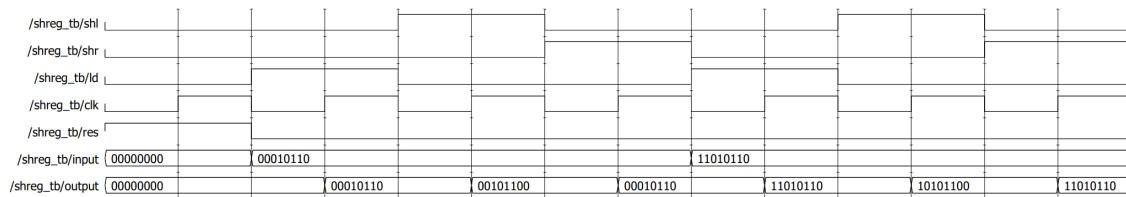


Figure 4: Shift register test bench

Adder

An 11-bit of a classical adder has been implemented, allowed to perform additions and subtractions by means of the Cin signal.

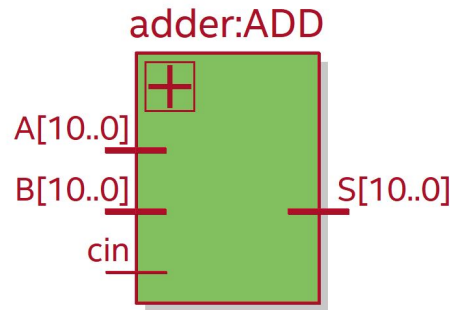


Figure 5: Adder

Below is shown the test bench of the designed adder.

The timing diagram shows four digital signals over time. The signals are labeled as follows:

- adder_8bit_tb/ci**: A clock signal (square wave).
- adder_8bit_tb/Ain**: An 8-bit input signal that transitions from 00010101 to 01101101.
- adder_8bit_tb/Bin**: An 8-bit input signal that transitions from 01001010 to 01101101.
- adder_8bit_tb/Sout**: An 8-bit output signal that transitions from 01110011 to 00100011.

The signals are shown as horizontal lines with vertical transitions. The output signal Sout transitions at approximately the same time as the input signals Ain and Bin.

Figure 6: Adder test bench

Data converters

Finally, since the circuit must be able to determine overflow/underflow and saturate the output if needed, it uses 2 blocks able to convert 8 bit signals to 11 bit and vice versa, as will be discussed in further sections.

The simulated test bench are shown below:

[illegible]

Figure 7: 8 to 11 bit converter; 11 to 8 bit converter

Counter

To be able to go through the various addresses an universal counter has also been implemented as shown: The counter can count up or down. Furthermore, the output

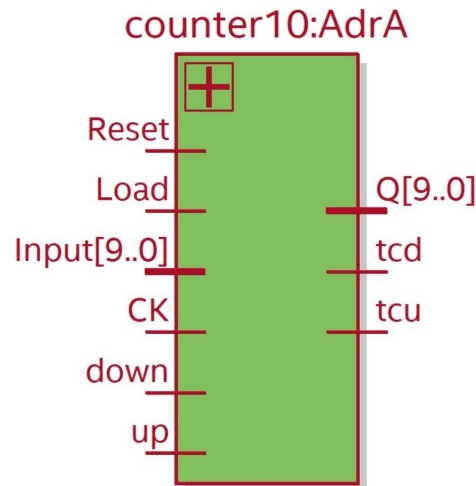


Figure 8: 10 bit counter

ports *tcd* and *tcu* have an high logical value when the counter reaches respectively its minimum or maximum possible value.

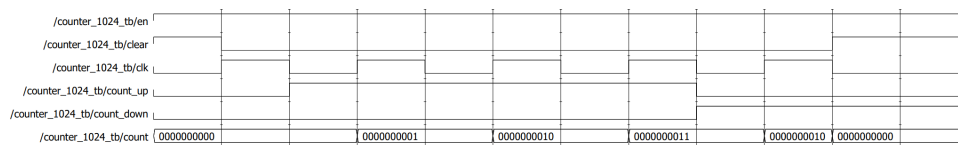


Figure 9: 10 bit counter test bench

Data path

The data path of the filter has been designed with two primary goals: use the minimum number of components and simplify the control unit.

All the computations in the circuits are made on 11 bits (range: 1023 to -1024), because the biggest absolute value of $Y(n)$ is 864.

The computation of $Y(n)$ is performed in several steps: 1) an 8 bits number is read from Memory A converted to 11 bits and stored in the shift register 2) the shift register applies the proper division/multiplication factor 3) the multiplied/divided number is saved in the register *Operand* 4) in the register *Result* is saved the content of the operation $Result + / - Operand$ with the proper sign 5) if the computation

of $Y(n)$ is not complete we restart from step 1, if it is complete the value stored in *Result* is converted to 8 bits and written in Memory B.

Control unit

The control unit has been implemented as a Moore finite state machine and is able to manage all the phases of the circuit: the acquisition and the computation.

In the acquisition phase (state:*Read1*) the memory is selected and configured to write, the *counterA* (which contains the address of A) is configured to count up. The cu remains in this state as long as the terminal counter up of *counterA* remains to 0.

The computation phase is done in two while loops: the first starts with *W1_start* and ends when the terminal counter up of *counterB* is 1 that means that all $Y(n)$ have been computed and written in Memory B. The second loop starts with *W2_start* and ends if the terminal counter down of *counterA* is 1 therefore there are no valid addresses to retrieve the $x(n-1/2/3)$ or if the terminal counter up of *loop* is 1. The *loop* counter